Research Article

Nonlinear Control for Bioprocesses with Model Uncertainties and External Disturbances

Cecilia Fernández, Nadia Pantano, Leandro Rodriguez, and Gustavo Scaglia

Instituto de Ingeniería Química, Universidad Nacional de San Juan (UNSJ), CONICET, Avenida Libertador General San Martin Oeste 1109, San Juan J5400ARL, Argentina

Correspondence should be addressed to Cecilia Fernández; mcfernandez@unsj.edu.ar

Received 21 February 2020; Revised 26 April 2021; Accepted 7 May 2021; Published 18 May 2021

Academic Editor: Shangce Gao

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In this paper, a new alternative for profiles tracking control considering additive uncertainties is proposed. Based on a previously presented work about a nonlinear and multivariable controller design for a fed-batch bioethanol production, parametric uncertainty and process disturbance are taken into account to find a more reliable control strategy for a successful industrial implementation. To decrease the uncertainties effect, an approach based on the error estimation using Newton’s backward interpolation is included in the design equations. The proposed modification assures the error convergence to zero (demonstration is shown) despite the uncertainties, which is one of the main contributions of this work. A comparison between the new, the original proposal, and another methodology is exposed.

1. Introduction

Biological processes are becoming more frequent nowadays due to the wide variety of products obtained from them [1] and their possibility of making some processes environmentally friendly while high standard products are obtained [2, 3]. The fed-batch operation mode is which most interest awake for its main benefits [4, 5]. One of the most interesting advantages is the substrate concentration regulation in the cultivation medium by a suitable feed rate profile [6], obtaining better production yields and minimizing the production costs [5, 7]. However, bioprocesses control is required to follow a certain feed flow rate and get stability and the best productivity [8]. Furthermore, the mathematical representation of the process is the key to achieve good results.

A mathematical model provides a map from inputs to responses. The quality of a model depends on how closely its responses match those of the true plant. However, a model set that includes the true physical plant can never be constructed [9]. Generally, a bioprocess modeling presents particular difficulty in their parameters determination caused by the poorly understood microorganism’s dynamics (multivariable and highly nonlinear dynamics), the strongly coupled variables, and the presence of numerous external disturbances, which leads to having many modeling uncertainties [10]. Furthermore, sometimes those parameters are determined without a previous model analysis, or their values are not informed with their respective confidence intervals [11, 12]. Moreover, time-varying parameters are usually assumed as constants [13]. Also, the uncertainties related to the processing technologies parameters are rarely considered [14]. The dismissal of all these aspects leads to a poor real-life representation and, consequently, to a bad performance with severe risks [15, 16]. Therefore, the main task to guarantee the bioprocess quality implies finding a way to control these distortions [17–19]. For this reason, the development of new control schemes that reduce the effect of uncertainties in the tracking error has become an attention focus for the scientific community [11, 15].

One of the problems to take into account is the feasibility of the occurrence of events in systems with uncertainty, for this, the uncertainty theory was introduced [20]. On the other hand, a very interesting way to describe uncertainty is
using the uncertain fractional differential equation (UFDE), which allows keeping a record of some properties to be considered a posteriori, which is fundamental in the evolution of uncertainties [21, 22]. Furthermore, the contemplation of uncertainties is approached with different strategies, depending on the nature of the system under study [22–24]. One of the most used strategies for model parameter identification and/or estimation involves an offline optimization using a nominal model of the process [25–30]; the main disadvantage of this methodology is that the variability of microorganisms decreases the possibility of batch-to-batch repeatability. To improve the results of nominal optimization, a methodology called “run to run” optimization appears, which uses previous runs information to optimize the operation of subsequent ones [31–37]. Another strategy is the online optimization of the model parameters [38–45]. This kind of optimization is difficult to perform since the available models might only be locally valid and thus inappropriate for predicting final concentrations [46].

On the other hand, many scientists have developed some feedback control strategies to deal with bioprocess uncertainties. Optimal control, nonlinear model predictive control, hybrid control, adaptive control, fuzzy control, tracking control, and neural network are examples of them [47, 48]. But, due to the online implementation difficulty, the high computational cost, the imprecise mathematical models, and online solutions, their use for bioprocesses is limited [7] and is still a research topic [49].

For the specific case of ethanol production defined by Hunag et al. [36], Fernández et al. [50] presented a controller design focused on looking for control actions, to track predefined concentration profiles. As the controller structure comes from the mathematical model of the process, it can be implemented in many systems. The procedure is characterized by its simplicity, versatility, and precision. Besides, only basic knowledge of numerical methods and linear algebra is needed to implement it. One of the main contributions of that work was to achieve the tracking error convergence to zero. Also, the technique was tested against different disturbances and compared with a typical PID controller.

