Plane Wave Diffraction by a Finite Plate with Impedance Boundary Conditions

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Abstract

In this study we have examined a plane wave diffraction problem by a finite plate having different impedance boundaries. The Fourier transforms were used to reduce the governing problem into simultaneous Wiener-Hopf equations which are then solved using the standard Wiener-Hopf procedure. Afterwards the separated and interacted fields were developed asymptotically by using inverse Fourier transform and the modified stationary phase method. Detailed graphical analysis was also made for various physical parameters we were interested in.

Introduction

Diffraction theory can be applied successfully to reduce the noise due to heavy traffic, environmental pollution and industrial growth by means of barriers in heavily built up areas. A barrier should be a good attenuator of sound and inexpensive at the same time. Such barriers may have absorbing lining on the surfaces and satisfy impedance boundary conditions as well. The scattering of sound and electromagnetic waves has been studied extensively since the half plane problems were investigated by Poincare [1] and Sommerfeld [2]. Many classical problems related to electromagnetic waves diffraction due to line source and point source have been studied so far. These problems constitute a canonical problem for the GTD (geometrical theory of diffraction). Scattering analysis by metallic tapes on paneled compact range reflectors [3] and the line source diffraction of electromagnetic waves by a perfectly conducting half plane was investigated by Jones [4]. Rawlins [5] then considered line source diffraction by an acoustically penetrable or an electromagnetically dielectric half plane having smaller width as compared to the incident wave length. In continuation to this, diffraction by an absorbent semi-infinite plane having different impedance faces [6] is also examined. Rawlins’ idea is extended to calculate the diffraction by finite strip [9] and diffraction of spherical acoustic wave from an absorbing plane [10]. A related study of diffraction by a finite airfoil in uniform flow is presented by Jeon et al. [11]. Myers’ presented an improved form of impedance boundary conditions [12] which were used in [13,14] for sound wave diffraction problems.

Diffraction by strips is a significant and classical subject both in electromagnetic and acoustic wave theory. In specific, scattering from resistive, conductive and impedance strips have been considered by Herman et al. [15], while, Senior has also made an attempt to solve a problem related to resistive strip configuration [16]. Many analytical, numerical or approximate analytical methods have been used to study a single or multiple diffraction patterns from a strip. To name a few for example, geometrical theory of diffraction [17], Kobayashi’s potential method [18,19] spectral iteration technique (SIT) [20], method of successive approximations [21] and the W-H technique [22] have positively been utilized. Some recent advances in the literature are also found on Bessel’s potential spaces [23] and Maliuhinets-Sommerfeld integral representation [24].

Keeping in view the aforementioned studies, the major aim of this article is to discuss a wave diffraction problem relating field and its normal derivative as first order impedance conditions which are referred to as standard impedance conditions. These conditions are used as they introduce simplification in calculations to make the problem tractable and to achieve a solution simple enough to use. The impedance conditions can be used effectively for the problems dealing with material surfaces whose solution would be impractical without them. Such conditions are widely used to give analytical solutions to canonical problems. This article provides the comprehensive treatment of impedance boundary conditions applied to electromagnetic. The analytical solutions are amenable to develop high frequency electromagnetic scattering codes and should therefore be of interest to practicing engineers as well as researchers concerned with high frequency diffraction by impedance structures. As mentioned earlier, the finite strip problems have been solved by many researchers who considered different impedance boundary conditions. In the present analysis, the solution to plane wave diffraction by a finite conducting plate with impedance type boundary conditions is produced by the two edges of the finite plate.

The structure of the paper is organized as follows. In Section 0, governing problem which is composed of Helmholtz’s equation,
impedance boundary conditions and continuity conditions, is stated along with its geometrical configuration. The integral transforms are introduced to convert the problem in complex $\mathbb{C}$-plane so that two unknown functions are defined. The three part boundary value problem is simplified in terms of two Wiener-Hopf functional equations in Section 2 from which we derive integral equations in Section 3 using the standard Wiener-Hopf procedure. The procedure is inspired by the book of Noble [22] which is concerned with the application of the Wiener-Hopf technique to the problems involving semi-infinite and finite geometries and discusses a wide range of extensions. In Section 4, analytic approximation for the two unknown functions are derived using asymptotic analysis of the integral equation for large complex argument. The analytical expressions for separated and interacted fields at both edges are computed. In Section 6 the amplitude of the separated field (which contributes in a physical situation) versus observation angle is tested graphically while problem is concluded in Section 7. It is mentioned that the time factor is supposed to be $\exp(i\omega t)$ and neglected throughout the analysis.

