Research Article

Distributed Adaptive Fixed-Time Tracking Consensus Control for Multiple Uncertain Nonlinear Strict-Feedback Systems under a Directed Graph

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1. Introduction

In the past decade, the finite-time consensus control of the multiagent has become one of the hot topics, since the finite-time control has a faster convergence rate, higher precision, and more robustness. However, there is a limitation for finite-time control that the convergence time increases accordingly when the initial state of the system is far from the equilibrium point. To overcome this limitation, fixed-time stability control [1] was proposed in 2012 by Polyakov et al.

In 1965, finite-time control was first proposed, and many related research results emerged in the following decades. Due to the different agent dynamics in multiagent systems (MASs), there are various finite-time control developed [2–7]. To list a few, in [8, 9], by utilizing the principle of homogeneity [10], the authors proposed finite-time consensus control protocols for first-order and second-order MASs, respectively. In [11], power integrator technique was employed to solve the problem of finite-time consensus control for MASs with double integrator dynamics. Furthermore, in [12], by combining the homogeneous domination method with the power integrator method, the authors proposed two classes of finite-time consensus protocols for MASs composed of first-order and second-order integrator agents. Thereafter, by using the Lyapunov function-based method, several finite-time consensus control strategies were proposed for high-order uncertain MASs [13–15]. Even so, there is a limitation for the finite-time consensus control results that the convergence time is seriously dependent on the MASs’ initial states. That is to say, once the initial state is far away from the equilibrium point, the convergence time will increase as a result. In practice, it is much more reasonable for a predictable convergence time.

To overcome the limitation of finite-time control, a serial of research results [16–26] on fixed-time consensus control for MASs has been carried out since its setting time can be
bounded and predefined. Note that most of the fixed-time control results [16–20] were designed for MASs with first order or second order. Only a few results [21–26] solved the fixed-time consensus problems for high-order MASs. On the contrary, strict-feedback or lower triangular systems have been widely regarded as a control target due to the description of various physical systems [27]. In fact, the strict-feedback systems require the recursive and systematic control design procedure since the control input is not matched with system nonlinearities. In [28], the adaptive backstepping design was proposed to satisfy this requirement. Furthermore, in [29, 30], the adaptive neural network backstepping controllers were proposed for a class of strict-feedback nonlinear systems. Nevertheless, there is a problem with using backstepping to design a controller for the strict-feedback systems; it can cause “explosion of complexity.” To overcome this problem, in [31], a dynamic surface design technique was proposed for strict-feedback systems. Furthermore, in [32], neural network-based adaptive dynamic surface control was proposed. Subsequently, this method was extended to MASs in [33–35], and a distributed adaptive fixed-time tracking consensus control protocol is proposed to solve this problem. (4) RBF neural networks are introduced to approximate the unknown functions $f_i(\cdot)$ in MASs to make our designed fixed-time control can suitable more systems with complex dynamics.

The organization for the remaining part of this paper is given below. Section 2 presents preliminaries and problem description. Sections 3 and 4 give the detailed process of distributed fixed-time control protocol and stability analysis, respectively. In Section 5, the proposed control scheme is proved to be effective through a simulation experiment. Finally, conclusions are summarized in Section 6.

2. Preliminaries and Problem Description

2.1. Graph Theory. Let $G = (\mathcal{V}, \mathcal{E})$ be a directed graph with the set of nodes or vertices $\mathcal{V} = \{1, \ldots, M\}$ and the set of edges or arcs $\mathcal{E} \subseteq \mathcal{V} \times \mathcal{V}$. An edge $(j, i) \in \mathcal{E}$ means that agent $j$ can obtain information from agent $i$, but agent $i$ cannot obtain agent $j$’s. The set of neighbors of a node $i$ is $N_i = \{j \mid (j, i) \in \mathcal{E}\}$, which is the set of nodes with edges incoming to node $i$. The weighted adjacency matrix $A = [a_{ij}] \in \mathbb{R}^{M \times M}$ is defined as $a_{ij} > 0$ if $(j, i) \in \mathcal{E}$ and $a_{ij} = 0$ otherwise. Self-loop is not allowed, i.e., $a_{ii} = 0$. The Laplacian matrix $L$ is defined as $L = D - A \in \mathbb{R}^{M \times M}$, where $D = \mathrm{diag}[d_1, \ldots, d_M]$; $d_i = \sum_{j=1, j \neq i}^{M} a_{ij}$ is the diagonal element of the degree matrix $D$.

A directed path from node $i_1$ to node $i_k$ is a sequence of edges of the form $(i_1, i_2), (i_2, i_3), \ldots, (i_{k-1}, i_k)$ in a directed graph. A directed tree is a directed graph where every node has exactly one parent except for the root and the root has directed paths to every other node. A directed graph has a directed spanning tree if there exists at least a node having a directed path to all the other nodes.

2.2. RBFNNs. An RBF neural network [40] is applied in this paper to approximate arbitrary continuous functions. The RBF neural network is defined as follows:

$$\tilde{f}(Z) = W^T S(Z),$$

where $W = [w_1, \ldots, w_l]^T \in \mathbb{R}^l$ is the weight vector, $l > 0$ is the number of nodes of the neural network, $Z \in \Omega_Z \subset \mathbb{R}^l$ is the input of the RBF neural network, $q$ is the input dimension, $S(Z) = [s_1(Z), \ldots, s_l(Z)]^T \in \mathbb{R}^l$ is the basis vector function, and $s_i(Z)$ is the output of the $i$ th neural node. A Gaussian function is always chosen as $s_i(Z)$, i.e.,

$$s_i(Z) = \exp[-(Z - \xi_i)^T (Z - \xi_i)/r_i^2],$$

where $r_i$ is the width of the basis function and $\xi_i = [\xi_{i1}, \ldots, \xi_{iD}]$ is the center of the basis function. With a sufficient number $l$ of neural nodes selected, an RBF neural network can approximate arbitrary continuous function $f(Z)$ in a compact set $\Omega_Z \subset \mathbb{R}^l$ with arbitrary accuracy $\epsilon$:

$$f(Z) = W^* S(Z) + \delta(Z), \quad \forall Z \in \Omega_Z \subset \mathbb{R}^l,$$

where $\delta(Z)$ is the approximation error with $|\delta(Z)| \leq \epsilon$ and $W^*$ is the given ideal constant weight vector, which is defined as follows:
Consider the following nonlinear system:

\[ \dot{x}(t) = f(t,x), \]

where \( x \in \mathbb{R}^n \) and \( f: \mathbb{R}^n \times \mathbb{R}^n \to \mathbb{R}^n \), and assume that the origin is an equilibrium point.

**Lemma 1** (see [41]). Consider the Gaussian function (1); \( \|S(Z)\| \) has an upper bound such that

\[ \|S(Z)\| \leq \sum_{k=0}^{\infty} 3q(k+2)^{r-1}e^{-(2\pi k^2)r^2} = s, \]

where \( p = (1/2)(\min_{i,p}\|\xi_i - \xi_j\|) \).

**Remark 1.** Lemma 1 has been proved in [41, 42]. Since \( \sum_{k=0}^{\infty} 3q(k+2)^{r-1}e^{-(2\pi k^2)r^2} \) is convergent, \( s \) is a limited value. In addition, \( s \) is independent of the neural network node numbers \( l \) and the neural network inputs \( Z \).

**Assumption 1** (see [36]). There are unknown constants \( Q_i \) that make \( |\theta_i| \leq Q_i < \infty, i = 1, 2, \ldots, n \).

