Ball on plate platform is a great tool for testing of various control techniques. This study provides a detailed design and comparison of three control structures for automatic stabilization of the ball’s location on plate and ball trajectory tracking. Namely PD controller, state space feedback and state space observer with state space feedback. All three methods are designed and practically compared. All structures have a common feature. It is the methodology for calculating the parameters of the controllers. Ball - on - plate mathematical model derived by Newton’s method and DC motor position loop control are also part of this article.

For the authors of this article, this study is an opportunity to create an educative setup to explain the principles of basic control system for non - technical person and offers motivation to future students.

**Keywords:** ball on plate, trajectory tracking, parameter analysis, position control systems

1. **Introduction**

In application practice, the properties of new control theories are often investigated at inverted pendulum. Balancing a ball of on two - axis platform is a variant of this testbed in which you can successfully find the limits and observe behavior of the control structure. Practical impact of these control structures is immense. Generally, in defense military technology such as camera stabilizations on moving vehicle or pointing camera angle control during drone flying are used. The same principle of operation like in this article is used for stabilizing platforms in sea ships and aircrafts. General idea is also used in high - end applications like vertical landing of space rockets.

Bench - marking system consists of two DC machines connected with a long stick at moving platform. DC motors are controlled via standard full bridge topology. The processing unit is Kinetis K60 which is a novel digital signal controller designed for the automotive industry. The position of subject placed at plate is obtained from the resistive touch panel in terms of voltage drop. In some articles, the camera [1, 2] is used to estimate the position of the globule. Optical detection requires heavy processing for position estimation. This method leads to huge computation power. Another approach is based on laser technology or basic sonar rangefinders. For stabilizing one degree of freedom applications gyro and accelerometers are used to measure system states [3]. Resistive touch surface excel over others due to its positive features. From the perspective of the automatic regulation conductive surface represents zero order transfer function. This means that the response will be recovered immediately without undesirable delay [4].

2. **Ball on plate mathematical model**

To be able to analyze the behavior of an ideal ball - on - plate balancing system it is necessary to create mathematical model and equivalent transfer function.

Our testbed is biaxial. It means that it has two degrees of freedom X and Y, and in a center is anchored on a rotating joint. The ball is stabilized and controlled to the demanded position through the movements of the plate, which are performed by two DC motors with permanent magnets. Each of the motors is responsible for the movements of the platform just in one axis. This is the reason why the same scheme is applied to both degrees of freedom Figure 1b).
Substituting Equation (1) and Equation (5) into the Equation (6) leads to:

$$m \ddot{x} + \frac{2}{3} m \dot{x} = m g \sin \alpha$$  \hspace{1cm} (7)$$

After mathematical corrections of Equation (7) we can get the relationship between ball acceleration and $\sin(\alpha)$. Due to $\sin$ function, nonlinear system model was created. In the area of angles up to $\pm 5^\circ$ it is possible to neglected the influence of $\sin$ function and substitute expression $\sin(\alpha) = \alpha$. By applying Laplace transformation, we can get transfer function of the ball on plate system. Relation between ball position $x$ and inclination angle of the plate $\alpha$ is as follows:

$$F_x(\alpha) s = \frac{x(s)}{a(s)} = \frac{\frac{g}{s}}{\frac{\frac{1}{s}}{s}}$$  \hspace{1cm} (8)$$

Where $g$ is gravitational acceleration. Relation between motor arm angle and inclination angle of the plate:

$$\phi = \frac{1}{d} \alpha$$  \hspace{1cm} (9)$$

Our main goal during stabilizing platform is to ensure a stable objective position and to achieve position reference tracking. Active force in this task is gravity. The ball is capable of rolling on the beam under the action of gravity. Combination of gravitational force and inclination of plate leads to unstable position change in an open loop system. Root locus plot Figure 2b) explains instability of derivation transfer function Equation (8). Closer look for steady state shows us the behavior of the system under the following condition $s = 0$. When $s$ approaches 0, the denominator of the transfer function approaches 0, but final value of Equation (8) approaches infinity Figure 2a). A well-known principle as the final value theorem applied at Equation (8), lead us directly to the reason for instability. The same theorem offers a huge space for feedback control by algorithms ensuring stability.
3. Control of DC machines

