Noise of a single-electron emitter

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I analyze the correlation function of currents generated by the periodically driven quantum capacitor emitting single electrons and holes into the chiral waveguide. I compare adiabatic and non-adiabatic, transient working regimes of a single-electron emitter and find the striking difference between the correlation functions in two regimes. Quite generally for the system driven with frequency $\Omega$ the correlation function depends on two frequencies, $\omega$ and $\ell \Omega - \omega$, where $\ell$ is an integer. For the emitter driven non-adiabatically the correlation functions for different $\ell$ are similar and almost symmetric in $\omega$. While in the case of adiabatic drive the correlation functions for $\ell \neq 0$ are highly asymmetric in $\omega$ and exceed significantly the one corresponding to $\ell = 0$. Under optimal operating conditions the correlation function for odd $\ell$ is zero.

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I. INTRODUCTION

Non-stationary transport in quantum systems\textsuperscript{1} attracts recently increasing interest. In the dynamic regime the quantum conductors show properties revealing such fundamental features of individual carriers, which are not accessible in the stationary regime. Thus, recent experiments demonstrate that even a regular stream of electrons, see Fig. 1, is noisy on a short time scale due to quantum fluctuations of the emission time.\textsuperscript{2} The fluctuations of energy of emitted electrons\textsuperscript{3} as well as the random waiting times between charge transfers\textsuperscript{4} are also consequences of the probabilistic nature of the emission process.

The experimental implementation of high-speed single-electron sources not relying on an electron-electron interaction\textsuperscript{5–9} opens up perspectives for the development of quantum coherent electronics (or, as it is often referred to, electron quantum optics\textsuperscript{10}), an emerging field aimed at quantum processing applications,\textsuperscript{11–15} manipulation\textsuperscript{16–21} and transportation\textsuperscript{22–23} of single to few particles wave-packets in the condensed matter. To characterize the coherence properties of a single-electron state the few protocols were already suggested.\textsuperscript{10,24–26}

One of the most important questions for quantum information processing applications\textsuperscript{27} is whether the emitted state is a genuine single-particle state. In quantum optics the single-photon state is probed via the (no-)coincidence measurement.\textsuperscript{28} In mesoscopics the current noise measurement is more suitable for this purpose. It is known that the zero-frequency noise vanishes if the stream of electrons is regular.\textsuperscript{29} Indeed, the measurements\textsuperscript{30} show a significant reduction of the low-frequency noise power in the quantized emission regime, when the source emits the same number of electrons, in particular one electron, during each working period. The direct counting of the number of particles emitted in each cycle was demonstrated by measuring the partition noise of the beam-splitter\textsuperscript{31} and with the help of a charge detector\textsuperscript{32}.

As the experiment reveals, the noise spectrum provides an accurate characterization of a single-electron emission regime.\textsuperscript{2} If the source works as a single-electron emitter then the excess finite-frequency noise becomes minimal though not zero. To shed light onto the nature of this fundamental noise limit, the semiclassical model of a single-electron emitter, consisting in a particle periodically attempting to leave the cavity was put forward.\textsuperscript{2,33,34}

This model, shows clearly that the uncertainty in the emission time puts the lower limit to the excess finite-frequency noise. This limiting noise is referred to as the phase noise. The phase noise limit was achieved in Ref. 2 for the source driven by the pulsed potential and thus emitting electrons non-adiabatically.

Our aim here is to analyze the entire spectrum of the noise generated by the single-particle source. In addition to the noise present already in the stationary case\textsuperscript{35} and modified by the working single-particle source, - this modification is referred to as the excess noise, - there is also the photon-induced noise completely absent in the stationary case. To calculate the noise spectrum we use the Floquet scattering matrix theory\textsuperscript{36} for both adiabatic and non-adiabatic emission.\textsuperscript{37} For non-adiabatic...
optimal working conditions our analytical results for the phase noise are in agreement with both the solutions of the semiclassical model and the results of numerical calculations based on the Floquet approach to photon assisted noise, which were contrasted in Ref. 34. As we will show the phase noise in the adiabatic regime exhibits similar behavior: At low measurement frequencies \( \omega \) the phase noise grows quadratically while with increasing frequency it gets suppressed. The photon-induced noise generated in the non-adiabatic regime resembles the phase noise. In contrast, the photon-induced noise in the adiabatic regime shows completely different behavior: It exists within the finite frequency window, is asymmetric in frequency, and far exceeds the phase noise in magnitude. This is one more example illustrating a striking difference between the quantum states of electrons emitted adiabatically and non-adiabatically, see also Ref. 37.

The paper is organized as follows. In Sec. II we use the Floquet scattering matrix formalism to derive the correlation function for currents generated by the quantum capacitor attached to the chiral electron waveguide. The scattering amplitude of a periodically driven capacitor is presented and analyzed in Sec. III. The spectrum of the current correlation function in adiabatic and non-adiabatic regimes is presented in Secs. IV and V for the excess noise and the photon-induced noise, respectively. We conclude in Sec. VI. Some details of calculations are given in appendices.

II. FINITE-FREQUENCY NOISE OF A CAPACITOR ATTACHED TO THE HALL BAR

The cartoon of the set-up of Ref. 2 is shown in Fig. 2. The cavity, a circular edge state, driven by the periodic potential of the top-gate, \( U(t) = U(t + 2\pi/\Omega) \), emits electrons and holes into the chiral edge state of the Hall bar. The working principle of this single-electron source is as follows. The sample is made in a two-dimensional electron gas in the quantum Hall effect regime\(^{38}\). In this regime the bulk of the sample is not conducting. However there exist conducting states along the edges of a sample. Electrons propagate within edge states without backscattering.\(^{39,40}\) Therefore, these states are referred to as the chiral edge states. The emitter is a small cavity with quantized electron spectrum which is side-coupled to the chiral electron waveguide with the chemical potential \( \mu \). The periodic potential \( U(t) \) drives cavity’s levels up and down. When \( eU(t) \) increases and some cavity’s level crosses \( \mu \) an electron is emitted to the waveguide, since in the waveguide only the states with energy larger than \( \mu \) are empty (for simplicity we consider zero temperature). In the reverse process an electron enters the cavity, a hole is emitted to the waveguide. In total during the period no net charge is emitted to the waveguide. Therefore, such a cavity can be referred to as a quantum capacitor.\(^{35,41}\)

The essential feature of the measurement set-up is the presence of two independent metallic contacts, 1 and 2 playing the role of the source and drain, respectively, for an electron waveguide. These contacts have the same chemical potential \( \mu \), the same temperature, \( T \), and thus the same Fermi distribution function \( f(E) \). Only one of contacts, namely the contact 2, is used to measure the current \( I(t) \), associated with particles emitted by the capacitor, and its fluctuations. Importantly, since the current flowing into and out of the contact 2 are not correlated, it is possible to extract the contribution to current fluctuations solely due to emitted particles.\(^{34}\)

A. Current correlation function

The current fluctuations at contact 2 we characterize with the help of the current correlation function\(^{29}\)

\[
P_{22}(\omega_1, \omega_2) = \frac{1}{2} \left\langle \delta I_2(\omega_1) \delta I_2(\omega_2) + \delta I_2(\omega_2) \delta I_2(\omega_1) \right\rangle.
\]

(1)

Here the angle brackets \( \langle \ldots \rangle \) stands for quantum-statistical average over the equilibrium state of electron reservoirs; \( \delta I_2(\omega) = \bar{I}_2(\omega) - \langle \bar{I}_2(\omega) \rangle \) is an operator of current fluctuations in the contact 2. The current operator, \( \bar{I}_n(2) \), is expressed in terms of the second quantization operators \( \hat{b}_n(2) \), \( \hat{b}_n^\dagger(2) \), and annihilation, \( \hat{b}_2 \), operators for particles moving in the wave guide and entering the reservoir 2 and the operators \( \hat{a}_n \) and \( \hat{a}_n^\dagger \) for particles leaving the reservoir 2. If the relevant energy scales (such as the voltage applied, the temperature difference, the energy quantum \( \hbar \Omega \), etc.) characterizing the non-equilibrium state are small compared to the Fermi energy \( \mu \), then the current operator reads,\(^{42}\)
\[ I_2(\omega) = e^{\int_0^\infty dE \left\{ \hat{b}_2^\dagger(E) \hat{b}_2(E + \hbar\omega) - \hat{a}_2^\dagger(E) \hat{a}_2(E + \hbar\omega) \right\} }. \tag{2} \]

In the case shown in Fig. 2, the particles entering the contact 2 were originally emanated by the contact 1 and subsequently scattered by the capacitor. Since the capacitor is driven by the periodic potential with frequency \( \Omega \), an electron scattered off the capacitor can gain or lose one or several energy quantum \( \hbar\Omega \). Such a scattering is described by the Floquet scattering matrix, which we will denote as \( S_F \). Therefore, we write,\(^{36}\)

\[ \hat{b}_2(E) = \sum_{n=-\infty}^{\infty} S_F(E, E_n) \hat{a}_1(E_n), \tag{3} \]

where \( E_n = E + n\hbar\Omega \). The element \( S_F(E, E_n) \) is a forward-scattering photon-assisted amplitude for electrons propagating in the waveguide and passing through the place where the capacitor is attached to. We stress, the initial equations are written in terms of operators for electrons propagating in the waveguide, however the final equations can be interpreted as such which describe the properties of particles emitted by the source attached to the waveguide.

The electron reservoirs, contacts 1 and 2, are assumed to be in equilibrium. Therefore, the operators, \( \hat{a}_\alpha \), \( \alpha = 1, 2 \), for particles emanating by these contacts satisfy the following relations for equilibrium fermions:

\[ \hat{a}_{\alpha}^\dagger(E) \hat{a}_{\beta}(E') + \hat{a}_{\beta}(E') \hat{a}_{\alpha}^\dagger(E) = \delta_{\alpha, \beta} \delta(E - E'), \]
\[ \langle \hat{a}_{\alpha}^\dagger(E) \hat{a}_\alpha(E') \rangle = f(E) \delta(E - E'). \tag{4} \]

Here \( \delta_{\alpha, \beta} \) is the Kronecker symbol and \( \delta(E - E') \) is the Dirac delta function. \( f(E) \) is the Fermi distribution function the same for both contacts, \( f_\alpha(E) = f(E), \alpha = 1, 2 \).

For the reference purposes we give a time-dependent current, \( I_2(t) = \int d\omega/(2\pi) e^{-i\omega t} \langle \hat{I}_2(\omega) \rangle \), generated in the contact 2. Substituting Eq. (3) into Eq. (2) and calculating quantum-statistical averaging with the help of Eq. (4) we obtain,\(^{43}\)

\[ I_2(t) = \sum_{\ell=-\infty}^{\infty} e^{-\ell\omega t} \frac{\hbar}{\pi} \int_0^\infty dE \times \sum_{n=-\infty}^{\infty} S_F(E_n, E) S_F(E_n+\ell, E) \{ f(E) - f(E_n) \}. \tag{5} \]

Now we calculate the current correlation function. Substituting Eq. (2) into Eq. (1) and calculating quantum-statistical averaging as before we arrive at the following general equation for the current correlation function,\(^{44,45}\)

\[ P_{22}(\omega_1, \omega_2) = \sum_{\ell=-\infty}^{\infty} 2\pi \delta(\omega_1 + \omega_2 - \ell\Omega) P_\ell(\omega_1), \tag{6a} \]

with the noise power

\[ P_\ell(\omega) = \frac{e^2}{\hbar} \int_0^\infty dE \left\{ \delta_{\ell 0} F(E, E + \hbar\omega) + \sum_{n,m,p=-\infty}^{\infty} F(E_{\ell+n}, E_m + \hbar\omega) \times S_F^* (E, E_{\ell+n}) S_F(E + \hbar\omega, E_m + \hbar\omega) \right. \]
\[ \left. \times S_F(E_{\ell+p}, E_{\ell+n}) S_F^* (E_p, E_m + \hbar\omega), E_m + \hbar\omega) \right\}, \tag{6b} \]

and the symmetrized combination of the Fermi functions

\[ F(E_1, E_2) = \frac{f(E_1) [1 - f(E_2)] + f(E_2) [1 - f(E_1)]}{2}. \tag{6c} \]

Notice with small \( \omega_1 \) and \( \omega_2 \) we denote frequencies which the correlation function is measured at, while with a large \( \Omega \) we denote the frequency of the potential driving the source. Within the Floquet scattering matrix formalism the component of the noise power with \( \ell = 0 \) was addressed in Ref. 46.