**2.3. Fixed Time**

**Definition 1.** Consider the following nonlinear system:

\[ \dot{x}(t) = f(t,x), \]

where \( x \in \mathbb{R}^n \) and \( f: \mathbb{R}^n \times \mathbb{R}^n \to \mathbb{R}^n \), and assume that the origin is an equilibrium point.

**Lemma 2** (see [22]). If there exist design constants \( \phi_1 > 0, \phi_2 > 0, \alpha \in (0,1), \) and \( \beta \in (1, +\infty) \) such that

\[ \dot{V} \leq -\phi_1 V^\alpha(x) - \phi_2 V^\beta(x), \]

where \( V(x) \) is a continuous differentiable positive definite function; system (5) is global fixed-time stable, and the fixed convergence time satisfies

\[ T \leq T_{\max} := \frac{1}{\phi_1(1-\alpha)} + \frac{1}{\phi_2(\beta-1)}. \]

**Remark 2.** The sufficient conditions and convergence time for finite-time and fixed-time consensus control schemes are shown in Table 1. Where \( \mu_1, \mu_2, \rho_1, \) and \( \rho_2 \) are positive design parameters, \( \alpha > 1 \), and \( 0 < \beta < 1 \). From convergence time, it can be seen that the finite-time control is related to the initial state \( x(0) \), while the fixed-time control is not.

**Lemma 3** (see [36]). If there exist design constants \( \phi_1 > 0, \phi_2 > 0, \alpha \in (0,1), \beta \in (1, +\infty), \) and \( \rho \in (0,1) \) such that

\[ \dot{V} \leq -\phi_1 V^\alpha(x) - \phi_2 V^\beta(x) + \tau. \]

Then, the origin of system (5) is practical fixed-time stable and the fixed-time \( T \) can be estimated by

\[ T \leq T_{\max} := \frac{1}{\phi_1(1-\alpha)} + \frac{1}{\phi_2(\beta-1)}. \]

The residual set of the solution of system (5) is given by

\[ x \in \left\{ V(x) \leq \min\left\{ \left(\frac{\tau}{(1-\alpha)\phi_2}\right)^{\frac{1}{\beta}}, \left(\frac{\tau}{(1-\alpha)\phi_1}\right)^{\frac{1}{\alpha}} \right\} \right\}. \]

**Lemma 4** (see [43]). Let \( x_1, x_2, \ldots, x_n \geq 0 \). Then,

\[ \sum_{i=1}^{n} x_i^p \geq \left(\sum_{i=1}^{n} x_i\right)^p, \]

if \( 0 < p \leq 1 \),

\[ \sum_{i=1}^{n} x_i^p \geq n^{1-p} \left(\sum_{i=1}^{n} x_i\right)^p, \]

if \( 1 < p < \infty \).

**Lemma 5** (see [44]). For any variable \( x \in \mathbb{R} \) and any positive constant \( \kappa \), the following relationship holds:

\[ 0 \leq |x| < \kappa + \frac{x^2}{\sqrt{x^2 + \kappa^2}}. \]

**Lemma 6** For \( y \geq x > 0, x, y \in \mathbb{R} \) and any positive constant \( \delta \), then satisfying

\[ \frac{y}{\sqrt{y + \delta}} \geq \frac{x}{\sqrt{x + \delta}}. \]

**Proof**

\[ \frac{y}{\sqrt{y + \delta}} - \frac{x}{\sqrt{x + \delta}} = \frac{y + \delta}{\sqrt{y + \delta}} - \frac{x + \delta}{\sqrt{x + \delta}} - \left( \frac{\delta}{\sqrt{y + \delta}} - \frac{\delta}{\sqrt{x + \delta}} \right) \]

\[ = \left( \frac{\delta}{\sqrt{x + \delta}} - \frac{\delta}{\sqrt{y + \delta}} \right) \geq 0. \]

2.4. Problem Statement. Suppose that there exist one leader agent, labeled as 0, and \( M \) follower agents, labeled as 1 to \( M \), under a directed communication graph topology. The dynamic models of \( M \) followers in the strict-feedback form are considered as follows:

\[ \begin{align*}
\dot{x}_{i,k} &= x_{i,k+1} + f_{i,k}(x_{i,k}), \\
\dot{x}_{i,l} &= u_i + f_{i,n}(x_i), \\
y_i &= x_{i,1},
\end{align*} \]

where \( i = 1, \ldots, M, k = 1, \ldots, n_i - 1, x_{i,k} = [x_{i1,k}, \ldots, x_{ini,k}]^T; x_{i,l} \in \mathbb{R}^p \) with \( l = 1, \ldots, k, x_i = [x_{i1}, \ldots, x_{ini}]^T \in \mathbb{R}^{mp} \), and \( u_i \in \mathbb{R}^p \) are the state vector and the control input of the \( i \) th follower, respectively, and \( y_i \in \mathbb{R}^p \) is the output of the \( i \)th.
follower. $f_{i,k}(\mathbf{x}_{i,k}) \in \mathbb{R}^p$ are $C^1$ nonlinear function vectors. It is assumed that the leader’s motion is independent of the followers’ motion.

The communication topology for the $M+1$ agents is described by a directed graph $G = (\mathcal{V}, \mathcal{E})$ with $\mathcal{V} = \{0, 1, \ldots, M\}$. To represent the communications among followers, we define a subgraph as $\mathcal{G} = (\mathcal{V}, \mathcal{E})$ with $\mathcal{V} = \{1, \ldots, M\}$. The adjacency matrix $\mathbf{A}$ of the subgraph $\mathcal{G}$ is $\mathbf{A} = [a_{ij}] \in \mathbb{R}^{M \times M}$; $a_{ij} > 0$ if $(j, i) \in \mathcal{E}$, $a_{ij} = 0$ otherwise, and $a_{ii} = 0$. Then, the Laplacian matrix $L$ is defined as

$$L = \begin{bmatrix} 0 & 0_{1 \times M} \\ -h & \mathcal{L} + H \end{bmatrix}, \tag{16}$$

where $h = [h_1, \ldots, h_M]^T$, with $h_i > 0$ if the leader $0 \in \mathcal{N}_i$ and $h_i = 0$ otherwise, which denotes the communication weight from the leader to followers, $H = \text{diag}[h_1, \ldots, h_M]$, and $\mathcal{L} = \mathcal{D} - \mathbf{A}$ with $\mathcal{D} = \text{diag}[d_1, \ldots, d_M]$ is the Laplacian matrix of the subgraph denoting the communication among followers.

Remark 3. For the simplicity of analysis, in the following, we only consider the case where $p = 1$. The analysis and main results still hold for any dimension $p$ by using the Kronecker product.

Remark 4. If the directed graph $G$ has a spanning tree, $\text{rank}(L) = M$ [45]. Then, $\text{rank}(L + H) = M$ from $(L + H)1_M = h$, where $1_M$ is an $M$–vector of all ones. Therefore, $L + H$ is invertible.

Assumption 2. The $C^1$ nonlinear functions $f_{i,k}$ are unknown on a directed graph $G$.

Assumption 3. The leader output signal $r(t) \in \mathbb{R}^p$ is an $n_i$–order differentiable and bounded function and available for the $i$th followers satisfying $0 \in \mathcal{N}_i$, $i = 1, \ldots, M$.

