Transient Performance of Electric Power Networks under Colored Noise

T. Coletta, B. Bamieh, and Ph. Jacquod

Abstract—New classes of performance measures have been recently introduced to quantify the transient response to external disturbances of coupled dynamical systems on complex networks. These performance measures are time-integrated quadratic forms in the system’s coordinates or their time derivative. So far, investigations of these performance measures have been restricted to Dirac-δ impulse disturbances, in which case they can be alternatively interpreted as giving the long time output variances for stochastic white noise demand/generation fluctuations. Strictly speaking, the approach is therefore restricted to power fluctuating on time scales shorter than the shortest time scales in the swing equations. To account for power productions from new renewable energy sources, we extend this earlier works to the relevant case of colored noise power fluctuations, with a finite correlation time \( \tau > 0 \). We calculate a closed-form expression for generic quadratic performance measures. Applied to specific cases, this leads to a spectral representation of performance measures as a sum over the non-zero modes of the network Laplacian. Our results emphasize the competition between inertia, damping and the Laplacian modes, whose balance is determined to a large extent by the noise correlation time scale \( \tau \).

I. INTRODUCTION

Transmission system operators constantly monitor electric power grids and evaluate their potential response to possible faults and unexpected disturbances [1]. Standardly, frequency nadir and maximum rate of change of frequency (RoCoF) are the indicators of choice. This is so because when they exceed pre-defined values, they trigger corrective measures such as activation of primary control and disconnection of power plants from the grid, or even, in worst cases, of entire geographical areas. Frequency nadir and RoCoF however only partially characterize the transient excurses away from the previously operating synchronous state. Inspired by consensus and synchronization studies [3], [4], [5], [6], recent works have introduced new transient performance measures following disturbances on busses [7], [8], [9], [10], [11] and on power lines [12]. The method is particularly appealing because (i) these performance measures have a clear physical meaning, quantifying the additional ohmic losses [7], [8], [9], the primary control effort [10], [12] originating from the transient or the phase coherence of the grid [3], [4], [12]; (ii) they are integrated quadratic forms in phase or frequency excursions that can be expressed as \( L_2 \) norms of the system’s output. As such, they are conveniently calculated through an observability Gramian solution to a Lyapunov equation [13].

Technically speaking, the advantage of the \( L_2 \)-based approach over infinity norm measures (frequency nadir and RoCoF) is not only that it is mathematically tractable, but also that it provides combined information on both the amplitude and the duration of the transient in both voltage phases and frequencies. Its drawback is that so far it has been applied to Dirac-\( \delta \) impulse disturbances only [7], [8], [9], [10], [11], [12], which is equivalent to considering long time output variances for stochastic white noise power demand/generation fluctuations. Fluctuations of photovoltaic or wind turbine power generation have however finite correlation times [14], typically on the order of minutes or more, i.e. significantly longer than typical time scales in swing equations. To model them it is therefore desirable to go beyond white noise power fluctuations. In this work we extend the observability Gramian formalism to treat the swing dynamics under colored noise inputs. In the spirit of the method proposed in [15] in the different context of turbulent flows, we achieve this by means of a filter to generate colored noise from a white noise stochastic input [16]. We obtain fluctuating power generations with exponentially decaying correlations, with a tunable characteristic correlation time \( \tau \) which we take as a parameter in our model. We provide a closed form expression for any output representing a quadratic performance index. To illustrate our theory, we consider a performance measure quantifying voltage angle coherence in the network. For fluctuating injections localized on a single node, we show how different correlation time scales lead to qualitatively different behaviors of this quadratic performance measure.

This paper is organized as follows. Section II introduces some mathematical notations. Section III presents the model and the observability Gramian formalism for colored noise. In Sec. IV we diagonalize the dynamics and present a closed form expression for generic performance measures. In Sec. V we illustrate our theory for a specific case of an angle coherence performance measure and discuss the obtained results. A brief conclusion is given in Sec. VI.

