Design and analysis of CPW based shunt capacitive RF MEMS switch

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Abstract: This paper is about, the design and analysis of shunt capacitive RF MEMS switch with less actuation voltage, low insertion losses and high isolation losses. The switch design is incorporated the Electrostatics MEMS actuation technique with vertically deforming bridge. In terms of actuation voltage the switch performance is improved by choosing step type actuation structure with holes. The switch Radio Frequency (RF) performance is analysed over the frequency range from 0.6 to 40\,GHz. The major achievements in this work are actuation voltage is reduced to 4.2\,V for 0.9\,μm displacement, the return loss is below −16\,dB, the insertion loss is below −0.44\,dB, and the isolation loss is −20\,dB. The dielectric material used between the membrane and the CPW line is Aluminum Nitride (AlN) with dielectric constant 9.5. The substrate material used for the CPW transmission line is quartz with dielectric constant 3.9. The bridge is designed with meanders, step structure by using gold material with thickness 0.5\,μm. The switch upstate capacitance is capacitance ratio of the shunt capacitive switch is 65.22.

Subjects: Technology; Design; Electromagnetics & Communication; Electronic Devices & Materials

Keywords: RF MEMS switch; CPW transmission line; pull-in voltage; up capacitance; down capacitance; MEMS actuation mechanisms; electrostatic MEMS actuation; insertion losses; isolation losses

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PUBLIC INTEREST STATEMENT
RF MEMS Switches are mainly used to design the reconfigurable communication modules like antennas and filters. There are different types of RF MEMS switches i.e. series and shunt switches. Shunt switches has more operating frequency and offer high isolation. In this work we designed a shunt capacitive Radio Frequency (RF) operated MEMS Switch which offers more isolation, good insertion and requiring low actuation voltage. The switch is having good capacitance ratio. Here, the working of the proposed switch, if the membrane or beam is in upstate the switch is act like open circuit and \( R_{\text{fin}} \) is equal to zero, the membrane is downstate switch is acts like closed circuit and \( R_{\text{fout}} \) is equal to zero.
1. Introduction
The RF MEMS switches have a predominant role in the design of present day advanced communication applications. To design reconfigurable microwave antennas and filters RF MEMS switches are preferable than solid state devices like FET and PIN diode (Molaei & Ganji, 2017). MEMS technology has scope for miniaturization when compared to CMOS and GaAs technologies. The major advantages in MEMS technology based RF MEMS switch are better linearity, high isolation, low noise, low power consumption, and high operating frequency (Rebeiz & Muldavin, 2001). The RF MEMS switch performance depends on return losses, isolation losses, insertion losses, switching time and actuation voltage (Bakri-Kassem & Mansour, 2015). The materials used in the design also decides the performance, stiffness of the cantilever depends on the contact material used in the switch (Ravirala et al., 2017; Srinivasa Rao & Thalluri, 2016). MEMS technology offers different actuation mechanisms like electrostatic, magneto static, piezoelectric and thermal. In this electrostatic actuation technique is preferable because other techniques require more dc voltage to actuate the structure (Rebeiz & Muldavin, 2001). The RF MEMS switches are electrically classified as series type and shunt type. Based on contact type the switches are classified as capacitive and resistive (Barbato & Meneghesso, 2015). Capacitive switches are preferable for high frequency and resistive switches are preferable for low frequency applications (Guha, Kumar, Parmar & Baishya, 2016). In capacitive switches the isolation losses mainly depends on the dielectric material used between the electrodes, generally silicon nitride, silicon dioxide and Aluminum Nitride are used. The MEMS structure may be bridge type (or) cantilever type (or) diaphragm type (Mafinejad, Zarghami & Kouzani, 2013). The bridge structure with supporting meanders and step type will help to minimize the actuation voltage (Chawla & Khanna, 2014; Molaei & Ganji, 2017).

2. Theoretical analysis
In this paper an electrostatically actuated shunt type capacitive RF MEMS Switch is designed by adopting new techniques in the shape of structure like meanders, step, and holes to the structure as shown in Figure 4.

2.1. CPW transmission line
CPW and Microstrip transmission lines are used to design RF MEMS switches. The return losses, operating frequency depends on the dimensions of the transmission line. In this paper co-planar wave (CPW) transmission line is used to design the shunt capacitive RF MEMS switch. In CPW transmission line both the conductors are in same plane as shown in the Figure 1.