Definition 2 (see [33]). The distributed consensus tracking errors for nonlinear followers (15) under the communication graph are said to be cooperatively semiglobally uniformly ultimately bounded (CSUUB) if there exist adjustable constants $c_{1i} > 0$ and $c_{2i} > 0$, and the bounds $\beta_{1i} > 0$ and $\beta_{2i} > 0$, independent of $t_0$, for each $a_{ij} \in \{0, c_{1j}\}$ and $a_{ik} \in \{0, c_{2k}\}$, there is a time $T_0 \geq 0$, independent of $t_0$, such that $\|y_i(t_0) - r(t_0)\| \leq c_{1i} \Rightarrow \|y_i(t) - r(t)\| \leq \beta_{1i}$ and $\|y_i(t_0) - y_j(t_0)\| \leq c_{2i} \Rightarrow \|y_i(t) - y_j(t)\| \leq \beta_{2i}$ for all $t \geq t_0 + T_0$, where $i = 1, \ldots, M$, $j = 1, \ldots, M$, $i \neq j$ and $k = 1, \ldots, n_i$.

The objective of this brief is to design RBF-neural network-based distributed fixed-time consensus control laws $u_i$ for $M$ followers (15) with unknown nonlinearities so that, under the directed graph, the follower outputs $y_i$ synchronize to the dynamic leader output $r$ within fixed time while all signals in the total closed-loop systems are bounded.

Remark 5. The strict-feedback system (15) can describe many state-space models of nonlinear systems, i.e., various physical systems, such as flight systems, biochemical process, jet engine, and robotic systems [28]. Therefore, system (15) under a graph topology can represent multiagent systems consisting of several practical applications with different dynamics.

Remark 6. Notice that followers (15) can have various forms. That is, a group of the followers with different nonlinear functions and order of the dynamics can be considered in this brief.

Remark 7. Compared with the previous consensus works, this brief considers the consensus problem of a group of agents consisting of nonlinear followers with nonlinearities unmatched in the control input. Besides, the nonlinearities are unknown under the total communication topology, and the tracking convergence time independent of the initial state of system (15).

3. Distributed Fixed-Time Consensus Controller Design

The design procedure on the $i$th follower contains $n_i$ steps. The distributed backstepping design coordinate transformation is as follows:

$$z_{i,1} = \sum_{j=1}^{M} a_{ij} (y_j - y_i) + h_i (y_i - y_0), \tag{17}$$

$$z_{i,k} = x_{i,k} - v_{i,k-1}, \tag{18}$$

where $i = 1, \ldots, M$ and $k = 2, \ldots, n_i$ and $v_{i,k}$ is the virtual controller of the $k$th subsystem of $i$th follower.

An RBF neural network is used in this paper to approximate the unknown functions $f_{i,k}(Z_{i,k})$:

$$\tilde{f}_{i,k}(Z_{i,k}) = W_{i,k}^T S_{i,k}(Z_{i,k}) + \delta_{i,k}(Z_{i,k}), \tag{19}$$

and the inequalities involved in the following text are as below:

$$z_{i,k}^2 \leq b_{i,k}^2 c_{i,k}^2 + \frac{\eta_{i,k}^2}{4b_{i,k}^2}, \tag{20}$$

$$z_{i,k} W_{i,k}^T S_{i,k}(Z_{i,k}) \leq \frac{b_{i,k}^2 c_{i,k}^2}{2h_{i,k}^2} S_{i,k}(Z_{i,k})^2 Z_{i,k}^2 + \frac{\eta_{i,k}^2}{2}. \tag{21}$$
\[-b_i \zeta_{i,k} z_i^2 (1+\beta)/2 \left( \frac{1}{2} \frac{z_i^2}{z_{i,k}} \right)^{(1+\beta)/2} + \left( \frac{b_i \theta_{i,k}^2}{2y_i} \right)^{(1+\beta)/2} \]
\[\leq -2b_i \zeta_{i,k} \left( \frac{1}{2} \frac{z_i^2}{z_{i,k}} + \frac{b_i \theta_{i,k}^2}{2y_i} \right)^{(1+\beta)/2}, \]
\[-b_i \zeta_{i,k} z_i^2 (1+\alpha)/2 \left( \frac{1}{2} \frac{z_i^2}{z_{i,k}} \right)^{(1+\alpha)/2} + \left( \frac{b_i \theta_{i,k}^2}{2y_i} \right)^{(1+\alpha)/2} \]
\[\leq -2b_i \zeta_{i,k} \left( \frac{1}{2} \frac{z_i^2}{z_{i,k}} + \frac{b_i \theta_{i,k}^2}{2y_i} \right)^{(1+\alpha)/2}, \]
where \(Z_{i,k} = \{x_{i,1}, \ldots, x_{i,k}, \tilde{\theta}_{i,1}, \ldots, \tilde{\theta}_{i,k}, h_i y_0, h_i y_0, \ldots, h_i y_0, a_i y_{i,1}, \ldots, a_i y_{i,k}, \ldots, a_i y_{i,k} \} \) with \(i = 1, \ldots, M, k = 1, \ldots, n_i \) is the input vector, and \(b_i, \zeta_{i,k}, \xi_{i,k}, \xi_{i,k}, \epsilon_{i,k}, \eta_{i,k}, \) and \(y_i \) are positive design parameters.

**Step 1.** According to (15), (17), and (18), we have

\[
\dot{z}_{i,i} = (d_i + h_i) z_{i,2} + (d_i + h_i) y_{i,1} + \tilde{f}_{i,i}(Z_{i,1}) - \sum_{j=1}^{M} a_i y_{j,2},
\]

where \(\tilde{f}_{i,i}(Z_{i,1}) = (d_i + h_i) f_{i,i} - \sum_{j=1}^{M} a_i f_{j,i} - h_i y_0.\)

Construct a Lyapunov function as

\[
V_{i,i} = \frac{1}{2} z_i^2 + \frac{b_i \theta_{i,k}^2}{2y_i}.
\]

Time differentiation of \(V_i\) yields

\[
\dot{V}_{i,i} = (d_i + h_i) z_{i,1} z_{i,2} + (d_i + h_i) z_{i,1} y_{i,1} + z_{i,1} \tilde{f}_{i,i}(Z_{i,1}) - z_{i,1} \sum_{j=1}^{M} a_i y_{j,2} - \frac{b_i \theta_{i,k}^2}{y_i}.
\]

The virtual controller \(y_{i,1}\) is defined as

\[y_{i,1} = \frac{1}{d_i + h_i} \left( \frac{b_i \theta_{i,k}^2}{2y_i} \right) \left( S_{z_{i,1}} \right) (Z_{i,1}) z_{i,1} + \sum_{j=1}^{M} a_i y_{j,2} \]

where \(\zeta_{i,11}, \zeta_{i,12}, \zeta_{i,13}\), and \(\eta_{i,1}\) are positive design parameters.

In (27), \(S_{z_{i,1}}, \) and \(S_{z_{i,2}}\) are defined as

\[S_{z_{i,1}} = \left\{ \begin{array}{l}
z_i^{(1+\alpha)/2}, \\
\sum_{j=1}^{n} a_j \left( z_i^{(1+\alpha)/2} \right)^{-j} (z_{i,k})^{-j} (z_{i,k})^{-j}, \\
\sum_{j=1}^{n} a_j \left( z_i^{(1+\alpha)/2} \right)^{-j} (z_{i,k})^{-j} (z_{i,k})^{-j}, \\
\end{array} \right\}
\]

where \(k = 1, \ldots, n_i, \) and \(\epsilon_{i,k}\) is a positive design parameter, and coefficients \(a_j, j = 1, \ldots, n_i\) are calculated using the following equation:

\[\begin{bmatrix} 1 & 1 & \cdots & 1 \\
1 & 2 & \cdots & n_i - 1 \\
0 & 2 & \cdots & (n_i - 1)(n_i - 2) \\
\vdots & \vdots & \ddots & \vdots \\
0 & 0 & \cdots & n_i - 2 \\
0 & 0 & \cdots & 0 \\
\end{bmatrix} \times \begin{bmatrix} a_1 \\
\vdots \\
\vdots \\
\vdots \\
a_n \\
\vdots \\
\end{bmatrix} = \begin{bmatrix} b_1 \\
b_2 \\
b_3 \\
\vdots \\
b_n \\
\vdots \\
\end{bmatrix},
\]

where \(b_1 = 1, b_2 = 3/4, b_3 = (3/4)((3/4) - 1), \ldots, b_n = \prod_{j=0}^{n-2} ((3/4) - j).\)