One can check that the noise power satisfies the following symmetries,\(^{47}\)

\[ P_\ell(\omega) = P_\ell(\ell\Omega - \omega), \]
\[ P_\ell(\omega) = \{ P_{-\ell}(-\omega) \}^*. \tag{7} \]

### B. Excess noise

The central component, \( \ell = 0 \), of the correlation function, Eq. (6), is not zero even in equilibrium, when the driving potential is switched off. This is due to the quantum noise.\(^{48,49}\) To assess the contribution due to the working SES only, the difference between the correlation functions (corresponding to \( \ell = 0 \)) with the SES on and off was measured.\(^2\) This difference is referred to as the excess noise. We will denote it with the superscript “\( \text{ex} \)” After the simple algebra we get from Eq. (6b),
ing the role of an electron waveguide. The QPC has

\( \mathcal{P}^{\text{ex}}(\omega) = \frac{e^2}{\hbar} \int_0^\infty dE \sum_{m=-\infty}^\infty F(E, E_m + \hbar \omega) \times \left\{ \Pi_m(E, \omega) \right\}^2 - \delta_{m,0}, \) \tag{8}

where

\( \Pi_m(E, \omega) = \sum_{q=-\infty}^\infty S_F(E_q, E) S_F^*(E_q + \hbar \omega, E_m + \hbar \omega). \) \tag{9}

The similar equation was found in Ref. 34.

C. Photon-induced noise

In the stationary case the currents at \( \omega_1 \neq -\omega_2 \) are not correlated and the corresponding terms of the current correlation function, Eq. (6a), all are zero. In contrast, in the dynamic case the currents at \( \omega_1 \) and \( \omega_2 = t\Omega - \omega_1 \) are correlated due to the ability of a dynamic scatterer to provide (emit or absorb) photons with energy \( \hbar \Omega \). Therefore, the side components, \( t \neq 0 \), of the correlation function, Eq. (6), appear. We name the corresponding noise as the photon-induced noise (PIN). The spectrum of the PIN power reads,

\[ \mathcal{P}_t^{\text{PIN}}(\omega) = \frac{e^2}{\hbar} \int_0^\infty dE \sum_{m=-\infty}^\infty F(E, E_m + \hbar \omega) \times \Pi_m(E, \omega) \Pi_m^{\ast}(E, \omega), \] \tag{10}

where

\[ \Pi_m(E, \omega) = \sum_{q=-\infty}^\infty S_F(E_q + t, E) S_F^*(E_q + \hbar \omega, E_m + \hbar \omega). \] \tag{11}

To proceed with calculations we need to specify the model describing an electron source and to calculate the Floquet scattering matrix, \( S_F \), of the source.

III. FLOQUET SCATTERING MATRIX OF A DrIVEN CAPACITOR

In order to calculate the Floquet scattering matrix elements \( S_F(E_n, E) \) entering Eqs. (9) and (11) we use the following model. Electrons in the cavity propagate along a single-channel chiral state of length \( L \), which is coupled with the help of a quantum point contact (QPC) to a single-channel linear edge state playing the role of an electron waveguide. The QPC has energy-independent reflection, \( r \), and transmission, \( t \), amplitudes. If the electron spectrum can be linearized, \( k(E) \approx k(\mu) + (E - \mu) \partial k/\partial E \), then this model admits an exact solution: \( S_F(E, E_n) = S_{\text{out},-n}(E) \equiv \int_0^\tau dt e^{-i\Omega t} S_{\text{out}}(E, t) \). \tag{12}

with

\[ S_{\text{out}}(E, t) = \sum_{q=0}^\infty S(q)(t), \]

\[ S(0) = r, \quad S(q>0)(t) = r^2 e^{iql} e^{-i\Phi_q(t)}, \]

\[ \Phi_q(t) = \frac{e}{\hbar} \int_0^t dt' U(t'). \] \tag{13}

Here \( \tau \) is the time of one turn around the cavity. The physical meaning of the amplitude \( S_{\text{out}} \) is discussed in Ref. 53.

Given above general solution is not restricted to any particular amplitude and/or time-dependence of the driving potential \( U(t) \). However the scattering amplitude becomes especially simple in two limiting cases: (i) When the potential changes slow, adiabatically and (ii) when the time-dependent potential consists of the sequence of pulses. In the later case the potential changes non-adiabatically. The time scale differentiating slow and fast variations is put by the dwell time \( \tau_D \), the time necessary for an electron to leave the cavity starting from the time when the unoccupied states outside become available. Below we are interested in the case when the time interval between electron \( (t_-) \) and hole \( (t_+) \) emission is long enough such that the particles are definitely emitted,

\[ |t_- - t_+| \gg \tau_D. \] \tag{14}

This regime is referred to as the quantized emission regime. For the discussion of a short-period case see Refs. 2,33,34.

A. Adiabatic emission

When the potential changes slow, we can keep \( U(t') \) constant while integrating over \( t' \) in Eq. (13). Then the scattering amplitude becomes the frozen scattering amplitude, \( S_{\text{out}}(E, t) \), of a Fabry-Perot type: \( S(E, t) = \frac{-e^{i(\phi(E,t) + \theta_r)}}{1 - \sqrt{R} e^{-i\phi(E,t)}} \). \tag{15}

where \( \theta_r \) and \( R \) are the phase and the square absolute value of the QPC’s reflection amplitude, \( r = \ldots \)
The time-dependent phase acquired by an electron during one turn around the cavity is

\[ \phi(E,t) = \theta_t + k_\mu L + \frac{2\pi (E - \mu)}{h/\tau} - \frac{eU(t)}{h/\tau}, \tag{16} \]

with \( \tau = \Delta/h \) a duration of one turn and \( \Delta \) the level spacing in the cavity.

Here we are concerned with the case when the QPC’s transmission is small, \( T = 1 - R \ll 1 \) and, correspondingly, the width \( \delta \) of the levels in the cavity is small compared to the level spacing. For the equidistant spectrum it is \( \delta = \bar{T}\Delta/(4\pi) \). In this case the levels in the cavity can be modeled as the Breit-Wigner resonances and the scattering amplitude can be represented as follows,

\[ S(E,t) = e^{i\theta_t} \sum_n t - t_{\tau}^{(n)}(E) \pm i\Gamma^{(n)}(E) \left/ t - t_{\tau}^{(n)}(E) \mp i\Gamma^{(n)}(E) \right., \tag{17} \]

where \( n \) numbers the quantum levels in the cavity. The upper (lower) sign is for the case when the levels go up (down) under the action of the potential \( U(t) \). The time \( t_{\tau}^{(n)}(E) \) is a time when the \( n \)th quantum level has an energy \( E \), and \( \Gamma^{(n)}(E) \) is a parameter, which can be interpreted as the time interval during which the level of width \( \delta \) crosses the energy \( E \). Below we consider regime when only one level, \( n = 0 \), crosses the Fermi level. For the sake of short notation we will drop the level number superscript and use, \( t_{\tau} = t_{\tau}^{(0)} \) and \( \Gamma_{\tau} = \Gamma^{(0)}_{\tau} \).

At the adiabatic drive the Floquet scattering matrix element can be expressed (to the leading order in small \( \Omega \)) as the Fourier coefficient of the frozen scattering amplitude, \( S_F(E_n,E) \approx S_n(E) \). The use of this approximation in Eq. (5) results in the following current (at zero temperature): \( I_2(t) = ie/(2\pi)S^\ast(\mu,t)\partial S(\mu,t)/\partial t \). The frozen scattering amplitude given in Eq. (17) leads to the current consisting of the Lorentzian pulses. If only one level crosses the Fermi level during the period, the current is (see Fig. 3)

\[ I_{ad}(t) = \frac{e\Gamma_{\tau}/\pi}{(t - t_{-})^2 + \Gamma_{\tau}^2} - \frac{e\Gamma_{\tau}/\pi}{(t - t_{+})^2 + \Gamma_{\tau}^2}. \tag{18} \]

Here the superscript “\( ad \)” stands for the adiabatic regime; \( t_{-} \) (\( t_{+} \)) is the time of an electron (a hole) emission. In the above equation the current is given for the single period, \( 0 < t < T \) only. For other times it should be extended periodically, \( I_{ad}(t + nT) = I_{ad}(t) \), where \( n \) is an integer.

Given the current pulse shape we can formulate the adiabaticity condition more precisely,

\[ \Gamma_{\tau} \gg \tau_D. \tag{19} \]

That is, the time scale \( \Gamma_{\tau} \) over which the current varies has to be large compared to the time during which an electron escapes the cavity, the dwell time \( \tau_D \). Note, in the adiabatic regime the condition guaranteeing that the particle is emitted is more strict than Eq. (14). It is the following,

\[ |t_{-} - t_{+}| \gg \Gamma_{\tau}. \tag{20} \]

B. Non-adiabatic emission: Periodic step potential

Now we analyze the case when the potential \( U(t) \) changes fast compared to \( \tau_D \). As an example we choose the case when \( U(t) \) changes in a step-like manner between the values \( U_0 \) and \( U_1 \) during the period \( T \):

\[ U(t) = \begin{cases} U_0, & -T/2 < t < t_{-}, \\ U_1, & t_{-} < t < t_{+}, \\ U_0, & t_{+} < t < T/2. \end{cases} \tag{21} \]

Let us consider only the first potential jump and assume that before \( t < t_{-} \) an electrons system was in equilibrium corresponding to the cavity’s potential \( U = U_0 \). When the potential changes abruptly at \( t = t_{-} \) the system is driven into a non-equilibrium state. It takes a time of order \( \tau_D \) for the system to adjust to a new value of the
potential $U(t)$ and to equilibrate. During this equilibration transient process an electron (if $e(U_1 - U_0) > 0$) or a hole (if $e(U_1 - U_0) < 0$) is emitted. After this time, $t - t_\tau > \tau_D$, the system is in a new equilibrium state, which corresponds to the cavity’s potential $U = U_1$. Thus, if the time-delay between the two potential steps is long enough, see Eq. (14), then the electron system has enough time to react to each potential step independently. Therefore, the Floquet scattering matrix corresponding to the periodic step potential, Eq. (21), and under the condition set out in Eq. (14) we calculate the corresponding scattering amplitude. To make equation more compact, it is convenient to take as a period the following time interval, $t \in (t_+ - T, t_+)$.