II. MATHEMATICAL NOTATION

Given a vector \( \mathbf{v} \in \mathbb{R}^N \) and a matrix \( \mathbf{M} \in \mathbb{R}^{N \times N} \) we denote their transpose by \( \mathbf{v}^\top \) and \( \mathbf{M}^\top \). In terms of the components \( v_1, \ldots, v_N \), we also denote the vector \( \mathbf{v} \) as \( \mathbf{v} \equiv \text{vec}(\{v_i\}) \), while \( \text{diag}(\{v_i\}) \in \mathbb{R}^{N \times N} \) denotes the diagonal matrix having \( v_1, \ldots, v_N \) as diagonal entries. The \( \ell \)-th unit vector \( \mathbf{e}_\ell \in \mathbb{R}^N \) has components \( (\mathbf{e}_\ell)_i = \delta_{i\ell} \).
We denote undirected weighted graphs by \( G = (V, E, W) \) where \( V \) is the set of \( N \) vertices, \( E \) is the set of edges, and \( W = \{ w_{ij} \} \) is the set of edge weights, with \( w_{ij} = 0 \) whenever \( i \) and \( j \) are not connected by an edge, and \( w_{ij} = w_{ji} > 0 \) otherwise. The graph Laplacian \( L \in \mathbb{R}^{N \times N} \) is the symmetric matrix with components \( L_{ij} = -w_{ij} \) if \( i \neq j \) and \( L_{ii} = \sum_{i \neq j} w_{ij} \). We denote by \( \{ \lambda_1, \ldots, \lambda_N \} \) and \( \{ u^{(1)}, \ldots, u^{(N)} \} \) the eigenvalues and orthonormalized eigenvectors of \( L \). The orthogonal matrix \( T \in \mathbb{R}^{N \times N} \) having \( u^{(i)} \) as \( i \)th column diagonalizes \( L \), i.e. \( T^T L T = \Lambda \) where \( \Lambda = \text{diag}(\{ \lambda_i \}) \). The zero row and column sum property of \( L \) implies that \( \lambda_1 = 0 \) and that \( u^{(1)^T} = [1, \ldots, 1]/\sqrt{N} \). In connected graphs, all remaining eigenvalues of \( L \) are strictly positive, \( \lambda_i > 0 \) for \( i \geq 2 \). The column row vector product \( u^{(1)} u^{(1)^T} \) is the \( N \times N \) matrix having \( 1/N \) for all its entries.

### III. POWER NETWORK MODEL

We consider the swing dynamics of high voltage transmission power networks in the DC approximation. This approximation of the full nonlinear dynamics assumes uniform and constant voltage magnitudes, purely susceptive transmission lines and small voltage phase differences. We consider a Kron reduced network such that each of its \( N \) nodes models a synchronous machine (generator or consumer) of rotational inertia \( m_i > 0 \) and damping coefficient \( d_i > 0 \). The steady state power flow equations relating the active power injections \( P \) to the voltage phases \( \theta \) at every node define the nominal operating point \( \theta^* \) through \( P = L \theta^* \). Here, \( L \) is the Laplacian matrix of the graph modeling the Kron reduced electric network and whose edge weights are given by the effective susceptances \( b_{ij} \geq 0 \).

Subject to a power injection disturbance \( p(t) \), the system deviates from the nominal operating point according to \( \theta(t) = \theta^* + \varphi(t) \), and \( \omega(t) = \dot{\varphi}(t) \). In the DC approximation, and in a frame rotating at the nominal frequency of the network, the swing equations read [1]

\[
M \ddot{\varphi} = -D \dot{\varphi} - L \varphi + p(t),
\]

with \( M = \text{diag}(\{ m_i \}) \) and \( D = \text{diag}(\{ d_i \}) \). Given performance outputs of the form

\[
y = \begin{bmatrix} Q^{(1,1)} & 0 \\ 0 & Q^{(2,2)} \end{bmatrix}^{1/2} \begin{bmatrix} \varphi \\ \omega \end{bmatrix},
\]

we want to assess the long time output variance

\[
\lim_{t \to \infty} \mathbb{E}[y(t)^T y(t)]
\]

for the swing dynamics (1) subject to fluctuating power injection \( p(t) \) defined by the correlator

\[
\mathbb{E}[p_i(t_1)p_j(t_2)] = \delta_{ij} p_0^2 e^{-|t_1 - t_2|/\tau}.
\]

In (4), \( p_0^2 \) denotes the equal time variance of the power injection disturbance at node \( i \), \( \tau \) is the characteristic correlation time scale, and \( \mathbb{E} \) denotes the expectation value.

