Sensor noise in *LISA Pathfinder*: An extensive in-flight review of the angular and longitudinal interferometric measurement system

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In a previous article [1], we have reported on the first subpicometer interferometer flown in space as part of ESA’s LISA Pathfinder mission, and have shown the residual sensor noise to be on the level of $32.0_{-1.4}^{+0.7}$ fm/$\sqrt{Hz}$. This review provides a deeper and more complete overview of the full system and its interferometric mission performance under varying operational conditions, allowing a much more detailed view on the noise model. We also include the optical measurements of rotations through differential wave front sensing (DWS), which reached a sensitivity of as good as 100 prad/$\sqrt{Hz}$. We present more evidence for the long-term stability of the interferometric performance and components. This proves a solid foundation for future interferometry in space such as the LISA mission.

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I. INTRODUCTION

LISA Pathfinder (LPF) [1–18] was a European Space Agency (ESA) mission launched on December 3rd, 2015 with a VEGA launcher from Kourou, French Guiana. It was designed as a technology demonstrator for future space-based laser-interferometric gravitational wave detectors, in particular the Laser Interferometer Space Antenna (LISA) [19]. The core of the experiment was two free-floating test masses (TMs), one of which was used as a drag-free reference for the spacecraft and the other as a quiet reference. The primary objective of the mission was to verify and analyze in detail the performance of the free-falling TMs at a level of spurious accelerations below $3 \times 10^{-14}$ ms$^{-2}/\sqrt{Hz}$ at mHz frequencies.

An essential element of the associated drag-free control of the Spacecraft (SC) was the laser interferometric optical readout of the position and orientation of both TMs at a design noise level of a few pm/$\sqrt{Hz}$ and nrad/$\sqrt{Hz}$. This system, called the Optical Metrology System (OMS), represents the first precision laser interferometer operated in space. While the main results of the LISA Pathfinder mission have been reported in [12,13] and the subpicometer performance of the OMS longitudinal TM readout has been published in [1], this paper provides extensive details of the OMS design and in-flight performance across all degrees-of-freedom and under varying conditions from very low to very high contrast. We show that the sensitivity has been better than the requirements throughout the mission and would be sufficient for the local metrology of a LISA-like mission.

After achieving all objectives and a mission extension, the SC was finally turned off on July 18, 2017 and is now orbiting freely and unobserved in the inner solar system.

Please note that we provide additional descriptions of interferometer readouts and acronyms in Appendixes A and C to support the reader.

II. THE OPTICAL METROLOGY SUBSYSTEM

The purpose of the OMS was to continuously monitor, with low noise, the position and orientation of both TMs with respect to each other and to the spacecraft. These measurements were then used as input for the drag-free and attitude control system (DFACS) of the spacecraft, which included the suspension controller that maintains the second (reference) TM at the optimal working point [12,17,20]. The OMS output was also the primary measurement used to derive the residual differential acceleration of the TMs [14].

The optics of the OMS were mounted on an optical bench (OB) located between the two TMs and rigidly fixed to the spacecraft frame. The OMS directly measured six degrees of freedom, namely the variations in separation of the two TMs, $x_{12}$, the variation in separation of the first TM to the OB, $x_1$, and the corresponding pitch and yaw angles $\phi_{12}, \eta_{12}$ (differential tilts between the TMs), and $\phi_1, \eta_1$ (relative tilt of TM1 to the OB). Linear combinations of these measurements provided the position and orientation of the second TM with respect to the OB, $x_2, \phi_2$, and $\eta_2$ (see Fig. 1).
A. Requirements and environment

While the TMs were designed to be shielded from external forces and thus were very quiet, they could move by more than one laser wavelength such that one fundamental requirement on the OMS was the ability to track $x_1$ and $x_{12}$ with constant sensitivity over more than one wavelength (“fringe”).

In the initial design phase, the TM displacement noise requirement for the OMS was set at $9 \text{ pm/Hz} \times \text{NSF}(f)$ for $x_1$ and $x_{12}$, where $\text{NSF}(f) = \sqrt{1 + (3 \text{ mHz}/f)^4}$ is the noise shape function adapted to the expected cross-over between force noise induced TM motion and interferometric sensing noise. The required measurement band was $1 \text{ mHz} < f < 30 \text{ mHz}$.

The requirement for the measurement noise of $\phi_{12}$, $\eta_{12}$, $\phi_1$, and $\eta_1$ was $20 \text{ nrad/Hz} \times \text{NSF}(f)$ in the same band [21,22].

LPF was operated in a very quiet and stable Lissajous orbit around the Earth-Sun Lagrange point L1, about 1.5 million km from the Earth and at a nearly constant separation and orientation with respect to the Sun. This led to a very stable thermal and mechanical environment, better than what is possible to achieve on Earth.

B. Basic design and subsystems

Due to the requirement of tracking more than one fringe, the basic OMS design was a set of heterodyne Mach-Zehnder-derived interferometers all operating at the same fixed frequency difference of $f_{\text{het}} = 1 \text{ kHz}$ [23–25]. A single Nd:YAG NPRO laser delivered light with a wavelength of 1064 nm through a polarization-maintaining fiber to the modulation unit, where it was split into two parts, each of which was frequency-shifted by acousto-optic modulators (AOMs), operating at frequencies $f_1$ and $f_2$, respectively. While $f_1 \approx f_2 \approx 80 \text{ MHz}$, their difference is fixed to the preselected heterodyne frequency; $f_1 - f_2 = f_{\text{het}} = 1 \text{ kHz}$.

That difference between the modulation frequencies $f_{\text{het}}$ was locked with a phase locked loop (PLL) to a reference frequency generated by the Data Management Unit (DMU), which served as the data processing unit for the OMS. A piezo actuator in each beam in the modulation unit was used to control the phase difference between both beams on the optical bench. The light was delivered to the OB through two optical fibers, one for each frequency, to serve as “Measurement beam” (red traces in Figs. 2 and 3) and “Reference beam” (blue traces in Figs. 2 and 3), respectively. Fiber couplers mounted on the optical bench produced collimated nearly Gaussian beams with a diameter ($2\alpha_0$) of approximately 1.2 mm and in addition performed polarization cleaning. On the OB there were four separate interferometers (see Figs. 2 and 3) that measured:

(i) $x_1$, $\phi_1$, and $\eta_1$, i.e., the position and orientation of TM1 with respect to the OB, called the “X1” interferometer;

(ii) $x_{12}$, $\phi_{12} = \phi_2 - \phi_1$, and $\eta_{12} = \eta_2 - \eta_1$, i.e., the separation between the two TM and the difference of their orientations, called the “X12” interferometer;
(iii) $x_R$, the phase difference of the two beams on the OB without interaction with the TM. It measures any unwanted relative phase noise between the two beams that is accumulated in the optical paths through the AOMs and optical fiber links leading to the OB. Thus, it is used as a phase reference for the X1 and X12 interferometers and additionally as an error signal for the optical path length difference (OPD) stabilization loop explained below in Sec. II C. This interferometer is called “XR” for reference;

(iv) $x_F$, another TM independent phase difference but with an intentional path length difference of $\left(38.2 \pm 0.1\right)\text{ cm}$ [26] between the two arms, serving as a laser-frequency noise sensor and called “XF” for “frequency”.

The four interferometers were balanced in terms of beam powers, the number of reflections and transmissions in beam splitters each beam undergoes before interference, and in terms of optical path lengths, with the intentional exception of the frequency interferometer. In order to make the OB as compact as possible, this was achieved by moving part of the common optical path of the reference beam into the optical fiber of that beam, which was around 38 cm longer than its counterpart for the measurement beam.

Each of the four beam combiners has two complementary output ports, which were directed to a total of eight InGaAs quadrant photodiodes (QPDs) with a diameter of 5 mm. The photocurrent in each photodiode was dominated by a strong sinusoidal component at $f_{\text{het}}$, the phase of which encodes the path length difference between the reference and measurement beams. The second port of each beam splitter had the same signal with a phase shift of $\pi$ (the two outputs port photodiodes for the X1 interferometer, for example, are PD1A and PD1B in the upper panel of Fig. 2). The TM angles were sensed by the differences in the phase measurements of the four quadrants of the QPDs. A small tip or tilt motion of a TM caused not only a displacement of the respective measurement beam on the QPD (“beam walk”), but also an angle between the wave fronts. The former effect gives wide-range, but less sensitive, so-called differential power sensing (DPS) signals used for alignment and diagnostic purposes, while the latter produces phase shifts between the split segments of the QPDs and was used for accurate alignment sensing by the DWS [27,28] method. While it is possible to suppress the beam walk with imaging optics, this was not necessary in LPF due to the small expected TM angles. Not suppressing the beamwalk also enabled the use of the DPS signals for initial alignment of the TMs from conditions that could potentially be so strongly misaligned that the interferometers have no contrast and hence no usable DWS signal.

The X1 and X12 interferometers that sensed the TM motions and rotations used non-normal incidence
reflections off the TM surface at an angle of 4.5° to separate the returning beam. This was chosen because at the time of the design (2002) it was not clear whether normal incidence with beam separation by polarizing optics would be able to deliver the required pm/√Hz stability at mHz frequencies. In the meantime such doubts have been resolved with laboratory measurements [29] such that future designs might use normal incidence.

The QPDs were connected to two hot-redundant phasemeter (PM) units, one of which processed all “A” QPDs and the other one all “B” diodes, such that in case of a failure of one photodiode or one phasemeter all signals were still available. In normal operation all channels were used. To implement the so-called “balanced detection” two complementary QPD segments, one from each output port, were subtracted. This achieves better signal-to-noise ratio (SNR) and suppression of some noise sources [in particular laser relative intensity noise (RIN)]. Each phasemeter processed 16 channels in parallel, each consisting of an op-amp based transimpedance amplifier, followed by an antialiasing filter with 20 kHz bandwidth and a 16-bit successive-approximation analog to digital converter (ADC) sampling at 50 kHz, with a clock driven from the DMU. The low value of \( f_{\text{het}} = 1 \) kHz allowed moving all electronics and their associated heat dissipation away from the OB (in contrast to the interferometers in LISA, where MHz heterodyne frequencies require transimpedance amplifiers immediately adjacent to the photodiodes on the OB). Within the phasemeter, the digitized photocurrents or voltages, \( U_i \), were further processed by a field programmable gate array (FPGA) that performed a single-bin discrete Fourier transform (SBDFT) [24]. For each segment of \( N_{\text{FFT}} = 500 \) samples (10 ms and thus 10 cycles at 1 kHz), it computed the three quantities

\[
y_{\text{DC}} = \frac{1}{N_{\text{FFT}}} \sum_{i=0}^{N_{\text{FFT}}-1} U_i, \\
y_{\text{Re}} = \frac{2}{N_{\text{FFT}}} \sum_{i=0}^{N_{\text{FFT}}-1} U_i \cos(\omega_{\text{het}} t_i), \\
y_{\text{Im}} = \frac{2}{N_{\text{FFT}}} \sum_{i=0}^{N_{\text{FFT}}-1} U_i \sin(\omega_{\text{het}} t_i),
\]

and auxiliary diagnostic quantities (the actual processing was using unsigned integer arithmetic). These were transmitted to the DMU at a rate of 100 Hz through a payload-wide Mil-bus.

Within the DMU, phases were determined by computing \( \text{arctan}(y_{\text{Im}}/y_{\text{Re}}) \), followed by a phase-tracking algorithm to remove jumps of \( 2\pi \); further signal combinations were formed to derive the DWS signals. With error handling and calibration, the processing was in reality more complex, and details can be found in [30]. The measurements were scaled to optical path lengths using the factor \( \lambda/(2\pi) \), or equivalently to TM motion using the factor \( \lambda/(2\cos(4.5^\circ)) \), and to TM angles using calibration factors from electrical to optical phase of order 5000 rad/s/rad\(_{\text{opt}}\), which were determined on ground [see Eq. (19)]. The measurements were then downsampled to 10 Hz and transmitted to the spacecraft computer for use in the DFACS system and as science data telemetered to Earth.

C. Development and prototyping

First proof-of-principle prototypes of the interferometer and phasemeter were built at the AEI in Hannover and at the University of Glasgow starting in 2002. An engineering model (EM) of the optical bench was completed in 2004 by Rutherford Appleton Laboratories and the University of Glasgow [31]. Around the same time, advanced prototypes of the phasemeter and AOM driving electronics were built in Hannover.

Interferometer noise performance meeting the requirements was first demonstrated in 2004 on an early optical test bed with fixed mirrors in place of the TMs [11,32]. Detailed noise investigations in Glasgow identified an important noise source named Small Vector Noise (SVN). It originates from electrical sidebands introduced by the Radio Frequency (RF) signals driving the AOMs which then convert into optical sidebands. These sidebands, in turn, result in a spurious interferometric signal in conjunction with fluctuations of the differential path length between the two fibers that deliver the light to the OB [33]. A mitigation strategy of stabilizing the OPD between the two fibers using the phase measured in the reference interferometer as the error signal was experimentally demonstrated. Investigations in Hannover developed a firm theoretical understanding of this noise source. Two separate mitigation measures were subsequently introduced to the flight model requirements which led to a suppression of that noise to a negligible level. They consisted of placing very strict requirements on the spectral purity of the AOM driving frequencies and of implementing active stabilization of the OPD using a piezo behind the AOMs as actuators. Furthermore, the importance of laser RIN near the heterodyne frequency and its second harmonic was identified [34] and mitigated with an analog amplitude stabilization loop using dedicated power detection photodiodes on the OB (PDA1 and PDA2 in Fig. 2) as error signal and the RF amplitudes that drive the AOMs as actuators.