Then we can write

$$S_{out} = \begin{cases} S(E, U_0) + \delta S(E, t-t_\tau), & t_+ - T < t < t_\tau, \\ S(E, U_1) + \delta \tilde{S}(E, t-t_\tau), & t_\tau < t < t_+. \end{cases}$$

(26)

Here $\delta \tilde{S}$ is the same as $\delta S$ but with $U_0$ and $U_1$ being interchanged. Finally, to get the Floquet scattering matrix elements, Eq. (12) we have to calculate the Fourier coefficient for the function given above. It can be simply done in the general case.

However below we restrict ourselves to the experimentally important case when the potential is changed by exactly one level spacing, $e\delta U = \Delta$, and the potential $U_0$ is chosen in such a way that the Fermi level is positioned exactly in the middle between the two subsequent levels in the cavity. These working conditions are referred to as the optimal working conditions.6 At the non-optimal operating conditions the noise is enhanced indicating that the source works not as a single electron emitter.34,57

1. Optimal operating conditions

For the optimal operation conditions $\phi_1 = \phi_2 \equiv \phi^{opt}$ with

$$\phi^{opt}(E) = \pi + 2\pi (E - \mu) / \Delta,$$

(27)

and

$$\delta S_N^{opt} = -e^{i\theta_0} \tilde{T} \sqrt{R_N} \frac{e^{i(N+1)\phi^{opt}(E)} \left( e^{-2\pi i + \frac{\pi}{2}} - 1 \right)}{1 - \sqrt{R} e^{i\phi^{opt}(E)}}.

(28)

To calculate the Fourier coefficients of the function $S_{out}(E, t)$, Eq. (26), we represent the integral over time in Eq. (12) as the sum of integrals over the time interval of duration $\tau$. Under the condition set out in Eq. (14) this sum contains the infinite number of terms. Thus, substituting Eq. (28) into Eq. (26) and subsequently into Eq. (12) we find after a little algebra the Floquet scattering matrix elements for the optimal operating conditions,

$$S_S^{opt}(E, E_n) = S(E) \delta_{n,0} - A_n \left\{ \frac{e^{-i\Omega t -}}{1 + \frac{n\Omega t}{\Delta}} - \frac{e^{-i\Omega t +}}{1 - \frac{n\Omega t}{\Delta}} \right\},$$

(29)
where \( E_n = E + n\hbar\Omega \). \( S(E) \) is the stationary scattering amplitude at \( U = U_0 \),

\[
S(E) = -e^{i\phi(E) + \theta_1} \frac{1 - \sqrt{R} e^{-i\phi(E)}}{1 - \sqrt{R} e^{i\phi(E)}},
\]

and the factor

\[
A_n = \frac{\bar{T} S(E) e^{i\pi \frac{n\hbar\Omega}{\Delta}} \sin(\pi \frac{n\hbar\Omega}{\Delta})}{\left(1 - \sqrt{R} e^{-i\phi(E)}\right) \left(1 - \sqrt{R} e^{i\phi(E) + n\Omega}\right)}.
\]

For practical calculations it is convenient to exploit the fact that under the conditions set out in Eq. (14) the energy quantum \( \hbar\Omega \) is much smaller than any other energy scale over which the Floquet scattering matrix changes. Then one can ignore the discreteness on \( n \) and introduce a continuous variable \( \Omega_n = n\Omega \) instead.

2. Continuous frequency representation

The following relations establish correspondence between the discrete frequency representation we used before and the continuous frequency representation we are going to use from now on:

\[
\sum_{n=-\infty}^{\infty} \rightarrow \int_{-\infty}^{\infty} d\Omega_n, \quad \frac{\tau}{\Omega} \int_0^\tau e^{in\Omega t'} \rightarrow \int_0^\infty dt' e^{i\Omega_n t'}. \tag{32}
\]

In addition the Kronecker symbol has to be replaced with the Dirac delta function, \( \delta_{m,n} \rightarrow \Omega_0(\Omega_n - \Omega_m) \).

To simplify notations even more we normalize energies by the level spacing,

\[
\epsilon = \frac{E - \mu}{\Delta}, \quad \omega_n = \frac{n\Omega}{\Delta}. \tag{33}
\]

Then, for any function \( \psi(E) \) after the substitutions given in Eq. (33) we use the notation \( \psi(\epsilon) \). For instance, the Fermi function \( f(E) \) reads as follows,

\[
f(\epsilon) = \frac{1}{1 + e^{(E - \mu)/\Delta}}, \tag{34}
\]

where \( k_B \) is the Boltzmann constant. In addition we use the following short notation,

\[
\rho(\epsilon) = \frac{1 - \sqrt{R} e^{i\phi(E)}(\epsilon)}{\sqrt{T}}, \tag{35}
\]

with \( \phi(\epsilon) = \pi + 2\epsilon \), see Eq. (27). With new variables and short notations the equation (29) becomes

\[
S^\text{opt}(\epsilon, \epsilon + \omega_n) = S(\epsilon) \frac{\hbar \Omega \sin(\pi\omega_n)}{\pi\omega_n} e^{i\omega_n} π \exp\left(-\frac{\pi}{2} \frac{\omega_n}{\Omega}\right) + \frac{\exp\left(-\frac{\pi}{2} \frac{\omega_n}{\Omega}\right)}{(2\Omega)^{2}} \delta(\omega_n) \rho(\epsilon) \rho(\epsilon + \omega_n). \tag{36}
\]

We remind that \( \Omega \) is the frequency of the drive, \( \Delta \) is the level spacing, \( t_-(t_+) \) is the time of an electron (hole) emission. In Appendix A we show that this Floquet scattering matrix is unitary.

Using above equation in Eq. (5) we calculate a time dependent current which consists of asymmetric pulses decaying exponentially, \( \delta(\epsilon) \), see Fig. 3.

\[
I^{\text{na}}(t) = \frac{e\theta(t - t_-)}{\tau_D} e^{-\frac{\epsilon}{\tau_D}} - \frac{e\theta(t - t_+)}{\tau_D} e^{-\frac{\epsilon + \tau_D}{\tau_D}}. \tag{37}
\]

Here the superscript “na” stands for the non-adiabatic regime; \( \theta(t) \) is the Heaviside step function.

IV. EXCESS NOISE

A. Adiabatic regime

The adiabatic regime implies that the Floquet scattering matrix changes only a little with energy on the scale put by the frequency, i.e., on the scale \( \hbar\Omega \sim \hbar\omega \). Therefore, we can expand the scattering amplitudes in the powers of \( \omega \). Since the capacitor does not support a DC current, the current correlations function tends to zero at zero frequency. Taking into account the unitarity of the Floquet scattering matrix, Eq. (A1), we find from Eq. (8) that the zero-frequency noise, \( \omega = 0 \), is indeed zero.

At small frequency we expand \( \Pi_m(E, \omega) \), Eq. (9), up to the second order in \( \omega \) and find,

\[
|\Pi_m(E, \omega)|^2 = \delta_{m,0} \left(1 + \omega^2 \frac{\partial^2 \Pi_m(E, \omega)}{\partial \omega^2} \right)|_{\omega=0} \quad + \omega^2 \left|\frac{\partial \Pi_m(E, \omega)}{\partial \omega} \right|^2|_{\omega=0} + O(\omega^3). \tag{38}
\]

Here \( O(\omega^3) \) denotes the third and higher order in \( \omega \) terms. Note, due to the unitarity condition, Eq. (A1),
the linear in \( \omega \) term vanishes in the above equation. As it follows quite generally from Eqs. (8) and (38), the low frequency noise is quadratic in \( \omega \). Earlier the similar behavior was shown for the noise of a chaotic cavity.\(^{58}\)

To the leading order in \( \Omega \) the Floquet scattering matrix elements are given by the Fourier coefficients of the frozen scattering amplitude.\(^ {1}\) Thus we use

\[
S_F(E_q, E) \approx S_q(E), \tag{39}
\]

in Eqs. (38), (9) and then in Eq. (8) and find,

\[
\mathcal{P}_{\text{ex,ad}}(\omega) = \frac{e^2 \hbar^2}{2\pi} \int_0^\infty \sum_{m=-\infty}^{\infty} \left| \left( S \frac{\partial S^*}{\partial E} \right)_m \right|^2 \times \{ F(E, E + m\hbar\Omega + \hbar\omega) - F(E, E + \hbar\omega) \}. \tag{40}
\]

where the subscript “\( m \)” indicates the Fourier coefficient. Note, while calculating the above equation we used the following identities,

\[
\sum_m \left| \left( S \frac{\partial S^*}{\partial E} \right)_m \right|^2 = \left( \frac{\partial S}{\partial E} \frac{\partial S^*}{\partial E} \right)_0. \tag{41}
\]

which can be proven using the unitarity, \(|S(E, t)|^2 = 1\).

If the temperature is of the order of the frequency, \( k_B T \sim \hbar\omega \sim \hbar\Omega \), then the scattering amplitude can be kept constant over the entire energy interval relevant for the integration in Eq. (40). This follows from the fact mentioned already that in the adiabatic regime the scattering matrix can be kept as energy-independent on the scale of order \( \hbar\Omega \). Therefore, we can easily integrate in Eq. (40):

\[
\mathcal{P}_{\text{ex,ad}}(\omega) = \frac{\pi e^2 \hbar^2}{4\pi^2} \sum_{m=-\infty}^{\infty} |\nu_m|^2 \times \{ (m\Omega + \omega) \coth \left( \frac{m\Omega + \omega}{2k_B T/\hbar} \right) - \omega \coth \left( \frac{\omega}{2k_B T/\hbar} \right) \}. \tag{42}
\]

Here \( \nu_m \) is the \( m \)th Fourier coefficient of the frozen density of states of electrons in the cavity,

\[
\nu(E, t) = \frac{i}{2\pi} S(E, t) \frac{\partial S^*(E, t)}{\partial E}, \tag{43}
\]

evaluated at the Fermi energy, \( E = \mu \).

In the quantized emission regime, if the amplitude of the potential \( U(t) \) is chosen such that only one level crosses the Fermi energy [e.g., \( n = 0 \) in Eq. (17)], we find the Fourier coefficients of the frozen scattering matrix

\[
S_q = -2\Omega \Gamma_\tau e^{-iq\Omega\tau} e^{i\theta}, \{ e^{iq\Omega\tau}, \ q > 0, \ e^{-iq\Omega\tau}, \ q < 0 \}. \tag{44}
\]

and calculate the excess noise power,\(^ {59}\)

\[
\mathcal{P}_{\text{ex,ad}}(\omega) = \frac{2e^2}{\pi} \left( \omega\tau_D \right)^2 \left( \Omega\Gamma_\tau \right)^2 \sum_{m=-\infty}^{\infty} e^{-2|m|\Omega\Gamma_\tau} \times \{ (m\Omega + \omega) \coth \left( \frac{m\Omega + \omega}{2k_B T/\hbar} \right) - \omega \coth \left( \frac{\omega}{2k_B T/\hbar} \right) \}. \tag{45}
\]

where the dwell time \( \tau_D = \hbar/(T\Delta) \). Above equation is even in frequency, \( \mathcal{P}_{\text{ex,ad}}(\omega) = \mathcal{P}_{\text{ex,ad}}(-\omega) \), as it should be accordingly to Eq. (7) taken at \( \ell = 0 \).