This manuscript aims to improve the control strategy presented in [50]. In this sense, predefined trajectories can be tracked while estimating the difference between the model and the real plant in each sampling time (error estimation). To reach this objective, another term is incorporated into the controller structure, which symbolizes model uncertainties and outside instabilities that are evaded with Newton’s backward interpolation. Moreover, the higher the interpolation order, the better the estimation and the smaller the tracking error. Besides, this change guarantees uniformity in the signal and progressive reduction of tracking error, achieving improvements of up to 98% in some cases.

In [51], the problem of optimal profiles tracking control under uncertainties for a fed-batch bioprocess with two control actions is addressed with excellent results. In that manuscript, the authors add tracking error integrators in the control action calculation to reduce the additive uncertainties effect. However, the strategy presented in this paper has the advantage that adding the additive uncertainty term does not increase the order of the system, making mathematical development even simpler. Moreover, the ethanol bioprocess is an underactuated system with only one control action, so the control challenge is even greater than in [51]. Thus, a solution to the real trouble of multivariable and nonlinear tracking control in the presence of additive uncertainties is proposed, without increasing the system order.

The manuscript is presented as follows: first, a summary of the process under study and the original control technique is described to contextualize the problem. Second, the contribution of this work is detailed, including the relevant demonstrations. Third, algorithms are tested and compared. Lastly, conclusions are shown.

2. Process and Control Description

2.1. Mathematical Model of the Process. Hunag et al. [36] proposed a mathematical model for a fed-batch ethanol production, using Saccharomyces diastaticus yeast to carry out the fermentation. The temperature was fixed at 35.8°C, airflow at 1.5vvm, and pH at 5.0. The only system input is the feed rate (U), which is a 50% glucose and 50% fructose combination. The state variables are biomass (X), ethanol (P_1), glycerol (P_2), glucose (S_1), and fructose (S_2) concentration inside the bioreactor:

\[
\begin{align*}
\dot{X}(t) &= (\mu_1 + \mu_2)X - U \frac{V}{\lambda} X, \\
\dot{S}_1(t) &= -\left(\frac{q(S_1/P_1)}{Y(P_1/S_1)} + \frac{q(S_1/P_2)}{Y(P_1/S_2)}\right)X + U \frac{V}{\lambda} (S_f - S_1), \\
\dot{S}_2(t) &= -\left(\frac{q(S_2/P_1)}{Y(P_1/S_1)} + \frac{q(S_2/P_2)}{Y(P_1/S_2)}\right)X + U \frac{V}{\lambda} ((1 - \lambda)S_f - S_2), \\
\dot{P}_1(t) &= -\frac{q(S_1/P_1) + q(S_1/P_2)}{Y(P_1/S_1)}X - U \frac{V}{\lambda} P_1, \\
\dot{P}_2(t) &= -\left(\frac{q(S_2/P_1) + q(S_2/P_2)}{Y(P_1/S_1)}\right)X - U \frac{V}{\lambda} P_2,
\end{align*}
\]

where
\[ V(t) = U, \]

\[ \dot{q} = \frac{\gamma_{s1} S1}{K_{s1} + S1} \frac{k_{s1} p1}{k_{s1} p1 + P1}, \]

\[ \mu_1 = \frac{\mu_n S1}{(K_{s1} + S1 + S1^2/K_{s1})} \frac{k_{p1}}{(K_{p1} + P1 + P1^2/K_{p1})}, \]

\[ \dot{q} = \frac{\gamma_{s2} S2}{K_{s2} + S2} \frac{k_{s2} p2}{k_{s2} p2 + P2}, \]

\[ \mu_2 = \frac{\mu_n S2}{(K_{s2} + S2 + S2^2/K_{s2})} \frac{k_{p2}}{(K_{p2} + P2 + P2^2/K_{p2})}, \]

\[ \dot{q} = \frac{\gamma_{s1} S1}{K_{s1} p1 + S1} \frac{k_{s1} p1}{k_{s1} p1 + P1}, \]

\[ \dot{q} = \frac{\gamma_{s2} S2}{K_{s2} p1 + S2} \frac{k_{s2} p1}{k_{s2} p1 + P1}. \]

\[ (2) \]

\[ \sigma_{ref} = \frac{k_q (\sigma_{ref,n} - \sigma_n)}{error_n} \]

\[ (3) \]

\[ \Delta \sigma_{ref,n} + \sigma_{ref,n} - \sigma_n = \sigma_{ref,n} = \sigma_{ref,n+1} - k_q (\sigma_{ref,n} - \sigma_n) \]

\[ (4) \]

\[ \mu_1 + \mu_2 \]

\[ \Delta X_n = (\mu_1 + \mu_2) X_n - \frac{U_{n} X_n}{V_n}, \]