On the processing side, an analytically predictable phase error in the SBDFT-derived phase was found in the case of non-negligible phase shift occurring during one time segment (10 ms). If the changes are fast enough, they produce an equivalent frequency shift of the heterodyne signal. Since the SBDFT is always evaluated at the fixed heterodyne bin, sufficiently fast signals drift measurably out of...
this bin, resulting in a nonsymmetric Fourier transform. We call this effect “Doppler error”. It depends on the injection frequency and its offset to the heterodyne frequency for a given Fourier bin width. This error was mitigated by analytical correction formulas in the processing back end for the prototype measurements but was not visible under nominal in-flight operating conditions due to the achieved phase stability.

Prototypes of the data processing chain were developed in the laboratory [35] and served as model for the DMU software. It was also shown that angular jitter of the fixed pieces and an aspherical lens of focal length 6.24 mm creates an object history, providing full reproducibility of the operations that have been applied to it and store data inside a so-called analysis object which keeps its design was led by the Institut d’Estudis Espacials de Catalunya [49,50].

Since each TM was contained in a separate vacuum enclosure, the measurement beams were transmitted through an optical window [38,39] mounted in the TM enclosure. The windows were 6 mm thick and made from the special glass S-PHM52 from Ohara, which was chosen for its small value of the total temperature-driven pathlength sensitivity, \(dn/dT + (n-1)\alpha \approx 0.59 \text{ ppm} / \text{K}\). They were coated with a conductive indium tin oxide (ITO) layer on the side facing the TM to avoid electrostatic disturbances due to charge accumulation on the otherwise isolating glass.

D. Flight hardware construction

The flight model (FM) of the OB was built in Glasgow between August of 2009 and March 2010 using hydroxycatalysis bonding, a coordinate measurement machine and calibrated quadrant photodiode pairs to determine beam positions and directions in space with an accuracy of a few \(\mu\text{m}\) [40–42].

The laser beams were required to be well aligned to the center of mass of the TMs to minimize the coupling between SC rotation and apparent TM separation changes [tilt-to-length (TTL) coupling]. The alignment was measured prior to launch relative to the OB mechanical frame, i.e., relative to a virtual point against which also the electrode housing was aligned during assembly. For this, the ultimate alignment accuracy for the reflection point on the TM was found to be TM1\(_{\text{offset}}\) = \((-6, -15) \mu\text{m}\) and TM2\(_{\text{offset}}\) = \((-16, -7) \mu\text{m}\), in the horizontal and vertical direction, which was comfortably better than the corresponding requirement of \(\pm 25 \mu\text{m}\) for each value. The path length mismatches in the interferometers were controlled to better than 0.5 mm, well within the 1.0 mm requirement [43].

The beam splitters were rectangular cuboids of dimension (h/w/d) 20 \(\times\) 15 \(\times\) 7 mm made from fused silica coated with appropriate dielectric coatings on both sides [44]. The recombination beam splitters have dimension of 20 \(\times\) 20 \(\times\) 7 mm. The fiber collimators were quasimonolithic constructions made from rectangular fused silica pieces and an aspherical lens of focal length 6.24 mm [45,46], followed by a cubical polarizing beam splitter mounted under a small angle in order to clean the polarization of the beams before they enter the interferometer. The optical components were mounted on the baseplate of dimension 200 \(\times\) 212 \(\times\) 45 mm made from Expansion Class 0 Zerodur which had an arched lower surface for mass reduction and had a mass of approximately 4.23 kg [47].

The FMs of the phasemeter were built by the University of Birmingham [48]. Initial processing took places in Actel (now Microsemi) FPGAs. The DMU was built in Spain and its design was led by the Institut d’Estudis Espacials de Catalunya [49,50].

The beam modulation unit (“laser modulator” in Fig. 2) was jointly built by Kayser-Threde, Contraves, and APC Paris and uses AOMs from Gooch and Housengo in shear mode for reduced RF power consumption, lever-arm amplified piezo actuators for the OPD control and dedicated electronics from Contraves.

The laser was made by Tesat and delivered approximately 35 mW nominally at a single-mode, polarization maintaining fiber output. It had both a slow temperature input and a fast piezo input for laser frequency control, both of which were driven by signals from a digital control loop running in the DMU with a designed unity gain frequency of 240 mHz [51]. This control loop, which is separate from the beam modulation unit as shown in Fig. 2, used the phase measured in the Frequency interferometer as an error signal. A similar digital control loop also running in the DMU stabilized the OPDs with a designed unity gain frequency of 3.1 Hz [51].

The resulting optical power in each instance of the four measurement and reference beams that reached the photodiodes was approximately 1.2 mW. The power reflected from the TMs was around 2.4 mW for TM1 and around 1.2 mW for TM2, which is important for calculation of the TM forces from radiation pressure and its possible fluctuations.

After complete assembly of all interferometer elements in the spacecraft, numerous tests were performed [22,52], the most relevant one for this discussion being a thermal vacuum test performed at Industrieaengen-Betriebsgesellschaft mbH (IABG) in Ottobrunn for three weeks in 2011, which verified the functional and noise performance with fixed mirrors substituting the TMs in a representative thermal environment.

E. Dedicated data-analysis framework

To analyze the data from the LISA Pathfinder mission, the object-oriented MATLAB\textsuperscript{®} toolbox LTPDA has been developed and extensively tested [53]. Its key feature is to store data inside a so-called analysis object which keeps track of the operations that have been applied to it and creates an object history, providing full reproducibility of the results. It was adapted to the mission in two ways; first, the software allowed us to produce results fast enough to plan the following experiments on LISA Pathfinder while
remaining flexible enough to account for last-minute changes in the experiment and analysis designs, and second, it provided routines adapted to the LISA Pathfinder experiments which were developed before launch and tested on laboratory data. The results and figures presented in the following have all been generated using LTPDA.

**III. IN-ORBIT PERFORMANCE**

The OMS exhibited exceptionally stable performance throughout the mission with 32.0±1.2 fm/√Hz sensitivity above 200 Hz along x (as presented in [1]), substantially better than its requirement.

In this section we discuss the OMS in-orbit behavior in terms of its sensitivity and reliability in more detail. We show the performance under different operating conditions and also present our understanding of the contributing noise terms for both the longitudinal and the angular measurements.

The interferometer was first turned on during commissioning (13th January 2016) with still caged TMs and hence large misalignment. The propulsion module was also still attached to the SC at this time, resulting in a varying thermal environment under vibrational stress. Nevertheless, the OMS immediately demonstrated full functionality and noise levels well below 10 pm/√Hz at a frequency range between (1 to 1000) mHz, despite having a contrast of only 0.6% in the X12 interferometer, limited by misalignment of the beams with the TMs in their “launch locked” positions.

The TMs were released during commissioning on 15./16.02.2016; a few days later they were aligned and dynamically controlled by DFACS using the interferometer signals. A contrast of around 99% in X12 and 98.5% in X1 was immediately achieved (see Fig. 4) and science operations started on March 1st, 2016.

Overall, the operational duty cycle of the OMS, measured in terms of laser and phasemeter functionality, between the start of science operations and July, 1st 2017 at 08:00 UTC reached 99.4%, demonstrating the system’s reliability (see Sec. IV E).

One of the most important channels to be characterized in this publication is the differential TM to TM measurement, which is constructed onboard after taking the difference between the X12 and XR (reference) interferometer. We call this quantity

\[
o_{12} \equiv -\kappa (x_{12} - x_R),
\]

where \( \kappa = \lambda/(4\pi \cos(4.5^\circ)) \) is used to calibrate to TM displacement (note the minus sign due to coordinate conventions).

We also present the X1 interferometer measurement

\[
o_1 = \kappa (x_1 - x_R)
\]

in comparison, which is used for SC control.

**A. Typical spectra and comparison with ground measurements**

In this section we look at some typical noise measurements of the longitudinal and angular measurements and compare to data from the ground test campaign.

**1. Description of the data**

The in-flight measurements for these examples were taken in March 2016 during the first two days of mission operations (2016-03-01 08:05—2016-03-03 00:00 UTC), just after the commissioning phase was completed.

The on-ground measurements were taken more than 4 years earlier during the on-station thermal tests (OSTT) campaign at IABG in Ottobrunn from October 26th, 2011—November 6th, 2011, during the hot (used for angular DWS data here) and cold (used for the longitudinal data here) test phase with temperatures of between approximately (9.5 to 30.5) °C.
The spectral densities of the longitudinal readouts \( o_{12} \) and \( o_1 \) are shown in Fig. 5, expressed as an amplitude spectral density (ASD). We give one measurement from on-ground tests and one from in-flight data, respectively, compared to the overall requirement of the OMS. Similarly, Fig. 6 shows the corresponding DWS comparison.

2. Description of \( o_{12} \)

The in-flight \( o_{12} \) measurement of the relative \( x \) displacement of two TMs represents an “in-loop” sensor, used by DFACS to control the TM2 \( x \) electrostatic force to maintain constant TM separation. However, its unity gain frequency is around 1 mHz and has negligible effect on the data shown, especially the sensor noise dominated high frequencies.

We note that in general the sensing noise at those mHz frequencies cannot be measured but only estimated and bound by upper limits. As visible, the measured \( o_{12} \) displacement exceeds the optical metrology requirement at such frequencies less than 5 mHz, but this can most likely be attributed to true TM motion, since the sensor is designed to detect real distance fluctuations.

In addition, we provide measurements for mechanically fixed TMs by the Gravitational Reference Sensor (GRS) “grabbing positioning release” system during flight. These are shown along \( x \) in Figs. 17 and 18, while the angular DWS is presented in Fig. 19. This data suggest that also the in-flight noise at frequencies as low as 1 mHz is consistent with the level observed at higher frequencies.

![FIG. 5. Comparison of interferometric measurements of \( o_1 \) and \( o_{12} \) on-ground and in-flight. In the ground measurements, the noise in \( o_1 \) and \( o_{12} \) is comparable in the required measurement band. The in-flight measurement of \( o_{12} \) below about 200 mHz is dominated by cross-coupling from spacecraft jitter and real TM motion (see Sec. III C 2 and Fig. 12). At higher frequencies, the noise level is significantly below the ground measurements and OMS sensing noise is dominating (marked in red). The \( o_1 \) measurement in-flight is used to control the SC and as such represents an in-loop measurement, which is dominated by the force noise on the SC. This figure has been produced from a time series using the LPSD algorithm [54] with 66.1% overlapping BH92 windows [55], using as input parameters a desired 1000 Fourier frequencies with 100 averages.](image1)

![FIG. 6. Comparison of interferometric measurements on-ground and in-flight of the angular DWS TM readouts together with the noise requirement. The spectral shape is similar to the longitudinal measurements but has contributions from real TM movements (independently validated by low frequency (<mHz) capacitive sensing), SC jitter, star-tracker noise and OMS sensing noise. The latter is computed from the pseudoflat noise at high frequencies. This figure has been produced using the same LPSD algorithm as in Fig. 5.](image2)
For the data between (20 to 200) mHz we attribute the excess noise to pick-up of SC motion (TTL coupling), see Sec. III C 2 and [1,12,14].

As reported in the previous letter [1], the main $o_{12}$ measurement recorded in-flight shows a quasiflat sensing noise floor at frequencies above 1 Hz which is much lower than on ground, at frequencies above 400 mHz up to more than a factor of one hundred.

The small peaks in the in-flight $o_{12}$ data at 1 Hz, 2 Hz, 3 Hz, 4 Hz, and 5 Hz are probably due to electrical cross-coupling from the pulse-per-second timing signal present on the SC.

3. Description of $o_1$

The $o_1$ measurement shows a different spectral shape than the ground measurement. To explain this difference we consider that the $o_1$ measurement is also an in-loop measurement of another DFACS’s drag-free control loop. This loop, with unity gain frequency of about 0.2 Hz, ensures that the SC follows the x position of TM1. It has a higher unity gain frequency than the control loop which uses $o_{12}$. Consequently, the $o_1$ spectral density is dominated by force noise of the SC, which originates mainly from the thrusters, and the gain of the drag-free control loop which suppresses this noise [17].

The lines around 1.5 mHz and its multiples, and probably the feature around 70 mHz as well, are associated to the thrusters [15]. This idea is also supported by the discussion in [56]. When comparing the amplitude of this feature in $o_1$ and $o_{12}$, we find that the ratio of the two amplitudes is slightly higher than what we could attribute to common-mode rejection [14]. However, the feature around 70 mHz is also observed by a capacitive position sensor of TM1 (not shown here, for details about this sensor, we refer to [57]). This is also confirmed during a period when the OMS served as an out-of-loop sensor, while the control was switched to the independent but less sensitive capacitive sensing. Another important test was performed during disturbance reduction system (DRS) operations, when the main cold gas thrusters were in standby mode, while another set of thrusters was used to control the SC [56]. We find that the interferometric readouts show no feature at 70 mHz for two selected and comparable timespans. Consequently, a platform jitter originating from the cold gas thrusters is a more likely reason, and the OMS does not cause this artifact.

4. Description of DWS measurements

The angular DWS measurements, presented in Fig. 6, have very similar characteristics towards frequencies above 1 Hz. The low frequency behavior has been analyzed in [17] and shows a combination of star tracker noise, TM torque noise, capacitive sensing noises, and SC force noise, since DWS has been used as an in-loop sensor for the DFACS. As can be expected by design, the differential channels show more than an order of magnitude common-mode rejection of those contributions.

We find that the high-frequency sensitivity, limited by the OMS, reached even below 100 prad/$\sqrt{\text{Hz}}$ during the mission, more than a factor of 200 better than required.