**Statement of the Boundary Value Problem**

This section is dedicated to yield the geometric configuration, governing mathematical equation, corresponding boundary conditions and the transformation used to obtain standard Wiener-Hopf functional equations.

Consider the scattering of a time harmonic, plane wave incidence by an impedance finite plate having specific impedance say $\beta_1$. A plate of length $l$ ($-l<x<0$, $y=0$) is encountering a small gust with uniform flow $U$ parallel to the finite plate as shown in Figure 1.

The finite plate is assumed to be infinitely thin and straight. For analysis purpose, it is convenient to express the total electric field as follows

$$\Psi^e(x,y) = \Psi_i(x,y) + \Psi_s(x,y) + \Psi_r(x,y),$$

where $\Psi_i$ is the plane wave incident field and is given by

$$\Psi_i(x,y) = \exp(-ik(x \cos \theta_0 + y \sin \theta_0)),$$

while $\Psi_r$ denotes the field reflected from the finite plate at $y=0$ and is given by

$$\Psi_r(x,0) = \frac{1 - \beta_1 \sin \theta_0}{1 + \beta_1 \sin \theta_0} \exp(-ik(x \cos \theta_0 - y \sin \theta_0)).$$

![Figure 1. Geometrical configuration of the problem.](doi:10.1371/journal.pone.0092566.g001)

Figure 2. Representation of complex $\mathbb{C}$-plane.

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$$\Psi_s(x,y) = \left(\frac{1 - \beta_1 \sin \theta_0}{1 + \beta_1 \sin \theta_0}\right) \exp(-ik(x \cos \theta_0 - y \sin \theta_0)).$$

Here for an absorbent surface it is required that $\Re(\beta_1) > 0$. The diffracted electric field $\Psi^e$ satisfying the following Helmholtz's equation in the range $x \in (-\infty, \infty)$ is given by

$$\left(\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + k^2\right)\Psi^e(x,y) = 0,$$

where $k$ is the free space wave number and for analytic convenience we shall assume that $k = k_1 + ik_2$ ($0 < k_2 \ll k_1$). It is supposed that the medium is slightly lossy and the solution for real $k$ is obtained by letting $k_2 \to 0$. The under considered boundary value problem is expressed in terms of the reduced potential in dimensionalized form, and it is appropriate to denote separated field in the different regions.

As we are interested in determining the diffracted field due to plane wave incidence on the impedance finite plate, Neumann and Dirichlet conditions are imposed along the plate line in mixed type. Different impedance conditions are imposed on upper and lower faces of the conducting plate. Therefore the total diffracted field $\Psi^e$ (which may be named as diffracted electric field due to a conducting plate) is to be determined with the following boundary and continuity conditions

$$\left(1 + \frac{\beta_1}{ik} \frac{\partial}{\partial y}\right)\Psi^e(x,0^+) = 0, \quad -l \leq x \leq 0, \quad y = 0,$$

and

$$\Psi^e(x,0^+) = \Psi^e(x,0^-), \quad -\infty < x < -l, \quad x > 0, \quad y = 0,$$

$$\frac{\partial}{\partial y} \Psi^e(x,0^+) = \frac{\partial}{\partial y} \Psi^e(x,0^-), \quad -\infty < x < -l, \quad x > 0, \quad y = 0.$$

The boundary conditions (5) are the first order impedance conditions relating field and its normal derivative as outlined by Senior et al. [25]. These impedance boundary conditions were subsequently used to model radio waves propagation along the surface of earth and near conducting obstacles. The detail discussion and practical importance of these impedance conditions...
can be found in [26]. The obstacle (finite conducting plate) occupies $y=0, -l < x < 0$ with the velocity of the moving fluid parallel to the $x-$axis having magnitude $U > 0$. The fluid flow is considered as uniform flow moving along the plate. The governing equations are linearized and the special effects of viscosity, thermal conductivity and gravity are ignored while the fluid is assumed to have a constant density (incompressible fluid) and sound speed $c$.