Here \( t_h \) is the height of the substrate, \( t_w \) is the width of the CWP line and \( t_g \) is the gap between the CPW planes. \( \varepsilon_r \) is the substrate dielectric constant generally in between 3.3 and 4.7. The switches are designed for radio frequency applications so the CPW line metal thickness is considered as 0.0001 = 0 μm.

2.2. Beam structure
To design an RF MEMS switch different structures like cantilever, diaphragm, and beam or bridge are preferable. The shape and dimensions of the beam decides the magnitude of actuation voltage and isolation losses. In this paper an RF MEMS switches is design using beam structure with meanders, step and holes as shown in Figure 2.

The mass of the structure is associated with the mass of the meanders \( (m_d) \), mass of the steps \( (m_s) \) and mass of the membrane \( (m_m) \). The total effective mass \( (m_t) \) can be calculated by subtracting holes mass \( (m_h) \) from overall mass.
The switch is an electrostatically actuated switch, which require a actuation voltage \( V_p \) to deform the structure can be expressed as:

\[
V_p = \sqrt{\frac{8k_{eff}}{27\varepsilon_0}} \left( g_1 + \frac{t_d}{\varepsilon_r} \right) \frac{3}{8}
\]

where \( k_{eff} \) is the effective spring constant of the beam i.e. it is the overall spring constant of the meanders, step and membrane.

The spring constant will decide the required actuation voltage i.e. more spring constant means more actuation voltage is required. The deforming structure spring constant generally expressed as:

\[
k = \frac{EWt^3}{l^3}
\]

where \( E \) is the young's modules, \( W \) is the width, \( t \) is the thickness, \( l \) is the length of the beam.

The meander structure used in this paper is associated with different beams as shown in Figure 3 and Table 1. Each beam is associated with own spring constant i.e. \( k_1, k_2, k_3, k_4, k_5 \). The mean spring constant \( (k_m) \) can be expressed as:

\[
\frac{1}{k_m} = \frac{1}{k_1} + \frac{1}{k_2} + \frac{1}{k_3} + \frac{1}{k_4} + \frac{1}{k_5}
\]

The overall effective spring constant \( (k_{eff}) \) associated with actuation structure can be expressed as:

\[
k_{eff} = 4k_m
\]

Here the mean spring constant is multiplied by 4 because the structure is associated with for meanders on four sides as shown in Figure 2.

The capacitive RF MEMS switch performance can be improve by decreasing the upstate capacitance and increasing the downstate capacitance. The capacitance variation mainly depends on the dielectric material used between the bottom electrode and membrane, generally silicon nitride or silicon dioxide or Aluminum nitride is used as dielectric material. The dielectric constant of these materials is in between 3.3 and 9.5 (see Tables 2 and 3). The upstate capacitance \( (C_u) \) can be expressed as:
where $A = W \times w$ is effective area between membrane and bottom electrode as shown in Figure 5.

When an actuation voltage is applied the membrane associated with the structure start deforming and come to downstate. Under this condition the downstate capacitance ($C_d$) can be expressed as:

$$C_d = \frac{\varepsilon_0 \varepsilon_r A}{t_d}$$

### Table 1. Meander with non-uniform spring constants dimensions

| Name | Length (μm) | Width (μm) | Thickness (μm) |
|------|-------------|------------|----------------|
| K1   | 10          | 5          | 0.5            |
| K2   | 30          | 5          | 0.5            |
| K3   | 10          | 5          | 0.5            |
| K4   | 40          | 5          | 0.5            |
| K5   | 10          | 5          | 0.5            |

### Table 2. Switch materials and properties

| Name          | Material    | $\varepsilon_r$ | Young's modules ($E$) | Electrical conductivity ($\sigma$) |
|---------------|-------------|-----------------|------------------------|----------------------------------|
| Substrate     | Quartz      | 3.9             | -                      | -                                |
| Bridge        | Gold        | -               | 70 GPa                 | -                                |
| CPW lines     | Gold        | -               | -                      | 4.56e6 (S/m)                     |
| Dielectric material | AlN       | 9.5             | -                      | -                                |