The ground measurements show comparable characteristics to the longitudinal data but had ample margin to their requirement.

5. Comment on the difference between ground and in-flight measurements

The results of the on-ground test campaign reported in [52], visible in Fig. 5, showed a measured performance of around $3 \pm 8$ pm/$\sqrt{\text{Hz}}$ around 10 mHz, a significantly higher noise level than observed during science operations.

During the ground tests, Earth’s gravity did not allow for free-falling TMs under DFACS control as in-flight. Therefore, the TMs were replaced with fixed mirrors whose attitude and longitudinal position were hand-aligned. The alignment resulted in a somewhat nonoptimal phase offset between X12 and the reference interferometer, and a typically lower contrast than what was achieved in flight, resulting in some of the noise contributions being higher than in flight conditions.

However, this does not explain all of the extra noise.

The noise level may be that the two separated mirrors, rigidly connected via individual piezoelectric transducers (PZTs), may not have been strictly stable to pm level. These individual mounts may also explain the difference between the $o_1$ and $o_{12}$ noise levels above 30 mHz because of possible acoustic or seismic couplings that could cancel to some extent in the differential measurement. The origin of the feature in $o_1$ just above 100 mHz is discussed in more detail in [58]. However, the on-ground noise spectrum was never fully explained. Since the performance had been below the requirements, further investigations were not given high priority.

As we show in this article, the in-flight noise spectrum is well modeled (see the discussion below).

We present our noise model in Sec. III B and apply it to selected in-flight data under various conditions in Sec. III D. A complete mission overview is given afterwards in Sec. III F.

B. Noise model

Here, we summarize the noise model and sensing noise sources.

We extend our previously published description (see [11]) to include DWS measurements and the so called SVN that is minimized by the OPD. We also expand on indirect contributors like TTL and Brownian force noise and give more details on the two observed frequency noise states of the laser.
Our understanding of the system explains most of the sensing noise as a combination of noise sources associated with the laser (frequency and intensity noise), and noise sources associated with the phase measurement system (quantization noise, electronic noise).

Here, the focus is on the calibrated TM readouts of the X12 and X1 interferometer, which are computed on-board during the mission. Their sensitivity can best be modeled using data from all four interferometers. However, limitations apply as to how well we can model the data given the available measurements, see Sec. III E.

The total phasemeter noise is a combination of the noise introduced by both the analog and the digital stage. The noise in the analog part of the phasemeter is proportional to the equivalent input current noise of the ADCs and the electronic circuits and the noise in the digital part arises mainly from the quantization noise of the ADC converters used. In flight we cannot differentiate between the electronic circuit noise and the ADC noise and use the resulting total phasemeter noise instead.

Under nominal operating conditions, the phasemeter readout noise is the dominating noise term, scaling reciprocally with contrast and beam powers. Much smaller is shot noise due to the mW powers arriving at the photodiodes, for which the phase error scales reciprocally with contrast and the square root of the beam powers. Still, with a decreasing contrast also the impact of shot noise increases, since the signal has lower amplitude.

In terms of the laser, we analyzed two other noise sources that can dominate the performance, depending on the conditions: Laser frequency noise, and laser intensity noise.

The contribution of laser intensity or power noise is described by the so called RIN coupling (see [34]). RIN at the heterodyne frequency couples directly into phase noise but is suppressed by the use of balanced detection and an analog power stabilization control loop (see Sec. II C).

At twice the heterodyne frequency, 2 kHz, the laser-power control loop has a smaller gain. There also exists a coupling mechanism, which is, however, phase matched with the signal and thus cannot be canceled by balanced detection. Instead, this coupling appears correlated on all interferometers and therefore depends on the phase difference between the corresponding combinations. For the longitudinal measurements it is the difference between the X1 or X12 and the XR interferometer and for the angular readout via DWS it is the difference of the involved quadrants.

Accordingly, this noise source is minimized if the phase difference is around zero radian and thus contributes very little if the TMs are at the nominal longitudinal positions (close to zero offset). However, angular offsets were introduced during the mission to mitigate TTL coupling that in turn led to an increase of the RIN contribution. Furthermore, low frequency RIN couples as a direct force noise on the TMs via radiation pressure to the phase readout. This is suppressed by the power control loop.

The optical components are also affected by thermally driven effects, which have been explained in more detail in [1]. We apply the same upper limit on these path length noises in the noise budget plots given here.

In the following we analyze two noise sources in greater detail that have not been (fully) covered in [1].

I. Laser frequency fluctuations

Frequency fluctuations $\delta \nu$ of the laser introduce a phase error $\delta \phi$ due to unequal optical path lengths $\Delta L$ between the interfering beams, as described by [1]

$$\delta \phi = 2\pi \frac{\Delta L}{c} \delta \nu.$$  

$c$ denotes the speed of light.

In a first processing step aboard LPF, the measurement of the reference interferometer, $x_R$, is subtracted from that of the frequency interferometer, $x_F$. We define $\Psi_F$ as this difference,

$$\Psi_F = x_F - x_R.$$  

A dedicated control loop used $\Psi_F$ as an in-loop sensor to reduce this noise [24].

The coupling of the remaining frequency fluctuations to the longitudinal readouts is dictated by the difference in optical path length between the measurement and the reference beam, as described by Eq. (6). In the case of the LPF $o_1$ and $o_{12}$ measurements, we have to take the subtraction of the reference interferometer into account, so that the difference in path length difference between the respective interferometer and the reference interferometer determines the amount of coupling.

This quantity was found to be subject to small changes in the course of the mission, due to the absolute positioning of the two TMs. From dedicated laser frequency modulation experiments, the path length mismatch $\Delta L_{12} - \Delta L_R$ was estimated to be around 368 $\mu$m in June 2016 and around 329 $\mu$m in January 2017 [59]. These values have been used to obtain the respective frequency noise contributions in Sec. III D.

During the mission, we experienced periods of increased laser frequency noise which resulted in an increase of the total noise in $o_{12}$ (see Sec. III F).

Due to telemetry limitations, not the full frequency noise data was available throughout the mission. However, we still have reason to believe that our frequency noise estimation is consistent. We see that its subtraction from the $o_{12}$ measurement removes the periods of increased noise, as shown in Fig. 7. For clarity, we have band passed the data from (0.2 to 1.9) Hz. The coupling coefficient $H_{o_{12}}$ has been estimated from the transfer function during the periods of increased laser frequency shown and is consistent with the results.
Unfortunately, this verification could only be done for the first noise runs since in the subsequent noise measurements the laser frequency noise was either not telemetered at 10 Hz or no period of increased noise was occurring in the respective noise run. However, we have found no reason to assume a different behavior at other times, given that we continue to observe periods of increased noise in the two channels at the same time over the course of the mission, and the 1 Hz data indicates an increased spectral noise power.

This contribution is also visible in the noise budget for a similar period in Sec. III D 2.

2. Small vector noise (SVN)

As explained in Sec. II C, SVN originating from electric crosstalk between the two modulation frequencies of the AOMs was mitigated by design in two ways: via a stringent requirement on the spectral purity of the AOM driving frequencies and via the OPD control loop. In this paragraph, we will look at the remaining level of SVN in-flight. This analysis was not yet included in [1].

For characterizing the suppression of SVN we deliberately applied triangular path length differences through the OPD actuators. This appears in the \( x_R \) measurement, which, under nominal operating conditions, is used as an in-loop measurement of the OPD control loop.

In general, in-flight, the path length difference fluctuations in the fibers were below the requirement, see Fig. 8. The in-flight data shown here is from 2016-06-11/12 (inactive loop) and 2016-06-02 (active loop). The OB temperature sensors report temperatures around \( 21^\circ C \) for \( x_R \) and \( 2016-06-02 \) (active loop). The OB temperature sensors report temperatures around \( 21^\circ C \) for \( x_R \) and \( 2016-06-02 \) (active loop). The OB temperature sensors report temperatures around \( 21^\circ C \) for \( x_R \) and \( 2016-06-02 \) (active loop).

The blue trace in this figure is the ASD of the reference interferometer measurement \( x_R \) when the OPD loop was inactive. The level of fluctuations is most likely due to a relatively stable thermal environment of the SC. This was not the case for the on-ground measurements.

With an activated OPD control loop the noise in \( x_R \) is comparable to the example shown in red in Fig. 8.

By comparing these two measurements, we find that the control loop suppresses the noise at frequencies below 0.4 Hz and increases the noise above this frequency. We believe the OPD actuator is adding noise at high frequencies but the details are still under investigation, see also [60].

We observed that the piezo-controlled loop exhibits spikes at around its actuation frequency of 100 Hz which are aliased down to lower frequencies. This OPD actuator contributes its own noise and often leads to increased noise towards the end of the 10 Hz spectrum.

The related discrepancy to the designed unity gain frequency was already noted before launch and was attributed either to a change in noise level between different measurement times or a different control loop response when operated under fully closed-loop conditions [61].

In addition, a few dedicated OPD loop characterization experiments are under study. Nonetheless, the requirement on the noise in \( x_R \) is fulfilled and the SVN contribution is minimized. To measure the remaining SVN sideband strengths directly, a set of experiments that deliberately amplified this coupling have been performed and analyzed. The impact of SVN was modeled following [33]. For a given OPD injection at frequency \( f_{inj} \), this model considers two small contributing vectors; one at the frequency \( (1f_{inj}) \) and one at twice the frequency \( (2f_{inj}) \).

Like any other vector noise the magnitude of the coupling depends on the differential phase when subtracting correlated measurements and behaves therefore similar to RIN.

For \( o_{12}^{raw} \) the model reads in the time domain

\[
N_{x12,SVN} = \left( k_1 \sin \left( \frac{x_R + x_{12}}{2} \right) + k_2 \cos \left( \frac{x_R + x_{12}}{2} \right) \right) \sin \left( o_{12}^{raw} \right)
+ \left( k_3 \sin \left( x_R + x_{12} \right) + k_4 \cos \left( x_R + x_{12} \right) \right) \sin \left( o_{12}^{raw} \right).
\]
The in-flight experiments aim at estimating the sideband strengths $k_i$. These experiments are detailed in Fig. 9. “Fast” phase ramps were injected through the OPD loop and resulted in sinusoidal phase noise in the raw phase signals. One experiment took place before, and one during the cooldown of the SC in the beginning of 2017. When analyzing this experiment, we had to correct for the “Doppler error”, due to the fast phase changes (see Sec. II C). We were able to calculate and remove this effect, that leads to wrong amplitude estimates, see also the bottom panels of Fig. 9 and [33]. It was calculated specifically for the flight phasemeter and estimated based on the limited telemetry. However, the Doppler error is not relevant in nominal operations, when the phase changes are much slower. From fitting the SVN noise model to these injections in all channels with appropriate SNR, we found a similar set of coupling vector amplitudes for the different operating conditions, while likely only their phase relation changes. The main findings are that the 1f-SVN contributions are below $2 \mu$rad, while the 2f-SVN coefficients reach almost $40 \mu$rad, see Fig. 10. These sidebands are further minimized by the sin() of the differential phase during the mission, because the nominal offsets are almost zero for $o_{12}$. For the rotational degrees-of-freedom the contributions are slightly higher due to the DWS alignments.

For example, with an $o_{12}$ offset of 10 nm we estimate a contribution of around 0.1 fm/$\sqrt{\text{Hz}}$ at 1 Hz. Overall, the measurements and experiments confirm that SVN was well mitigated.

3. Summary of the longitudinal noise model
To summarize, our noise model for the noise power in units of power spectral density (PSD) consists of the uncorrelated sum of:
(i) X12 phasemeter noises (shot noise, ADC noise and electronic noise that depend reciprocally on the X12 contrast and power).
(ii) XR phasemeter noises (shot noise, ADC noise and electronic noise that depend reciprocally on the XR contrast and power).
(iii) XF frequency noise to $o_{12}$ coupling.
(iv) RIN to $o_{12}$ coupling.
(v) SVN to $o_{12}$ coupling.
(vi) Thermally driven noise on the OB as a conservative upper limit neglecting common-mode effects, see [1] for more details.
(vii) (Not from the sensor: Indirect contributions). The same holds true for the X1 interferometer by adjusting the inputs accordingly. Detailed equations were given in [1].

4. Description of the angular noise model
The angular noise model is similar to the longitudinal cases. Differences arise from the different calibration factors, the missing subtraction of the XR interferometer readout, the good suppression of frequency noise and the different channel combinations. We give the exact coupling formulas here, since they have not been published yet.
The parameters are described in detail in [1]. In short we have the contrast $C$, the elementary charge $e$, the normalized average total power per interferometer $\Sigma$, the voltage range of the ADC, $U_{\text{ADC}}$, and the effective transimpedance $R_{\text{TIA}}$.

For DWS readout $\gamma \in \{\phi_1, \phi_1, \eta_1, \eta_1\}$ in a corresponding interferometer $K \in \{X12, X1\}$ we find the shot noise power

$$S_{K,\text{shot,DWS}}^{1/2} = \frac{4}{C_K} \sqrt{\frac{e}{2\Sigma_K U_{\text{ADC}} R_{\text{TIA}}}}$$

(9)

and the phaseometer noise (labeled “PM”), $S_{K,\text{PM,DWS}}^{1/2}$ describes the noise power fluctuations over the mean power (at that frequency), expressed as an ASD called $r$, and contributes a phase error given by

$$S_{\alpha,\text{PM,RIN}}^{1/2} = \sqrt{2} |r | \sin(\gamma_{\text{raw}})|.$$

(11)

Here, $\gamma_{\text{raw}}$ is the uncalibrated phase of the electrical current in units of radian (i.e., representing wavefront differences).