Remark 8. One of the main contributions of this paper is to design suitable virtual controllers \(y_{i,k}\) so that nonlinear strict-feedback systems meet the requirement of fixed-time control. In [36-39], virtual controllers designed exhibit similar power functions \(z^{2\alpha - 1}\), where \(0 < \alpha < 1.\) If \(\alpha\) is not appropriately selected, it will make unsolvable at the origin and in the negative domain. As \(0 < \alpha < 1,\) therefore, power exponents \(2\alpha - 1 \in (-1, 1),\) which leads to the possibility of
negative power exponents. For example, if $2\alpha - 1 = -(1/3)$, $z^{-(1/3)}$ is unsolvable at $z = 0$, that is, $(0)^{-1/3}$ does not exist. Suppose $2\alpha - 1 = (q_1/q_2) \in (-1, 1)$, where $q_2$ is an even number, then $z^{(q_1/q_2)}$ is unsolvable at the negative domain. For example, if $(q_1/q_2) = 1/2$, $z^{1/2}$ is unsolvable at the negative domain, that is, $(-1)^{1/2}$ does not exist. The controller designed in this paper overcomes the abovementioned defect and promotes the application of fixed-time control in more common nonlinear systems.

Substituting (27) into (26) yields

$$
\dot{V}_{i,1} \leq (d_i + h_i)z_{i,1}^2 + \frac{b_i\hat{\theta}_{i,1}(Z_{i,1})}{2\eta_i} + \frac{b_i\hat{\theta}_{i,1}(Z_{i,1})}{2\eta_i} \frac{\|z_{i,1}\|^2}{\gamma_1} - \frac{b_i\hat{\theta}_{i,1}(Z_{i,1})}{\gamma_1} - \frac{b_i\hat{\theta}_{i,1}(Z_{i,1})}{\gamma_1} - c_{i,11}z_{i,1}S_{i,1} - c_{i,12}z_{i,1}S_{i,2}.
$$

(31)

According to (19) and inequalities (20)-(21), we have

$$
\dot{V}_{i,1} \leq \frac{b_i\hat{\theta}_{i,1}}{\gamma_1} \left( \frac{\gamma_i}{2\eta_i} \frac{S_i}{S_{i,1}} (Z_{i,1})^2 S_{i,1}(Z_{i,1}) - \frac{\hat{\theta}_{i,1}}{\gamma_1} \right) \leq \frac{b_i\hat{\theta}_{i,1}}{\gamma_1} \left( \frac{\gamma_i}{2\eta_i} \frac{S_i}{S_{i,1}} (Z_{i,1})^2 S_{i,1}(Z_{i,1}) - \frac{\hat{\theta}_{i,1}}{\gamma_1} \right).
$$

(32)

where $\alpha_{i,1} = (1/2)\eta_i^2 + ((\epsilon_i^2)/(4b_i\epsilon_i))$. The adaptive law $\dot{\theta}_{i,1}$ is then defined as

$$
\dot{\theta}_{i,1} = \frac{\gamma_i}{2\eta_i} \frac{S_i}{S_{i,1}} (Z_{i,1})^2 S_{i,1}(Z_{i,1}) - \lambda_{i,1} \hat{\theta}_{i,1},
$$

(33)

where

$$
C_{i,1} = \alpha_{i,1} + \beta_{i,1} + (\lambda_1 b_i/2\gamma_i)\theta_{i,1}^2, \quad \beta_{i,1} = \epsilon_i^2 (1/2),
$$

(34)

(1/2) + (1/2) \left( \frac{b_i}{\gamma_i} \right)^2, \quad \text{and}

$$
(\lambda_1 b_i/2\gamma_i) (\theta_{i,1}/\gamma_i^2) \leq (\lambda_1 b_i/2\gamma_i) ((\theta_{i,1}/2) - (\theta_{i,1}/2)) \leq (\lambda_1 b_i/2\gamma_i) (2\gamma_i/2\gamma_i).
$$

According to (22), (23), (28), and (29), if $\|z_{i,1}\| \geq \epsilon_1$, then (34) is rewritten as

$$
\dot{V}_{i,1} \leq \epsilon_i M + \frac{1}{2} \epsilon_i M - \epsilon_i M + (d_i + h_i)z_{i,1}z_{i,2} + C_{i,1},
$$

(35)

According to (22), (23), (28), and (29), if $\|z_{i,1}\| \leq \epsilon_1$, then (34) is rewritten as

$$
\dot{V}_{i,1} \leq \epsilon_i M + \frac{1}{2} \epsilon_i M - \epsilon_i M + (d_i + h_i)z_{i,1}z_{i,2} + C_{i,1}.
$$

(36)
Remark 9 (see [46]). Based on (28) and (29), when \(\|z_i,1\| < \epsilon_{10}\), there is an additional term in (36): \(\dot{c}_{i,11}(z_i,1)\left(1+\alpha/2\right) - c_{i,11}z_i,1 + \sum_{j=1}^{M} a_j (z_j,1)^{1+\beta/2} - \frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial \dot{\theta}_{i,1}} - \sum_{j=1}^{M} a_j (z_j,1)^{1+\beta/2} + c_{i,12}z_i,1\). Note that if \(\|z_i,1\| < \epsilon_{10}\), this additional term is obviously limited by some smaller constant \(\epsilon_{11}\), so the structure of (35) is retained, while the constant term \(C_{i,1}\) only slightly increases. Owing to page limitations and to avoid repetitive discussions, we will omit this part in the rest of the analysis.

**Step 2.** According to (15) and (18), we have

\[
\dot{V}_{i,2} \leq -c_{i,11} \dot{z}_i,1^{(1+\alpha)/2} \left(\frac{1}{2} \dot{z}_i,2 + \frac{b_i \theta_i,1^2}{2\gamma_i}\right)^{(1+\alpha)/2} - 2c_{i,12} \left(\frac{1}{2} \dot{z}_i,2 + \frac{b_i \theta_i,1^2}{2\gamma_i}\right)^{(1+\beta)/2} + z_i,2 \dot{f}_{i,2}(z_i,2) + z_i,2 z_i,3 + z_i,3 y_i,2 - z_i,3 M_{i,1}(z_i,2) + C_{i,1},
\]

where

\[
\dot{f}_{i,2}(z_i,2) = f_{i,2} + z_i,1 \left(\dot{d}_i + h_i - \frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial x_{i,1}} x_i,1 - \sum_{j=1}^{M} a_j (\frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial x_{j,1}} x_j,1 + \frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial x_{j,2}} x_j,2) - \frac{M_{i,1}(z_i,2)}{\gamma_i}\right).
\]

**Remark 10.** \(M_{i,1}(z_i,2)\) is a smooth function that is used to overcome the design difficulty of \(\frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial \dot{\theta}_{i,1}}\).

The virtual controller \(\dot{v}_{i,2}\) is defined as

\[
\dot{v}_{i,2} = -c_{i,21} y_i,2 - c_{i,22} y_i,2 - \frac{b_i \theta_i,2}{2\gamma_i} x_i,2 - b_i c_{i,23} x_i,2.
\]

where \(c_{i,21}, c_{i,22}, c_{i,23}\), and \(\eta_{i,2}\) are positive design parameters.