Note, for the case when the Fermi level aligns with a quantum level in the cavity at \( U = 0 \) and the amplitude of the time-dependent potential is equal to the level spacing, \( eU(t) = (\Delta/2) \cos(\Omega t) \), the half-duration of an emitted current pulse \( \Gamma_\tau \sim \delta/|eU/dt| \) becomes

\[
\frac{\Gamma_\tau}{T} = \frac{T}{4\pi^2}. \tag{46}
\]

Remember \( T = 2\pi/\Omega \) is the period of the drive and \( T \ll 1 \) is the transmission probability of the QPC connecting the cavity and the electron waveguide. Note the adiabaticity condition, Eq. (19), now reads, \( T^2 \gg 4\pi^2 h\Omega/\Delta \).\(^ {43}\)

1. **Low temperatures**

At low temperatures, \( k_B T \ll \hbar\omega, h\Omega \), the hyperbolic cotangent in Eq. (45) equals to 1 with an exponential accuracy and we can sum up over \( m \),

\[
\mathcal{P}_{\text{ex,ad}}(\omega) = \frac{2e^2}{T} \left( \omega\tau_D \right)^2 e^{-2\omega|\Gamma_\tau|. \tag{47}
\]

The factor 2 in front is the number of particles (one electron and one hole) emitted during the period \( T = 2\pi/\Omega \). The electrons and holes are emitted at different times and, therefore, they contribute to noise independently.

The frequency dependence of the excess noise in the adiabatic regime is shown in Fig. 4. At small frequencies, \( \omega\Gamma_\tau \ll 1 \), the noise is quadratic in frequency,
ergy of emitted particles \( E \) is the half-width of an emitted current pulse. The mean energy for emitted particles \( \bar{E} \) is directly related to the distribution function \( f \). The temperature is zero.

\[ \omega = \frac{\hbar}{(2\Gamma_\tau)} \]

FIG. 4: (Color online) Excess noise power in the adiabatic regime, Eq. (47), in units of \( 2(e^3/T) |\tau_D/(2\Gamma_\tau)|^2 \). The frequency \( \omega \) is in units of \( 1/(2\Gamma_\tau) \). \( \tau_D \) is the dwell time. \( \Gamma_\tau \) is the half-width of an emitted current pulse. The mean energy of emitted particles \( \bar{E} = \hbar/(2\Gamma_\tau) \) defines the frequency cut-off over which the excess noise power drops quickly. The temperature is zero.

\[ \mathbf{P}_{\text{ex,ad}}^0(\omega) = 2\frac{e^2}{T} (\omega \tau_D)^2. \] (48)

At high frequencies, \( \omega \Gamma_\tau \gg 1 \), it gets suppressed.

2. High frequency cut-off

To explain the noise suppression effect we note the following. Accordingly to the numerical calculations of Ref. 34 carried out for the non-adiabatic regime, the single-particle excess noise is cut off for frequencies exceeding the energy of emitted particle (counted above the Fermi energy for emitted electrons and below the Fermi energy for emitted holes). For simplicity we consider the zero temperature case only.

To account for this effect we calculate how many electrons (on average) have an energy larger than the energy quantum \( \hbar \omega \) and, therefore, are able to emit it. We denote this number as \( \mathcal{N}_{\text{e}}(\omega) \). To calculate it we need the probability \( p_{\text{e}}^{\text{ad}}(E) \) for a single electron to be emitted with energy \( E \). These two quantities are related as follows,

\[ \mathcal{N}_{\text{e}}(\omega) = \int_{\mu + \hbar \omega}^{\infty} dE p_{\text{e}}^{\text{ad}}(E). \] (49)

Since in the quantized emission regime only a single particle can be emitted at a time, the probability \( p_{\text{e}}^{\text{ad}}(E) \) is directly related to the distribution function \( f_{\text{e}}^{\text{ad}}(E) \) of electrons scattered off the dynamic cavity, see Sec. C:

\[ p_{\text{e}}^{\text{ad}}(E) = \frac{f_{\text{e}}^{\text{ad}}(E)}{\int_0^{\infty} dE f_{\text{e}}^{\text{ad}}(E)}. \] (50)

Using the distribution function given in Eq. (C5), we calculate the probability,

\[ p_{\text{e}}^{\text{ad}}(E) = \frac{2\Gamma_\tau}{\hbar} e^{-\frac{2\Gamma_\tau}{\hbar} |E - \mu|}. \] (51)

Using this probability we can evaluate the mean energy (counted from the Fermi energy) of emitted electrons, \( \mathcal{E} = \hbar/(2\Gamma_\tau) \), which was found in Ref. 47 on the base of the dc heat carried by the emitted electrons.

Using Eq. (51) in Eq. (49) we find

\[ \mathcal{N}_{\text{e}}^{\text{ad}}(\omega) = e^{-2\Gamma_\tau |\omega|}. \] (52)

For the holes we get the same answer, \( \mathcal{N}_{\text{h}}^{\text{ad}}(\omega) = \mathcal{N}_{\text{e}}^{\text{ad}}(\omega) \).

Thus, the zero temperature result (47) can be represented as the phase noise of two emitted particles \( \mathcal{P}_{\text{ex,ad}}^0(\omega) \), Eq. (48), times by the efficiency factor \( \mathcal{N}_{\text{e}}^{\text{ad}}(\omega) \equiv \mathcal{N}_{\text{e}}(\omega) \).

B. Non-adiabatic regime

If the cavity is driven by the pulsed potential, Eq. (21), the excess noise persists up to high frequencies and we cannot use the expansion given in Eq. (38). Instead, we use directly Eqs. (8) and (9). In the continuous frequency representation, see Sec. III B 2, the excess noise power reads,

\[ \mathcal{P}_{\text{ex}}^0(\bar{\omega}) = \frac{\delta^2(\omega_m)}{2\pi} \int_{-\infty}^{\infty} d\omega_m F(\epsilon, \epsilon + \omega_m + \bar{\omega}) \]

\[ \times \left\{ \left| \Pi_m(\epsilon, \bar{\omega}) \right|^2 - \delta^2(\omega_m) \right\}, \]

\[ \Pi_m(\epsilon, \bar{\omega}) = \left( \frac{T_\mu}{h} \right)^2 \int_{-\infty}^{\infty} d\omega_q S_{\text{e}}(\epsilon + \omega_q, \epsilon) \]

\[ \times S_{\text{f}}(\epsilon + \bar{\omega} + \omega_q, \epsilon + \bar{\omega} + \omega_m). \] (53)

By analogy with Eq. (33) here we introduced the dimensionless frequency \( \bar{\omega} = \hbar \omega / \Delta \).

Note in the quantized emission regime under optimal working conditions the excess noise reaches its minimal value and is referred to as the phase noise. As we already mention the phase noise is due to the uncertainty in the single-particle emission time.\(^2,33,34\)
1. Finite temperatures

We calculate the excess noise power of the cavity working under optimal operating conditions, see Sec. III B 1, and weakly coupled to an electron waveguide, T ≪ 1. We use the scattering matrix $S_F^p$, Eq. (36), and obtain after lengthy but straightforward calculations (see Appendix B),

$$\mathcal{P}^{ex,na}(\omega) = 2 \frac{e^2}{T} \frac{\omega^2 \tau_D^2}{1 + \omega^2 \tau_D^2} \mathcal{N}_T^{na}(\omega),$$  \hspace{1cm} (54)

with $T = 2\pi/\Omega$ the period of the drive, $\tau_D$ the dwell time, and the finite-temperature (the subscript “T”) efficiency factor

$$\mathcal{N}_T^{na}(\omega) = \sum_{a=-\infty}^{\infty} \left\{ -2F(a - 0.5, a - 0.5 + \omega) + F(a - 0.5, a + 0.5 - \omega) \right\}. \hspace{1cm} (55)$$

Notice the arguments of the function $F$, Eq. (6c), in above equation are normalized accordingly to Eq. (33). Using Eq. (B27) one can check that $\mathcal{N}_T^{na}(\omega) = \mathcal{N}_T^{na}(-\omega)$ and hence the excess noise power is even in frequency, $\mathcal{P}^{ex,na}(\omega) = \mathcal{P}^{ex,na}(-\omega)$ in agreement with Eq. (7) for $\ell = 0$.

The factor $2/T$ in front of Eq. (54) accounts for the particle rate emission. The factor $e^2 \omega^2 \tau_D^2 / (1 + \omega^2 \tau_D^2)$ accounts for the noise of a single particle. And finally the efficiency factor $\mathcal{N}_T^{na}(\omega)$ accounts for the fraction of particles able to emit the energy quantum $\hbar \omega$. Equation (55) was derived for the vanishing level width, $\delta \ll \Delta, k_B T$.

The frequency dependence of the excess noise power in the non-adiabatic regime is shown in Fig. 5 for several temperatures. This agrees qualitatively with the Floquet numerical calculations present in Ref. 34. At low temperatures the noise is cut at $\omega > \Delta/(2\hbar)$. As in the adiabatic case the frequency cut-off is related to the energy of emitted particles, $\sim \Delta/2$ (counted from the Fermi energy $\mu$). Since all the states below $\mu$ are fully occupied, an electron cannot emit more energy than $\Delta/2$. However, with increasing temperature the states below the Fermi level become partially unoccupied. As a consequence, after emission an electron can jump below $\mu$ and, therefore, it can emit more energy than at zero temperature. This explains why with increasing temperature the noise persists up to higher frequencies and why the decay of the noise power at high frequencies is governed by the temperature.

2. High-frequency cut-off at zero temperature

The efficiency factor $\mathcal{N}_T^{na}(\omega)$, Eq. (55), was calculated in the limit of vanishing width of the quantum levels in the cavity, $\delta \to 0$. In fact this limit is a high temperature limit, $k_B T \gg \delta$. To estimate the efficiency factor at zero temperature we follow the same procedure as in the adiabatic regime, see Sec. IV A 2 and calculate the fraction of emitted particles $\mathcal{N}_T^{na} (\omega)$ which are able to emit the energy quantum $\hbar \omega$. For this purpose we use the distribution function for the non-adiabatic regime, Eq. (C19), and find the corresponding probability,

$$\mathcal{P}_c^{na}(E) = C \frac{\delta/\pi}{(E - \mu - \Delta/2)^2 + \delta^2}, \hspace{1cm} 0 < E - \mu < \Delta$$  \hspace{1cm} (56)

The factor $C = \pi/2(\arctan(\Delta/(2\delta))$ accounts for the normalization. Then we calculate,

$$\mathcal{N}_c^{na}(\omega) \equiv \mu + \Delta \int_{\mu + \hbar \omega}^{\mu + \Delta} dE \mathcal{P}_c^{na}(E) = \frac{1}{2} - \frac{\arctan(2\omega \tau_D - \Delta/2\hbar)}{2\arctan(\delta/2\hbar)}.\hspace{1cm} (57)$$

For the emitted holes the efficiency factor is the same. The frequency dependence of the noise power, Eq. (54) with $\mathcal{N}_T^{na}$ replaced by $\mathcal{N}_c^{na}$, is shown in Fig. 5 by the thin black solid line.