The coupling $N_{\text{SVN}}$ of SVN to the phase readout is modeled by a linear combination of the corresponding sidebands with strengths $k_i$ as described above, adjusted and estimated for DWS given the available data,

$$N_{K,\alpha,\text{SVN}} = (k_1 \sin(x_K) + k_2 \cos(x_K)) \sin(\gamma_{\text{raw}} / 2)$$

$$+ (k_3 \sin(2x_K) + k_4 \cos(2x_K)) \sin(\gamma_{\text{rad}}).$$

(12)

Note that the longitudinal measurements $x_K$ that enter the equation above are used to approximate the unavailable quadrant phase data.

These sidebands have been estimated to be below $|k_i| < 40 \mu$rad by the dedicated experiments discussed before. We call the resulting noise power $S_{K,\alpha,\text{SVN}}$.

The total noise model for each DWS channel is a combination that reflects the data processing with the calibration factors $g_i$ (inverse of rad$_{ad}$/rad$_{opt}$ from Sec. II B),

$$g_1 = (-0.202 \pm 0.004) \times 10^{-3} \text{ rad/rad},$$

(13)

$$g_2 = (0.209 \pm 0.004) \times 10^{-3} \text{ rad/rad},$$

(14)

$$g_3 = (-0.175 \pm 0.006) \times 10^{-3} \text{ rad/rad},$$

(15)

$$g_4 = (-0.189 \pm 0.003) \times 10^{-3} \text{ rad/rad},$$

(16)

$$g_5 = (0.188 \pm 0.007) \times 10^{-3} \text{ rad/rad},$$

(17)

$$g_6 = (-0.193 \pm 0.004) \times 10^{-3} \text{ rad/rad},$$

(18)

with the uncertainties originating from on-ground calibration measurements with approximately $2\%$ errors, recalculated for the in-flight coefficients.

These are applied in the following way:

$$\phi_1 = g_1 \phi_{1,\text{raw}},$$

(19)

$$\eta_1 = g_2 \eta_{1,\text{raw}},$$

(20)

$$\phi_2 = g_3 \phi_{1,\text{raw}} + g_4 \phi_{12,\text{raw}},$$

(21)

$$\eta_2 = g_5 \eta_{1,\text{raw}} + g_6 \eta_{12,\text{raw}}.$$ 

(22)

Therefore, the total model for DWS in the X1 interferometer reads as

$$S_{\phi_1} = g_1^2 (S_{X1,\text{shot,DWS}} + S_{X1,\text{PM,DWS}})$$

$$+ g_2^2 (S_{X1,\text{PM,DWS}} + S_{X1,\phi_{1,\text{SVN}}}).$$

(23)

$$S_{\eta_1} = g_2^2 (S_{X1,\text{shot,DWS}} + S_{X1,\text{PM,DWS}})$$

$$+ g_2^2 (S_{X1,\phi_{1,\text{SVN}}}).$$

(24)

The model for the calibrated X12 DWS is slightly more convoluted due to the linear combinations,

$$S_{\phi_{12}} = (g_3 - g_1)^2 (S_{X1,\text{shot,DWS}} + S_{X1,\text{PM,DWS}})$$

$$+ g_4^2 (S_{X12,\text{shot,DWS}} + S_{X12,\text{PM,DWS}})$$

$$+ (g_3 - g_1)^2 S_{X1,\phi_{1,\text{SVN}}}$$

$$+ g_4^2 S_{X12,\phi_{1,\text{SVN}}}).$$

(25)

$$S_{\eta_{12}} = (g_5 - g_2)^2 (S_{X1,\text{shot,DWS}} + S_{X1,\text{PM,DWS}})$$

$$+ g_6^2 (S_{X12,\text{shot,DWS}} + S_{X12,\text{PM,DWS}})$$

$$+ (g_5 - g_2)^2 S_{X1,\phi_{1,\text{SVN}}}$$

$$+ g_6^2 S_{X12,\phi_{1,\text{SVN}}}).$$

(26)

Note that the RIN and SVN here is a correlated quantity, meaning that the $|\cdot|$ operator on the sine in Eq. (11) is dropped in $\eta_{12}$ and $\phi_{12}$.

We include no thermally driven noise in DWS since we do not have independent investigations and modeling regarding its influence.

C. Indirect sensor noise contributions

1. Brownian noise

Brownian force noise is not a limiting noise of the sensor, but rather the cause of real, low frequency TM motion that
is detected by the optical readout. It is the most relevant source of TM acceleration noise in the mHz to 100 mHz regime and thus becomes the dominant noise at the frequencies, below roughly 60 mHz, where true TM motion is larger than sensing noise.

We include a simple model based on the estimation in [12] with a residual white differential TM acceleration noise of $S_{B,12}^{1/2} = (5.2 \pm 0.1) \text{fm s}^{-2}/\sqrt{\text{Hz}}$ yielding an equivalent path length noise in early April 2016 (after integrating twice) of

$$S_{B,1o12}^{1/2} = \frac{S_B^{1/2}}{(2\pi f)^2}. \quad (27)$$

The level of $S_{B,1o12}^{1/2}$ decreased throughout the mission [13].

We note that this displacement noise is calculated at frequencies where control forces are not relevant.

For DWS, the TM torque noise is not solely dominated by the corresponding Brownian torque noise (which can be propagated from $\Delta g$ by following [62]) and is therefore not included here due to the more complex control forces as analyzed in [17].

2. Angular and lateral motion cross-coupling (TTL noise)

The angular and lateral motion of the three bodies (TMs and SC) involved in the X1 and X12 interferometric readouts can leak into the longitudinal measurements through misalignments between the TMs and the optical system. We often refer to this noise as TTL coupling [63].

Since the motion of the SC can always be interpreted as TM motion with respect to the SC, we can describe this leakage by the effect of TM motion on the beam path only. In the case of angular jitter of either of the TMs, the measurement beam tilts due to the reflection at the respective TM. A rotation of the TM around an arbitrary point of rotation additionally moves the TM surface in or out of the beam path. Lateral TM jitters, i.e., motion along its $y$- or $z$-axis, couples into the interferometric readout via static angular misalignments of the TMs with respect to the optical system. A lateral shift of a tilted TM moves its reflective surface in or out of the measurement beam path, decreasing or increasing its path length.

In both cases, the reference beam remains unchanged under this motion. Therefore, the TM motion yields an altered optical path length difference, relative angle, and relative beam offset in the X1 interferometer for X1 motion and the X12 interferometer for motions of both TMs. All these effects will change the longitudinal readout.

Changes in the relative angle due to TM angular jitter are additionally visible in DWS of the respective interferometers. However, we expect the angular motion of the SC to cancel in the X12 DWS readout due to the nearly free-fall of the TMs: The angular jitter of the SC can be interpreted as a rotation of both test masses around the center of mass of the SC by the negative SC angle. Therefore, the beam tilts due to the reflection at both TMs. Since the tilt of the second TM affects the beam inversely compared to the first TM, the beam tilt is compensated by its reflection at TM2.

During the LPF mission, TTL noise has been reduced via realignments of the TMs. In particular, rotations to new nominal positions have reduced or counteracted TTL noise contributions [64]. The residual noise has been minimized by fitting and subtracting a linear model in postprocessing [14,37].

In most of the results presented here (such as Fig. 5), this coupling is not subtracted and is often visible as a “bump” of noise between (20 to 200) mHz. Figure 11 shows an example of the sensing noise after the subtraction of this effect. It has been modeled similar to the description in [14]. Here, however, we applied a fit in the displacement domain instead of acceleration, which was the usual method during the mission. The model uses a linear combination of translational and rotational displacement estimates of the SC and corresponding coupling coefficients. We also add a cross-coupling term for $o_1$ that describes the amount of remaining common-mode translational jitter in $o_{12}$ due to imperfections in the setup.

The fit model reads

$$S_{\text{TTL}} = a_0 o_1 + a_\phi \tilde{\phi} + a_\eta \tilde{\eta} + a_y \tilde{y} + a_z \tilde{z}, \quad (28)$$

with the SC variables $\phi, \eta, y, z$ describing the motion relative to the differential TM measurement, calculated

![FIG. 11. The measurement data of a nominal run (same as presented in [1] and here in Fig. 12), corrected for the TTL contributions using a frequency domain fit. After the TTL subtraction, it becomes evident that the noise level of the corrected $o_{12}$ extends down to approximately 90 mHz. However, due to the subtraction, sensing noise from multiple degrees of freedom is added to the OMS model estimate as visible at higher frequencies in the red curves, increasing the OMS noise compared to Fig. 12. Note that only 1 Hz data was available for $y, z$ such that a downsampling filter reduces the visibility of the additional noise here. The raising spectral shape towards lower frequencies is consistent with the measured TM displacement due to Brownian force noise.](image)
from the (detrended) TM1 and TM2 readouts, e.g., \( \tilde{\phi} = (\phi_2 + \phi_1)/2 \), and the coupling coefficients \( a_i \).

The fit has been performed in frequency domain, using a iteratively reweighted least squares method in the range (22–100) mHz, avoiding the influence of frequency noise, and overlapping Blackman-Harris 92 windows with 24 averages.

Our fit result is consistent with the value given in [14], Table 1, for \( \delta_{1o,2} \) matching \( a_o \) for this time of the mission.

The capacitive sensors provide the \( y \) and \( z \) measurements; hence, their higher sensing noise increases the noise level in the data that has TTL contributions subtracted. Our ability to estimate this additional sensing noise is limited, however, and based on the fit result at high frequencies, which is filtered due to the onboard downsampling to 1 Hz. With higher sampled data available, the SC contributions could be low-pass filtered to reduce the additional noise towards higher frequencies where the TTL contribution is not limiting.

We do not attribute the TTL noise to the sensor itself since it does not depend on properties like beam characteristics or the measurement chain but can be described independently by SC motion. Subtracting it also adds sensing noise of the measurements of the contributing degrees of freedom. Further, not all the data channels needed for the subtraction were always available with a sampling frequency of 10 Hz. For these reasons, TTL has not been included in the OMS noise model in general.

Further details of the treatment of TTL in LPF can be found in [14,37]. For a detailed analysis of sources and physical origins we refer to [67,68].

**D. Testing and validation of the sensing model**

The performance observed during flight can be considered under various conditions that correspond to high, intermediate, and low contrast. They give rise to different total noise levels in the X12 interferometer, both in terms of the longitudinal readout and the DWS signals. In the following, the same model of the underlying noise sources is used and shows good agreement with the measured noise.

The first condition (shown in Secs. III D 1 and III D 2) is where the alignment of the test masses is optimal. This condition is achieved in the nominal science mode. There, TM1 is in free-fall along the sensitive \( x \)-axis and TM2 is controlled to follow it in this degree of freedom. The angular orientation along \( \phi \) and \( \eta \) of both TMs is controlled using the DWS signals. By choosing an appropriate working point of the DFACS controller the relative phase \( \alpha_{12} \) of the X12 and reference interferometers is kept close to zero. In this condition, the extremely low noise of the X12 interferometer means that motion of the TM dominates any measurement below approximately 20 mHz. At higher frequencies, the measurement of the relative position of the two TMs is limited by the sensing noise in the X12 interferometer, which is dominated by a combination of RIN coupling at \( 2 \times f_{het} \), ADC quantization noise, and frequency noise.

The second condition (shown in Secs. III D 3 and III D 4) corresponds to a medium-level contrast which was achieved by deliberately misaligning the TMs or by reducing one of the beam powers. Some of the noise contributions depend on the contrast. This gives us the possibility of examining the sensitivity of an interferometer as a function of contrast, and therefore of validating the intrinsic noise levels discussed above. Since the RIN (and SVN) coupling to \( \alpha_{12} \) does not depend on the contrast but rather only on the TM position it is possible to disentangle contributions of correlated and uncorrelated noise sources.

The third condition under which we can study the interferometer performance is when the TMs are both mechanically grabbed (shown in Sec. III D 5). In this condition, the relative motion of the two TMs is restricted to that induced by thermomechanical fluctuations of the grabbing fingers that hold the test masses. In addition, the alignment of the two TMs (both in attitude and longitudinally) is far from optimal, with tens of micrometer and hundreds of microradian offsets. This nonoptimal alignment results in two effects; a nonzero phase offset between the X12 and reference interferometers, and a very low contrast (due to misalignment of the measurement beam’s wave front with respect to the fixed wave front of the reference beam).

### 1. Noise budget under nominal conditions

Figure 12 shows an \( \alpha_{12} \) measurement in nominal configuration. The noise contributions are estimated according to the noise model and show good agreement with the measurement.

This is one of the few occasions when the frequency interferometer data were transmitted to ground with 10 Hz sampling.

![FIG. 12. Noise model applied to an \( \alpha_{12} \) measurement from June 1st, 2016 starting at 18:46 UTC and lasting until 20:29 UTC in nominal configuration. Shown is the ASD computed with the same LPSD algorithm as in Fig. 5. Below 20 mHz TM motion caused by Brownian force noise dominates the noise spectrum. Modified from [1] by including SVN. We see that this noise contribution has negligible impact due to the implemented mitigation strategies.](082001-15)
Our model explains the measured noise density between (0.2 to 5) Hz. Between (20 to 200) mHz the measured noise is increased due to TTL coupling. We find that phasemeter noise (and within that, the ADC noise) is the dominating term in this condition.

Frequency noise causes the spectrum to increase slightly between (0.15 to 2) Hz. RIN and SVN coupling is highly suppressed due to the controlled TM position.

Below 20 mHz TM motion caused by Brownian force noise is detected. The estimate is based on the $\Delta g$ measurement of April 2016 [13].