![Figure 3](image-url)

**Figure 3.** Separated field versus observation angle for different values of incidence angle $\theta_0$ when $k = 2$.

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![Figure 4](image-url)

**Figure 4.** Separated field versus observation angle for different values of incidence angle $\theta_0$ when $k = 3$.

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Figure 5. Separated field versus observation angle for different values of incidence angle $\theta_0$ when $k = 3$ and plate length $l = 50$.

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Figure 6. Separated field versus observation angle for different values of admittance parameter $\beta_1$ when $k = 2$.

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Figure 7. Separated field versus observation angle for different values of admittance parameter $\beta_1$ when $k = 3$.

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Figure 8. Separated field versus observation angle for different values of admittance parameter $\beta_1$ when $k = 2$ and plate length $l = 50$.

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Figure 9. Separated field versus observation angle for different values of wave number $k$ when $\beta_1 = 0.5$.
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Figure 10. Separated field versus observation angle for different values of wave number $k$ when $\beta_1 = 0.9$.
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Problem in Transform Domain

For the solution of boundary value problem (4–7), let us introduce Fourier transform with respect to variable \( x \) as:

\[
\begin{align*}
\Psi_\varepsilon(x,y) &= \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \Psi_\varepsilon(x,y) \exp(i\varepsilon x) dx \\
&= \Psi_\varepsilon^+(x,y) + \exp(-i\varepsilon l)\Psi_\varepsilon^-(x,y) + \Psi_{\varepsilon l}(x,y),
\end{align*}
\]  

(8)

Figure 11. Separated field versus observation angle for different values of wave number \( k \) when \( \beta_1 = 0.9 \) and the plate length \( l = 50 \).

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Figure 12. Separated field versus observation angle for values of \( k, \beta_1 \) and \( \theta_0 \) with plate length \( l = 1000 \).

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for $\alpha = \Re\{\alpha\} + \Im\{\alpha\} = \sigma + \ii \tau$. While taking into account the asymptotic behaviors of $\Psi_\alpha(x,y)$ for $x \to \pm \infty$ as

$$\Psi_\alpha(x,y) = \begin{cases} O(\exp(-\ii k x)), & \alpha < 0; \\ O(\exp(\kappa x \cos \theta_0)), & \alpha > 0. \end{cases}$$ (9)

$\Psi_{e-}(x,y)$ is a regular function of $x$ in $\Im\{x\} < \Im\{k\}$ and $\Psi_{e+}(x,y)$ to be regular in $\Im\{x\} > -\Im\{k\}$ and $\Psi_i(x,y)$ to be analytic in the common region $-\Im\{k\} < \Im\{x\} < \Im\{k\}$ which will provide the analytic region for the use of Wiener-Hopf technique, hence.

$$\Psi_{e+}(x,y) = \frac{1}{\sqrt{2\pi}} \int_0^{\infty} \Psi_e(x,y) \exp(\ii x z) \, dz,$$

$$\Psi_{e-}(x,y) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^0 \Psi_e(x,y) \exp(\ii y z) \, dz,$$

$$\Psi_i(x,y) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \Psi_e(x,y) \exp(\ii x z) \, dz.$$ (10)

Now for a plane wave incident on a finite plate, $\Psi_e(x,y) = \exp(-\ii k x \cos \theta_0 - \ii k y \sin \theta_0)$ and the incident field in the transformed domain in the region $-l \leq x \leq 0$, $y = 0$ gives.

$$\Psi_e(x,0) = \frac{\exp(\ii k l (\cos \theta_0 - z)) - 1}{\ii k (\cos \theta_0 - z)}.$$ (11)

Also the reflected field $\Psi_r$ in the transformed domain $z$ is given by.

$$\Psi_r(x,0) = \left(1 - \frac{\beta_i \sin \theta_0}{1 + \beta_i \sin \theta_0}\right) \ii \exp(\ii k l (\cos \theta_0 - z)) - \frac{i}{k(\cos \theta_0 - z)}.$$ (12)

The Fourier transform of Eqs. (4–7) yields.