Obtaining (3) is equivalent to considering Dirac--\( \delta \) impulse disturbances in the augmented dynamical system (see Appendix VII-A)

\[
\begin{bmatrix} \ddot{\varphi} \\ \dot{\omega} \end{bmatrix} = \begin{bmatrix} -M^{-1} L - M^{-1} D & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \varphi \\ \omega \end{bmatrix} + \begin{bmatrix} 0 \\ \delta(t) \end{bmatrix},
\]

with \( \eta_0 = \sqrt{2/\tau} \), and measuring the transient performance by evaluating the quadratic measure

\[
P = \int_0^\infty \left[ \varphi^T \omega^T \eta \right] Q \begin{bmatrix} \varphi \\ \omega \end{bmatrix} dt,
\]

which remains finite under the assumption that \( u^{(1)} \in \ker(Q^{(1,1)}) \). Following [11] and [12], we introduce the change of variables \( \overline{\varphi} = M^{1/2} \varphi \) and \( \overline{\omega} = M^{1/2} \omega \). This allows us to rewrite (5) as

\[
\begin{bmatrix} \ddot{\varphi} \\ \dot{\omega} \end{bmatrix} = \begin{bmatrix} 0 & I \\ -L_M & -M^{-1} D - M^{1/2} p \end{bmatrix} \begin{bmatrix} \varphi \\ \omega \end{bmatrix} + \begin{bmatrix} 0 \\ \delta(t) \end{bmatrix},
\]

where \( L_M \) is the symmetric matrix

\[
L_M = M^{-1/2} L M^{-1/2}.
\]

For initial conditions \( (\varphi(0), \omega(0), \eta(0)) = (0, 0, 0) \), solving (7) yields

\[
\begin{bmatrix} \varphi(t) \\ \omega(t) \end{bmatrix} = e^{At} \begin{bmatrix} 0 \\ \eta_0 \end{bmatrix}.
\]

The performance measure (6) can be expressed as

\[
P = B^T X B,
\]

with the observability Gramian \( X = \int_0^\infty e^{At} Q M e^{At} dt \), and

\[
Q_M = \begin{bmatrix} M^{-1/2} Q^{(1,1)} M^{-1/2} & 0 \\ 0 & M^{-1/2} Q^{(2,2)} M^{-1/2} \end{bmatrix}.
\]

In what follows we denote the non zero blocks of \( Q_M \) by \( Q_M^{(1,1)} = M^{-1/2} Q^{(1,1)} M^{-1/2} \) and \( Q_M^{(2,2)} = M^{-1/2} Q^{(2,2)} M^{-1/2} \) respectively.

When the matrix \( A \) is Hurwitz, the system is asymptotically stable and the observability Gramian \( X \) satisfies the Lyapunov equation

\[
A^T X + X A = -Q_M.
\]

In the present case however, the system is Laplacian and it follows from \( L u^{(1)} = 0 \) that \( A \) has a marginally stable mode \( A[M^{-1/2} u^{(1)}, 0]^T = 0 \). Nevertheless, this unobservable mode does not carry any relevant physical information and only reflects the model’s invariance under a global shift of all voltage phases. Standard approaches to deal with this marginally stable mode include: (i) considering performance measures \( Q \) such that \( u^{(1)} \in \ker(Q^{(1,1)}) \), in which case

\[
X = \int_0^\infty e^{At} Q M e^{At} dt.
\]
the observability Gramian is well defined by (12) with the additional constraint $X[M^{1/2}u(1),0,0] = 0$ [7], [9], [10], and (ii) introducing a regularizing parameter, $\epsilon$, in the Laplacian making it nonsingular, and taking the limit $\epsilon \to 0$ only at the very end of the calculation of a performance measure [12]. In the derivations of this manuscript we will follow the latter approach.

IV. CLOSED FORM EXPRESSION FOR QUADRATIC PERFORMANCE MEASURES

Before providing a closed form expression for performance measures of the type (6), we first recall two results proven in [12].

**Proposition 1 (Laplacian regularization):**
Under the transformation $L \to L + \epsilon I$, with regularizing parameter $\epsilon > 0$, the system defined in (7) is asymptotically stable and has no marginally stable mode.

**Proposition 2 (Solution of the Lyapunov equation):**
Let $A$ be a non symmetric, diagonalizable matrix with eigenvalues $\mu_i \neq 0$. Let $T_R$ ($T_L$) denote the matrix whose columns (rows) are the right (left) eigenvectors of $A$. The observability Gramian $X$, solution of the Lyapunov equation (12) is given by

$$X_{ij} = \sum_{l,q=1}^{2N+1} \frac{-1}{\mu_l + \mu_q} (T_L)_l (T_R)_{qj} (T_R^T Q^T R_R)_l q.$$  \hspace{1cm} (13)

The proofs of Propositions 1 and 2 are given in [12].