Choose the adaptive law \(\dot{\theta}_{i,2}\) as

\[
\dot{\theta}_{i,2} = \frac{\eta_{i,2}}{2\gamma_i} S_{i,2}(z_i,2) S_{i,2}(Z_i,2) z_i,2 - \lambda_i \theta_{i,2}.
\]

Combining (19)–(21) and Assumption 1 and substituting (41) and (42) into (38) and adopting the same design method as in Step 1 yields

\[
\dot{V}_{i,2} \leq -c_{i,21} z_i,2 S_{i,2} - c_{i,22} z_i,2 S_{i,2} - c_{i,23} z_i,2 \left(1+\alpha/2\right) \left(\frac{b_i \theta_i,2^2}{2\gamma_i}\right)^{(1+\alpha)/2} - c_{i,22} \left(1+\beta/2\right) \left(\frac{b_i \theta_i,2^2}{2\gamma_i}\right)^{(1+\beta)/2} - c_{i,11} \left(1+\alpha/2\right) \left(\frac{1}{2} \dot{z}_i,2 + \frac{b_i \theta_i,1^2}{2\gamma_i}\right)^{(1+\alpha)/2} - c_{i,22} \left(1+\beta/2\right) \left(\frac{b_i \theta_i,2^2}{2\gamma_i}\right)^{(1+\beta)/2} - c_{i,11} \left(1+\alpha/2\right) \left(\frac{1}{2} \dot{z}_i,2 + \frac{b_i \theta_i,1^2}{2\gamma_i}\right)^{(1+\alpha)/2} \sum_{j=1}^{M} C_{i,j} z_i,1 + z_i,2 z_i,3 + z_i,3 M_{i,1}(z_i,2) - \frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial \dot{\theta}_{i,1}} - \sum_{j=1}^{M} a_j (\frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial x_{j,1}} x_j,1 + \frac{\partial \dot{v}_{i,1}(z_i,1)}{\partial x_{j,2}} x_j,2) - \frac{M_{i,1}(z_i,2)}{\gamma_i},
\]

where \(C_{i,2} = \sigma_{i,2} + \beta_i,2 + (\lambda_i b_i/2\gamma_i) \beta_i,2, \beta_i,2 = c_{i,21} \left(1+\alpha/2\right) \left(\frac{b_i Q_i,2^2}{2\gamma_i}\right)^{(1+\alpha)/2} + c_{i,22} \left(1+\beta/2\right) \left(\frac{b_i Q_i,2^2}{2\gamma_i}\right)^{(1+\beta)/2}, \quad \sigma_{i,2} = \left(1/2\right)\eta_{i,2}^2 + (\epsilon_{i,2}^2/4b_i c_{i,22}), \text{ and } (\lambda_i b_i \theta_i,2 \beta_i,2/\gamma_i) \leq (\lambda_i b_i/\gamma_i) ((\theta_{i,2}^2/2) - (\theta_{i,2}^2/2) \leq (\lambda_i b_i/2\gamma_i) \theta_{i,2}^2.\)
According to (22), (23), and Remark 9, we consider only \( \|z_{i,2}\| \geq \varepsilon_{20} \). Then, (43) can be written as

\[
\dot{V}_{i,2} \leq -2^{(1+n)/2} \sum_{j=1}^{2} \left( \frac{1}{2} \dot{z}_{i,j} + \frac{b_i \dot{\theta}_{i,j}^2}{2y_i} \right)^{(1+n)/2} - 2 \sum_{j=1}^{2} \zeta_{i,j} \left( \frac{1}{2} \dot{z}_{i,j} + \frac{b_i \dot{\theta}_{i,j}^2}{2y_i} \right)^{(1+n)/2} + \sum_{j=1}^{2} C_j + \dot{z}_{i,2} \left( M_{i,1}(Z_{i,2}) - \frac{\partial v_{i,1}}{\partial \theta_{i,1}} \right).
\]  

(44)

It can be seen from (44) that defining the design smooth function \( M_{i,1}(Z_{i,2}) \) to overcome the design difficulty of \( (\partial v_{i,1}/\partial \theta_{i,1})\dot{\theta}_{i,1} \) is one of the difficulties of designing the controllers in this paper.

\[
-\frac{\partial v_{i,1}}{\partial \theta_{i,1}} \leq - \frac{\partial v_{i,1}}{\partial \theta_{i,1}} \left( \frac{\gamma_i}{2\eta_n^2} S_{i,1}(Z_{i,1}) \right) z_{i,1}^2 - \lambda \dot{\theta}_{i,1} \leq \frac{z_{i,1}^2 (\partial v_{i,1}/\partial \theta_{i,1})^2 (\gamma_i/2\eta_n^2) S_{i,1}(Z_{i,1}) z_{i,1}^2}{\sqrt{z_{i,1}^2 (\partial v_{i,1}/\partial \theta_{i,1})^2 (\gamma_i/2\eta_n^2) S_{i,1}(Z_{i,1})}} + \kappa^2 \quad + z_{i,2} \frac{\partial v_{i,1}^2}{\partial \theta_{i,1}} \dot{\theta}_{i,1}.
\]

(45)

Therefore, \( M_{i,1}(Z_{i,2}) \) can be defined as

\[
M_{i,1}(Z_{i,2}) = -\frac{z_{i,1} (\partial v_{i,1}/\partial \theta_{i,1})^2 (\gamma_i/2\eta_n^2) S_{i,1}(Z_{i,1}) z_{i,1}^2}{\sqrt{z_{i,1}^2 (\partial v_{i,1}/\partial \theta_{i,1})^2 (\gamma_i/2\eta_n^2) S_{i,1}(Z_{i,1})}} - \frac{\partial v_{i,1}}{\partial \theta_{i,1}} \dot{\theta}_{i,1}.
\]

(46)

Step \( k (3 \leq k \leq n_t - 1) \): according to (15) and (18), we have

\[
\dot{z}_{i,k} = x_{i,k+1} + f_{i,k}(x_{i,k}) - y_{i,k-1},
\]

(49)

where \( \dot{y}_{i,k-1} = \sum_{j=1}^{k-1} (\partial y_{i,k-1}/\partial x_{i,j}) \dot{x}_{i,j} + \sum_{j=1}^{k} \sum_{j=1}^{M} a_{i,j} \sum_{j=1}^{k-1} (\partial y_{i,k-1}/\partial x_{i,j}) x_{i,j} + \sum_{j=1}^{k-1} (\partial y_{i,k-1}/\partial \theta_{i,j}) \dot{x}_{i,j} \).
where

\[
\tilde{f}_{ik}(Z_{i,k}) = f_{ik}(x_{i,k}) - \frac{k-1}{2} \sum_{j=1}^{k} \partial y_{i,k-1} \hat{x}_{i,j} - \frac{M}{2} \sum_{j=1}^{k} \sum_{q=1}^{M} \left( \frac{\partial y_{i,k-1} \hat{x}_{j,q}}{\partial x_{j,q}} \right) - h \frac{k}{2} \sum_{j=1}^{k} \frac{\partial y_{i,k-1} \hat{x}_{i,j}}{\partial x_{i,j}}.
\]

(52)

Defining virtual controller \( v_{i,k} \) as

\[
v_{i,k} = -c_{i,k1} S_{i,k} - c_{i,k2} S_{i,k} - \frac{b_i \hat{\theta}_{i,k}^2}{2\eta_{i,k}} Z_{i,k} \hat{Z}_{i,k} z_{i,k} - b_i c_{i,k} z_{i,k},
\]

(53)

where \( c_{i,k1}, c_{i,k2}, c_{i,k3} \), and \( \eta_{i,k} \) are positive design parameters.