![FIG. 5: (Color online) Excess noise power in the non-adiabatic regime, Eq. (54), in units of $2(e^2/T)$. The frequency $\omega$ is in units of $\Delta/\hbar$. The noise power is given for different temperatures: $k_B T/\Delta$ = 0.05 (black solid line), 0.1 (green dashed line), and 0.15 (red dotted line). The thin black solid line is the noise power calculated with a zero-temperature efficiency factor, Eq. (57). The cut-off frequency at $\omega \sim 1/2$ is due to the energy $\sim \Delta/2$ of emitted particles. The transmission of the cavity’s QPC $T = 0.1$ and, correspondingly, the level width is $\delta = 0.008\Delta$. The dwell time $\tau_D\Delta/\hbar \approx 62.8$.](https://www.aip.org/pubservs/ajp/ajp5993/ajp5993_05_hyp.png)
V. PHOTON-INDUCED NOISE

A. Adiabatic regime

Unlike the case considered in Sec. IV A, the leading order term of the correlation function for $\ell \neq 0$ is linear in $\omega$. Therefore, to calculate the noise power, which is bi-quadratic in $\omega$ and $\Omega$, we have to keep the first order in $\Omega$ term in the expansion of the Floquet scattering matrix. Instead of Eq. (39) we now use the following,\cite{foot:57}

$$S_F(E_{p+\ell}, E) \approx S_{p+\ell}(E) + \frac{(p + \ell)h\Omega}{2} \frac{\partial S_{p+\ell}(E)}{\partial E},$$

$$S_F^p(E', E_m') \approx (S^*)_{m-p}(E) + \frac{(p + m)h\Omega + 2\hbar\omega \partial(S^*)_{m-p}(E)}{2} \frac{\partial S^{*}}{\partial E}, \quad (58)$$

where $E' = E + \hbar\omega$ and $S_n(E)$ is the Fourier coefficient of the frozen scattering amplitude $S(E,t)$. Substituting above equations into Eq. (11) we find,

$$\Pi_m^\ell(\omega) \approx \left\{ 1 - \frac{i\hbar}{2} S \frac{\partial^2 S^*}{\partial \omega \partial E} + \frac{i\hbar}{2} S^* \frac{\partial^2 S}{\partial \omega \partial E} \right\} + (\hbar\omega + m\hbar\Omega) S \frac{\partial S^*}{\partial E} \right\}_{m+\ell}. \quad (59)$$

Thus to the lowest order in frequency the PIN power, Eq. (10), reads,

$$\mathcal{P}_{\ell}^{PIN, ad}(\omega) = \frac{e^2}{2\pi} \int_0^{\infty} \frac{d\omega}{\omega} \left\{ S^* \left[ \frac{\partial S}{\partial E} \right] \right\}_{\ell} \times \{ F(E, E-\ell + \hbar\omega) \omega + F(E, E + \hbar\omega) (\ell\Omega - \omega) \}. \quad (60)$$

Note in the equation above we used the identity

$$\frac{i\hbar}{2} \left\{ S^* \frac{\partial^2 S}{\partial \omega \partial E} - S \frac{\partial^2 S^*}{\partial \omega \partial E} \right\}_{\ell} = \ell\hbar\Omega \left\{ S^* \frac{\partial S}{\partial E} \right\}_{\ell}, \quad (61)$$

following from the unitarity of the frozen scattering amplitude, $|S(E,t)|^2 = 1$.

At low temperatures, $k_B T \ll \hbar\omega, \hbar\Omega$, we put $S(E,t) \approx S(\mu,t)$ and integrate over energy in Eq. (60). In the regime when the capacitor works as a single-particle source we use the scattering amplitude given in Eq. (17) (with one level of the cavity crossing the Fermi level) and calculate the PIN power as follows,

$$\mathcal{P}_{\ell}^{PIN, ad}(\omega) = C_\ell \frac{\epsilon^2}{T} \Omega^2 \Gamma_T e^{-|\ell|\Omega} \xi_\ell (\frac{\omega}{\Omega}),$$

$$\xi_\ell(x) = x|x - \ell| + (\ell - x)|x|, \quad C_\ell = -i \{ e^{i\ell\Omega}\tau - e^{i\ell\Omega}\tau} \} \quad (62)$$

Remember, $t_-$ ($t_+$) is the time of an electron (hole) emission, $\tau_D$ is the dwell time, and $\Gamma_\tau$ is the half-width of the current pulse corresponding to an emitted particle, an electron or a hole. If electrons and holes are emitted equally distributed in time, $t_+ = t_- + \tau/2$, then,

$$C_\ell = -i e^{i\ell\Omega}\tau - \{ 1 + (-1)^\ell \} \quad (63)$$

Thus the PIN power is zero for odd $\ell$. Under the same conditions the Fourier harmonics of a current,

$$I_\ell = \int_0^\tau dt e^{-i\ell\Omega} I(t), \quad (64)$$

are zero for even $\ell$. For instance, for the adiabatic current $I^{ad}(t)$, Eq. (18), we have

$$I_\ell^{ad} = \frac{e}{T} e^{-i\ell\Omega}\tau e^{i\ell\Omega}\tau - \{ 1 + (-1)^\ell \} \quad (65)$$

The even-odd alternation for the current Fourier harmonics demonstrates clearly the presence of two carriers having opposite charge, while the even-odd alternation for the PIN power is a clear signature of statistical independence of emitted carriers and, therefore, is a signature of the quantized emission regime with no spurious electron-hole pairs emitted.

The PIN power $\mathcal{P}_{\ell}^{PIN, ad}$, see Eqs. (62) and (63), is shown in Fig. 6. It differs essentially from the excess noise power $\mathcal{P}_{\ell}^{(ex, ad)}$, see Eq. (47) and Fig. 4. First, the PIN power is larger by the factor of $\Gamma_\tau/\tau_D \sim (\Delta/\hbar\Omega)(\bar{T}/\bar{T}_{\pi})^2 \gg 1$. Second, unlike the phase noise power, the PIN power is independent of the transparency $T$ of the QPC connecting the cavity and the electron waveguide provided that the adiabaticity condition, $\Gamma_\tau \gg \tau_D$, holds, see Eq. (19). It’s maximum value is $\mathcal{P}_{\ell,max}^{PIN, ad} \sim (e^2/T)(\hbar\Omega/\Delta)^2/2$. And, third, the PIN power is highly asymmetric in frequency: It exists within the finite frequency window, which is $0 < \omega < \Omega$ for positive $\ell$, see Fig. 6, and $0 > \omega > -|\ell|\Omega$ for negative $\ell$. Within its range of existence the PIN power is quadratic in frequency. The function $\xi_\ell (\frac{\omega}{\Omega})$, defining the frequency dependent of the PIN power, reads,

$$\xi_\ell (\frac{\omega}{\Omega}) = \begin{cases} 2\Omega^2 (\ell\Omega - \omega); & \ell > 0, \ 0 < \omega < \ell\Omega, \\ 2\Omega^2 (\omega - \ell\Omega); & \ell < 0, \ -|\ell|\Omega < \omega < 0. \end{cases} \quad (66)$$
B. Non-adiabatic regime

For the potential $U(t)$, Eq. (21), driving the cavity non-adiabatically, we use the continuous frequency representation, see Sec. III B 2, and calculate the PIN power, Eq. (10), as follows,

$$
P^\text{PIN,na}_\ell(\omega) = \frac{e^2}{2\pi} \int \frac{de}{\omega_{m}} \int d\omega_{q} F(\epsilon, \omega_{m} + \omega_{q})$$

$$\times \Pi_{m,\ell}^a(\epsilon, \omega_{q}) \Pi_{m,\ell}^{a\dagger}(\epsilon, \omega_{q}),$$

where

$$\Pi_{m,\ell}^a(\epsilon, \omega_{q}) = \left(\frac{\Delta}{\hbar \Omega}\right)^2 \int d\omega_{q} S_{\ell}(\epsilon, \omega_{q}, \omega_{m}, \epsilon)$$

$$\times S_{\ell}^{\dagger}(\epsilon, \omega_{q}, \omega_{m}, \epsilon),$$

where $\omega_{m} = \ell \hbar \Omega / \Delta$.

To calculate above equations we use the Floquet scattering matrix $S_{\ell}^{opt}$, Eq. (36). Along the lines of Appendix B we calculate,

$$P^\text{PIN,na}_\ell(\omega) = \frac{e^2}{2\pi} e^{i \pi \omega_{m} / \hbar \Omega} \sin \left(\frac{\pi \Omega T}{\Delta}\right) \frac{\pi \Omega}{1 + (\ell \Omega / \Delta)^2}$$

$$\times \frac{1}{(1 - i \ell \Omega D)(1 + i \ell \Omega D)} \left\{ e^{i \ell \Omega D} N_{\ell,-}(\omega) + e^{i \ell \Omega D} N_{\ell,+}(\omega) \right\},$$

where

$$N_{\ell,\mp}(\omega) = \sum_{a = -\infty}^{\infty} \left\{ -F(a - 0.5, a - 0.5 + \omega) - F(a - 0.5, a - 0.5 - \omega + \Omega T) \right\}$$

$$+ \frac{1}{2} (1 - \omega_{\ell}) \left[ F(a - 0.5, a + 0.5 + \omega) + F(a - 0.5, a + 0.5 - \omega + \Omega T) \right]$$

$$+ \frac{1}{2} (1 + \omega_{\ell}) \left[ F(a - 0.5, a + 0.5 - \omega) + F(a - 0.5, a + 0.5 + \omega - \Omega T) \right].$$

If electrons and holes are emitted at regular intervals, $t_{\pm} = t_{\mp} + T / 2$, then the PIN power, Eq. (69), is non-zero for even $\ell = 2\lambda$ only, while the current, Eq. (37), has only odd Fourier harmonics,

$$I_{\ell}^\text{na} = \frac{e}{T} \frac{e^{i \ell \Omega D} \left(1 - (-1)^{\ell}\right)}{1 - i \ell \Omega D}. \quad (71)$$

The magnitude of $P^\text{PIN,na}_\ell(\omega)$ is of the same order as the magnitude of the phase noise power $P^\text{ex,na}_\ell(\omega)$, Eq. (54). This is easy to verify, since at $\ell = 0$ we have $N_{\ell,-}(\omega) = N_{\ell,+}(\omega) = N_{0}^\text{na}(\omega)$, see Eq. (55), and the equation (69) becomes Eq. (54). However, in contrast to the phase noise power, the PIN power is asymmetric in frequency. To characterize this asymmetry we calculate the difference $\delta P_{2\lambda}$ between the PIN power for some $2\lambda \neq 0$ and the phase noise power. For small $\ell$, such that $\ell \hbar \Omega / \Delta \ll 1$, we obtain the following: (for simplicity the origin of time is taken at the time of an electron emission, $t_{-} = 0$)

$$\delta P_{2\lambda}(\omega) \equiv P^\text{PIN,na}_\ell(\omega) - P^\text{ex,na}_\ell(\omega) \approx \frac{e^2}{2\pi} \left\{ \frac{\omega(\omega - 2\lambda \Omega)^2 T_{D}^2}{1 + \omega(\omega - 2\lambda \Omega)^2 T_{D}} - \frac{\omega^2 T_{D}^2}{1 + \omega^2 T_{D}} \right\}. \quad (72)$$

This function is plotted down in Fig. 7 for several positive $\lambda$. Its behavior for negative $\lambda$ can be anticipated on the base of the symmetry properties of the noise power, Eq. (7).
VI. CONCLUSION

Within the Floquet scattering matrix formalism I addressed the finite-frequency noise spectrum of a quantum capacitor working as a single-particle emitter. I considered two working regimes, adiabatic and non-adiabatic. In the former/later case the potential driving the capacitor varies slow/fast on the scale of internal dynamics of the capacitor, which is characterized by the dwell time $\tau_D$, the time necessary for an electron to tunnel from the capacitor into the waveguide of vice versa. I calculated analytically the correlation function in the adiabatic regime and in the non-adiabatic regime under optimal operating conditions, see Sec. III B 1.