Under the transformation of Proposition 1, $A$ has no marginal modes and the Lyapunov equation (12) suffices to define the observability Gramian. For the regularized Laplacian, Proposition 2 specifically provides a closed form expression for the observability Gramian of the eigenvalues and eigenvectors of $A$. In this approach we use (13) to compute $P = B^T X B$ and discuss in what circumstances the limit $\epsilon \to 0$ can be taken safely to recover the physically relevant quantities.

**Assumption 1 (Uniform damping to inertia ratio):**
All synchronous machines have uniform damping over inertia ratios $d_i/m_i = \gamma > 0 \forall i$. This assumption makes the computation of quadratic performance measures analytically tractable, and is a standard one in the literature [7], [8], [9], [10], [11], [12]. Machine measurements indicate that the ratio $d_i/m_i$ varies by at most an order of magnitude from rotating machine to rotating machine [18].

**Proposition 3 (Diagonalization of $A$):**
Under the assumption of uniform damping to inertia ratios, the left and right transformation matrices $T_L$ and $T_R$ diagonalizing $A$ can be expressed in terms of the eigenvectors of $L_M$ through the linear transformations given below in (19), (20), and (21).

**Proof:** For uniform damping to inertia ratios one has that $M^\top D = \gamma I$. Thus $M^\top D$ and $L_M$ commute and share a common eigenbasis. Since $L_M$ is symmetric, it has a real spectrum with eigenvalues denoted by $\lambda_i^M$, and it is diagonalized by an orthogonal matrix $T_M$

$$T_M^\top L_M T_M = \Lambda_M := \text{diag}\{\lambda_i^M\}. \hspace{1cm} (14)$$

From the similarity transformation

$$\begin{bmatrix} T_M & 0 & 0 \\ 0 & T_M & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} A & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} T_M^\top & 0 & 0 \\ 0 & T_M^\top & 0 \\ 0 & 0 & 0 \end{bmatrix} = \begin{bmatrix} 0 & \gamma I & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} T_M & T_M & T_M \end{bmatrix}^{1/2} \begin{bmatrix} \gamma I & 0 & 0 \\ 0 & -\epsilon I & 0 \\ 0 & 0 & -\epsilon I \end{bmatrix} \begin{bmatrix} T_M & T_M & T_M \end{bmatrix}^{1/2}, \hspace{1cm} (15)$$

one easily obtains the eigenvalues of $A$ which are

$$\{\mu_1^+, \ldots, \mu_N^+, \mu_1^-, \ldots, \mu_N^-\} \hspace{1cm} (16)$$

with

$$\mu_i^\pm = \frac{1}{2} (\gamma \pm \Gamma_i), \hspace{0.5cm} \Gamma_i = \sqrt{\gamma^2 - 4\lambda_i^M}. \hspace{1cm} (17)$$

From the last row of the right-hand side of (15) one straightforwardly concludes that $-\epsilon^1$ is an eigenvalue. The remaining $\mu_i^\pm$’s eigenvalues actually are the eigenvalues of the upper left $2N \times 2N$ block of the right-hand side of (15). This can be easily seen after the appropriate index reordering of this block and using that $I, -\Lambda_M$, and $-\gamma I$ are all diagonal.

For $\Gamma_i \neq 0$, the full transformation which diagonalizes $A$

$$T_L A T_R = \begin{bmatrix} \text{diag}\{\mu_i^+\} & 0 & 0 \\ 0 & \text{diag}\{\mu_i^-\} & 0 \\ 0 & 0 & -\epsilon^1 \end{bmatrix}, \hspace{1cm} (18)$$

and which fulfills the bi-orthogonality condition $T_L T_R = T_R T_L = I$ is given by

$$T_R = \begin{bmatrix} T_M & 0 & 0 \\ 0 & T_M & 0 \\ 0 & 0 & 0 \end{bmatrix}, \hspace{0.5cm} T_L = \begin{bmatrix} T_M^\top & 0 & 0 \\ 0 & T_M^\top & 0 \\ 0 & 0 & 0 \end{bmatrix}, \hspace{1cm} (19)$$

with

$$S_R = \begin{bmatrix} \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{vec}\left(\frac{2[p^2 T_M^\top M^{1/2} p]}{1 - \gamma^2 + 2\lambda_i^M}\right) \\ \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{vec}\left(\frac{2[p^2 T_M^\top M^{1/2} p]}{1 - \gamma^2 + 2\lambda_i^M}\right) \\ 0 \end{bmatrix}, \hspace{1cm} (20)$$

and

$$S_L = \begin{bmatrix} \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{vec}\left(\frac{\tau^2 + 2\lambda_i^M}{\sqrt{\gamma^2}}\right) \\ \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{diag}\left(\frac{1}{\sqrt{\gamma^2}}\right) \text{vec}\left(\frac{\tau^2 + 2\lambda_i^M}{\sqrt{\gamma^2}}\right) \\ 0 \end{bmatrix} \hspace{1cm} (21)$$