$$\left(\frac{d^2}{dy^2} + \gamma^2\right) \Psi(z,y) = 0,$$ (13)

where $\gamma(z) = \sqrt{k^2 - z^2}$ with $\Re\{\gamma(z)\} > 0$, also equation (13) is valid for any $x$ in the strip $-k_2 < \Im\{x\} < k_2 \cos \theta_0$. The Fourier transform of boundary conditions (5–7) gives

$$\Psi_e(z,0^+) = -\ii \frac{\beta_i}{\kappa} (\Psi_e(z,0) + \Psi_i(z,0)) - \ii \frac{\beta_i}{\kappa} \Psi_i(z,0^+),$$ (14)

$$\left(\Psi_e(z,0) + \Psi_i(z,0)\right) = \ii \frac{\beta_i}{\kappa} \Psi_i(z,0^-),$$ (15)

and

$$\Psi_e(z,0^+) = \Psi_e(z,0^-) = \Psi_e(z,0),$$

$$\Psi_e(z,0^+) = \Psi_e(z,0^-) = \Psi_e(z,0),$$

$$\Psi_e(z,0^+) = \Psi_e(z,0^-) = \Psi_e(z,0).$$ (16)

**Derivation of Wiener-Hopf Equations**

The solution of Eq. (13) satisfying radiation condition as $x \to \infty$ is given by.

$$\Psi_e(x,y) = \begin{cases} A_1(x) \exp(\ii y), & y \geq 0, \\ A_2(x) \exp(-\ii y), & y < 0. \end{cases}$$ (17)

Now with the help of Eqs. (14–17), the following Wiener-Hopf functional equations are computed

$$\Psi_e'(z,0) + \exp(-\ii z) \Psi_e'(z,0) - \beta_i(z) L(z) J_i(z,0) = \Psi_i(z,0),$$

$$\Psi_e'(z,0) + \exp(-\ii z) \Psi_e'(z,0) + \beta_i(z) L(z) J_i(z,0) = \Psi_i(z,0),$$ (19)

where

$$J_i(z,0) = \frac{1}{2} \left(\Psi_e(z,0^+) - \Psi_e(z,0^-)\right),$$ (20)

$$J_i'(z,0) = \frac{1}{2} \left(\Psi_e'(z,0^+) - \Psi_e'(z,0^-)\right),$$ (21)
\[ A_1(z) = J_1(z,0) + \frac{J_1'(z,0)}{i \gamma L(z)}, \tag{22} \]

\[ A_2(z) = -J_1(z,0) + \frac{J_1'(z,0)}{i \gamma L(z)}, \tag{23} \]

\[ L(z) = \left( 1 + \frac{k}{\beta_1 \gamma(z)} \right), \tag{24} \]

Since Eqs. (18–19) are the Wiener-Hopf equations therefore we proceed to find the solution for these equations in next section.

**Wiener-Hopf Method**

In order to solve the model problem, the intention was, and is, to see the effect of incident wave (which ultimately produces a diffracted field) on a finite conducting plate while considering the impedance boundary conditions. The functional equations (18–19) for the three part boundary value problem are analyzed rigorously using the Wiener-Hopf technique. The main feature of this technique is that it is not fundamentally numerical in nature and thus allows additional insight into the mathematical and physical structure of the diffracted field. The kernel factor appearing in (24) is factorized as

\[ L(z) = \left( 1 + \frac{k}{\beta_1 \gamma(z)} \right) = L_+(z)L_-(z), \tag{25} \]

and

\[ \gamma(z) = \gamma_+(z)\gamma_-(z), \tag{26} \]

where \( L_+(z) \) and \( \gamma_+(z) \) are regular for \( \Im\{z\} > -\Im\{k\} \), i.e., for upper half plane and \( L_-(z) \) and \( \gamma_-(z) \) are regular for \( \Im\{z\} < \Im\{k\} \), i.e., lower half plane. The factorization of such factors is discussed in [21]. Injecting the value of \( J_1(z,0) \) and \( J_1'(z,0) \) from Eqs. (18–19), into (22) and (23), it is found

\[ A_1(z) = \frac{1}{i \gamma L(z)} \left( \Psi_+(z,0) + \exp(-izl)\Psi'_-(z,0) \right. \]

\[ -\Psi_+(z,0) - \Psi_-(z,0) \right) - \frac{ik}{\beta_1 L(z)} \left( \Psi_+(z,0) + \exp(-izl)\Psi_-(z,0) \right. \]