Choose the adaptive law \( \hat{\theta}_{i,k} \) as

\[
\hat{\theta}_{i,k} = \frac{y_i}{2\eta_{i,k}} S_{i,k} Z_{i,k} Z_{i,k} \hat{Z}_{i,k}^2 - \lambda_i \hat{\theta}_{i,k}.
\]

(54)

Combining (19)–(21) and Assumption 1 and substituting (53) and (54) into (51) yields

\[
V_{i,k} \leq -2^{(1+a)/2} \sum_{j=1}^{k} c_{i,j1} \left( \frac{1}{2} z_{i,j}^2 + \frac{b_i \hat{\theta}_{i,j}^2}{2\gamma_j} \right)^{(1+a)/2} - 2 \sum_{j=1}^{k} c_{i,j2} \left( \frac{1}{2} z_{i,j}^2 + \frac{b_i \hat{\theta}_{i,j}^2}{2\gamma_j} \right)^{(1+a)/2} - c_{i,k1} z_{i,k} S_{i,k} - c_{i,k3} S_{i,k} Z_{i,k}^2 - c_{i,k2} z_{i,k} Z_{i,k}^2 - \alpha \frac{1}{2} \left( \frac{\theta_{i,k}^2}{\sigma_i^2} \right) - \frac{1}{2} \left( \frac{\theta_{i,k}^2}{\sigma_i^2} \right) - \frac{1}{2} \left( \frac{\theta_{i,k}^2}{\sigma_i^2} \right) \leq (\lambda_i b_i / y_i) \left( (\theta_{i,k}^2 / 2) - (\hat{\theta}_{i,k}^2 / 2) \right) \leq (\lambda_i b_i / y_i) \hat{\theta}_{i,k}^2.
\]

(55)

According to (22), (23), and Remark 9, we consider only \( \|z_{i,k}\| \geq \varepsilon_{k,i} \). Then, (55) can be written as

\[
V_{i,k} \leq -2^{(1+a)/2} \sum_{j=1}^{k} c_{i,j1} \left( \frac{1}{2} z_{i,j}^2 + \frac{b_i \hat{\theta}_{i,j}^2}{2\gamma_j} \right)^{(1+a)/2} - 2 \sum_{j=1}^{k} c_{i,j2} \left( \frac{1}{2} z_{i,j}^2 + \frac{b_i \hat{\theta}_{i,j}^2}{2\gamma_j} \right)^{(1+a)/2} \leq \sum_{j=1}^{k} C_j + z_{i,k} z_{i,k+1} \leq M_{i,k-1}(Z_{i,k}) \leq \frac{k}{2} \sum_{j=1}^{k} \frac{\partial y_{i,k-1} \hat{x}_{i,j}}{\partial x_{i,j}}.
\]

(56)
From Lemma 1, Lemma 5, Lemma 6, and (54), it follows that

\[
-z_{ik} \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \leq -z_{ik} \sum_{j=1}^{k-1} \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \left( -\frac{y_{ij}}{2\eta_{ij}} S_{ij}^T \left( Z_{ij}, S_{ij}(Z_{ij})z_{ij}^2 - \lambda_i \theta_{ij} \right) \right)
\]

\[
\leq \sum_{j=1}^{k-1} \frac{z_{ik}^2 (\partial y_{ik-1}/\partial \theta_{ij})^2 \left( (y_{ij}/2\eta_{ij}) S_{ij}^T \left( Z_{ij}, S_{ij}(Z_{ij})z_{ij}^2 \right) \right)^2}{z_{ik}^2 (\partial y_{ik-1}/\partial \theta_{ij})^2 \left( (y_{ij}/2\eta_{ij}) S_{ij}^T \left( Z_{ij}, S_{ij}(Z_{ij})z_{ij}^2 \right) \right)^2 + \kappa^2} + \lambda_i z_{ik} \sum_{j=1}^{k-1} \frac{\partial y_{ik-1}}{\partial \theta_{ij}}\theta_{ij},
\]

(57)

Therefore, \(M_{ik-1}(Z_{ik})\) can be defined as

\[
M_{ik-1}(Z_{ik}) = \sum_{j=1}^{k-1} z_{ij} \left( \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right) \leq -z_{ik} \sum_{j=1}^{k-1} \frac{\partial y_{ik-1}}{\partial \theta_{ij}}
\]

(58)

with the result that

\[
z_{ik} \left( M_{ik-1}(Z_{ik}) - \sum_{j=1}^{k-1} \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right) \leq 0.
\]

(59)

Substituting (59) into (56) yields

\[
\dot{V}_{ik} \leq -2^{(1+\alpha)/2} \sum_{j=1}^{k-1} \frac{1}{2} \left( \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right) \left( \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right)^{(1+\alpha)/2} - 2 \sum_{j=1}^{k-1} \frac{1}{2} \left( \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right) \left( \frac{\partial y_{ik-1}}{\partial \theta_{ij}} \right)^{(1+\beta)/2} + \sum_{j=1}^{k-1} C_j + z_{ik} z_{ik-1}
\]

(60)

Step \(n_i\): according to (15) and (18), we have

\[
\dot{z}_{in} = u_i + f_{in}(x_i) - \dot{\nu}_{in-1},
\]

(61)

where \(\dot{\nu}_{in-1} \triangleq -1 + \sum_{j=1}^{n_i-1} \left( (\partial y_{inj-1}/\partial x_i) x_{ij} + \sum_{j=1}^{n_i} (\partial y_{inj-1}/\partial x_i) x_{ij,0j} + \sum_{j=1}^{n_i-1} \left( (\partial y_{inj-1}/\partial \theta_{ij}) \dot{\theta}_{ij} \right) \right)
\]

Construct a Lyapunov function as

\[
V_{in} = V_{in-1} + \frac{1}{2} \dot{z}_{in}^2 + \frac{b_i \dot{\theta}_{in}}{2y_i}
\]

(62)

The derivative of \(V_{in}\) is written as

\[
\dot{V}_{in} \leq -2^{(1+\alpha)/2} \sum_{j=1}^{n_i-1} \frac{1}{2} \left( \frac{\partial y_{inj-1}}{\partial \theta_{ij}} \right) \left( \frac{\partial y_{inj-1}}{\partial \theta_{ij}} \right)^{(1+\alpha)/2} - 2 \sum_{j=1}^{n_i-1} \frac{1}{2} \left( \frac{\partial y_{inj-1}}{\partial \theta_{ij}} \right) \left( \frac{\partial y_{inj-1}}{\partial \theta_{ij}} \right)^{(1+\beta)/2} + \sum_{j=1}^{n_i-1} C_j + z_{in} f_{in}(Z_{in})
\]

(63)
where

\[
\tilde{f}_{i,k}(Z_{i,n}) = f_{i,n}(x_i) - \sum_{j=1}^{n_1} \frac{\partial y_{i,n-1}}{\partial x_{i,j}} \tilde{z}_{i,j} - \frac{1}{\hat{\beta}_{i,n}} \sum_{j=1}^{M} a_{ij} \sum_{q=1}^{n_2} \left( \frac{\partial y_{i,n-1}}{\partial x_{j,q}} \right) - k_i \sum_{j=1}^{n_2} \frac{\partial y_{i,n-1}}{\partial x_{a,j}} \tilde{x}_{0,j} + z_{i,n-1} - M_{i,n-1}(Z_{i,n}).
\]

Defining virtual controller \( u_i \) as

\[
u_i = -\zeta_{i,n,1} S_{i,n,1} - \zeta_{i,n,2} S_{i,n,2} - \frac{b_i \tilde{\beta}_{i,n} S_{i,n}(Z_{i,n}) S_{i,n}(Z_{i,n}) \tilde{z}_{i,n}}{2 \eta_{i,n}} - b_i \xi_{i,n} \tilde{z}_{i,n},
\]

(65)

where \( \zeta_{i,n,1}, \zeta_{i,n,2}, \zeta_{i,n,3} \), and \( \eta_{i,n} \) are positive design parameters.