Equations (19), (20) and (21) relate the eigenvectors of $A$ to those of $L_M$. Combining this result with the result of Proposition 2 we express the observability Gramian of (13) in terms of the eigenvectors of $L_M$.

**Proposition 4 (Generic performance measure):**
Consider the power system model defined in (7) and satisfying Proposition 1. Under the assumption of uniform damping
to inertia ratios $d_i/m_i = \gamma \forall i$, the quadratic performance measure $P$ defined in (6) is given by
\[
P = \frac{\gamma^2}{m^2} \sum_{l,q=1}^N \left[ T_M^l M^{-1/2} p_l \right] \left[ T_M^q M^{-1/2} p_q \right] \times \left\{ (T_M^l Q_{M}^{(1,1)} T_M^q)_{lq} f + (Teter \tau \approx p^2 \sum_{l\geq 2} u(l)^2 \frac{m + \delta r}{\lambda l d (\tau^2 m + d + \lambda_1 \tau)}.
\]

We note that since the summation index $l \geq 2$, this expression is well behaved also if the regularizing parameter $\epsilon$ is set to zero. In the specific case where the noise injection fluctuation is localized at a single node labeled $\alpha$, i.e. $p = \hat{p}_\alpha$, we have
\[
P = \frac{\gamma^2}{m^2} \sum_{l,q=1}^N u_l^{(1)}^2 \frac{m + \delta r}{\lambda l d (\tau^2 m + d + \lambda_1 \tau)}.
\]

where $u^{(l)}$ is the eigenvector of the network’s Laplacian $L$, associated to the eigenvalue $\lambda_l$.

We next interpret our result (26) from a graph-theoretic perspective. We show that depending on the correlation time scale $\tau$ and on the measure considered, the transient performance is either independent of the location of the noisy node or is determined by the resistance closeness centrality of the noisy node.

The effective resistance distance between any two nodes $i$ and $j$ of the network is defined as $\Omega_{ij} = L_{ii}^{-1} + L_{jj}^{-1} - 2L_{ij}$, where $L$ is the Moore-Penrose pseudoinverse of the network’s Laplacian matrix $L$ [17], [20]. It is known as the resistance distance because if one replaces the network edges with resistors with a resistance $R_{ij} = 1/b_{ij}$, then $\Omega_{ij}$ is equal to the equivalent network resistance when a current is injected at node $i$ and extracted at node $j$ with no injection anywhere else. The pseudoinverse of $L$ is given by $L^\dagger = T \text{diag}(\{0, \lambda_1, \ldots, \lambda_N\}) T^\dagger$. This allows to rewrite the resistance distance in terms of the eigenvalues and eigenvectors of $L$ [21], [22]

\[
\Omega_{ij} = \sum_{l \geq 2} \lambda_l^1 (u_l^{(i)} - u_l^{(j)})^2.
\]

The resistance distance closeness centrality of node $\alpha$, $C_{\alpha}$, is the inverse average distance separating node $\alpha$ from the rest of the network $C_{\alpha}^{-1} = \sum_{j=1}^N \Omega_{\alpha j} / N$ [23]. Using (27), the inverse closeness centrality is given by

\[
C_{\alpha}^{-1} = \sum_{l \geq 2} \frac{u_l^{(i)}^2}{\lambda_l} + \frac{1}{N} \sum_{l \geq 2} \frac{1}{\lambda_l},
\]

where we have used that $\sum_{j=1}^N u_l^{(j)} = u^{(l)\top} u^{(i)} = 1$, and that $\sum_{j=1}^N u_l^{(j)} = 0$ for $l \neq 1$, since $u_l^{(i)} \perp u^{(j)}$. We note that only the first term in the right-hand side of (28) depends on $\alpha$. The second term is proportional to the networks Kirchoff’s index $Kf_1 = N \sum_{l=2}^N \lambda_l$, and is thus independent of the location of the noisy node.