\[ -\Psi_+(z,0) - \Psi_-(z,0), \tag{27} \]

and

\[ A_2(z) = -\frac{1}{i \gamma L(z)} \left( \Psi'_+(z,0) + \exp(-izl)\Psi'_-(z,0) \right. \]

\[ -\Psi_+(z,0) - \Psi_-(z,0) \right) - \frac{ik}{\beta_1 L(z)} \left( \Psi_+(z,0) + \exp(-izl)\Psi_-(z,0) - \Psi_+(z,0) - \Psi_-(z,0) \right). \tag{28} \]

Making use of Eqs. (11) and (12) in Eqs. (18) and (19) to obtain

\[ \Psi_+(z,0) + \exp(-izl)\Psi_-(z,0) + S(z)J_1(z,0) \]

\[ = \frac{b\left( \exp(ikl(\cos \theta_0 - z)) - 1 \right)}{(\cos \theta_0 - z)}, \tag{29} \]

and

\[ \Psi_+(z,0) + \exp(-izl)\Psi_-(z,0) + \frac{\beta_1}{ik} L_+(z)L_-(z)J_1'(z,0) \]

\[ = -\frac{b\left( \exp(ikl(\cos \theta_0 - z)) - 1 \right)}{ik(\cos \theta_0 - z)}, \tag{30} \]

where \( b = \frac{1 - \beta_1 \sin \theta_0}{1 + \beta_1 \sin \theta_0} - 1 \) and \( h = \sin \theta_0 \) with

\[ S(z) = -i \gamma(z)L(z) = S_+(z)S_-(z). \tag{31} \]

Here \( S_+(z) = (k + \frac{1}{2}L_+(z) \) and \( S_-(z) = (k - \frac{1}{2})L-(z) \) are regular in upper and lower half plane \( (z = \sigma + iT, \quad -k_2 < T < -k_2 \cos \theta_0) \), respectively (complex \( z \) plane is shown in Figure 2).

Fortunately equations of types (29) and (30) have been considered by Noble [22] and an analysis based on the same path, may be utilized to obtain an approximate solution for large \( kr \) \( (r = \sqrt{x^2 + y^2}) \). On equating the terms of Eqs. (29–30) with negative sign on one side of the equation and the terms with positive sign on the other side results into a same function say \( J(z) \). Analytic continuation and extended form of Liouville’s theorem extended the function \( J(z) \) throughout the complex \( z \)-plane so that the entire function \( J(z) \) which appears in terms of a polynomial is equated to be zero. Omitting the details of calculations and following the procedure given in [22], it is found that

\[ \Psi_+(z,0) = \frac{bhS_+(z)}{\sqrt{2\pi}} \left( G_1(z) + T(z)C_1 \right), \tag{32} \]

\[ \Psi_-(z,0) = \frac{bhS_-(z)}{\sqrt{2\pi}} \left( G_2(z) + T(-z)C_2 \right), \tag{33} \]
\[ \Psi_{+}(x,0) = -\frac{b\beta L_{+}(z)}{\sqrt{2\pi k}} \left( G'_{+}(z) + T(x)C'_{+} \right), \quad (34) \]

\[ \Psi_{-}(x,0) = \frac{b\beta L_{-}(z)}{\sqrt{2\pi k}} \left( G'_{-}(-z) - T(-z)C'_{-} \right), \quad (35) \]

where

\[ G_{1}(z) = \frac{1}{\cos \theta_{0} - \cos \theta(z)} \left( \frac{1}{S_{+} \left( \cos \theta(z) \right)} - \frac{1}{S_{+} \left( \cos \theta_{0} \right)} \right) \quad (36) \]

\[ T(x) = \frac{1}{2\pi i} E_{-1} W_{-1}(-i(k+x)l), \quad (45) \]

\[ R_{1,2}(z) = E_{-1} \left[ W_{-1}(-i(k+\cos \theta_{0})l) - W_{-1}(-i(k+z)l) \right] / 2\pi i(z+\cos \theta_{0}) \quad (44) \]

\[ G_{2}(z) = \exp \left( -i \cos \theta_{0} \right) \left( \frac{1}{S_{+} \left( \cos \theta(z) \right)} - \frac{1}{S_{+} \left( \cos \theta_{0} \right)} \right) \quad (37) \]