Choose the adaptive law \( \tilde{\theta}_{i,n} \) as

\[
\tilde{\theta}_{i,n} = \frac{\gamma_i S_{i,n}(Z_{i,n}) S_{i,n}(Z_{i,n}) \zeta_{i,n} - \lambda_i \tilde{\theta}_{i,n}}{2 \eta_{i,n}}.
\]

(66)

Combining (19)–(21) and Assumption 1 and substituting (65) and (66) into (63) yields

\[
\dot{V}_{i,n} \leq -2^{(1+a)/2} \sum_{j=1}^{n_1} \zeta_{i,j,1} \left( \frac{1}{2} \theta_{i,j}^2 + \frac{b_i \tilde{\beta}_{i,n}^2}{2 \gamma_i} \right) - 2^{(1+a)/2} \sum_{j=1}^{n_1} \zeta_{i,j,2} \left( \frac{1}{2} \theta_{i,j}^2 + \frac{b_i \tilde{\beta}_{i,n}^2}{2 \gamma_i} \right) - \zeta_{i,n,1} S_{i,n,1} - \zeta_{i,n,2} S_{i,n,2} - \zeta_{i,n,3} \left( \frac{b_i \tilde{\beta}_{i,n}}{2 \gamma_i} \right)
\]

(67)

where \( C_{i,n} = \sigma_{i,n} + \beta_{i,n} + (\lambda_i b_i / 2 \gamma_i) \tilde{\theta}_{i,n}^2 \), \( \beta_{i,n} = \zeta_{i,n,2} \left( \frac{1}{2} \theta_{i,n}^2 + \frac{b_i \tilde{\beta}_{i,n}^2}{2 \gamma_i} \right) \), \( \sigma_{i,n} = (1/2) \eta_{i,n}^2 + (e_{i,n}^2 / 4 b_i \xi_{i,n}) \), and \( (\theta_{i,n}^2 - (\tilde{\theta}_{i,n}^2 / 2)) \leq (\lambda_i b_i / 2 \gamma_i) \tilde{\theta}_{i,n}^2 
\)

According to (22), (23), and Remark 9, we consider only \( \|z_{i,n}\| \geq \varepsilon_{n,0} \). Then, (67) can be written as

\[
\dot{V}_{i,n} \leq -2^{(1+a)/2} \sum_{j=1}^{n_1} \zeta_{i,j,1} \left( \frac{1}{2} \theta_{i,j}^2 + \frac{b_i \tilde{\beta}_{i,n}^2}{2 \gamma_i} \right) - 2^{(1+a)/2} \sum_{j=1}^{n_1} \zeta_{i,j,2} \left( \frac{1}{2} \theta_{i,j}^2 + \frac{b_i \tilde{\beta}_{i,n}^2}{2 \gamma_i} \right) + \sum_{j=1}^{n_3} C_{i,j} + z_{i,n} \left( M_{i,n-1}(Z_{i,n}) - \frac{n_1}{\hat{\theta}_{i,n}} \frac{\partial y_{i,n-1}}{\partial \tilde{\theta}_{i,n}} \right).
\]

(68)
4. Stability Analysis

Theorem 1. Consider the multiple uncertain nonlinear strict-feedback system (15) controlled by the distributed fixed-time consensus controller (65) and assume that the leader has
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directed paths to all followers 1 to M. Under Assumptions 1–3, virtual controllers (27), (41), (53), and (65), and the adaptive laws (33), (42), (54), and (66), the consensus tracking errors between the leader and the followers in the overall closed-loop system are CSUUB and can be made arbitrarily small within fixed-time, and the upper limit of the fixed convergence time is irrelevant to the initial state.

Proof. For the stability analysis, choose the total Lyapunov candidate function $V$ as

$$V = \sum_{i=1}^{M} V_{i,n_i},$$

(76)

Differentiating (76) and substituting (75) into it, we have

$$\dot{V} \leq -\sum_{i=1}^{M} \phi_{i,i} V_{i,n_i}^{(1+\alpha)/2} - \sum_{i=1}^{M} \phi_{i,i} V_{i,n_i}^{(1+\beta)/2} + \sum_{i=1}^{M} \tau_i.$$

(77)

Making $\rho_1 = \min \{ \phi_{i,i}, \phi_{i,i}^{(1+\alpha)/2}, \phi_{i,i}^{(1+\beta)/2} \}$ and $\rho_2 = \min \{ \phi_{i,i}^{(1+\alpha)/2}, \phi_{i,i}^{(1+\beta)/2} \}$, in accordance with Lemma 4 yields

$$-\sum_{i=1}^{M} \phi_{i,i} V_{i,n_i}^{(1+\alpha)/2} \leq -\rho_1 \left( \sum_{i=1}^{M} V_{i,n_i} \right)^{(1+\alpha)/2},$$

(78)

$$-\sum_{i=1}^{M} \phi_{i,i}^{(1+\beta)/2} V_{i,n_i}^{(1+\beta)/2} \leq -\rho_2 \left( \sum_{i=1}^{M} V_{i,n_i} \right)^{(1+\beta)/2},$$

where $\rho_1 > 0, \rho_2 > 0, \alpha \in (0,1], \text{and } \beta \in (1,\infty).$

Then, (77) can be rewritten as

$$\dot{V} \leq -\rho_1 V^{(1+\alpha)/2} - \rho_2 V^{(1+\beta)/2} + \tau.$$  

(79)

In accordance with Lemma 3, design proper parameters to make (79) satisfy the following situation.

Case 1. If $V > (1/(1-\omega) \rho_1)^{(1+\alpha)/2}, \omega \in (0,1)$, then (79) can be written as

$$\dot{V} \leq -\omega \rho_1 V^{(1+\alpha)/2} - \rho_2 V^{(1+\beta)/2},$$

(80)

and in that manner, the solution of system (15) converges on the compact set

$$x \in \left\{ V(x) \leq \left( \frac{\tau}{(1-\omega) \rho_1} \right)^{2/(1+\alpha)} \right\}.$$  

(81)

The fixed convergence time is

$$T \leq T_{\max} := \frac{2}{\omega \rho_1 (1-\alpha)} + \frac{2}{\rho_2 (\beta - 1)}.$$  

(82)

Case 2. If $V > (1/(1-\omega) \rho_2)^{(1+\beta)/2}, \omega \in (0,1)$, then (79) can be written as

$$\dot{V} \leq -\rho_1 V^{(1+\alpha)/2} - \omega \rho_2 V^{(1+\beta)/2},$$

(83)

and in that manner, the solution of system (16) converges on the compact set

$$x \in \left\{ V(x) \leq \left( \frac{\tau}{(1-\omega) \rho_2} \right)^{2/(1+\beta)} \right\}.$$  

(84)

The fixed convergence time is

$$T \leq T_{\max} := \frac{2}{\omega \rho_1 (1-\alpha)} + \frac{2}{\rho_2 (\beta - 1)}.$$  

(85)

When cases 1 and 2 are combined, the solution of system (15) converges on

$$x \in \left\{ V(x) \leq \min \left\{ \left( \frac{\tau}{(1-\omega) \rho_1} \right)^{2/(1+\alpha)}, \left( \frac{\tau}{(1-\omega) \rho_2} \right)^{2/(1+\beta)} \right\} \right\}.$$  

(86)

The fixed convergence time is

$$T \leq T_{\max} := \frac{2}{\omega \rho_1 (1-\alpha)} + \frac{2}{\rho_2 (\beta - 1)}.$$  

(87)

It can be seen from (80) and (83) that $V$ is bounded, with the result that $z_{i,j}$ and $\theta_{i,j}$ are bounded, where $i=1, \ldots, M$ and $j=1, \ldots, n_i$, as $\tilde{\theta}_{i,j} = \theta_{i,j} - \theta_{i,j}$ is also bounded. Because $z_{i,j} = \sum_{k=1}^{M} a_{i,j} (y_j - y_i) + b_i (y_i - y_0)$ and $z_{i,j}$ and $y_0$ are bounded, $y_j$ are bounded. Because $x_{i,j} = x_{i,j} - y_{i,k}$ and $x_{i,j}$ and $y_{i,k}$ are bounded. Similarly, $y_{i,k}$ and $x_{i,j}, k=1, \ldots, n_i$ are bounded. Therefore, all signals in the overall closed-loop system are CSUUB.