3.1 Mathematical model of DC machine with permanent magnet

Inclination of the plate, respectively the angle of platform, is controlled by standard position loop of DC machine. Permanent magnet DC machine is represented by following equations:

\[
\frac{d\omega}{dt} = \frac{1}{L_i} (U_v - R_i i - \omega \psi_{ns}) \quad \frac{di}{dt} = \frac{P}{J} (M_v - M_{in}),
\]

Parameters of DC machines are listed in the appendix.

3.2 Position loop design

Angular position control with PI controllers was implemented in to Kinetis K60. The most important in this step is to ensure the prescribed behavior and set up correct dynamic response. Standard Dodd’s formula was applied in prescribed band width finding [8] in combination with pole placement method. The results of this approach describe comparison between simulation and real measurement Figure 4. Step response and bode plots for speed and position loop are shown below.

From a practical standpoint, knowing how the system responds to Heaviside step function is important because this concept shows us systems transient phenomenon. Fast and stable systems such as in Figure 3 and Figure 4 with settling time 0.05 sec are able to damp fast deviations. Bode plot analysis shows ability of the trajectory following. Any trajectory can be divided into axes X, Y. The resulting curve is decomposed by offline FFT and then compared with the original position loop bode plot Figure 4b). If the resulting frequencies are part of the position loop bandwidth, then trajectory will be followed without error.

4. Master control algorithm

The proposed testbed is controlled by the Kinetis K60 digital signal controller. K60 device detects five feedbacks. Voltage drop on resistive touch screen, two currents from hall effect sensors and pulses from two encoders. Each DC machine has its own full bridge topology of power supply converter integrated in BTN8982/86TA.

5. Stabilization design

The first experimentally verified structure is standard PD controller. By adding the PD controller (the green part of Figure 6a)), a stable structure is created to provide the both objectives. The transfer function Equation (11) of the whole structure was derived from the Masson’s rule formula [8]. Parameters of PD controller \( K_p \), \( K_d \) were determined by pole placement method. Equation (11) shows experimental results of PD structure.

Closed loop transfer function of Figure 6a) PD controller:

\[
F_{vo}(s) = \frac{F_{vo}(s)}{1 + F_{vo}(s)} = \frac{a_1 \cdot (K_p + K_d \cdot s)}{s^2 + a_1 \cdot K_v \cdot s + a_1 \cdot K_p},
\]

\[
K_v = \frac{2 \cdot \xi \cdot \omega_n}{a_1}, \quad K_p = \frac{\omega_n^2}{a_1}
\]

Figure 2 a) Step response of Equation (8); b) Pole placement of Equation (8)

Figure 3 Axis inclination control via DC machine position loop
The second control algorithm is based on a full state feedback. A state variable is one of the variables used to describe ability to accumulate energy which is speed and position of the ball. The general equation of the linear system with the state feedback is Equation (12). After the following adjustment occurs Equation (13).

$$
\dot{X} = A \cdot X + B \cdot (K_s \cdot Ball\_position\_dem - K_s \cdot X) \quad (12)
$$

$$
\dot{X} = (A - B \cdot K_s) \cdot X + B \cdot K_s \cdot Ball\_position\_dem \quad (13)
$$

Where $X$ is a matrix of state variables (ball position, ball velocity), $A$ is a system matrix, $B$ is a matrix of input variables,
\[ M_{c}\alpha = [A - B \cdot K_e], \]
\[ [A - B \cdot K_s] = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ -K_{s1} \end{bmatrix} = \begin{bmatrix} 0 \\ -K_{s1} \end{bmatrix} \] (14)

stability of the system, the left side of the complex plane is selected. Eigenvalues are obtained by solving characteristic equation \( \det [\lambda - A] = 0 \). Subsequently, the result is compared to the polynomial with the prescribed behavior Equation (15). Searched constants \( K_s, K_z \) Equation (16) have finally emerged.