\[ E_{-1} = 2e^{\frac{ik}{2}} \exp \left( ikl \right) \left( i \right)^{-1} h_{-1} \quad (46) \]

\[ C_{1} = S_{+}(k) \left( \frac{G_{2}(k) + S_{+}(k)G_{1}(k)T(k)}{1 - S_{+}^{2}(k)T^{2}(k)} \right), \quad (38) \]

\[ \frac{G_{1}(z)}{S_{+}(k)} = \frac{ik}{L_{+} \left( \cos \theta_{0} \right)} \left( \frac{1}{L_{+}(z)} - \frac{1}{L_{+}(\cos \theta_{0})} \right) \quad (40) \]

\[ C_{2} = S_{+}(k) \left( \frac{G_{1}(k) + S_{+}(k)G_{2}(k)T(k)}{1 - S_{+}^{2}(k)T^{2}(k)} \right), \quad (39) \]

\[ \begin{aligned} \frac{G_{2}(z)}{L_{+}(z)} &= \frac{x}{L_{+}(z)} + \frac{\cos \theta_{0}}{L_{+}(\cos \theta_{0})} - R_{2}(z) \quad (41) \end{aligned} \]

\[ C_{1}' = L_{+}(k) \left( \frac{G_{2}(k) + L_{+}(k)G_{1}(k)T(k)}{1 - L_{+}^{2}(k)T^{2}(k)} \right), \quad (42) \]

\[ \frac{G_{1}(z)}{L_{+}(z)} = \frac{ik}{\sqrt{2\pi i\lambda_{L}(z)}} \left( \frac{1}{L_{+}(z)} - \frac{1}{L_{+}(\cos \theta_{0})} \right) \quad (48) \]

\[ \begin{aligned} \frac{G_{2}(z)}{L_{+}(z)} &= \frac{x}{L_{+}(z)} + \frac{\cos \theta_{0}}{L_{+}(\cos \theta_{0})} - R_{2}(z) \quad (41) \end{aligned} \]

\[ \frac{G_{1}(z)}{L_{+}(z)} = \frac{ik}{\sqrt{2\pi i\lambda_{L}(z)}} \left( \frac{1}{L_{+}(z)} - \frac{1}{L_{+}(\cos \theta_{0})} \right) \quad (48) \]

where \( A_{1}(z) \) corresponds to \( y \geq 0 \) and \( A_{2}(z) \) corresponds to \( y < 0 \). Now, we shall derive a diffracted field expression explicitly in the real space by using the results obtained in (48). The diffracted field \( \Psi_{+}(x,y) \) is obtained by taking the inverse Fourier transform of Eq. (17), that is

\[ \Psi_{+}(x,y) = \frac{1}{\sqrt{2\pi i}} \int_{-\infty}^{\infty} \left\{ A_{1}(z) \right\} \exp \left( i|y| - ixy \right) d\lambda_{z}, \quad (49) \]
two parts
\[ \Psi_e(x, y) = \Psi_{e \text{op}}(x, y) + \Psi_{e \text{int}}(x, y), \quad (50) \]

where
\[ \Psi_{e \text{op}}(x, y) = -\frac{\text{sgn}(y)}{2\pi} \int_{-\infty}^{\infty} \frac{\exp(-il(x, y) - ixz)}{i\gamma L(x)} S_+(z) \exp(i\gamma|y| - i\gamma x) \, dz \]
\[ + \frac{\text{sgn}(y)}{2\pi} \int_{-\infty}^{\infty} \frac{\exp(-il(x, y) - ixz)}{i\gamma L(x)} S_-(z) \exp(i\gamma|y| + i\gamma x) \, dz \]
\[ + \frac{1}{2\pi} \int_{-\infty}^{\infty} L_+(z) \exp(-il(x, y) - ixz) \frac{\partial}{\partial z} \left( \frac{C_1}{\gamma L(z)} \right) \binom{S_+(z)}{R_1(z)} e^{-il|y| - i\gamma x} \, dz \]
\[ + \frac{1}{2\pi} \int_{-\infty}^{\infty} L_-(z) \exp(-il(x, y) - ixz) \frac{\partial}{\partial z} \left( \frac{C_1}{\gamma L(z)} \right) \binom{S_-(z)}{R_2(z)} e^{-il|y| - i\gamma x} \, dz. \quad (51) \]