We find that the reference interferometer has almost the same noise levels in shot and phasemeter noise. This is expected due to its very similar design and parameters. Since the reference interferometer is subtracted, its noise adds to the combined readout $o_{12}$.

Figure 13 shows the sensitivity for DWS by example of $\eta_1$ for the same timespan as the nominal $o_{12}$ budget plot given above.

The noise model includes the described noise sources, using the appropriate coupling factors for DWS signals. We find that the noise model has slightly larger deviations from the measurements in DWS compared to the longitudinal $o_{12}$ data. This will also become visible later in Fig. 23.

We note that $\eta_1$ is an in-loop measurement of the DFACS with a unity gain frequency of around 1 mHz.

The pick-up from SC jitter as well as other force and torque noises is visible between (0.001 to 1) Hz, see [17], and therefore in a larger band compared to $o_{12}$.

2. Noise budget for $o_{12}$ during a period of increased laser frequency noise

As explained in Sec. III B 1, we observed periods of increased laser frequency fluctuations during the LPF operations.

We show one example of the OMS noise model under these circumstances in Fig. 14.

During these periods, laser frequency noise becomes the dominating noise source in the $o_{12}$ measurement in the frequency range from approximately (0.1 to 1.5) Hz.

This supports the previously stated hypothesis that the two levels of noise in $o_{12}$ observed over the course of the mission (clearly visible in Sec. III F), are indeed due to an increase in laser frequency noise.

As anticipated in Sec. III B 1, this plot makes use of the path length mismatch $\Delta L_{12} - \Delta L_R \approx 329 \mu m$, resulting from a dedicated experiment [59].

3. Noise behavior at intermediate contrasts

The data in Fig. 15 originates from an experiment in which both TMs were commanded to a range of angular tilts, which resulted in an overall reduced contrast due to imperfectly aligned wave fronts.

We selected stable timespans during which the measured noise floor has a flatness comparable to actual white noise. The TM orientation also affects the RIN coupling effect to $o_{12}$, because a geometric piston couples TM tilt into the longitudinal path measurement of the X12 interferometer, causing the RIN coupling to vary depending on the piston crosstalk strength. The piston arises when the TM angles are controlled to certain offsets, because the rotation has its origin in the center of mass of the TM (and not at the point of reflection). This is visible in the fluctuations of the light blue RIN contribution in Fig. 15.

The resulting noise floor in $o_{12}$ is estimated from a frequency average in the band (1.2 to 2.8) Hz (from PSD) as described in Fig. 20 and plotted over the X12 contrast.

We can see the influence of the contrast on shot and PM noise and the variation of the two frequency noise states, whereas RIN is only coupling via the longitudinal phase...
FIG. 15. Noise model at intermediate contrasts. For this figure we fitted the noise model as stated in [1] to an $o_{12}$ measurement between (1.2 to 2.8) Hz with two free parameters (levels of RIN and ADC noise). The measurement is from a dedicated experiment (performed 2017-02-06 23:06:00 UTC—2017-02-08 22:55:00 UTC) in which the TMs were deliberately tilted. The TM orientation influences the RIN coupling strength and the interferometer contrast. Here, the fit yields RIN $r = (2.59 \pm 0.09) \times 10^{-6}/\sqrt{\text{Hz}}$ and ADC $S_{12}^{1/2} = (108.0 \pm 1.6) \text{ pA}/\sqrt{\text{Hz}}$. The SVN contribution is not shown, because it reaches only subfemtometer level. Overall, we find good agreement between the model and measurement.

offset. In this condition, RIN becomes important at high contrasts because of the phase offset and geometric piston.

The RIN value is estimated from a fit to the data and varies between this and other experiments. We also get different results depending of the type of experiment, which is still under investigation. The variations are small enough such that our estimates given here are sufficient to explain the good performance within requirements.

We also fit the equivalent noise current introduced by the ADC, which is likely to vary slightly compared to the ground estimate from years ago. The fit result confirms this by deviating only by about 5%.

The noise model (given in [1] and used in the fit) explains the observed behavior for contrasts between (30 to 98)% in the X12 interferometer. The contrast in the reference interferometer remains unchanged.

4. Noise behavior for varying beam powers

We also investigated the effect of unequal powers between the measurement and reference beam. Instead of changing the wave front overlap due to TM tilts, this also lowers the strength of the heterodyne signal and thus the contrast.

The experiment consisted of stepwise reduction on the reference beam power via the power stabilization loop setpoints. After reaching the lowest power, TM2 was shifted via electrostatic actuation to an 132 nm offset, increasing correlated noise sources.

FIG. 16. Noise behavior for varying beam powers (2017-06-17 08:20:00 UTC—2017-06-17 15:00:00 UTC). Top panel: The reference beam power was decreased stepwise, thus lowering the contrast. Bottom panel: Directly after lowering the reference beam power, the second TM was moved to a new (constant) position with maximal 2f-RIN coupling, and the power was increased again with the same steps in reverse. We applied a fit to the total data (all set points combined) with both RIN and ADC noise contributions as free parameters. The coefficients yield RIN $r = (1.08 \pm 0.01) \times 10^{-6}/\sqrt{\text{Hz}}$ and ADC $S_{12}^{1/2} = (110.4 \pm 2.6) \text{ pA}/\sqrt{\text{Hz}}$ and are within our previously assumed limits [1]. In the first panel, ADC noise is dominant, while at the 132 nm offset in the second panel RIN is the major contributor. Note: One segment at the lowest power level showed particularly high and unexplained noise and was excluded from the bottom panel.

Compared to Sec. III D 3, this experiment also changes the noise couplings in the fixed interferometers regarding power and contrast dependent noise sources.

Figure 16 shows the results for the noise measurements and the model at the two TM positions. As expected, the noise increases towards lower powers due to the 1/C dependence. Similarly, at the offset, RIN maximizes its effect and thus decreases the sensitivity to above 150 fm/$\sqrt{\text{Hz}}$.

As in Fig. 15 we fit both the ADC noise and the RIN level using the same model from [1]. While the ADC yields compatible results, the RIN estimate is almost 60% lower, yet still within sensible ranges compared to ground measurements.

One outlier at the lowest power has been removed, which showed a highly increased noise compared to the rest of the data. We suspect nonlinearities in the power stabilization and thermal effects for causing that.

In conclusion, our model explains the observed noise behavior for two of the main parameters of the optical system—the contrast and the power ratio of the two beams for a wide parameter range.

5. Noise budget with grabbed TMs and very low contrast

In April 2017 both TMs were mechanically grabbed. The two longitudinal measurements $o_3$ and $o_{12}$ and their
corresponding noise models are shown in Figs. 17 and 18. DWS is presented in Fig. 19.

The grabbed case corresponds to a very low contrast of 0.6% in X12 and 2% in X1, while the contrast in the reference interferometer remains unchanged. During this measurement period, an upper limit for the RIN level of $r \approx 3 \times 10^{-6}/\sqrt{\text{Hz}}$ was estimated.

The contribution of laser frequency noise is also shown as an upper limit. This limit stems from an experiment to characterize the laser frequency control loop which was executed rather shortly before this noise measurement. While the details of this experiment are under investigation, it can still be used to provide an upper limit for both the $a_1$ and $a_{12}$ measurements.

In this condition the model reproduces the measured PSD down to mHz frequencies, even with very low contrast, for the longitudinal measurements. As expected while being grabbed, no TTL coupling or Brownian force noise is visible. As previously, the PM noise is dominating.

The thermally driven noise, as visible at low frequencies in Figs. 17 and 18, suggests that the upper limit stated in [1] and applied here is too conservative below 3 mHz. This implies that the temperature effects of the window transmission, the expansion of the Zerodur baseplate, or the thermal impact of the beam splitters are less significant. Another probable explanation is the presence of common-mode rejection for the balanced beam paths. We previously excluded common-mode effects to obtain an upper limit for these thermally driven effects.

Also note that the data during the grabbed TMs case was recorded during a low temperature period (SC cooling) of about 11.5°C on the OB.

These cases support our understanding of the sensing noise model, being nominally only visible at high frequencies, extending down to the mHz range.

The peak in $a_1$ around 70 mHz was already observed and discussed in Fig. 5 and Sec. III A.

In principle, due to grabbing, one would expect the platform jitter (and thus the peak) to be common mode and thus suppressed. Comparing with Fig. 6 we find that the peak is indeed reduced. However, we observe a drift of a few nm between TM1 and the SC, showing that the mechanical grabbing is not perfectly stable. Therefore, some residual coupling seems plausible. The same may hold true for $a_{12}$, where the differential mode further minimizes the coupling so that it becomes invisible.

These measurements indicate that the sensing noise floor can be well explained in all observable circumstances for both low and high contrasts and the full measurement band.

FIG. 17. Noise model applied to an $a_{12}$ measurement with grabbed TMs starting from 2017-04-08 19:55:00 UTC and lasting for 2 hours. Note that the SVN is not visible, since it is around $1 \times 10^{-16} m/\sqrt{\text{Hz}}$. For the RIN and laser frequency noise contribution, only an upper limit could be estimated.

FIG. 18. Noise model applied to an $a_1$ measurement with grabbed TMs. This data is recorded at the same time as the data shown in Fig. 17. Due to X1 interferometer data measuring a different TM offset and contrast, the shot noise, RIN and frequency noise contributions differ to those of Fig. 17. However, the RIN and laser frequency noise contribution is only provided as an upper limit. Note that the SVN is not visible, since it is well below $1 \times 10^{-15} m/\sqrt{\text{Hz}}$.

FIG. 19. Noise model applied to DWS with grabbed TMs. This data is recorded at the same time as the data shown in Fig. 17. We observe that the noise level is pseudoflat in all cases down to low frequencies, but has higher magnitude for the differential angles due to the much lower contrast. While the noise model works reasonably well for the X1 interferometer angles, this is not the case for the X12 interferometer angles. Especially the $\phi_{12}$ channel is deviating by about a factor of 2 from the model. We do not know the reason for this behavior. However, we note that the model still seems to work relatively well with such low contrast that should result in questionable signal quality.
We observed variations from roughly 32 fm/√Hz above 1 Hz for high contrast to 3 pm/√Hz at 1 mHz for very low contrast (details see above).

For DWS, see Fig. 19, the model underestimates the noise in \( \phi_{12} \) significantly, while still somewhat matching most of the other channels. The reason for this is unknown.

We conclude that the noise model reflects the observed behavior at frequencies where the sensing noise is expected to be limiting for most of the nominal flight operations.

### 6. Noise behavior below 1 mHz

In the nominal conditions, white Brownian force noise is dominating at mHz frequencies, with other sources of TM acceleration noise becoming dominant at even lower frequencies [13]. Even with grabbed TMs, at frequencies < 1 mHz, thermomechanical induced motions of the structure holding the TMs is likely dominating the measurement, see also [16].

We have no indication of additional interferometer noise sources, nor of any deterioration of the performance at frequencies below 1 mHz, but we cannot completely exclude them. Based on Fig. 17 we report that even with grabbed TM we reach a sensitivity of approximately 3 pm/√Hz at frequencies just below 1 mHz, which would already enable the local TM interferometry on LISA.

### E. Limitations of the sensing noise model

Even though the model explains the observed behavior to a good degree, limitations arise mainly due to the fact that LPF, having only a limited telemetry budget, cannot provide all measurements at all times and not at the rate and resolution needed to further improve the model, as well as the fact that we measure a strong real signal at low frequencies (by design).

The physical noises are all well-known contributions that depend on parameters such as contrasts and beam powers, some of which are only known after additional processing took place. For example, individual QPD data was not transmitted to ground, but just averaged channels over interferometer output ports and their quadrants. A more exact model using individual quadrant contributions is therefore unavailable.

In the high frequency range, we are likely affected by aliasing. The processing used aboard LPF employs moving average filters before decimation, which have a limited antialias capability. Thus, we do not always measure the flat spectrum that we understand well and the amount of aliasing cannot be assessed. In addition, we can only extrapolate the noise spectrum for frequencies below about 200 mHz and have to rely on measurements taken with grabbed TMs to indicate the correctness of the model during these relatively high noise states.

Furthermore, measurements of the frequency and reference interferometer at 10 Hz are not available for most of the noise runs, thus rendering estimation of frequency noise and SVN difficult, especially since we cannot differentiate the latter from RIN for most of the mission. At the same time, the actuator noise of the OPD is not measured independently at high frequencies, which is known to affect the spectra.

We also neglect losses arising from realistic beam combiners that are not exactly 50/50 such that the balanced detection will not be ideal.

The estimation of RIN at kHz frequencies is difficult and its level has been estimated to vary throughout the mission. This is likely due to changes in the environmental conditions and can only be observed indirectly. We have only a few more specific indirect RIN measurements that are subject to ongoing analysis.

In addition, we have a large (and difficult to estimate) uncertainty on the ADC noise contribution, which is the dominating noise term in many conditions.

### F. Long-term noise behavior over the mission

This part addresses the measurements over the course of the mission for periods where the SC was operated in nominal science mode and only noise was measured.

An overview for the longitudinal and angular readouts is given in Fig. 20 together with a noise model fit and as a histogram in Fig. 22.