### Table 3. Switch upstate and downstate capacitance

| Parameter       | Equation                                           | Theoretical value (in F) | Practical value (in F) |
|-----------------|---------------------------------------------------|--------------------------|------------------------|
| Upstate capacitance ($C_u$) | $C_u = \frac{\varepsilon_0 A}{g_1 + \frac{t_d}{\varepsilon_r}} + C_f$ | $46 \times 10^{-15}$ | $57.08 \times 10^{-15}$ |
| Downstate capacitance ($C_d$) | $C_d = \frac{\varepsilon_0 A}{t_d} + \frac{t_d}{\varepsilon_r} C_f$ | $3.78 \times 10^{-12}$ | $3.74 \times 10^{-12}$ |
| Capacitance ratio ($C_r$) | $C_r = \frac{C_{max}}{C_{min}} = \frac{C_d}{C_u}$ | 82                       | 65.22                  |

$$C_u = \frac{\varepsilon_0 A}{g_1 + \frac{t_d}{\varepsilon_r}} + C_f$$

(6)
Generally the capacitance ratio is defined as the ratio of downstate capacitance to upstate capacitance i.e. $C_d/C_u$.

The switch speed is decided by the switching time of the switch, the switching time depend on the actuation voltage ($V_p$), supply voltage ($V_s$), and the resonant frequency ($\omega_0$). The actuation structure resonant frequency is given as:

$$\omega_0 = \sqrt{\frac{k_{\text{eff}}}{m}}$$ \hspace{1cm} (8)

where $k_{\text{eff}}$ is effective spring constant, $m$ is the total mass associated with the deforming structure.

Generally the RF MEMS switches switching time is in milliseconds. The switching time of the switch is expressed as:

$$t_s = \frac{3.67}{\omega_0} \frac{V_p}{V_s}$$ \hspace{1cm} (9)

The RF MEMS switch designed in this paper is shown in Figures 4 and 5 in top view and side view respectively.

3. Design and simulation

A micro level RF MEMS switch can be design and simulate using Finite Element Method (FEM) or Finite Element Analysis (FEA) tool’s. The capacitive shunt RF MEMS switch designed in this paper is using COMSOL FEM tool. The designed switch performance is analysed over the frequency range from 0.6 to 40 GHz. The overall switch is designed on a quartz die with dimensions 220 $\mu$m length, 220 $\mu$m width, 30 $\mu$m height. The switch dementions are shown in Table 4 and Figure 6.

The switch designed in this paper, works depending on electrostatic actuated, i.e. if the actuation voltage is not applied the actuation structure is in upstate and the capacitance offered by the shunt switch is very low in the order of femto farad under this condition the input Radio frequency input signal will go to the output ($RF_{\text{out}} = RF_{\text{in}}$). If an actuation voltage is applied to the switch electrodes then the actuation structure will come to downstate and the switch offer a capacitance in the order of pico farad under this condition the input Radio frequency input signal will not go to the output ($RF_{\text{out}} = 0$).
The RF MEMS switch designed in this paper, is designed with a membrane structure associated with the non uniform meander’s which is helped to reduce the actuation voltage. Here we achieved a displacement of 0.9 μm for actuation voltage of 4.2 V as shown in Figure 7. The switch actuation structure is a step type structure because of this also the actuation voltage is reduced significantly. The switch Radio Frequency (RF) properties are analyzed over 0.6–40 GHz frequency range, and we observed that the switch is offering a return losses in the range −42 to −16 dB as shown in Figure 8(a), insertion losses in the range −0.01 to 0.45 dB as shown in Figure 8(b), isolation losses of −20 dB at 21 GHz is shown in Figure 8(c).

Uniform rectangular holes are formed in the beam structure to minimize the mass of the beam, each hole dimension is 5 μm width and 10 μm length. A step is taken in the structure, the dimensions of the anchors used in this is 2 μm height, 5 μm length and 5 μm width.

The overall mass of the actuation structure is $57514 \times 10^{-15}$ kg, the mass removed by the rectangular holes is $7237 \times 10^{-15}$ kg, so final mass of the structure is $50276 \times 10^{-15}$ kg, because of the holes in the structure 12.58% of the mass is removed from the overall mass.