\[ \Delta S_{1,n} = \frac{\dot{q}(s_1/p_1)}{Y(p_1)} + \frac{\dot{q}(s_2/p_2)}{Y(p_2)} X_n + \frac{U_n}{V_n} (\lambda S_f - S_{1,n}), \]

\[ \Delta S_{2,n} = \frac{\dot{q}(s_1/p_1)}{Y(p_1)} + \frac{\dot{q}(s_2/p_2)}{Y(p_2)} X_n + \frac{U_n}{V_n} ((1 - \lambda) S_f - S_{2,n}), \]

\[ \Delta P_{1,n} = \frac{\dot{q}(s_1/p_1) + \dot{q}(s_2/p_2)}{Y(p_1)} X_n - \frac{U_n}{V_n} P_{1,n}, \]

\[ \Delta P_{2,n} = \frac{\dot{q}(s_1/p_1) + \dot{q}(s_2/p_2)}{Y(p_1)} X_n - \frac{U_n}{V_n} P_{2,n}. \]

\[ (6) \]
System equation (7) must have an exact solution to find $U_n$. Consequently, $b_n$ have to be a linear combination of $A_n$ [54]; that is to say, $A_n$ and $b_n$ must be parallel and with the same sense. One way to accomplish this is

$$\cos(A_n, b_n) = \frac{\langle A_n, b_n \rangle}{\|A_n\| \|b_n\|} = 1.$$  \hspace{1cm} (8)

In (8), the operation between $\langle >$ and $\|\|$ represent the inner product and the vectors norm in $R^n$ space, respectively. The angle between $A$ and $b$ is $\theta = 0^\circ$; this implies a positive $U_n$ value. If $\theta = 180^\circ$, $U_n$ would be negative, which does not make physical sense because $U_n$ is a flow.

As stated in the Introduction, bioprocess systems generally involve several control objectives that may be conflicting, to balance these objectives during the design of the presented controller, the selection of a “sacrificed variable” is required. It is denoted as $S_{ref}$, which guarantees that (7) has an exact solution and that the references are followed. For more details on its selection and calculation, see [50]. Finally, $U_n$ is obtained using least squares [54]:

$$\|e_n\| = \left(\frac{(X_{ref,n} - X_n)^2}{\max X_{ref,n}} + \frac{(P_{1ref,n} - P_{1,n})^2}{\max P_{1ref,n}} + \frac{(P_{2ref,n} - P_{2,n})^2}{\max P_{2ref,n}} + \frac{(S_{2ref,n} - S_{2,n})^2}{\max S_{2ref,n}}\right)^{\frac{1}{2}},$$  \hspace{1cm} (11)

$$E_{p,1} = T_s \sum_{n=1}^{j} \|e_n\|,$$  \hspace{1cm} (12)

$$E_{p,2} = T_s \left(\sum_{n=1}^{j} \|e_n\|nT_s\right).$$  \hspace{1cm} (13)

2.3. Controller Tuning. In [50], the Monte Carlo algorithm was used to find the controller parameters that make the accumulated error minimum. It consists of simulating the bioprocess $N$ times with random $k_o$ [55]:

$$N \geq \frac{\log(1/\delta)}{\log(1/(1 - \varepsilon))}.$$  \hspace{1cm} (10)

In (10), $\delta$ is the confidence and $\varepsilon$ is the accuracy.

However, many other strategies can be used to tune this controller. In [56], a genetic algorithm and a hybrid one are proposed.

Next, the “tracking error ($\|e_n\|$)” and “total error ($E_p$)” concepts are introduced:
The matrix system generically, error convergence to zero when $n$ tends to infinity is achieved. If the discrete system is given by equation (1), the control action is calculated with equation (9), and $k_\sigma$ take values between zero and one ($0 < k_\sigma < 1$); then, the tracking error convergence to zero when $n$ tends to infinity is achieved.

Demonstration [50]:
Substituting the sacrificed variable in (7) and expressing the matrix system generically,

$$
\begin{bmatrix}
  a_1 \\
  a_2 \\
  a_3 \\
  a_4 \\
  a_5 \\
\end{bmatrix}
U_n =
\begin{bmatrix}
  b_1 \\
  b_2 \\
  b_3 \\
  b_4 \\
  b_5 \\
\end{bmatrix}
$$

$$
U_n = \frac{a_1b_1 + ((a_2^2b_1)/a_1) + ((a_2^2b_2)/a_1) + ((a_2^2b_3)/a_1)}{a_1^2 + a_2^2 + a_3^2 + a_4^2 + a_5^2} = \frac{(b_1)(a_1)(a_1^2 + a_1^2 + a_3^2 + a_4^2 + a_5^2)}{a_1^2 + a_2^2 + a_3^2 + a_4^2 + a_5^2} = \frac{b_1}{a_1}
$$