#### 1. Data selection and analysis

We selected times during the mission where the SC was kept in nominal operating mode without (known) interfering actions. This excludes commissioning activities, experiments and spacecraft maintenance, DRS (NASA) operations and the spacecraft cooling towards the end of the mission. In a second filtering step, we split the data into 12 min long stretches and discarded segments that include glitches or transients and do not allow a stationary noise power spectrum estimation. The algorithm to filter these transients has a threshold on the amplitude change, i.e., the first derivative in time domain. We used the median absolute deviation (MAD) estimator to specify an equivalent 5σ threshold. From the initial 12916 segments, 140 have been discarded in \( \sigma_{12} \) (1.1%), 283 in \( \eta_1 \) (2.2%), 112 in \( \eta_{12} \), (0.9%), 381 in \( \phi_1 \) (2.9%), and 331 in \( \phi_{12} \) (2.6%).

The selected and filtered times are listed in Table III. For more information on nonstationarities and glitches during the mission see [18,69,70].

We report that even in the presence of short-time nonstationarities the sensitivity was below 2 pm/√Hz at high frequencies in the worst case (not shown here, including even more noise only measurements in different configurations).

Every data point in the long-term plots corresponds to those 12 min of data sampled with 10 Hz and averaged over the pseudoflat frequency band between (1.2 to 2.8) Hz.
This band has been chosen because we expect the measurement to be sensor noise limited here.

For each of the 12 minute segments the ASD is calculated using 50% overlapping BH92 windows with 36 averages. Every fourth bin is kept to avoid correlations between neighboring bins [71]. At 2 Hz the center bin and two surrounding bins are masked to remove the influence of on-board 1 Hz harmonics as described above.

However, this noise floor may be overestimated due to an imperfect antialiasing caused by the moving average filters used in-flight. Therefore we treat this estimation as an upper bound of the real sensor noise, as there was no access to continuous data with a higher sampling rate.

2. Long-term noise fit

Shown in Fig. 20 are the noise predictions by our models. They use the data from Fig. 21 and fit the RIN value and the frequency noise contribution, the latter based on an estimate of the 1 Hz data. Since we cannot differentiate RIN from SVN here, the estimate can be seen as the effective contribution from both of these.

The fit allows the RIN parameter to vary per noise run, and each noise run is divided into another set of segments, if there is a correlation between the 1 Hz frequency noise estimates and the noise level in $o_{12}$, to account for the two frequency noise states. Afterwards, all five data channels are jointly fitted per time segment.

While the RIN and SVN contribution are mainly determined from the DWS measurements, the frequency noise contribution estimates relies on the $o_{12}$ data.

Figure 21 shows the main noise model contributions for which we do have independent measurements, showing in what ranges the parameters were stable.

3. Discussion of long-term results

For $o_{12}$, there are more than two noise regimes apparent. However, one broad level is visible around the median and one upper level around 40 fm/$\sqrt{\text{Hz}}$. The upper noise level represents 9% of the shown data. The lower-noise state has a mean of 31.9 fm/$\sqrt{\text{Hz}}$.

The upper noise levels correlate in time with increased laser frequency noise due to the nonstationary (two-state) behavior of the laser. For the first three measurement timespans, the laser frequency telemetry was available at a sampling frequency of 10 Hz and we found that with the subtraction of the laser frequency noise, the periods of
FIG. 21. Main OMS sensor contributions to the noise model over the mission noise runs. In the top five panels we show the time series data for the relevant model contributors. For this plot the measured data channels are sampled with 1 Hz, except for the XF and XR contrast, which is estimated from ADC housekeeping data (reason for higher contrast in $c_R$ than stated in [1]). The bottom two panels show the joint fit results for the two uncorrelated fit parameters from Fig. 20. Note that in segment 2016-11-16 11:05:23 UTC—2016-11-26 07:59:25 UTC the two parameters $\Sigma_R, \Sigma_F$ were not available and during 2016-12-26 08:14:59 UTC—2017-01-13 19:57:57 UTC, $\Sigma_R, \Sigma_{12}$ had some missing samples. In both cases we interpolated those for the noise model fit.
increased noise can be removed. At later times, the laser
frequency telemetry was only available at 1 Hz which does
not allow for this subtraction.

Figure 22 shows the histogram of the noise spectrum
estimation. The $y$-axis shows the percentage of time spent
at a specific noise level with respect to the total measure-
ment length per observable. The bin width is $0.5 \text{ fm} = \sqrt{\text{Hz}}$
for $o_{12}$ and $1 \text{ prad}/\sqrt{\text{Hz}}$ for DWS.

Based on the segment analysis defined above we find
that the measurement in $o_{12}$ varies between $27.8 \text{ fm}/\sqrt{\text{Hz}}$
and $44.1 \text{ fm}/\sqrt{\text{Hz}}$ with a median of $32.0^{+2.4}_{-1.7} \text{ fm}/\sqrt{\text{Hz}}$. The
values are based on the 16th, 50th, and 84th percentiles of
the histogram.

The DWS signals show a much broader distribution of
varying modes with maxima between $(86 \text{ to } 290) \text{ prad}/\sqrt{\text{Hz}}$. We attribute this behavior (as far as we can explain)
to offset dependent noise couplings (such as RIN, SVN).
During specific times of the mission, the TM orientation
was changed, causing varying noise levels. See Secs. III B
and III D for details.

In DWS, the frequency noise is highly correlated and
well phase-matched on all photodiode quadrants and
therefore subtracts to a large degree when forming the
DWS signals (by differencing the left/right, up/down
quadrant phases). As expected, we see no effect from
changing frequency noise levels.

The noise levels of Fig. 20 show that the OMS
measurements have been much more sensitive than
expected, enabling a level of investigation unpredicted at
the beginning of the mission.
weak correlation with changing RF amplitudes used to drive the AOMs may be present.

Even though we could observe a few periods during which the pump current and the laser temperature appear correlated to the noise in some of the observed channels, the exact mechanism from a laser setting to a change of the noise above 1 Hz is not clear, appears nonlinear and no definite, repeating pattern could be found.

Furthermore, in between two noise runs, we found a period where the laser heater set point was correlated to the noise level. However, at other times in between noise runs, we cannot confirm this observation and thus we conclude that this is not the main driver of the observed changes in noise level.

In addition, it should be noted that the DWS spectra were not always perfectly flat. This could possibly point to aliasing which is not modeled well enough or a higher influence of the OPD actuator with the observed spikes (Sec. III B 2). But again, we are limited by the data rate of the telemetry. There are just a few dedicated experiments where 100 Hz data was available, which are very short and always only cover a small portion of the channels.

Since the reason for the observed changes in the DWS noise levels over the mission duration could not be found yet, we can also not estimate whether this behavior is related to any specifics of LPF and the OMS or if it is a more general aspect limiting the common-mode noise rejection in DWS compared to the much less affected longitudinal measurement.

Nonetheless, the agreement between the model and the data is good in most cases and much below any required level.

**IV. LONG-TERM STABILITY OF THE OMS**

The long-term stability of the OMS is of relevance with regard to the three to six times longer LISA mission. Besides typical lifetime limiting effects like dust and operational abrasion, the cosmic radiation is additionally penetrating the components.

The long-term performance of the photodiodes (PDs) is of special interest, since they are relevant for the shot noise contribution in LISA and, as electro-optical components, especially sensitive to cosmic particle radiation. Furthermore, the mechanical stability of the OB is important, since the interferometry alignment relies on it.

In the following, stability analyses of different subsystems of the OMS are described.

**A. OB stability**

The mechanical stability of the OB was actively monitored during flight, using dedicated, so-called spot position measurements.

For this, the beams were turned off one after another to measure the beam spot positions on all OB QPDs and identify possible distortions of the OB baseplate or changes in the alignment.

The procedure was executed multiple times covering more than 1.5 years of the mission (see Fig. 24).

During all of these measurements, the spot positions of both beams changed less than 5 μm in horizontal “x” direction and for the reference beam less than 7 μm in vertical “y” direction.

Only the measurement beam’s spot positions showed a larger deviation in vertical direction. This is especially interesting in combination with a frequent monitoring during the so-called cool-down phase. Here, temperature changes of more than 20 °C occurred without an impact on the beam spot positions above 43 μm. In consequence, the OB was found to be highly stable against temperature changes.

A detailed examination of the spot position changes due to the cool down of the SC shown in Fig. 24 reveals that the beams’ spot positions are more stable in the horizontal direction, i.e., parallel to the optical bench surface.

When the SC was cooled down, the spot positions decreased in along the vertical axis, i.e., the beams’ incident point came closer to the OB surface. This effect was stronger for the measurement than for the reference beam. This beam walk most likely results from a down-tilt of the beams at their fiber injector optical subassemblies (FIOSs). From ground tests of these, we know that the beams tilt less than 3 μrad per °C temperature difference. Since the measurement beams propagate a longer path between its FIOS and diode PDRA, the resulting spot position change is expected to be larger.
Additionally, the particular spot position readouts suggest a bending of the entire OB yielding additional beam tilts and beam walk on the photodiodes. In pre-launch tests it was demonstrated that the center of the OB can lift or sink up to 1.7 μm with respect to its outer edges due to mechanical stresses of its mounts.

Such a behavior is likely a consequence of the thermally induced stress regarding the mechanical structures holding the OB, as also discussed in [69].

In summary, both, the beam tilt at the FIOS and the OB deformation are small even for the strong temperature changes intentionally applied during the cooling period of LPF.

### B. PD responsivity

The responsivity of the LPF interferometer diodes was monitored with a dedicated experiment using the radiation pressure to the TMs as a power reference signal. The experiment was executed three times during the mission resulting in a monitoring duration of 1.5 years. During this monitoring period no degradation could be observed within errors of 1%.

A comparison to preflight measurements, in contrast, showed a small degradation of the responsivity of 4% for the whole flight phase. A systematic error of up to 3% had to be taken into account in-flight. It is assumed to originate from two minor solar energetic particle (SEPs) events between launch and the PD monitoring period that emitted solar protons at relevant energies.

However, there was no direct observation of an SEP event during the monitoring duration. In consequence, the “background” cosmic radiation in the declining solar cycle was found to be of low relevance for the PD responsivity.

Nonetheless, a significant impact of SEP events with low particle energies can be supposed, although no direct observation during the monitoring duration was possible. By that, the results further agree with preflight radiation tests with flight spare PDs.

We also refer to [72–74].

### C. TM reflectivity and optical window transmittance

In addition to the radiation pressure measurement, the TM reflectivity in combination with the optical window transmittance was monitored. This joint parameter could be extracted from the beam spot position measurements. It was found that the TM reflectivity and optical window transmittance changes less than 0.2% during the whole monitoring duration of 392 days [72].

Thus, no degradation of the TM surface or the optical window (OW) transmittance could be observed from cosmic radiation or dust, within the measurement accuracy.

### D. Beam polarization

The spot position measurements showed unexpected significant fluctuations in the single beam powers, despite the active beam power control loop. This is for example visible in \( \Sigma_F \) in Fig. 21. During certain periods of the mission, this noise was correlated to temperature measurements.

The observation was best explained by changes of the polarization of the light (see the in-depth modeling in [72]). To be more precise, in the measurement beam, the ratio of the undesired P-polarization, compared to the S-polarization aimed for, amounts to 1.7% and in the reference beam to 4.5%. It is assumed that the off-nominal polarized light originates in the fibers that deliver the light to the OB. However, the fiber couplers mounted on the OB were designed to clean the polarization (as explained in Sec. II B).

The detailed investigation found a decrease in the polarization extinction ratio of the polarizing beam splitter inside the fibre couplers in vacuum due to the out-gassing of water to be the best explanation for the polarization fluctuations. However, this explanation is not sufficient for all of the P-polarized light on the OB.

The P-polarized light may impact the OMS measurements via two mechanisms. The first of these is the decrease of the balancing detection efficiency which can then impact the suppression of RIN at the heterodyne frequency.

However, with the observed RIN noise levels during the mission, this effect is negligible.

The second mechanism concerns additional phase noise resulting from the interference of the two P-polarized beams in each interferometer. This can yield a readout noise depending on the relative phase of the S- and P-polarized light but the LPF measurements restrict this effect to be below mHz frequencies.

### E. Laser and phasemeter performance

The laser and the phasemeter worked reliably throughout the mission.

We chose a measured contrast in the reference interferometer above 85% as a parameter to indicate that these two units are operating well. This choice was made because the reference interferometer measurement is unaffected during a large number of manoeuvres and this level of contrast can only be reached if both beams are operational and well centered on the PDs. In addition, the ADC utilization parameters are a result of the phasemeter and thus the availability of reasonable data indicates it was operating.

We defined the mission duration to begin with the beginning of nominal operations on March 1st, 2016 at 08:00 UTC and to end on July 1st, 2017 at 08:00 UTC.

We find that the laser and phasemeter unit reached a duty cycle of 99.4% with regard to the total mission duration. If we consider only the times of the contrast in the reference
interferometer above 85% with respect to the total mission duration, we find a duty cycle of 99.0%. The latter also includes experiments that deliberately lowered the contrast.

In general the laser power stability (RIN) was within requirements and not the limiting noise source, as far as we were able to estimate from indirect measurements.

F. Contamination

The optical transmission from the laser to the PDs was estimated by looking at the pump current, the RF amplifier powers and the measured optical power on the QPDs during the radiation pressure measurements [72].

However, since the polarization state coming out of the optical fibers is known to fluctuate, the main source of power losses is expected to originate from reflections at the polarizing beam splitters. Since the reflected power is not monitored, the accuracy of an estimate for the optical transmission is limited.

Nevertheless, no common increase of the pump current and RF amplifier power could be observed within the three radiation pressure measurements, although the OB power stayed comparably constant. A contamination of the optical component surfaces on the OB would further lead to a common decrease of the power on the interferometer diodes, since the power control pick off is located directly after the fiber out-couplers.

The specific OB model that was used to fit the level of parasitic polarization, however, showed no common decrease of the power at the QPDs.