The current correlation function depends on two frequencies, $\omega_1$ and $\omega_2$. In the stationary case it is non-zero for $\omega_1 = -\omega_2$ only. If the system (in our case the capacitor) is driven with frequency $\Omega$, then the correlation function becomes non-zero also for $\omega_1 + \omega_2 = \ell \Omega$, where $\ell$ is an integer. Therefore, the periodic drive manifests itself, first, by correlating currents at frequencies shifted by one or few frequencies of the drive and, second, by the modification of the correlation function for $\omega_1 = -\omega_2$. The correlation function for $\ell \neq 0$ is referred to as the photon-induced noise power, since for the currents at $\omega_1$ and $\omega_2 = -\omega_1 + \ell \Omega$ to become correlated the electron system has to exchange $\ell$ energy quantum $\hbar \Omega$ with the external potential driving the capacitor. The difference between the correlation function for $\ell = 0$ and the stationary one is referred to as the excess noise, since the drive generally increases the noise.

The excess noise in the non-adiabatic regime was recently measured. In the quantized emission regime the excess noise reaches its minimal value (referred to as the

phase noise), which is due to the quantum uncertainty in the emission time. I calculated the excess noise power analytically in the non-adiabatic regime and found it to be suppressed at frequencies (times $\hbar$) exceeding the energy, $\sim \Delta/2$, of emitted particles, see Fig. 5. This is in agreement with the results of numerical calculations of Ref. 34. I predicted similar properties for the excess noise in the adiabatic regime with the trivial difference due to smaller characteristic energy, $E = \hbar/(2\Gamma)$, of emitted particles, see Fig. 4.

In contrast the photon-induced noise (PIN) shows striking difference for two discussed regimes. In the adiabatic regime the PIN power exists within the finite frequency window, see Fig. 6, and its magnitude far exceeds the excess noise power, compare Eqs. (62) and (47). While in the non-adiabatic regime the PIN power deviates from the phase noise power only a little, compare Eqs. (54) and (69). At the frequencies where it exists, the adiabatic PIN power is much larger than the non-adiabatic one. The last observation has a simple explanation. The PIN involves an energy exchange between the electron system and the driving time-periodic potential. This exchange is described by the Floquet scattering amplitude $S_F(E + \hbar \Omega, E)$. In the adiabatic case $S_F^{\ell \rightarrow \Gamma_\tau} = \Gamma_\tau/\mathcal{T}$ provided that $\ell \bar{\Omega} \Gamma_\tau \ll 1$, see Eq. (44). Here $\Gamma_\tau$ is the (half-)width of the current pulse corresponding to an adiabatically emitted electron and $\mathcal{T} = 2\pi/\Omega$ is the period of a drive. In the non-adiabatic case $S_F^{\ell \rightarrow \Gamma_\tau} = \bar{\hbar} \bar{\Omega} / \mathcal{D}$ provided that both $\hbar \bar{\Omega} / \mathcal{D}$ and the level in the dot is not aligned with energy $E$, see Eq. (29). Remember that the dwell time $\tau_D$ defines also the width of the current pulse corresponding to an electron emitted non-adiabatically. The ratio of the scattering amplitudes, $S_F^{\ell \rightarrow \Gamma_\tau}/S_F^{\ell \rightarrow \Gamma_\tau} = \Gamma_\tau/\tau_D \gg 1$, defines the ratio between the adiabatic PIN power, $P_{\ell}^{\text{PIN,ad}} = \omega^2 \tau_D \Gamma_\tau$, see Eq. (62) and the non-adiabatic one, $P_{\ell}^{\text{PIN,na}} = \omega^2 \tau_D \Gamma_\tau$, see Eq. (69). Therefore, the PIN power provides a direct access to the Floquet scattering amplitudes.

I also found that under optimal operating conditions the PIN power is zero for odd $\ell$ and is non-zero for even $\ell$. This parity-dependent behavior if being observed would indicate clearly the statistical independence of emitted electrons and holes and, therefore, would verify the quantized emission regime with no spurious electron-hole pairs emitted. Some advantage of the measurement of PIN compared to the measurement of the excess noise is that no subtracting procedures is needed, since the PIN is absent in the stationary case.

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Appendix A: Unitarity of the Floquet scattering matrix for the periodic step potential

I give the proof of unitarity since it is helpful for subsequent calculations of the noise power. Here I follow close to Ref. 37.

The unitarity condition for a single orbital channel case reads:

$$\sum_{p,m,n} S^*_p(E_p,E_m)S_F(E_p,E_n) = \delta_{m,n}, \quad (A1)$$

where $p$, $m$, and $n$ all are integer. In the continuous frequency representation, see Sec. III B 2, we rewrite above identity as follows,

$$\int d\omega_p S^*_p(\epsilon + \omega_p, \epsilon + \omega_m) S_F(\epsilon + \omega_p, \epsilon + \omega_n) = \left(\frac{\hbar \Omega}{\Delta}\right)^2 \delta(\omega_m - \omega_n). \quad (A2)$$

Here we use dimensionless quantities accordingly to Eq. (33).

For the scattering matrix given in Eq. (36) we find

$$\int_{-\infty}^{\infty} d\omega_p \frac{\sin(\pi[\omega_p - \omega_m]) \sin(\pi[\omega_p - \omega_n])}{\pi^2[\omega_p - \omega_m][\omega_p - \omega_n]} e^{i\pi[\omega_n - \omega_m]} \times \left\{ \delta(\omega_p - \omega_m) - \frac{e^{-i2\pi[\omega_p - \omega_m]}}{1+i^{n-1}} + \frac{e^{-i2\pi[\omega_p - \omega_m]}}{1+i^{n+1}} \right\}$$

$$\times \left\{ \delta(\omega_p - \omega_n) - \frac{e^{i2\pi[\omega_p - \omega_m]}}{1+i^{n-1}} + \frac{e^{i2\pi[\omega_p - \omega_m]}}{1+i^{n+1}} \right\}$$

$$= \delta(\omega_m - \omega_n). \quad (A3)$$

Here we took into account that $|S|^2 = 1$. Next we open the curly brackets and integrate over $\omega_p$ using the Dirac delta function,

$$\int_{-\infty}^{\infty} d\omega_p \frac{\sin(\pi[\omega_p - \omega_m]) \sin(\pi[\omega_p - \omega_n])}{|\rho(\epsilon + \omega_p)|^2} \frac{1}{\pi^2[\omega_p - \omega_m][\omega_p - \omega_n]} \times \left\{ \xi_p + \frac{e^{-i2\pi[\omega_n - \omega_m]}}{1+i^{n-1}} \right\}$$

$$+ \frac{e^{-i2\pi[\omega_n - \omega_m]}}{1+i^{n+1}} = \frac{2}{\pi} \sin(\pi[\omega_n - \omega_m]) \left| \frac{\omega_p - \omega_m}{\omega_p - \omega_n} \right| \frac{\omega_p - \omega_m}{\omega_p - \omega_n} \times \frac{1}{1 + \frac{\omega_p - \omega_m}{\omega_p - \omega_n}} \right\}, \quad (A4)$$

with

$$\xi_p = -e^{-i2\pi[\omega_p - \omega_m]} \frac{1}{\omega_p - \omega_m} + e^{i2\pi(\omega_p - \omega_m)} \frac{1}{\omega_p - \omega_m}.$$
proven by expanding the left hand side into the simple fractions and then using the following text-book sum

\[
\sum_{a=-\infty}^{\infty} \frac{1}{a + \gamma} = \pi \cot(\pi \gamma), \quad (A11)
\]
taken at proper \(\gamma\)’s. As a result we find

\[
\Sigma_q = \frac{2 \sin(\pi \omega_q)}{\pi \omega_q (1 - \omega_q^2)}. \quad (A12)
\]

Since \(\Sigma_q\) is independent of \(\omega_q\), we can integrate in Eq. (A8). With \(\rho\) given in Eq. (35) the integral is one. Thus the identity (A8) and hence the unitarity condition, Eq. (A2), is proven.

Alternatively one can use the following unitarity condition, \(^{60}\)

\[
\sum_{p=-\infty}^{\infty} S_F^*(E_m, E_p) S_F(E_n, E_p) = \delta_{m,n}, \quad (A13)
\]

which in the continuous frequency representation reads,

\[
\int d\omega_p S_F^*(\epsilon + \omega_m, \epsilon + \omega_p) S_F(\epsilon + \omega_n, \epsilon + \omega_p) = \left( \frac{\hbar \Omega}{\Delta} \right)^2 \delta(\omega_m - \omega_n). \quad (A14)
\]

Here we use normalized quantities introduced in Eq. (33). We will use Eq. (A14) below, see Sec. C2.

**Appendix B: Excess noise for the periodic step potential**

We calculate the excess noise, Eq. (53), of the SES working under optimal operating conditions. The Floquet scattering matrix elements necessary for calculations are following, see Sec. III B 1:

\[
S_F^{opt}(\epsilon + \omega, \epsilon) = S(\epsilon + \omega) \frac{\hbar \Omega \sin(\pi \omega_q)}{\pi \omega_q} e^{-i \pi \omega q} \times \left\{ \frac{\exp(i2\pi \omega_q \frac{t}{1-\omega_q}) + \exp(i2\pi \omega_q \frac{t}{1+\omega_q})}{\rho(\epsilon + \omega_q) \rho(\epsilon)} \delta(\omega_q) - \frac{\exp(2\pi i \omega \frac{t}{1+\omega}) + \exp(2\pi i \omega \frac{t}{1-\omega})}{(1+\omega) \rho(\epsilon + \omega) \rho(\epsilon)} \right\}, \quad (B1)
\]

where we use the short notation \(\rho(\epsilon)\) introduced in Eq. (35) and rewrite the stationary scattering amplitude, Eq. (30), as follows:

\[
S(\epsilon) = e^{i(2\pi + \theta)} \frac{\rho^*(\epsilon)}{\rho(\epsilon)}. \quad (B3)
\]

Here we took into account that under optimal conditions the phase is, \(\phi^{opt}(\epsilon) = \pi + 2\pi \epsilon\), see Eq. (27).

Using Eqs. (B1) and (B2) we calculate

\[
\Pi_m(\epsilon, \omega) = e^{-i\pi \omega_m} \int_{-\infty}^{\infty} d\omega_q S(\epsilon + \omega_q) S^*(\epsilon + \omega + \omega_q) \times \left\{ \frac{\sin(\pi \omega_q)}{\pi \omega_q} \sin(\pi [\omega_m - \omega_q]) \delta(\omega_q) - \frac{\exp(2\pi i \omega \frac{t}{1+\omega}) + \exp(2\pi i \omega \frac{t}{1-\omega})}{(1+\omega) \rho(\epsilon + \omega) \rho(\epsilon)} \right\}, \quad (B4)
\]

After opening the brackets we use Eq. (B3) and get

\[
\Pi_m(\epsilon, \omega) = e^{-i\pi \omega_m} \sum_{j=1}^{3} A_j \delta(\omega_m) + \frac{\sum_{j=1}^{3} A_j}{\rho(\epsilon + \omega) \rho^*(\epsilon + \omega + \omega_m)} \quad (B5)
\]

\[
A_1 = - \frac{\sin(\pi \omega_m)}{\pi \omega_m} \left\{ \frac{e^{i2\pi \omega \frac{t}{1+\omega}}}{(1+\omega_m)} + \frac{e^{i2\pi \omega \frac{t}{1-\omega}}}{(1-\omega_m)} \right\}, \quad (B6)
\]

\[
A_2 = - \frac{\sin(\pi \omega_m)}{\pi \omega_m} \left\{ \frac{e^{i2\pi \omega \frac{t}{1-\omega}}}{(1-\omega_m)} + \frac{e^{i2\pi \omega \frac{t}{1+\omega}}}{(1+\omega_m)} \right\}, \quad (B7)
\]
\[ A_3 = \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon)} \int_{-\infty}^{\infty} \frac{d\omega_q}{\rho(\epsilon + \omega_q)\rho^*(\epsilon + \bar{\omega} + \omega_q)} \times \frac{\sin(\pi \omega_q)}{\pi \omega_q} \sin(\pi [\omega_q - \omega_m]) \times \{ \zeta_q + B_1 e^{i2\pi\omega_m} \frac{1}{\bar{\omega} + B_2 e^{i2\pi\omega_m} \frac{1}{\bar{\omega}}} \}, \] (B8)

where

\[ B_1 = \frac{1}{(\omega_q - 1)(\omega_q - \omega_m - 1)}, \]
\[ B_2 = \frac{1}{(\omega_q + 1)(\omega_q - \omega_m + 1)}. \] (B9)

The quantity \( \zeta_q \) contains a factor oscillating fast in \( \omega_q \) and, therefore, under the condition set out in Eq. (14) \( \zeta_q \) vanishes upon integration over \( \omega_q \).