$V_n((X_{ref,n+1} - k_b(X_{ref,n} - X_n) - X_n)T_s) - (\mu_1(S_{lez,n}, P_{1,n}) + \mu_2(S_{2,n}, P_{1,n})X_n)$

Replacing (17) in (7),

$$
X_{n+1} = X_{ref,n+1} - k_b(X_{ref,n} - X_n) + T_s[\mu_1(S_{lez,n}, P_{1,n}) - \mu_1(S_{1,n}, P_{1,n})]X_n
$$

Before, the tracking error for $X$ is defined as

$$
e_{X_{n+1}} = X_{ref,n+1} - X_{n+1}
$$

Substituting (18) in (19),

$$
e_{X_{n+1}} = k_b(X_{ref,n} - X_n) - T_s[\mu_1(S_{lez,n}, P_{1,n}) - \mu_1(S_{1,n}, P_{1,n})]X_n
$$

The $\mu_i(S_{inz}, P_{inz})$ Taylor approximation in the desired value $\mu_i(S_{lez}, P_{inz})$ is

Applying least squares to (14),

$$
U_n = (A^T A)^{-1} A^T b = \frac{a_1b_1 + a_2b_2 + a_3b_3 + a_4b_4 + a_5b_5}{a_1^2 + a_2^2 + a_3^2 + a_4^2 + a_5^2}
$$

From (14),

$$
a_1 \frac{b_1}{a_2} \rightarrow b_2 = a_2b_1, \\
a_3 \frac{b_1}{a_3} \rightarrow b_3 = a_2b_1, \\
a_4 \frac{b_1}{a_4} \rightarrow b_4 = a_2b_1, \\
a_5 \frac{b_1}{a_5} \rightarrow b_5 = a_2b_1.
$$

Placing (16) in (15),

$$
\mu_1(S_{1,n}, P_{1,n}) = \mu_1(S_{lez,n}, P_{1,n}) + \frac{d\mu_i(S_{1,n}, P_{1,n})}{ds_i} \frac{(S_{inz} - S_{lez,n})}{S_{inz} - S_{lez,n} + \theta}, \quad \theta \rightarrow 0 < \theta < 1
$$

Placing (21) in (20),

$$
e_{X_{n+1}} = k_b(X_{ref,n} - X_n) - T_s[\mu_1(S_{lez,n}, P_{1,n}) + \frac{d\mu_i(S_{1,n}, P_{1,n})}{ds_i} \frac{(S_{inz} - S_{lez,n})}{S_{inz} - S_{lez,n} + \theta} - \mu_1(S_{lez,n}, P_{1,n})]X_n.
$$

$$
e_{X_{n+1}} = k_b(X_{ref,n} - X_n) + T_s \left[ \frac{d\mu_i(S_{1,n}, P_{1,n})}{ds_i}(S_{inz} - S_{lez,n}) \right] \frac{X_n}{S_{inz} - S_{lez,n} + \theta}.
$$
Table 2: Nomenclature, description, and values of parameters.

| Parameter    | Description                                      | Value     |
|--------------|--------------------------------------------------|-----------|
| $\mu_{m1}$   | Maximum specific growth rate coefficient for yeast on glucose ($h^{-1}$) | 1.8823    |
| $\mu_{m2}$   | Maximum specific growth rate coefficient for yeast on fructose ($h^{-1}$) | 1.7098    |
| $Y_{P1/S1}$  | Yield coefficient for ethanol from glucose       | 0.5085    |
| $Y_{P2/S1}$  | Yield coefficient for glycerol from glucose      | 0.5331    |
| $Y_{P1/S2}$  | Yield coefficient for ethanol from fructose      | 0.5098    |
| $Y_{P2/S2}$  | Yield coefficient for glycerol from fructose     | 0.4462    |
| $K_{S1}$     | Saturation coefficient for cell growth on glucose (g/L) | 159.75    |
| $K_{S1I}$    | Inhibition coefficient for cell growth on glucose (g/L) | 94.233    |
| $K_{P1}$     | Saturation coefficient for cell growth on ethanol (g/L) | 238.39    |
| $K_{P1I}$    | Inhibition coefficient for cell growth on ethanol (g/L) | 2.7378    |
| $K_{S2}$     | Saturation coefficient for cell growth on fructose (g/L) | 9.0726    |
| $K_{S2I}$    | Inhibition coefficient for cell growth on fructose (g/L) | 9.0048    |
| $K_{P2}$     | Saturation coefficient for cell growth on glycerol (g/L) | 35.958    |
| $K_{P2I}$    | Inhibition coefficient for cell growth on glycerol (g/L) | 9.9722    |
| $K_{S1P1}$   | Saturation coefficient for ethanol production on glucose (g/L) | 1.3409    |
| $K_{S1P1I}$  | Inhibition coefficient for ethanol production on glucose (g/L) | 18.612    |
| $K_{S2P1}$   | Saturation coefficient for ethanol production on fructose (g/L) | 0.9129    |
| $K_{S2P1I}$  | Inhibition coefficient for ethanol production on fructose (g/L) | 1000      |
| $K_{S1P2}$   | Saturation coefficient for glycerol production on glucose (g/L) | 6.7116    |
| $K_{S1P2I}$  | Inhibition coefficient for glycerol production on glucose (g/L) | 0.5863    |
| $K_{S2P2}$   | Saturation coefficient for glycerol production on fructose (g/L) | 0.4310    |
| $K_{S2P2I}$  | Inhibition coefficient for glycerol production on fructose (g/L) | 1.150     |
| $\gamma_{S1P1}$ | Coefficient of maximum specific ethanol production rate for yeast on glucose ($h^{-1}$) | 1.5051    |
| $\gamma_{S2P1}$ | Coefficient of maximum specific ethanol production rate for yeast on fructose ($h^{-1}$) | 0.3321    |
| $\gamma_{S1P2}$ | Coefficient of maximum specific glycerol production rate for yeast on glucose ($h^{-1}$) | 0.0023    |
| $\gamma_{S2P2}$ | Coefficient of maximum specific glycerol production rate for yeast on fructose ($h^{-1}$) | 0.1609    |
| $\lambda$    | Proportion of glucose and fructose                | 0.5       |
| $S_f$        | Sugar total feed concentration (g/L)              | 300       |