We next consider the performance measures $P_\varphi$ in the limit when the correlation time $\tau$ is much shorter than any characteristic time scale of the swing equation. Expanding (26) in the limit $\tau \ll 1$ we get

\[
P_\varphi \approx \frac{\tau^2 p^2}{d} \sum_{l \geq 2} \frac{u_l^{(i)}^2}{\lambda_l} + O(\tau^2)
\]

\[
= \frac{\tau^2 p^2}{d} \left[ C_{\alpha}^{-1} - Kf_1/N^2 \right] + O(\tau^2).
\]

We see that for colored noise injection at node $\alpha$ with fast decaying correlations, the phase coherence is proportional to the inverse closeness centrality of the noisy node.

In the opposite limit $\tau \gg 1$ a Taylor expansion of Equation (26) gives

\[
P_\varphi \approx p^2 \sum_{l \geq 2} \frac{u_l^{(i)}^2}{\lambda_l} + O(\tau^{-1}).
\]
The phase coherence measure $P_{\phi}$ still depends on the location of the noisy node $\alpha$, but this time with a more involved expression of network related quantities.

VI. CONCLUSION

Our results illustrate how finite-time correlations in power fluctuations affect the transient performance. We have shown how, depending on the correlation time scale $\tau$, performance measures can change qualitatively form being network independent to network dependent. Our analytical results clearly emphasize that the resistance distance, and the associated resistance closeness centrality are the physically relevant measures of node criticality.

Compared to white noise, colored noise inputs provide a better description of the stochastic fluctuations of renewable generation. Future works should try to improve this modeling assumption and incorporate the non Gaussian character of renewable generation [14].

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VII. APPENDIX

A. Colored Noise from Gaussian white noise

In this section we illustrate how the augmented dynamical model (5), provides the framework to treat exponentially decorrelating noise. Consider the differential equation

$$\dot{\eta}(t) = -\tau^{-1}\eta(t) + \eta_0 \xi(t),$$  \hspace{1cm} (31)

where $\eta_0 = \sqrt{2}/\tau$ and $\xi(t)$ is a Gaussian white noise signal, such that $E[\xi(t_1)\xi(t_2)] = \delta(t_1-t_2)$. Solving (31), with initial condition $\eta(0) = 0$ leads to

$$\eta(t) = \eta_0 \int_0^t \xi(s)e^{(s-t)/\tau} \, ds,$$  \hspace{1cm} (32)

from which one obtains

$$E[\eta(t_1)\eta(t_2)] = \eta_0^2 \frac{T}{2} \left[ e^{-|t_1-t_2|/\tau} - e^{-|t_1+t_2|/\tau} \right],$$  \hspace{1cm} (33)

which simplifies to

$$E[\eta(t_1)\eta(t_2)] = e^{-|t_1-t_2|/\tau},$$  \hspace{1cm} (34)

for $t_1,t_2 \gg \tau$.

B. Generic performance measure coefficients

The functions $f$ and $g$ in (22) are given by

$$f(\tau, \gamma, \lambda_M, \lambda_q) = \left(1 + \gamma \tau + \lambda_M^2 \right)^{-1} \left(1 + \gamma \tau + \lambda_q^2 \right)^{-1} \frac{\tau^2}{2}$$

$$\times \left[ 8\gamma^2 \tau^2 + 4\gamma + 2\gamma^2 \tau^2 \left(2\gamma^2 + \lambda_M^2 + \lambda_q^2 \right) \right]$$

$$\times \left[ 2\gamma^2 \left(\lambda_M^2 + \lambda_q^2 \right) + \left(\lambda_M - \lambda_q \right)^2 + \tau^2 \right],$$  \hspace{1cm} (35)

and

$$g(\tau, \gamma, \lambda_M, \lambda_q) = \left(1 + \gamma \tau + \lambda_M^2 \right)^{-1} \left(1 + \gamma \tau + \lambda_q^2 \right)^{-1} \frac{\tau^2}{2}$$

$$\times \left[ 2\gamma^2 \tau \left(\lambda_M^2 + \lambda_q^2 \right) - \tau \left(\lambda_M - \lambda_q \right)^2 \right]$$

$$\times \left[ 2\gamma^2 \left(\lambda_M^2 + \lambda_q^2 \right) + 2\gamma \left(\lambda_M^2 + \lambda_q^2 + 2\gamma^2 \lambda_M \lambda_q \right) \right].$$  \hspace{1cm} (36)

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