In consequence, no significant long-term contamination of the optical components during flight could be observed.

V. OPERATIONAL CHALLENGES

The impressive performance reported here maintained throughout the mission. Nevertheless, some issues were observed during flight and had to be dealt with operationally, leading to “lessons-learned” for future similar sensor implementations.

An example of such an issue arose due to the fact that the laser system was more sensitive to the environmental temperature than expected, and so consideration of the platform environment for such units should be more carefully accounted for in future designs.

In this case, implementation of a dedicated temperature control system, for example, would have reduced the reliance on complex operational procedures. In addition, as predicted from on-ground tests, the slow power control loop as implemented was too strongly coupled to the temperature of the laser, and as such was not employed during operations.

The analysis of the in-flight data has allowed us to characterize the OMS to a significant level, however, the lack of simultaneous access to all of the raw phasemeter data (which was not possible given the telemetry restrictions of the mission) and the missing on-board clock synchronization made certain (noncritical) diagnostic and scientific investigations difficult and sometimes impossible.

An out-of-loop monitor of the single optical beams was not part of the design, which limits what we can say about the force noise produced by power fluctuations in the beam. In the flight system we have to infer this from isolated offline measurements of the single beams made by turning one beam off.

The ability to align the TMs prior to release to allow full validation of the OMS readout in the measurement band would have been desirable. In the event, we can only surmise that the system performance at low frequencies was consistent with that which we observed at high frequencies.

VI. CONCLUSIONS

The LPF OMS was designed to meet a differential displacement noise level of $9 \text{ pm}/\sqrt{\text{Hz}} \times \sqrt{1 + (3 \text{ mHz}/f)^3}$ over a frequency range of (1 to 30) mHz, and in ground testing it met this goal.

The in-flight performance was significantly better than this goal, with a residual median noise level of $3.2 \pm 2.4 \text{ fm}/\sqrt{\text{Hz}}$ at frequencies where TM motion was not limiting. Above 0.4 Hz the noise was typically smaller by more than a factor 100 compared to on-ground tests. A similar improvement is true for DWS angular signals.

This excellent performance during science operations of LISA Pathfinder has allowed us to investigate in great detail the noise behavior of the system. We could develop and verify a comprehensive model which is valid over a wide variety of sensor operating conditions. This analysis significantly increases the confidence in the model. When applying the sensing noise model to LPF noise measurements, we explain the averaged noise in the frequency range from (1.2 to 2.8) Hz in the $o_{12}$ measurement to better than 20% for most of the mission. A number of additional investigations (see Table I) has led to deeper understanding of the system, but also to some questions which could not be fully solved, partly due to the operational constraints (see Sec. V). One of these open questions entails the detailed understanding of the OPD, including the discrepancy between measured and expected loop behavior.

Another one is the discrepancy between the RIN fit results from TM tilt experiments and those from longitudinal step experiments, which stays well below the requirements. We only found a single angular measurement in the grabbed TMs case for very low contrast, where we could not identify why the model for the $\phi_{12}$ data was explaining only a fraction of the noise. Regarding the sensitivity over the mission, a few changes in DWS could not be fully explained by our model. As we could not identify the mechanism behind these changes, it is not possible to estimate the relevance of this observation for similar optical
setups. However, all of these measurements are orders of magnitude below the required accuracy level.

The deep understanding generated through the detailed model is very beneficial in the design of the OMS for LISA which has performance characteristics very similar to those demonstrated on LPF.

The long term performance of the OMS showed no major degradations that could harm the LISA mission when extrapolated to longer mission durations. Solely the LPF polarization stability would not be good enough for LISA. However, since the origin of the polarization fluctuations could be identified, the effect can easily be avoided by an appropriate polarization filter.

LPF has shown that high precision optical systems such as the OMS can work even better in the stable environment of a specialized space platform than demonstrated on ground. Moreover, the OMS has proven to be highly stable and reliable in itself, which is a cornerstone for the much longer LISA mission.

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APPENDIX A: PARAMETER SUMMARY

\( x_{12} \): Displacement measurement between the two TMs.
\( x_1 \): Displacement measurement between TM1 and the SC.
\( x_R \): Longitudinal reference interferometer measurement.
\( x_F \): Longitudinal frequency interferometer measurement.
\( \Psi_F \): Longitudinal frequency interferometer measurement with subtraction of \( x_R \) processed onboard.
\( o_{12} \): Main telemetry channel of the differential displacement between the two TMs, processed onboard via \( x_{12} - x_R \), with a calibration factor to scale to [m].
\( o_1 \): \( x_1 \) with subtracted reference phase, processed onboard via \( x_1 - x_R \), with a calibration factor to scale to [m].
\( \phi_1 \): Calibrated \( xy \)-plane DWS TM1 angle, measured in X1.
\( \eta_1 \): Calibrated \( xz \)-plane DWS TM1 angle, measured in X1.
\( \phi_{12} \): Calibrated \( xy \)-plane DWS differential TM angle, measured in X12.
\( \eta_{12} \): Calibrated \( xz \)-plane DWS differential TM angle, measured in X12.
\( \phi_2 \): Calibrated \( xy \)-plane DWS TM2 angle, processed onboard from linear combinations of X1 and X12.
\( \eta_2 \): Calibrated \( xz \)-plane DWS TM2 angle, processed onboard from linear combinations of X1 and X12.

APPENDIX B: OMS EXPERIMENTS AND SELECTED “NOISE RUNS” DURING THE LPF MISSION

In this appendix we give a list of dedicated experiments that were performed during the mission with the aim to characterize the OMS in Tables I and II. Some few investigations have missing data and are not mentioned. Commissioning activities are also not listed. Most of the experiments in this list have been used in the analysis for this publication. Furthermore, the segments used in the long-term analysis are stated in Table III.
| Nr. | Experiment                                | Start (UTC)            | End (UTC)         | Description                                                                                                                                 |
|-----|-------------------------------------------|------------------------|------------------|---------------------------------------------------------------------------------------------------------------------------------------------|
| 1   | Engineering days                          | 2016-03-14 08:00:00    | 2016-03-17 20:00:00 | Sinusoidal injections to the lateral Degree-of-Freedom (DOF) of both TMs at different set-points and final TM realignments for TTL suppression. |
| 2   | Thermal stability experiments              | 2016-03-16             | 2017-06-17        | Experiments testing various OMS components, see [66], given for completeness.                                                                |
| 3   | Single quadrant investigations 1           | 2016-04-23 13:30:00    | 2016-04-23 13:50:22 | Attempt to characterize limited number of single quadrants with short 100 Hz segments.                                                        |
| 4   | Longitudinal step experiment 1             | 2016-04-25 08:00:00    | 2016-04-26 08:00:00 | Longitudinally shift TM2 to different set points to analyze noise behavior and fit RIN.                                                   |
| 5   | Single quadrant investigations 2           | 2016-04-25 12:50:02    | 2016-04-25 13:40:22 | Attempt to characterize limited number of single quadrants with short 100 Hz segments.                                                        |
| 6   | Beam power modulation                      | 2016-05-30 08:00:00    | 2016-05-31 08:00:00 | Modulation of reference beam, during thermal experiments.                                                                                  |
| 7   | Beam power modulation                      | 2016-05-31 20:00:00    | 2016-06-01 16:00:00 | Modulation of measurement beam.                                                                                                            |
| 8   | OPD loop characterization experiment       | 2016-06-01 16:35:08    | 2016-06-01 17:05:30 | Estimate controller transfer function and open-loop transfer function.                                                                       |
| 9   | Frequency loop characterisation experiment | 2016-06-01 17:35:10    | 2016-06-01 18:06:00 | Estimate transfer functions from laser frequency modulation.                                                                               |
| 10  | Frequency loop open                        | 2016-06-02 00:42:00    | 2016-06-02 05:40:00 | Data to compare to ground and closed-loop measurement.                                                                                       |
| 11  | DWS angular tilt experiment 1              | 2016-06-02 08:30:00    | 2016-06-03 07:25:00 | Tilt TM2 to different angular set-points to analyze noise behavior and fit RIN.                                                            |
| 12  | OPD loop open                              | 2016-06-11 12:02:31    | 2016-06-12 00:00:59 | Data to compare to ground and closed-loop measurement.                                                                                       |
| 13  | Single quadrant investigations 3           | 2016-06-12 15:10:00    | 2016-06-12 16:30:21 | Attempt to characterize limited number of single quadrants with short 100 Hz segments.                                                        |
| 14  | OPD loop characterization experiment       | 2016-06-13 04:35:05    | 2016-06-13 05:32:05 | Estimate controller transfer function and open-loop transfer function.                                                                       |
| 15  | Frequency loop characterization experiment | 2016-06-13 06:05:10    | 2016-06-13 06:55:49 | Estimate transfer functions from laser frequency modulations.                                                                                |
| 16  | Beam power modulation                      | 2016-06-13 21:27:30    | 2016-06-14 21:27:30 | Modulation of reference beam, during thermal experiments.                                                                                   |
| 17  | Path length mismatch experiment            | 2016-06-14 21:20:00    | 2016-06-15 06:42:00 | Estimate the path length mismatch from laser frequency modulations.                                                                       |
| 18  | Frequency loop open                        | 2016-06-15 02:53:00    | 2016-06-15 05:29:59 | Data to compare to ground and closed-loop measurement.                                                                                       |
| 19  | OPD loop open                              | 2016-07-07 10:57:01    | 2016-07-07 11:20:20 | Data to compare to ground and closed-loop measurement.                                                                                       |
| 20  | OPD loop open                              | 2016-07-07 11:23:00    | 2016-07-07 11:44:27 | Data to compare to ground and closed-loop measurement.                                                                                       |
| 21  | DWS derisk with large tilts                | 2017-01-17 07:00:00    | 2017-01-18 07:25:00 | Tilt TM1 and TM2 to large angular set points in preparation for DWS tilt 2 experiment, check for too low contrast.                         |

*(Table continued)*
| Nr. | Experiment                                | Start (UTC)       | End (UTC)        | Description                                                                 |
|-----|-------------------------------------------|-------------------|------------------|-----------------------------------------------------------------------------|
| 22  | SVN experiment 1                          | 2017-01-18 06:27:05 | 2017-01-18 06:53:32 | Injection of phase ramps via the OPD loop to measure SVN.                  |
| 23  | Short cross-talk experiment               | 2017-01-21 00:30:00 | 2017-01-21 05:30:00 | Sinusoidal injections to the lateral DOF of both TMs at one set point.     |
| 24  | Beam power modulation                     | 2017-01-21 08:00:00 | 2017-01-22 10:30:00 | Multiple modulations and counter modulation experiments with varying depths.|
| 25  | Frequency loop characterisation experiment| 2017-01-22 10:35:10 | 2017-01-22 11:26:49 | Estimate transfer functions from laser frequency modulations.              |
| 26  | Path length mismatch experiment           | 2017-01-22 12:05:00 | 2017-01-22 14:59:00 | Estimate the path length mismatch from laser frequency modulations.       |
| 27  | Frequency loop open                       | 2017-01-22 16:45:00 | 2017-01-23 02:59:59 | Data to compare to ground and closed-loop measurement.                    |
| 28  | SVN experiment 2                          | 2017-02-02 20:35:00 | 2017-02-04 21:30:00 | Injection of phase ramps via the OPD loop to measure SVN, with longitudinal offsets. |
| 29  | Longitudinal step experiment 2            | 2017-02-02 20:36:00 | 2017-02-04 21:36:00 | Longitudinally shift TM2 to different set-points to analyze noise behavior and fit RIN. Includes OPD ramps at offsets for SVN analysis and unbalanced segments. |
| 30  | Long cross-talk experiment                | 2017-02-04 21:00:00 | 2017-02-06 23:00:00 | Series of short cross-talk experiments performed at different set-points of the various DOFs. |
| 31  | DWS angular tilt experiment 2             | 2017-02-06 23:06:00 | 2017-02-08 22:55:00 | Tilt TM1 and TM2 to different angular set-points to analyse noise behavior and fit RIN. Includes unbalanced segments. |
| 32  | Frequency loop characterization experiment| 2017-04-06 22:30:08 | 2017-04-06 23:50:42 | Estimate transfer functions from laser frequency modulations.              |
| 33  | Reference beam power modulation           | 2017-06-17 08:20:00 | 2017-06-17 15:00:00 | Steps to lower the reference beam power, shift TM2 and increase the power again. |
| 34  | OPD loop open                             | 2017-06-25 16:43:10 | 2017-06-25 17:55:10 | Data to compare to ground and closed-loop measurement.                    |
| 35  | Direct RIN measurement at 2f              | 2017-07-11 00:00:00 | 2017-07-12 03:32:00 | Attempt to measure 2f-RIN directly by shifting the SBDFT to 2 kHz.         |
TABLE II. Spot position measurements; MB Measurement beam, RB reference beam. In-flight more measurements have been taken for which a few had missing data. Those are not given here.