Figure 1: Cells, ethanol, glycerol, and fructose reference concentrations along the process. Reference feed flow rate.
Following the same procedure for the other variables and joining the final expressions,

\[
\begin{bmatrix}
e_{X,n+1} \\
e_{S_1,n+1} \\
e_{S_2,n+1} \\
e_{P_1,n+1} \\
e_{P_2,n+1}
\end{bmatrix}
= \begin{bmatrix}
k_X & 0 & 0 & 0 & 0 \\
0 & k_{S_1} & 0 & 0 & 0 \\
0 & 0 & k_{S_2} & 0 & 0 \\
0 & 0 & 0 & k_{P_1} & 0 \\
0 & 0 & 0 & 0 & k_{P_2}
\end{bmatrix}
\begin{bmatrix}
e_{X,n} \\
e_{S_1,n} \\
e_{S_2,n} \\
e_{P_1,n} \\
e_{P_2,n}
\end{bmatrix}
\]

In equation (23), \( L \) is a linear system and \( NL \) is a bounded nonlinearity [50]. Note that if \( k_\sigma = 0 \), the reference is reached in one step. Thus, if \( 0 < k_\sigma < 1 \), the tracking error tends to zero when \( n \to \infty \) [50, 57].

2.4. Steps to Implement the Controller

Step 1. Define \( T_s \), \( \sigma_{n_0} \) and \( \sigma_n \).

Step 2. Discretize differential equations using some numerical methods, equation (3).

Step 3. Obtain the state variables in \( n+1 \) with equation (4).

\[
\Delta X_n - (\mu_1 + \mu_2)X_n
\]

\[
\begin{bmatrix}
\Delta S_{1,n} + \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta S_{2,n} + \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta P_{1,n} - \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta P_{2,n} - \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n
\end{bmatrix}
\]

Step 4. Define and calculate the sacrificed variable.

Step 5. Determine \( U_n \) with least squares, equation (9). Figure 2 outlines the control diagram.

3. Control Structure Design under Additive Uncertainties

3.1. Controller Design under Uncertainty. Next, uncertainties effect in the tracking error are considered by adding the terms \( E_{\sigma,n} \) in (7):

\[
\Delta X_n - (\mu_1 + \mu_2)X_n
\]

\[
\begin{bmatrix}
\Delta S_{1,n} + \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta S_{2,n} + \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta P_{1,n} - \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n \\
\Delta P_{2,n} - \left( \frac{q(S/P_i)}{Y(P/S_i)} + \frac{q(S/P_j)}{Y(P/S_j)} \right) X_n
\end{bmatrix}
\]

Additive uncertainty (\( E_n \)) can be used to model several kinds of uncertainties as well as external perturbations (measurement errors are not considered). It might depend on the state variables and the system input. Moreover, considering a real plant like \( z_{n+1} = f(z_n, u_n) \), therefore the additive uncertainy can be expressed as \( E_n = f(z_n, u_n) - \hat{f} \)
where \( f(z_n, u_n) \) is the discrete-time nonlinear system model. If \( z \) and \( u \) are assumed to be bounded and \( f \) is Lipschitz [58], then \( E_{\sigma,n} \) can be modeled as a bounded uncertainty [59, 60].