1. **Zero-frequency limit**

At \( \omega = 0 \) we calculate \( A_3 \), Eq. (B8), in the same way as we calculated Eq. (A4): We use Eq. (A6) with the following modifications: \( \omega_p \to \omega_q, \omega_n = 0 \). In addition we change the sign of \( \omega_m \). The result is, \( A_3 = -(A_1 + A_2) \). Then we use Eq. (B5) and find the noise power, Eq. (53), to be zero as it is expected.

2. **Finite-frequency noise**

In the small transmission limit, \( \bar{T} \ll 1 \), we can calculate \( A_3 \), Eq. (B8), analytically. For this purpose we represent \( 1/\rho(\epsilon) \), Eq. (35), as the sum of the Breit-Wigner resonances. For optimal operating conditions, Eq. (27), it reads,

\[ \frac{1}{\rho(x)} = \sum_{a=-\infty}^{\infty} \frac{i\sqrt{g/\pi}}{x + 0.5 - a + ig}, \] (B10)

where we introduce a short notion \( g \equiv \bar{\delta}/\Delta = \bar{T}/(4\pi). \)

Since the width \( \delta \) of peaks is small and the adjacent peaks (separated by the level spacing \( \Delta \)) do not overlap, we can keep only the pair products of closest peaks in the product of \( \rho \)-functions entering the integral in Eq. (B8):

\[ \frac{1}{\rho(\epsilon + \omega_q)\rho^*(\epsilon + \bar{\omega} + \omega_q)} \approx \sum_{a=-\infty}^{\infty} \frac{g/\pi}{(\eta - a + ig)(\eta + \bar{\omega} - a - ig)}, \] (B11)

where we use a short notation \( \eta = \epsilon + \omega_q + 0.5 \). Notice for frequencies close to \( \Delta \) we have to re-pair peaks. This can be done with the help of the shift \( \bar{\omega} \to \bar{\omega} - 1 \).

Furthermore we make a shift, \( \omega_q \to \omega_q - \epsilon \pm 0.5 + a \) (the upper/lower sign is used in the term with \( B_1/B_2 \)), and get

\[ A_3 = \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon)} \int_{-\infty}^{\infty} \frac{d\omega_q}{\rho(\epsilon + \omega_q)\rho^*(\epsilon + \bar{\omega} + \omega_q)} \times \frac{\sin(\pi \omega_q)}{\pi \omega_q} \sin(\pi [\omega_q - \omega_m]) \times \{ \zeta_q + B_1 e^{i2\pi\omega_m} \frac{1}{\bar{\omega} + B_2 e^{i2\pi\omega_m} \frac{1}{\bar{\omega}}} \}, \] (B8)

where

\[ B_1 = \frac{1}{(\omega_q - 1)(\omega_q - \omega_m - 1)}, \]
\[ B_2 = \frac{1}{(\omega_q + 1)(\omega_q - \omega_m + 1)}. \] (B9)

The integration in Eq. (B12) becomes trivial and we obtain

\[ A_3 = \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon)} \frac{2\sin(\pi \omega_m)}{\pi \omega_m (1 - \omega_m^2)} e^{i2\pi\omega_m} \frac{1}{\bar{\omega}} + e^{i2\pi\omega_m} \frac{1}{\bar{\omega}}, \] (B14)

where \( \varpi = \bar{\omega}/(2g) \equiv \varpi T_D \).

Substituting all \( A_j \)'s into Eq. (B5) we find

\[ \Pi_m(\epsilon, \bar{\omega}) = e^{-i\pi\omega_m} S(\epsilon) S^*(\epsilon + \bar{\omega}) \times \left\{ \delta(\omega_m) + \frac{\sin(\pi \omega_m)}{\pi \omega_m (1 - \omega_m^2)} \left( C_1 e^{i2\pi\omega_m} \frac{1}{\bar{\omega}} + C_2 e^{i2\pi\omega_m} \frac{1}{\bar{\omega}} \right) \right\}, \] (B15)

with

\[ C_1 = \frac{2}{1 + i \varpi} \rho(\epsilon + \bar{\omega}) \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon + \bar{\omega} + \omega_m)} \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon + \bar{\omega} + \omega_m)} - \frac{\rho^*(\epsilon + \bar{\omega})}{\rho(\epsilon + \omega_m)} \frac{\rho^*(\epsilon + \bar{\omega} + \omega_m)}{\rho^*(\epsilon + \bar{\omega} + \omega_m)} \] (B16)

\[ C_2 = \frac{2}{1 + i \varpi} \rho(\epsilon + \bar{\omega}) \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon + \bar{\omega} + \omega_m)} \frac{\rho^*(\epsilon + \bar{\omega})}{\rho^*(\epsilon + \bar{\omega} + \omega_m)} - \frac{\rho^*(\epsilon + \bar{\omega})}{\rho(\epsilon + \omega_m)} \] (B17)
As the next step we square $\Pi_m(\epsilon, \bar{\omega})$, Eq. (B15), and calculate

$$|\Pi_m(\epsilon, \bar{\omega})|^2 = \delta^2(\omega_m) + 2\delta(\omega_m) \text{Re}[C_1 + C_2] + \frac{\sin^2(\pi \omega_m)}{\pi^2 \omega_m^2 (1 - \omega_m^2)^2} + \zeta_m \cdot (B18)$$

Here $\zeta_m$ is oscillating fast in $\omega_m$ and thus it does not contribute to the noise power, Eq. (53).

Next we calculate the necessary quantities,

$$|C_1|^2 = \frac{(\omega_m + 1)^2}{|\rho(\epsilon)|^2|\rho(\epsilon + \omega_m)|^2} \frac{1}{(\omega_m - 1)^2} \frac{1}{|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} + \frac{4}{1 + \omega^2} \frac{1}{|\rho(\epsilon)|^2|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} - 2\text{Re} \frac{\rho(\epsilon) \rho^*(\epsilon + \omega_m) \rho^*(\epsilon + \bar{\omega}) \rho(\epsilon + \bar{\omega} + \omega_m)}{2(\omega_m + 1)}$$

(B19)

$$|C_2|^2 = \frac{(\omega_m - 1)^2}{|\rho(\epsilon)|^2|\rho(\epsilon + \omega_m)|^2} \frac{1}{(\omega_m + 1)^2} \frac{1}{|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} + \frac{4}{1 + \omega^2} \frac{1}{|\rho(\epsilon)|^2|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} - 2\text{Re} \frac{\rho(\epsilon) \rho^*(\epsilon + \omega_m) \rho^*(\epsilon + \bar{\omega}) \rho(\epsilon + \bar{\omega} + \omega_m)}{2(\omega_m + 1)}$$

We sum up these equations

$$|C_1|^2 + |C_2|^2 = \frac{2(\omega_m^2 + 1)}{|\rho(\epsilon)|^2|\rho(\epsilon + \omega_m)|^2} \frac{1}{(\omega_m + 1)^2} \frac{1}{|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} + 8 \frac{1 + \omega^2}{|\rho(\epsilon)|^2|\rho(\epsilon + \bar{\omega} + \omega_m)|^2} - 4\text{Re} \frac{\rho(\epsilon) \rho^*(\epsilon + \bar{\omega}) \rho^*(\epsilon + \omega_m) \rho(\epsilon + \bar{\omega} + \omega_m)}{1 - i\omega\tau_D}$$

(B21)

Also we need the following

$$\delta(\omega_m) \text{Re}[C_1 + C_2] = \text{Re} \left[ \frac{4}{1 + i\omega\rho(\epsilon + \bar{\omega}) \rho^*(\epsilon)} \frac{1}{|\rho(\epsilon)|^2} \frac{2}{|\rho(\epsilon + \omega_m)|^2} \delta(\omega_m) \right] \cdot (B22)$$

We use Eqs. (B21) and (B22) in Eq. (B18) and represent the noise power, Eq. (53), as the sum of seven terms, $\mathcal{P}^{ex} = \sum_{j=1}^{7} \mathcal{P}_j(\omega)$. The first six terms originate from six terms in Eq. (B21) and the seventh one results from Eq. (B22). In the first three terms we integrate over both $\epsilon$ and $\omega_m$ using the series of the Breit-Wigner peaks, see, Eq. (B10). Ignoring small terms of order $g^2 = T/(4\pi) \ll 1$ we treat these peaks as delta-function peaks,

$$\frac{1}{|\rho(\epsilon)|^2} \approx \sum_{a=-\infty}^{\infty} \delta(x + 0.5 - a), \quad (B23)$$

and calculate,

$$\mathcal{P}_1(\bar{\omega}) = \frac{\alpha^2 \Omega}{2\pi} \sum_{a=-\infty}^{\infty} \{2F(a - 0.5, a - 0.5 + \bar{\omega}) + F(a - 0.5, a + 0.5 + \bar{\omega}) + F(a + 0.5, a - 0.5 + \bar{\omega}) + F(a + 0.5, a + 0.5 - \bar{\omega}) - F(a - 0.5, a - 0.5 - \bar{\omega}) \} \cdot \quad (B24)$$

$$\mathcal{P}_2(\bar{\omega}) = \frac{\alpha^2 \Omega}{2\pi} \sum_{a=-\infty}^{\infty} \{2F(a - 0.5, a - 0.5 + \bar{\omega}) + F(a - 0.5, a + 0.5 + \bar{\omega}) + F(a + 0.5, a - 0.5 + \bar{\omega}) + F(a + 0.5, a + 0.5 - \bar{\omega}) \} \cdot \quad (B25)$$
\begin{align}
\mathcal{P}_3(\omega) &= \frac{e^2 \Omega}{2 \pi} \frac{8}{1 + \omega^2} \sum_{a=-\infty}^{\infty} \sum_{b=-\infty}^{\infty} \sin^2(\pi \omega) \\
F(a-0.5, a+b-0.5) \frac{\sin^2(\pi \omega)}{\pi^2 |b - \omega|^2 (1 - |b - \omega|^2)^2}. \\
\text{(B26)}
\end{align}

Notice in Eqs. (B24) and (B25) we keep only the terms corresponding to \( \omega_m = 0, \pm 1 \), since others vanish due to the factor \( \sin^2(\pi \omega_m) \), see Eq. (B18). In the term with \( \omega_m = -1 \) we made a shift \( a \rightarrow a + 1 \) in the infinite sum over \( a \). In addition we used the following symmetries [see Eqs. (6c) and (34)]:

\[ F(\epsilon_1, \epsilon_2) = F(\epsilon_2, \epsilon_1), \quad F(\epsilon_1, \epsilon_2) = F(-\epsilon_1, -\epsilon_2). \quad \text{(B27)} \]

To calculate the remaining contributions we use the expansions similar to Eq. (B11) and proceed as follows:

\begin{align}
\int \frac{dk \Psi(k)}{\rho^*(k) \rho(k + \omega)} &= \int \frac{dx g/\pi}{(x - ig)(x + \omega + ig)} \\
& \times \sum_{a=-\infty}^{\infty} \frac{\Psi(a-0.5) + \Psi(a-0.5 - \omega)}{2} \\
&= \frac{1}{2} \sum_{a=-\infty}^{\infty} \frac{\Psi(a-0.5) + \Psi(a-0.5 - \omega)}{1 - i \omega}, \quad \text{(B28)}
\end{align}

where \( \Psi(k) \) is a function varying slow on the scale of \( g \ll 1 \). Notice the poles of \( \rho^* \) are at \( \epsilon^p = a-0.5 + ig \) while those of \( \rho \) are at \( \epsilon^p = a-0.5 - \omega - ig \). In the argument of \( \Psi \) we ignore the term \( x \sim g \) compared to others. This allowed us to integrate over \( x \).