\[ \det [1 - A] = \lambda^2 + \lambda K_s \alpha_1 + K_s \alpha_1; \]
\[ \lambda^2 + \lambda K_s \alpha_1 + K_s \alpha_1 = s^2 + 2 \xi \omega_1 + \omega_1^2; \]
\[ K_1 = \frac{\omega_1^2}{\alpha_1}; \quad K_z = \frac{2 \xi \omega_1}{\alpha_1}; \quad \frac{K_s}{K_1} = 1 \Rightarrow K_s = K_1 \] (15)

At this point, we are again using the final value theorem. There is a steady state error which will be compensated by the \( K_z \) constant Equation (16).

The last structure uses speed feedback from the state space observer. In the previous structures, the speed was obtained by differentiation of position. For the last structure, the derivation was replaced by a more complex observer. State observer reconstructs state variables by using a mathematical model of a controlled system. Observer equations are created by adding measurable deviation \( \varepsilon = x - \hat{x} \) into each system Equation (17).

\[ \begin{bmatrix} \dot{x} \\ \dot{\varepsilon} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ \varepsilon \end{bmatrix} + \begin{bmatrix} g_z \end{bmatrix} (x - \hat{x}) \] (17)

\[ a_1 = \frac{2}{T_g} = \frac{g_z}{h} \]

\[ \dot{\varepsilon} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \varepsilon \\ \dot{\varepsilon} \end{bmatrix} - \begin{bmatrix} g_z \end{bmatrix} \varepsilon = \begin{bmatrix} \dot{\varepsilon} \\ \dot{\varepsilon} \end{bmatrix} \begin{bmatrix} -g_z & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \varepsilon \\ \dot{\varepsilon} \end{bmatrix} \] (18)

This stability condition is met when eigenvalues of the dynamic error system are located in the left part of the complex plane, \( \det [\lambda - A] = 0 \). The result is compared to the second order polynomial with the prescribed behavior Equation (19). The calculated constants of the observer are according to the Equation (20).

\[ \det [\lambda - A] = \begin{bmatrix} \lambda + g_z - \frac{1}{\lambda} \\ g_z \end{bmatrix} = \lambda^2 + \lambda g_z + g_z, \]
\[ \lambda^2 + \lambda g_z + g_z = s^2 + 2 \xi \omega_1 + \omega_1^2; \]
\[ g_z = 2 \cdot \xi \cdot \omega_0; \quad g_z = \omega_0^2 \] (19)

The same approach for designing the parameters was used in all three stabilization techniques.

All techniques were set using pole placement method. Settling time resp., the bandwidth was chosen by Dodd’s formula. In the Figure 7 [9] there is a comparison of stabilization structures, which were implemented on real platform Figure 5b). All three structures have the same settling time \( T_{sett} = 1 \text{sec} \). On the left side
Due to settling dynamics of the state space observer, the last structure shows the best results. The entire control diagram used for the testbed regulation is in the Figure 9b). The black part is position loop of DC machine. The blue belongs to the mathematical model of the ball at beam. The orange part represents the state space control low, and with purple color, the state space observer is highlighted. The two most important graphs are plotted for the analysis.

In step response, the initial intersection in the negative position is seen. This is mostly due to the mechanical backlash of the system. The following increase and settlement is in a very good (Figure 8) accordance with the simulations of the structure in Figure 9b). Simulated and measured frequency characteristics are just as good as shown in Figure 8. The reference trajectory tracking is shown in Figure 9a).
behavior of the most successful structure is on the following link: https://www.youtube.com/watch?v=Rr90hb_Rn3M.

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Appendix

Table 1 Parameters of used PMDC

| Sign | Definition                              | Value   | Unit |
|------|----------------------------------------|---------|------|
| $R_a$ | Armature winding resistance            | 5.32    | Ω    |
| $L_a$ | Armature winding inductance            | 0.01534 | H    |
| $\Psi_{PM}$ | Linkage magnetic flux of PM   | 0.1432  | Wb   |
| $J$   | Rotor moment of inertia                | 0.0000414 | kg.m² |
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