The uncertainties terms in (24) affect the error convergence to zero of the tracking error [50]. This can be observed following the same procedure as in Section 2.3 Theorem demonstration:

\[
\begin{bmatrix}
    e_{X,n+1} \\
    e_{S_1,n+1} \\
    e_{S_2,n+1} \\
    e_{P_1,n+1} \\
    e_{P_2,n+1}
\end{bmatrix}
= \begin{bmatrix}
    k_X & 0 & 0 & 0 & e_{X,n} \\
    0 & k_{S_1} & 0 & 0 & e_{S_1,n} \\
    0 & 0 & k_{S_2} & 0 & e_{S_2,n} + TSX_n \\
    0 & 0 & k_{P_1} & 0 & e_{P_1,n} \\
    0 & 0 & 0 & k_{P_2} & e_{P_2,n}
\end{bmatrix}
\begin{bmatrix}
    e_{X,n} \\
    e_{S_1,n} \\
    e_{S_2,n} \\
    e_{P_1,n} \\
    e_{P_2,n}
\end{bmatrix}
\]

Comparing (25) with (23), the error nonconvergence to zero is noticed due to \( E_n \) presence.

Therefore, the next step is \( E_{\sigma,n} \) estimation to reduce their effect on the tracking error, achieving the convergence to zero.

3.2. Uncertainty Estimation. The following procedure, as the main contribution of this work, develops a control strategy based on linear algebra that reduces the effect of uncertainty in tracking error by its estimation using Newton’s backward interpolation [61]. The advantage is the estimation development with an easy-to-understand numerical procedure, which does not add further complexity to the control methodology previously presented. Thus, \( E_n \) is estimated with \( \hat{E}_n \). Then, \( \hat{E}_n \) is added in the control action calculation:
Taking into account that $E_{\sigma,n}$ is unknown, but it is assumed as a polynomial, its differences can be defined as

$$\delta E_{\sigma,n} = E_{\sigma,n+1} - E_{\sigma,n},$$

$$\delta^2 E_{\sigma,n} = \delta (\delta E_{\sigma,n}) = \delta (E_{\sigma,n+1} - E_{\sigma,n}) = E_{\sigma,n+2} - 2E_{\sigma,n+1} + E_{\sigma,n},$$

$$\delta^3 E_{\sigma,n} = \delta (\delta^2 E_{\sigma,n}).$$

(27)

3.2.1. Constant Uncertainty. If $E_{\sigma,n}$ is assumed as a constant, $\delta E_{\sigma,n} = 0$. The uncertainty estimation is represented by

$$U_n = \begin{bmatrix}
-(X_n/V_n) \\
(\lambda S_f - S_{1,n})/V_n \\
((1 - \lambda)S_f - S_{2,n})/V_n \\
-P_{1,n}/V_n \\
-P_{2,n}/V_n \\
\end{bmatrix}.$$

$$\begin{aligned}
\Delta X_n - (\mu_1 + \mu_2)X_n \\
\Delta S_{1,n} + \left( \frac{q(S_i/P_i)}{Y(P_i/S_i)} + \frac{q(S_i/P_2)}{Y(P_2/S_i)} \right)X_n \\
\Delta S_{2,n} + \left( \frac{q(S_i/P_i)}{Y(P_i/S_i)} + \frac{q(S_i/P_2)}{Y(P_2/S_i)} \right)X_n \\
\Delta P_{1,n} - \left( \frac{q(S_i/P_i)}{Y(P_i/S_i)} \right)X_n \\
\Delta P_{2,n} - \left( \frac{q(S_i/P_2)}{Y(P_2/S_i)} \right)X_n \\
\end{aligned}
$$

where $\sigma$ is the real state variable and $\tilde{\sigma}$ represents the state variable calculated with equation (3). Then, the uncertainty is estimated with a zero-order approximation:

$$\tilde{E}_{\sigma,n} = E_{\sigma,n}.\quad (29)$$

Demonstration of error convergence to zero when uncertainties are considered:

Replacing the control action given by equation (26) in (24) and applying the same steps described in Section 2.3 demonstration,

$$\begin{aligned}
\frac{d\mu_1(S_i, P_{1,n})}{dS_1} &+ T_S X_n \\
\frac{dq(S_i/P_i)(S_i, P_{1,n})}{dS_1} &+ T_{NL} X_n \\
\end{aligned}\quad (30)$$

Remark 1. If $E_{\sigma,n} = c$, where $c$ is a constant value, and $\tilde{E}_{\sigma,n}$ is calculated with equation (29), then $\delta E_{\sigma,n} = 0$.