and
\[ \Psi_{e \text{int}}(x, y) = \frac{\text{sgn}(y)}{2\pi} \int_{-\infty}^{\infty} \frac{1}{i\gamma L(x)} \left( S_+(z) R_1(z) e^{-il|y| - i\gamma x} \right) \, dz - C_1 S_+(z) T(z) + S_+(-x) e^{-ilx} R_2(-x) - C_2 T(-x) S_+(z) e^{-ilx} \]
\[ + \frac{1}{2\pi} \int_{-\infty}^{\infty} \left( T(z) L_+(z) \right) C_1^T \]
\[ + T(-z) L_-(z) C_2^T - L_+(z) R_1(z) e^{-ilx} \]
\[ - L_-(z) R_2(z) e^{-ilx} \] \[ \, dx. \quad (52) \]

Hence, for large \( kr \), Eqs. (51–52) take the following forms
\[ \Psi_{e \text{op}}(x, y) = \frac{ik}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \frac{\pi}{2kr} \left( A_1(-k \cos \theta) \right) \sin \theta \exp(ikr + i\pi/4), \quad (53) \]
with
\[ \Psi_{e \text{op}}(x, y) = -\text{sgn}(y) \frac{ikf_1(-k \cos \theta)}{4nk} \sqrt{\frac{1}{kr}} \exp(ikr + i\pi/4), \quad (54) \]
and
\[ \Psi_{e \text{int}}(x, y) = -\text{sgn}(y) \frac{ikf_2(-k \cos \theta)}{4nk} \sqrt{\frac{1}{kr}} \exp(ikr + i\pi/4), \quad (55) \]
where \( A_1(-k \cos \theta) \) and \( A_2(-k \cos \theta) \) can be found from Eq. (48), while
\[ f_1(-k \cos \theta) = \frac{S_+(-k \cos \theta)}{L(-k \cos \theta) S_+(k \cos \theta)} \]
\[ \frac{\exp(-il(x, y) + k \cos \theta)) S_+(k \cos \theta)}{L(-k \cos \theta) S_+(k \cos \theta) k \cos \theta}, \quad (56) \]
\[ g_1(-k \cos \theta) = \frac{\exp(-il(x, y) + k \cos \theta))}{(k \cos \theta)} \]
\[ \frac{L_+(k \cos \theta)}{L(-k \cos \theta) L_+(k \cos \theta)} \quad (57) \]
\[ f_2(-k \cos \theta) = \frac{1}{L(-k \cos \theta)} \left( S_+(-k \cos \theta) R_1(-k \cos \theta) e^{-il \cos \theta} \right) \]
\[ + S_+(k \cos \theta) e^{ik \cos \theta} R_2(k \cos \theta) - C_1 S_+(-k \cos \theta) T(-k \cos \theta) \]
\[ - C_2 T(k \cos \theta) S_+(k \cos \theta) e^{ik \cos \theta}), \quad (58) \]
\[ g_2(-k \cos \theta) = \frac{1}{L(-k \cos \theta)} \left( L_+(-k \cos \theta) R_1(-k \cos \theta) e^{-il \cos \theta} \right) \]
\[ + L_+(k \cos \theta) R_2(k \cos \theta) e^{ik \cos \theta} - T(-k \cos \theta) L_+(k \cos \theta) C_1^T \]
\[ - T(k \cos \theta) L_+(k \cos \theta) C_2^T. \quad (59) \]

Acquisition of Diffracted Field

The diffracted field in the far zone may now be calculated by evaluating the integrals appearing in Eqs. (51–52), asymptotically [27]. For that introducing the polar coordinates as \( x = r \cos \theta \), \( y = r \sin \theta \) and deforming the contour by the transformation
\[ z = -k \cos (\theta + i\xi) \quad \text{where} \quad 0 < \theta < \pi, \quad -\infty < \xi < \infty. \]

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is noted that the separated field represents the field diffracted by
the edges at $x = -l$ and $x = 0$ plus the additional contribution to
the geometrical wave field not included in the incident field. The
separated terms are the resultant wave field which will contribute
in the physical insight. Whilst interacted field gives the interaction
of one edge with the other but It gives no physical insight. Only
the separated field is discussed numerically as it provides the
physical insight to the diffraction problem at a defined boundary.
Moreover the interaction field is produced due to double
diffraction of the two edges which is already considered as
separated field is calculated by the edges at $x = -l$ and $x = 0$. Also
the contribution coming from the interacted terms is eliminated
when the plate length is widened up to infinity and only the
separated terms will result into the diffracted field. Therefore, we
only discuss separated field computationally in the subsequent
section.