### Table 4. Switch dimensions

| Parameter                                      | Value (μm) |
|------------------------------------------------|-------------|
| CPW substrate height ($t_h$)                   | 30          |
| CPW substrate dielectric constant ($\varepsilon_r$) | 3.9         |
| Gap between strip and ground ($t_s$)           | 15          |
| Width of the strip ($t_w$)                     | 60          |
| Length of CPW lines                            | 200         |
| Membrane width ($W$)                           | 80          |
| Bottom electrode width ($w$)                   | 60          |
| Gap between membrane and bottom electrode ($g_1$) | 0.9     |
| Overlap Area ($A = W \times w$)               | 80 $\times$ 60 |
| Dielectric thickness ($t_d$)                   | 0.1         |
| Dielectric constant ($\varepsilon_r$)          | 9.5         |
| Gap between transmission line and membrane ($g_2$) | 2.5     |
| Bridge thickness ($b_t$)                       | 0.5         |
If the beam is in upstate the gap \( g_1 \) between the membrane and the bottom electrode is 1 μm and a parallel plate capacitance of 0.05708 pF is achieved. When an actuation voltage of 4.2 V is applied the beam deforms and come to downstate therefore the resultant gap \( g_1 \) is 0.1 μm, and the resultant capacitance is 3.74 pF. So, the capacitance ratio \( \frac{C_d}{C_u} \) for the proposed switch is 65.22 (see Table 5).
4. Conclusion
In this paper a capacitive shunt RF MEMS Switch is designed and analysed over the frequency range from 0.6 to 40 GHz using FEM Tool. The actuation structure used in the design is anchored with the meanders, having step and holes to the membrane these all things helped to reduce the actuation voltage to 4.2 V. The beam and the CPW lines are designed using Gold (Au) material. The dielectric used between the electrodes is an Aluminum nitride (AlN) which is helped to improve the quick change capacitance. The structure is in upstate the capacitance is 0.05708 pF, and structure is in downstate the capacitance is 3.74 pF. The capacitance ratio for the proposed switch is 65.22. The RF MEMS switch designed in this paper is achieved a displacement of 0.9 μm for actuation voltage of 4.2 V. Return losses in the range −42 to −16 dB, insertion losses in the range −0.01 to −0.45 dB, isolation losses of −20 dB. Based on the switch performance the switch can be used in X, Ku band applications.

Authors’ contributions
All the authors contributed in the design and analysis of RF MEMS capacitive shunt Switch for the current study by COMSOL and HFSS. T. Lakshmi Narayana and K. Girija Sravani performed the coding in MATLAB and drafted the manuscript. K. Srinivasa Rao supervised the study and advised on the draft and the corrections in the manuscript. All authors read and approved the final manuscript.

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Competing interests
The authors declare that they have no competing interests.

Table 5. Different capacitive shunt RF MEMS switches comparison

| Switches @ 40 GHz | Rebeiz (2003) | Blondy et al. (2004) | Ziaei, Dean, and Mancuso (2005) | Fernández-Bolaños, Tsamados, Dainesi, and Ionescu (2009) | Proposed |
|-------------------|---------------|---------------------|-------------------------------|----------------------------------------------------------|----------|
| Suspenders material | 0.9 μm Ti/Au | 0.35/7.5 μm Ti–Au/Au | 0.5 μm Al | 0.9 μm Au/Ni/Au | 0.5 μm Au |
| Dielectric | 150 nm Si3N4 εr = 7.6 | Dielectric Less | 400 nm PZT εr = 190 | 300 nm AlN εr = 9.8 | 100 nm AlN εr = 9.5 |
| Air gap | 1.5 μm | 0.3 μm | 2.5–3 μm | 2–2.5 μm | 0.9 μm |
| Upstate capacitance (C_u) | 70 fF | 224 fF | – | 40 fF | 57.05 fF |
| Downstate capacitance (C_d) | 2.7 pF | 2.2 pF | – | 1.55 pF | 3.74 pF |
| Capacitance ratio (C_d/C_u) | 38 | 10 | 400 | 38 | 65.22 |
| Insertion Losses | 0.1 dB | 1.5 dB | 0.1 dB | 0.2 dB | 0.1–0.4 dB |
| Isolation losses | 35 dB | 20 dB | 38 dB @ 10 GHz | 38.5 dB | 20 dB |
| Actuation voltage | 25–30 V | 30 V | 35–40 V | 12 V | 4.2 V |

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