Thus, if $E_n$ is unknown and each component is an m-order polynomial, $E_n = \tilde{E}_n$, then $E_n$ and $\tilde{E}_n$ can be simplified from (30). Therefore, it is demonstrated that the tracking error tends to zero with the process progress.

3.2.2. Linear Uncertainty. If $E_{\sigma,n}$ is supposed as a linear function, then $\delta^2 E_{\sigma,n} = 0$. Following the same reasoning as in 3.2.1, $\tilde{E}_{\sigma,n}$ is defined as

$$\tilde{E}_{\sigma,n} = E_{\sigma,n} + E_{\sigma,n} - E_{\sigma,n-1}.\quad (31)$$

Following the same steps of 3.2.1, the error convergence to zero is demonstrated.
3.2.3. Polynomial Uncertainty. In this case, $E_{\sigma,n}$ is supposed as an $m$-order polynomial function. If $q > m$, $\delta^{q}E_{\sigma,n} \neq 0$. Using the same procedure explained before, the representation of $\hat{E}_{\sigma,n}$ is generically expressed as

$$\hat{E}_{\sigma,n} = \sum_{j=0}^{m} \sum_{i=0}^{j} \binom{j}{i} (-1)^{i} \frac{E_{\sigma,n-i-1}}{j!}.$$

(32)

3.3. Controller Parameter Selection. To select the best $k_{r}$ values (between zero and one for stability guarantee), as many authors recommend [62–64], the Monte Carlo algorithm is used following the procedure described in 2.3. Therefore, 1000 simulations were done to find the parameters for three different controllers, the original, described in [50] (C1), other with a zero-order estimator (C2), and the last with a first-order estimator (C3). Figure 3 shows the difference between the three controllers. Note the error improvement evidenced with both indexes ([12] and [13]), with C2, the total error $E_{p,1}$ decreases by 22.22% and $E_{p,2}$ by 19.31% compared to C1, while with C3, the total error is reduced by 52.38% and 48.73%, respectively, concerning C1. Due to the results similarity with both indexes, from now on only $E_{p,1}$ will be used to evaluate the algorithms.

4. Results and Discussion

In the following section, two important tests are developed to demonstrate the proposed estimation effectiveness. First, the Monte Carlo Algorithm is applied to test the controller operation under parametric uncertainty. Second, two different perturbations in the control action are added. Furthermore, both above tests are performed simultaneously and are compared with the performance of two other controllers using another methodology [65]. In this section, the original controller is noted as C1, the controller with a zero-order estimation is C2, and the controller with a first-order estimator is C3.

4.1. Simulation under Parametric Uncertainties. In all bioprocess, model parameters may vary in an unpredictable way [66]. This can lead to structural instability in the system dynamical behavior [15]. Therefore, a strict and efficient control system is required to deal with this problem.

In several research fields, probabilistic methods are useful for dealing with problems related to the robustness of systems affected by uncertainties [55]. Particularly, Monte Carlo Randomized Algorithm has been used for uncertainty quantification in many applications [67–69]. In this paper, the Monte Carlo Randomized Algorithm is applied using the procedure described in 2.3. The number of simulations, $N = 1000$, is obtained with equation (10), adopting a confidence ($\delta$) of 0.01 and an accuracy ($\varepsilon$) of 0.005. Therefore, the following test demonstrates the technique success from a statistical point of view [70–72].

In a simulation, a way to quantify uncertainties and perturbations is to specify the parameters real range of variation instead of using a constant value with greater error [73]. Hunag et al. [36] specified, in Table 1 of their work, the parameters confidence intervals for the ethanol process under study. For this test, in each simulation, all the system parameters are randomly changed by $\pm 10\%$ of their original range value (Table 1 of [36]). Then, the total error is calculated with equation (12). Table 3 shows the error range for each controller. Note how the total error range considerably decreases and maintains bounded to lower values when the estimation order increases. Figure 4 shows the total error for 1000 simulations to evaluate the three-controller
Figure 4: Total error for 1000 simulations under parametric uncertainty (±10%). (C1) Original controller; (C2) controller with a zero-order estimation; and (C3) controller with first-order estimator.

Figure 5: Controllers comparison of maximum total error.

Figure 6: Total error decrease when the controllers are tuned considering the worst situation of C1 (total error: \( C1 = 0.1334 \), \( C2 = 0.0633 \), \( C3 = 0.0585 \)).
Figure 7: Errors comparison for the three controllers: (C1) original controller; (C2) controller with a zero-order estimation; (C3) controller with the first-order estimator.

Table 4: Total error comparison under normal conditions.

|       | C1   | C2   | C3   |
|-------|------|------|------|
| Total error |      |      |      |
| a)    | 0.0063 | 0.0049 | 0.0030 |
| b)    | 0.1416 | 0.0062 | 0.0031 |

a) Tuned in normal conditions; b) tuned considering the worst situation of 4.1 for C1.