| Nr. | Experiment | Start (UTC)    | End (UTC)    |
|-----|------------|----------------|--------------|
| 1   | RB off     | 2016-04-30 12:00:32 | 2016-04-30 12:04:09 |
| 2   | MB off     | 2016-04-30 12:05:03 | 2016-04-30 12:08:32 |
| 3   | RB off     | 2016-05-29 12:00:32 | 2016-05-29 12:04:09 |
| 4   | MB off     | 2016-05-29 12:05:03 | 2016-05-29 12:08:32 |
| 5   | RB off     | 2016-06-12 00:01:32 | 2016-06-12 00:05:09 |
| 6   | MB off     | 2016-06-12 00:06:03 | 2016-06-12 00:09:32 |
| 7   | RB off     | 2016-06-25 12:05:57 | 2016-06-25 12:09:41 |
| 8   | MB off     | 2016-06-25 12:10:32 | 2016-06-25 12:14:09 |
| 9   | RB off     | 2016-07-09 18:06:37 | 2016-07-09 18:10:07 |
| 10  | MB off     | 2016-07-09 18:11:07 | 2016-07-09 18:14:37 |
| 11  | RB off     | 2016-07-30 18:06:37 | 2016-07-30 18:10:07 |
| 12  | MB off     | 2016-07-30 18:11:07 | 2016-07-30 18:14:37 |
| 13  | RB off     | 2016-08-20 12:03:25 | 2016-08-20 12:07:19 |
| 14  | MB off     | 2016-08-20 12:07:58 | 2016-08-20 12:11:45 |
| 15  | RB off     | 2016-09-03 12:03:26 | 2016-09-03 12:07:20 |
| 16  | MB off     | 2016-09-03 12:07:55 | 2016-09-03 12:11:50 |
| 17  | RB off     | 2016-09-04 12:03:26 | 2016-09-04 12:07:20 |
| 18  | MB off     | 2016-09-04 12:07:55 | 2016-09-04 12:11:50 |
| 19  | RB off     | 2016-09-17 12:03:26 | 2016-09-17 12:07:20 |
| 20  | MB off     | 2016-09-17 12:07:55 | 2016-09-17 12:11:50 |
| 21  | RB off     | 2016-09-18 12:03:26 | 2016-09-18 12:07:20 |
| 22  | MB off     | 2016-09-18 12:07:55 | 2016-09-18 12:11:50 |
| 23  | RB off     | 2016-10-01 12:03:26 | 2016-10-01 12:07:20 |
| 24  | MB off     | 2016-10-01 12:07:55 | 2016-10-01 12:11:50 |
| 25  | RB off     | 2016-10-02 11:03:26 | 2016-10-02 11:07:20 |
| 26  | MB off     | 2016-10-02 11:07:55 | 2016-10-02 11:11:50 |
| 27  | RB off     | 2016-10-15 12:03:26 | 2016-10-15 12:07:20 |
| 28  | MB off     | 2016-10-15 12:07:55 | 2016-10-15 12:11:50 |
| 29  | RB off     | 2016-10-16 11:03:26 | 2016-10-16 11:07:20 |
| 30  | MB off     | 2016-10-16 11:07:55 | 2016-10-16 11:11:50 |
| 31  | RB off     | 2016-10-29 12:03:26 | 2016-10-29 12:07:20 |
| 32  | MB off     | 2016-10-29 12:07:55 | 2016-10-29 12:11:50 |
| 33  | RB off     | 2016-10-30 12:03:26 | 2016-10-30 12:07:20 |
| 34  | MB off     | 2016-10-30 12:07:55 | 2016-10-30 12:11:50 |
| 35  | RB off     | 2016-11-12 12:03:26 | 2016-11-12 12:07:20 |
| 36  | MB off     | 2016-11-12 12:07:55 | 2016-11-12 12:11:50 |
| 37  | RB off     | 2016-11-13 15:33:26 | 2016-11-13 15:37:20 |
| 38  | MB off     | 2016-11-13 15:37:55 | 2016-11-13 15:41:50 |
| 39  | RB off     | 2016-11-26 12:03:26 | 2016-11-26 12:07:20 |
| 40  | MB off     | 2016-11-26 12:07:55 | 2016-11-26 12:11:50 |
| 41  | MB off     | 2016-11-27 12:03:15 | 2016-11-27 12:07:20 |
| 42  | MB off     | 2016-11-27 12:07:26 | 2016-11-27 12:11:25 |
| 43  | MB off     | 2016-11-27 12:11:31 | 2016-11-27 12:15:30 |
| 44  | MB off     | 2016-11-27 12:15:36 | 2016-11-27 12:19:35 |
| 45  | RB off     | 2016-12-11 12:03:26 | 2016-12-11 12:07:20 |

(Table continued)
| Nr. | Experiment | Start (UTC)       | End (UTC)        |
|-----|------------|-------------------|------------------|
| 93  | RB off     | 2017-01-24 03:18:22 | 2017-01-24 03:22:18 |
| 94  | MB off     | 2017-01-24 03:23:08 | 2017-01-24 03:26:52 |
| 95  | RB off     | 2017-01-24 04:18:22 | 2017-01-24 04:22:18 |
| 96  | MB off     | 2017-01-24 04:23:08 | 2017-01-24 04:26:52 |
| 97  | RB off     | 2017-01-24 05:18:22 | 2017-01-24 05:22:18 |
| 98  | MB off     | 2017-01-24 05:23:08 | 2017-01-24 05:26:52 |
| 99  | RB off     | 2017-01-24 06:18:22 | 2017-01-24 06:22:18 |
| 100 | MB off     | 2017-01-24 06:23:08 | 2017-01-24 06:26:52 |
| 101 | RB off     | 2017-01-24 07:18:22 | 2017-01-24 07:22:18 |
| 102 | MB off     | 2017-01-24 07:23:08 | 2017-01-24 07:26:52 |
| 103 | RB off     | 2017-01-24 09:18:23 | 2017-01-24 09:22:18 |
| 104 | MB off     | 2017-01-24 09:23:08 | 2017-01-24 09:26:52 |
| 105 | RB off     | 2017-01-24 13:18:23 | 2017-01-24 13:22:18 |
| 106 | MB off     | 2017-01-24 13:23:08 | 2017-01-24 13:26:52 |
| 107 | RB off     | 2017-01-24 17:18:23 | 2017-01-24 17:22:18 |
| 108 | MB off     | 2017-01-24 17:23:08 | 2017-01-24 17:26:52 |
| 109 | RB off     | 2017-01-24 21:18:23 | 2017-01-24 21:22:18 |
| 110 | MB off     | 2017-01-24 21:23:08 | 2017-01-24 21:26:52 |
| 111 | RB off     | 2017-01-25 01:18:23 | 2017-01-25 01:22:18 |
| 112 | MB off     | 2017-01-25 01:23:08 | 2017-01-25 01:26:52 |
| 113 | RB off     | 2017-01-25 05:18:23 | 2017-01-25 05:22:18 |
| 114 | MB off     | 2017-01-25 05:23:08 | 2017-01-25 05:26:52 |
| 115 | RB off     | 2017-01-25 08:18:23 | 2017-01-25 08:22:18 |
| 116 | MB off     | 2017-01-25 08:23:08 | 2017-01-25 08:26:52 |
| 117 | RB off     | 2017-01-25 13:48:48 | 2017-01-25 13:52:43 |
| 118 | MB off     | 2017-01-25 13:53:33 | 2017-01-25 13:57:17 |
| 119 | RB off     | 2017-01-25 14:18:23 | 2017-01-25 14:22:18 |
| 120 | MB off     | 2017-01-25 14:23:08 | 2017-01-25 14:26:52 |
| 121 | RB off     | 2017-01-25 20:18:23 | 2017-01-25 20:22:18 |
| 122 | MB off     | 2017-01-25 20:23:08 | 2017-01-25 20:26:52 |
| 123 | RB off     | 2017-01-26 02:18:23 | 2017-01-26 02:22:18 |
| 124 | MB off     | 2017-01-26 02:23:08 | 2017-01-26 02:26:52 |
| 125 | RB off     | 2017-01-26 08:18:23 | 2017-01-26 08:22:18 |
| 126 | MB off     | 2017-01-26 08:23:08 | 2017-01-26 08:26:52 |
| 127 | RB off     | 2017-01-26 14:18:23 | 2017-01-26 14:22:18 |
| 128 | MB off     | 2017-01-26 14:23:08 | 2017-01-26 14:26:52 |
| 129 | RB off     | 2017-01-26 20:18:23 | 2017-01-26 20:22:18 |
| 130 | MB off     | 2017-01-26 20:23:08 | 2017-01-26 20:26:52 |
| 131 | RB off     | 2017-01-27 02:18:23 | 2017-01-27 02:22:18 |
| 132 | MB off     | 2017-01-27 02:23:08 | 2017-01-27 02:26:52 |
| 133 | RB off     | 2017-01-28 12:03:26 | 2017-01-28 12:07:20 |
| 134 | MB off     | 2017-01-28 12:07:55 | 2017-01-28 12:11:50 |
| 135 | RB off     | 2017-02-11 12:03:26 | 2017-02-11 12:07:45 |
| 136 | MB off     | 2017-02-11 12:08:18 | 2017-02-11 12:12:15 |
| 137 | RB off     | 2017-02-12 19:07:45 | 2017-02-12 19:11:43 |
| 138 | MB off     | 2017-02-12 19:12:15 | 2017-02-12 19:16:14 |
| 139 | RB off     | 2017-03-04 12:03:26 | 2017-03-04 12:07:45 |

(Table continued)
APPENDIX C: ACRONYMS

In this appendix, we provide a list of acronyms used in the main text.

| Acronym | Description |
|---------|-------------|
| ADC     | Analog to digital converter |
| AOM     | Acousto-optic modulator |
| ASD     | Amplitude spectral density |
| BH92    | Blackmann-Harris-92 |
| CQP     | Calibrated quadrant photodiode |
| DFACS   | Drag-free and attitude control system |
| DMU     | Data management unit |
| DOF     | Degree-of-freedom |
| DPS     | Differential power sensing |
| DRS     | Disturbance reduction system |
| DWS     | Differential wave front sensing |
| EM      | Engineering model |
| ESA     | European space agency |
| FIOS    | Fibre injector optical subassembly |
| FM      | Flight model |
| FPGA    | Field programmable optical array |
| GRS     | Gravitational reference sensor |
| IABG    | Industrieanlagen-Betriebsgesellschaft mbH |
| ITO     | Indium tin oxide |
| LISA    | Laser interferometer space antenna |
| LPF     | LISA Pathfinder |
| LPSD    | Logarithmic frequency axis power spectral density |
| MAD     | Median absolute deviation |
| OB      | Optical bench |
| OBF     | Optical bench frame |
| OMS     | Optical metrology system |
| OPD     | Optical path length difference |
| OSTT    | On-station thermal tests |
| OW      | Optical window |
| PD      | Photodiode |
| PLL     | Phase locked loop |
| PM      | Phasemeter |
| PSD     | Power spectral density |
| PZT     | Piezoelectric transducer |
| QPD     | Quadrant photodiode |
| RF      | Radio frequency |
| RIN     | Relative intensity noise |
| SBDFT   | Single-bin discrete Fourier transform |
| SC      | Spacecraft |
| SEP     | Solar energetic particles |
| SNR     | Signal-to-noise ratio |
| SVN     | Small vector noise |
| TM      | Test mass |
| TTL     | Tilt-to-length |

TABLE III. Selected and filtered “noise runs” for the long-term noise behavior analysis of the OMS, see Sec. III F.

| Nr. | Start (UTC) | End (UTC) |
|-----|-------------|-----------|
| 1   | 2016-03-01 08:05:00 | 2016-03-02 23:59:59 |
| 2   | 2016-03-03 15:00:00 | 2016-03-04 20:59:58 |
| 3   | 2016-03-06 18:15:00 | 2016-03-08 08:00:00 |
| 4   | 2016-03-13 16:30:00 | 2016-03-15 06:59:59 |
| 5   | 2016-03-16 20:00:00 | 2016-03-18 23:59:59 |
| 6   | 2016-03-21 02:00:00 | 2016-03-26 07:59:59 |
| 7   | 2016-03-27 14:00:00 | 2016-03-28 07:59:58 |
| 8   | 2016-03-29 08:00:00 | 2016-03-30 07:59:59 |
| 9   | 2016-03-31 08:00:00 | 2016-04-02 01:59:59 |
| 10  | 2016-04-04 00:00:00 | 2016-04-14 07:59:59 |
| 11  | 2016-04-26 08:04:00 | 2016-04-28 07:59:59 |
| 12  | 2016-05-01 08:05:00 | 2016-05-02 23:54:58 |
| 13  | 2016-05-03 08:00:00 | 2016-05-05 15:29:59 |
| 14  | 2016-05-13 00:50:00 | 2016-05-13 07:29:59 |
| 15  | 2016-05-13 08:30:00 | 2016-05-14 07:59:58 |
| 16  | 2016-05-16 00:00:00 | 2016-05-19 04:59:59 |
| 17  | 2016-05-19 11:30:00 | 2016-05-21 10:59:58 |
| 18  | 2016-05-21 15:30:00 | 2016-05-23 13:59:59 |
| 19  | 2016-05-23 17:03:00 | 2016-05-25 16:43:59 |
| 20  | 2016-06-06 11:05:00 | 2016-06-09 07:59:58 |
| 21  | 2016-06-15 13:35:00 | 2016-06-18 07:58:59 |
| 22  | 2016-06-19 13:00:00 | 2016-06-24 07:59:59 |
| 23  | 2016-09-19 05:00:55 | 2016-09-21 12:59:33 |
| 24  | 2016-09-21 13:45:00 | 2016-09-22 05:55:58 |
| 25  | 2016-09-28 13:35:00 | 2016-10-01 07:55:58 |
| 26  | 2016-11-09 00:45:19 | 2016-11-12 07:58:20 |
| 27  | 2016-11-16 11:05:00 | 2016-11-26 07:59:59 |
| 28  | 2016-12-26 08:00:00 | 2017-01-13 19:57:59 |
| 29  | 2017-05-28 23:43:55 | 2017-06-05 14:59:00 |
| 30  | 2017-06-08 12:00:45 | 2017-06-17 02:55:58 |
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