Remark 11. The distributed fixed-time consensus control algorithm in this paper is different from previous control algorithms. The principles are as follows. (1) Compared to the fixed-time control algorithm proposed in [16–26], our control strategy is applicable to higher-order agents with various forms. That is, a group of the followers with the different unknown nonlinear functions and order of the dynamics can be considered in this brief. (2) To solve the problem of “explosion of complexity”, instead of using dynamic surface design technique [33], we construct a smooth function $M(Z)$, from Lyapunov stability theorem, to compensate the part of Virtual Controller’s derivative ($\partial V/\partial \theta$)$\partial \theta$, and using RBFNNs to approximate the rest of it. (3) RBF neural networks are used here to approximate unknown complex nonlinear functions, obviating the designed control protocol just work for special systems.

5. Simulation Results

In this section, we present two simulation results to validate the proposed theoretical result.

5.1. Mathematical Example. We consider a group of one leader and four followers with the following nonlinear strict-feedback dynamics in the 2D space (i.e., $p=2$) described by [33]
The design parameters are chosen as follows: \( \alpha = 1/3, \beta = 5, \gamma_i = 5, \eta_i = 1/3, \xi_{i,1} = \xi_{i,2} = 0.1, \) and \( \lambda_i = 0.04, \) where \( i = 1, \ldots, 4. \) It is worth noting that \( \xi_{i,11} = \xi_{i,12} = \xi_{i,21} = \xi_{i,22} \) should be at least ten times that of \( \xi_{i,13} = \xi_{i,23} \) and \( \beta_i \) setting of 5 is much better than setting it to 2. For the simulation, RBF neural networks are used to estimate unknown nonlinear functions, where \( W_{i,1}S_{i,1}(Z_{i,1}) \) and \( W_{i,2}S_{i,2}(Z_{i,2}) \) contain nine nodes, and the center of the Gaussian function are set as

\[
\begin{align*}
\xi_{i,1} &= \begin{bmatrix}
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2
\end{bmatrix}^T, \\
\xi_{i,2} &= \begin{bmatrix}
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
\end{align*}
\]

5.2. Practical Example. Consider multiple one-link robotic manipulators consisting of four followers and a leader. The dynamic of each follower manipulator is described as

\[
M\ddot{q}(t) + \frac{1}{2}mgl \sin q(t) = u(t), \quad y(t) = q(t),
\]

where \( q(t) \) represents angular position, \( M \) denotes the instantaneous inertia, \( u(t) \) is the input torque, \( g \) is the gravitational torque, and \( m \) and \( l \) are the mass and the length of the rod, respectively.

Perform coordinate transformation on system (90) and set \( x_{i,1}(t) = q(t) \) and \( x_{i,2}(t) = \dot{q}(t) \); then, system (90) can be rewritten as follows:

\[
\begin{align*}
\dot{x}_{i,1}(t) &= x_{i,2}(t), \\
\dot{x}_{i,2}(t) &= \frac{1}{M} \left( u(t) - \frac{1}{2}mgl \sin x_{i,1}(t) \right), \\
y_i(t) &= x_{i,1}(t),
\end{align*}
\]

where \( i = 1, \ldots, 4. \) The directed network topology for this practical simulation is also shown in Figure 1. We choose \( a_{ij} = 1 \) on \( j \in N_i, a_{ij} = 0 \) otherwise, \( h_i = h_2 = h_3 = 0, \) and \( h_4 = 1. \) The leader output \( r(t) \) is \( y_{l,2}(t) = 0.25 \sin(t) \) and \( \beta_i = 5, \gamma_i = 5, \eta_i = 1/3, \xi_{i,1} = 0.01, \) and \( \lambda_i = 0.04, \) where \( i = 1, \ldots, 4. \) For the simulation, RBF neural networks are used to estimate unknown nonlinear
functions, where \( W_{i,1} S_{i,1}(Z_{i,1}) \) and \( W_{i,2} S_{i,2}(Z_{i,2}) \) contain nine nodes, and the center of the Gaussian function are set as

\[
\begin{bmatrix}
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2 \\
-2 & -1.5 & -1 & -0.5 & 0 & 0.5 & 1 & 1.5 & 2
\end{bmatrix}
\]

(92)

where \( i = 1, \ldots, 4 \). Then, by using the scheme in [36], simulation results are given in Figures 6(a) and 7(a). Meanwhile, by employing our method, simulation results are given in Figures 6(b), 7(b), 8, and 9.

It is shown in Figures 6 and 7 that the tracking performance by using our method is better than [36], since our proposed scheme has faster convergence and higher precision. Figure 8 shows the states \( x_{i,j} \) of the four followers. Figure 9 shows the response curves of adaptive laws \( \hat{\theta}_{i,j} \). These figures reveal that the consensus tracking between the leader and followers is achieved satisfactorily under a directed network topology.

6. Conclusion

In this paper, the distributed adaptive fixed-time tracking consensus control problem is investigated for multiple strict-feedback systems with uncertain nonlinearities under a
Figure 4: The states of the four followers. (a) $x_{1,1}$ and $x_{1,2}$, (b) $x_{2,1}$ and $x_{2,2}$, (c) $x_{3,1}$ and $x_{3,2}$, (d) $x_{4,1}$ and $x_{4,2}$. 
Figure 5: The adaptive laws of the four followers. (a) $\hat{\theta}_{1,1}$ and $\hat{\theta}_{1,2}$. (b) $\hat{\theta}_{2,1}$ and $\hat{\theta}_{2,2}$. (c) $\hat{\theta}_{3,1}$ and $\hat{\theta}_{3,2}$. (d) $\hat{\theta}_{4,1}$ and $\hat{\theta}_{4,2}$.
Figure 6: Output trajectories of the followers with one leader by using two methods. (a) Output trajectories by the method in [36]. (b) Output trajectories by our method.

Figure 7: The tracking errors $e_i$ between followers and leader by using two methods. (a) Error curves by the method in [36]. (b) Error curves by our method.

Figure 8: Continued.
directed graph topology. The function approximation technique using RBF neural networks was employed to compensate for the unknown nonlinearities unmatched in the control input of followers. From Lyapunov stability and fixed-time control theory, by using distributed coordinate conversion, designed virtual controllers, and actual controllers. Based on Lyapunov stability theory, it is proved that the developed distributed consensus protocol can guarantee all the signals in multiple strict-feedback systems are bounded, and the tracking consensus errors converge to a small neighborhood of zero in fixed time. Finally, the effectiveness of the proposed approach was proved through two simulations. And in the following research, we will consider the application of fixed-time control in pure-feedback nonlinear systems with time delay.

**Data Availability**

No data were used to support this study.

**Conflicts of Interest**

The authors declare that they have no conflicts of interest.
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