Figure 8: (a) Controllers response to a step perturbation addition in the control action. (b) Total error comparison (C1 = 1.5968, C2 = 0.1523, C3 = 0.0801).
performance under parametric uncertainty. As can be seen, the total tracking error range is visibly reduced.

Expressing the total error for each case into a percentage can easily quantify the controller improvement when estimators are incorporated. Figure 5 suggests the better performance of the controllers C2 and C3 over C1. The worst case presented for C2 has a total error of 43.83% lower than that of the worst situation of C1, while the maximum error presented with C3 is 84.58% less than that of C1.
Considering Theorem 1 of [55], as the tracking error (equation (11)) for 1000 simulations remains bounded, C2 and C3 controllers operation will be satisfactory with 99% of probability while the parameters vary within a ±10% range.

4.2. Simulation for Uncertainties Prevention. In the following test, controllers are tuned taking into account the worst possible situation presented in 4.1 for C1, to prevent it. Figure 6 shows the total error decreases when each controller is tuned. Note how even considering the most problematic situation of the system, the performance error can be reduced by 52.55% just by adding a zero-order estimator and by 56.15% with a first-order one.

Then, the controllers using the chosen parameters were tested under normal operation conditions, demonstrating that their performance is not negatively affected. Figure 7 compares the accumulated error for the three controllers. Moreover, total errors (C1 = 0.1416, C2 = 0.0062, C3 = 0.0031) contrast is shown.

This test demonstrates another advantage of the proposed estimation technique. Table 4 compares the errors of Figure 3 with those of Figure 7. In both cases, the controllers are tested under normal conditions; however, the difference lies in the tuning conditions considered. Analyzing Table 2, it can be observed that C2 and C3 controllers present similar results, which does not happen with C1. Therefore, it can be said that the operation of C2 and C3 will have a minimum error over the entire range of possible uncertainties between the nominal system parameters values and their worst variation.

4.3. Simulation Adding Perturbations in the Control Action. In this test, a hypothetical situation that may produce an unexpected variation in the production is simulated. Firstly, a -30% step perturbation in the bioreactor feed rate is added to evaluate the response of controllers. Secondly, a ramp disturbance is added to the step perturbation previously presented. Figures 8 and 9 show the control action variation compared to the reference and the percentage error, considering C1 error as 100%. Note how the total error improves when the new algorithm is applied. In the first test, it is reduced by 90.46% with C2 and by 94.99% with C3, while in the second test, the results are improved 91.87% with C2 and 97.15% with C3.

4.4. Simulation Adding Parametric Uncertainties and Perturbations in the Control Action. This latest test aims to demonstrate how the controllers can fix a large drift. In this way, 4.1 and 4.3 disturbances are considered simultaneously. Besides, the methodology proposed in [55] was also implemented to this ethanol bioprocess [65] and the results obtained are compared with those of the proposed technique. Figure 10 shows the biomass and ethanol profiles obtained with each controller throughout the process and a comparison of the errors. Here, C4 refers to a controller with one integrator and C5 with two ones. An improvement can be seen with the use of estimators concerning integrators (C1 = 100%; C2 = 12.5%; C3 = 1.9%; C4 = 13.1%; C5 = 2.4%). Furthermore, the error estimation has the advantage that it does not increase the order of the system by incorporating the estimation term and does not modify the form tuning as it does the integrators methodology, making mathematical development simpler.

5. Conclusions

This paper presents an improvement for a tracking control strategy previously published [50]. This technique lets tracking reference concentration profiles, even in the presence of model uncertainties and external perturbations. To consider those uncertainties, a new term is included in the mathematical model. The error estimation is approximated with Newton’s backward interpolation. In this manuscript, the main contribution is to decrease additive uncertainties effect on the tracking error without increasing the controller mathematical complexity. Moreover, the controller tuning is simpler than in conventional controllers, since varying the parameters between zero and one the error convergence to zero is achieved.

The Monte Carlo Randomized Algorithm is used to tune the controllers and carry out the tests with parametric uncertainty. Those tests show the effectiveness of this methodology, which was demonstrated in Sections 3 and 4.

Comparing with other methodologies that deal with similar uncertain control problems, such as [74–79], the proposed controller presents the advantage of avoiding the stochastic modeling needed to deal with parameters under perturbation of white noise. Besides, this nonlinear control does not require a great mathematical effort and does not add significant complexity to the original controller.

Data Availability

The data used to support the findings of this study are included within the article.

Conflicts of Interest

The authors declare that they have no conflicts of interest.

Acknowledgments

The National Council of Scientific and Technological Research (CONICET) and the Chemical Engineering Institute (IIQ) from the National University of San Juan contributed to this investigation.

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