Numerical Results and Discussion

The effect of the values of incident angle, plate length, specific
impedance and the wave number on the diffraction phenomenon
is observed computationally by displaying the variation of the
separated field ($20 \log \rho$) versus observation angle. In Figures 3–4,
the results are presented for the different values of incident angle
$\theta_0$ by fixing all other parameters. On increasing the value of
incident angle, the amplitude of the separated field is increased
and is maximum for the value of incident angle where $\theta \to 0$. Also
while deforming the contour into a hyperbola, we make the
transformation.

$$x = -k \cos(\theta + i\xi) \quad \text{where} \quad 0 < \theta < \pi, \quad -\infty < \xi < \infty$$

so that the contour over $x$ goes into a hyperbola. The two
hyperbolae will not cross each other if

$$a = \sqrt{c^2 + b^2}.$$

The two hyperbolae will not cross each other if $a < c$ but if the inequality
is inverted, there will be a contribution from the pole which actually
cancels the incident wave in the shadow region. In Figure 4, the
value of wave number is slightly increased which results in little
oscillations in the graph while Figure 5 is plotted by extending the
plate length. Obviously the overall behavior remains same but the
amplitude of the separated field decreases by increasing the plate
length and the incident angle. Figures 6–8 are plotted for the
different values of specific impedance by fixing all other parameters. It can be seen that by increasing the specific
impedance the amplitude of the separated field reduces consid-
erably. It is observed from all the figures that the scattered far field
has maximum peaks at different values of observation angle
between 0 and $\pi$. In Figures 9–11, the amplitude of the separated
field versus observation angle is observed for different values of
wave number $k$. The separated field increases by increasing the
value of wave number, $i.e.$, the wave dimension then moves
towards the high frequency range.

It is interesting to note that, while increasing the length of plate,
the scattered far-field intensity has sharp peaks for almost every
value of observation angle. If we keep on increasing this length, the
graphs for all the values of $k$, $\beta_1$ and $\beta_0$ will behave like Figure 12.
That is, the overall behavior for different parameters remains alike
but the intensity of separated field got sharp oscillating peaks and
possibly this could be the results for the half plane. Consideration
on the structure of a finite conducting plate may offer a physical
understanding of the scattering phenomenon at these particular
values of observation angle.

Concluding Remarks

The principle concern of this investigation is to discuss plane
wave diffraction by a finite conducting plate with different
impedance boundary conditions. The two edges of the finite plate
give rise to two diffracted fields (one from each edge), that is, the
separated field and the interacted field (due to interaction of one
degree upon the other). Explicit expressions for the separated and
the interacted terms are obtained asymptotically using the
modified stationary phase method. The final solution obtained
here is rigorous and uniformly valid for impedance boundaries.
According to Rawlins [28] the impedance half plane gives better
attenuation results than a completely rigid semi-infinite plane for
singly diffracted fields. Therefore the impedance finite strip (plate)
gives better attenuation results for the separated and interacted
fields as compared to a completely rigid strip. As the mod value of
given field is directly proportional to the perturbation sound
pressure which ultimately yields a measure of sound intensity, few
graphs showing the effects of various parameters on the separated
diffracted field are presented and discussed. The consideration
of plane wave diffraction by finite plate will go a step further to
complete the discussion for line source/point source and slit/half
plane. If the obstacle (whether finite or semi-infinite) is taken as a
noise barrier, the consideration of trailing edge situation can also be
more efficient to reduce the noise in this region.

Author Contributions

Conceived and designed the experiments: RN MA. Performed the
experiments: RN MA. Analyzed the data: RN MA. Contributed
reagents/materials/analysis tools: RN AJ. Wrote the paper: RN AJ.
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