By analogy we use,

\begin{align}
\sum_{a=-\infty}^{\infty} \sum_{b=-\infty}^{\infty} \frac{d\epsilon d\omega_m \Psi(\epsilon, \omega_m)}{\rho(\epsilon) \rho^*(\epsilon + \omega) \rho(\epsilon + \omega + \omega_m)} &= \int \frac{dx g/\pi}{(x + ig)(x + \omega - ig)} \frac{dy g/\pi}{(y - ig)(y + \omega + ig)} \\
& \times \sum_{a=-\infty}^{\infty} \sum_{b=-\infty}^{\infty} \frac{\Psi(a-0.5, b-a) + \Psi(a-0.5 - \omega, b-a)}{2 (1 + i \omega)(1 - i \omega)} \\
&= \sum_{a=-\infty}^{\infty} \sum_{b=-\infty}^{\infty} \frac{\Psi(a-0.5, b-a) + \Psi(a-0.5 - \omega, b-a)}{2 (1 + i \omega)(1 - i \omega)}. \quad \text{(B29)}
\end{align}

Notice, we calculate \( \Psi(\epsilon, \omega_m) \) at the poles of \( 1/\rho(\epsilon) \rho^*(\epsilon + \omega_m) \) lying at \( \epsilon^p = a-0.5 - ig \) and \( \omega_m^p = b-a+2ig \), and at the poles of \( 1/\rho^*(\epsilon + \omega) \rho(\epsilon + \omega + \omega_m) \) lying at \( \epsilon^p = a-0.5 - \omega + ig \) and \( \omega_m^p = b-a-2ig \).

Thus the remaining contributions are following,

\[ \mathcal{P}_4(\omega) = \frac{e^2 \Omega}{2 \pi} \frac{4}{1 + \omega^2} \sum_{a=-\infty}^{\infty} F(a-0.5, a-0.5 + \omega), \quad \text{(B30)} \]

\[ \mathcal{P}_5(\omega) = \mathcal{P}_6(\omega) = -\frac{e^2 \Omega}{2 \pi} \frac{1}{1 + \omega^2} \sum_{a=-\infty}^{\infty} \left\{ 4F(a-0.5, a-0.5 + \omega) + F(a-0.5, a+0.5 - \omega) + F(a+0.5, a-0.5 + \omega) + \sum_{b=-\infty}^{\infty} \frac{4 \sin^2(\pi \omega) F(a-0.5, a+b-0.5)}{\pi^2 |b - \omega|^2 (1 - |b - \omega|^2)^2} \right\}. \quad \text{(B31)} \]

Note that the double sums in \( \mathcal{P}_5 \) and \( \mathcal{P}_6 \) cancel out the contribution \( \mathcal{P}_3 \).

And finally the contribution due to Eq. (B22) reads,

\[ \mathcal{P}_7(\omega) = \frac{e^2 \Omega}{2 \pi} \sum_{a=-\infty}^{\infty} \left\{ 8F(a-0.5, a-0.5 + \omega) - 4F(a-0.5, a-0.5 - \omega) - 4F(a-0.5, a+0.5 + \omega) \right\}, \quad \text{(B32)} \]

Summing up all the contributions and taking into account that \( \overline{\omega} = \omega \overline{\tau}_D \), we arrive at the equation (54).

**Appendix C: Distribution function for single-particle excitations**

The distribution function for electrons, which are scattered off the dynamic single-channel sample, is calculated as follows,

\[ f_{\text{out}}(E) = \sum_{q=-\infty}^{\infty} |S_F(E, E_q)|^2 f(E_q), \quad \text{(C1)} \]

where \( f(E) \) is the Fermi distribution function for electrons incoming from the reservoir. For zero temperature the Fermi function is \( f(E) = \theta(\mu - E) \) and Eq. (C1) becomes

\[ f_{\text{out}}(E) = \sum_{q=-\infty}^{\infty} |S_F(E, E_q)|^2. \quad \text{(C2)} \]

Here \( [X] \) stands for the integer part of \( X \).

The dynamic scatterer excites an electron system and makes it non-equilibrium even if before scattering electrons were in equilibrium. The basic process leading to this non-equilibration is an energy exchange between the dynamic scatterer and electrons. Generally electrons absorb energy from the scatterer. As a result some unoccupied in equilibrium states become occupied and vice
versa. The non-equilibrium excitations with $E > \mu$ are referred to as quasi-electrons and the ones with $E < \mu$ are referred to as holes. The distribution function for quasi-electrons can be defined as

$$f_e(E) = f_{\text{out}}(E) - f(E), \quad (C3)$$

and for holes as

$$f_h(E) = f(E) - f_{\text{out}}(E). \quad (C4)$$

Note, in the quantized emission regime instead of quasi-particle excitations of the Fermi sea in the electron waveguide it is more natural to speak about the particles, electrons and holes, emitted by the cavity.

1. Adiabatic regime

In the adiabatic regime we use for the Floquet scattering matrix elements the following, $S_F(E, E_q) \approx S_{-q}(E)$, Eq. (39), where the Fourier coefficients $S_q$ for the frozen scattering amplitude are given in Eq. (44). Using these equations we calculate the distribution function at zero temperature. For $E > \mu$ the equation (C3) gives:

$$f^{\text{ad}}_e(E_n) = 2\Omega^2 e^{-2\Omega \tau (n+1)} , \quad (C5)$$

where $n\hbar \Omega < E_n - \mu < (n+1)\hbar \Omega$ with $n = 0, 1, \ldots$. The superscript “$\text{ad}$” stands for the adiabatic regime. With this distribution function we can calculate the number of emitted electrons per unit length of the waveguide:

$$\delta n^{\text{ad}}_e = \int_0^\infty \frac{dE}{\hbar v_\mu} f^{\text{ad}}_e(E) = \frac{1}{\hbar v_\mu} \sum_{n=0}^\infty f^{\text{ad}}_e(E_n)$$

$$\approx \frac{1}{\hbar v_\mu} e^{-2\Omega} \approx \frac{1}{\hbar v_\mu}. \quad (C6)$$

So we have one electron emitted during each period $\mathcal{T}$, as expected. Notice in above equation $v_F$ is the velocity of an electron with Fermi energy and $1/(\hbar v_\mu)$ is the density of states in the electron waveguide at the Fermi energy.

By analogy for $m\hbar \Omega < \mu - E_m < (m+1)\hbar \Omega$, with $m = 0, 1, \ldots$, we calculate the hole distribution function, Eq. (C4), in the adiabatic regime:

$$f^{\text{ad}}_h(E_m) = 2\Omega^2 e^{-2\Omega \tau (m+1)} . \quad (C7)$$

2. Non-adiabatic regime

With notation introduced in Sec. III B 2 the distribution function $f_{\text{out}}$, Eq. (C1), reads

$$f_{\text{out}}(\epsilon) = \frac{\Delta}{\hbar \Omega} \int_{-\infty}^\infty d\omega q |S_F(\epsilon, \epsilon + \omega q)|^2 f(\epsilon + \omega q) . \quad (C8)$$

where we use $\epsilon = (E - \mu)/\Delta$ and $\omega q = q\hbar\Omega/\Delta$.

At zero temperature we have for quasi-electron excitations,

$$f_e(\epsilon) = \frac{\Delta}{\hbar \Omega} \int_{-\infty}^\epsilon d\omega q |S_F(\epsilon, \epsilon + \omega q)|^2 , \quad \epsilon > 0 , \quad (C9)$$

and for holes,

$$f_h(\epsilon) = \frac{\Delta}{\hbar \Omega} \int_{-\infty}^{\infty} d\omega q |S_F(\epsilon, \epsilon + \omega q)|^2 , \quad \epsilon < 0 , \quad (C10)$$

where we used

$$\int_{-\infty}^{\infty} d\omega q |S_F(\epsilon, \epsilon + \omega q)|^2 = \frac{\hbar \Omega}{\Delta} . \quad (C11)$$

Above identity follows from the unitarity condition, Eq. (A14), with $\omega_n = \omega_m = 0$ and $\omega_p$ replaced by $\omega_q$.

We also used there

$$\delta(0) = \frac{\Delta}{\hbar \Omega} , \quad (C12)$$

which can be proven by re-introducing the sum over $q$ instead of the integral over $d\omega q$.

a. Optimal operating conditions

For the cavity driven by the periodic step potential and working under optimal operating conditions, the Floquet scattering amplitude is given in Eq. (36). Its square, we need to calculate Eq. (C8), is the following,

$$|S_F^{\text{opt}}(\epsilon, \epsilon + \omega q)|^2 = \left( \frac{\hbar \Omega}{\Delta} \right)^2 \frac{\sin^2(\pi \omega q)}{\pi^2 \omega_q^2}$$

$$\times \left[ \exp \left( i2\pi \omega_q \frac{\epsilon}{\omega_q} \right) - \exp \left( i2\pi \omega_q \frac{\epsilon}{\omega_q + 1} \right) \right]^2 . \quad (C13)$$

the superscript “$\text{opt}$” stands for optimal operating conditions.
For our present purposes we can simplify above equation: Since $\omega_q = 0$ is out of the integration interval in both Eq. (C9) and Eq. (C10), we can safely relax terms containing the factor $\delta (\omega_q)$ and use,

$$
\left| S_F^{\text{opt}} (\epsilon, \epsilon + \omega_q) \right|^2 \sim \left( \frac{h \Omega}{\Delta} \right)^2 \frac{\sin^2 (\pi \omega_q)}{\pi^2 \omega_q^2} \times \frac{1}{|\rho(\epsilon)|^2} \frac{1}{|\rho(\epsilon - \omega_q)|^2} \left\{ \frac{1}{(\omega_q - 1)^2} + \frac{1}{(\omega_q + 1)^2} + \xi_q \right\}.
$$

(C14)

Here $\xi_q$ contains terms that oscillate fast in $\omega_q$ and which we neglect upon integration over $\omega_q$.

First, we use above equation and calculate the distribution function for electrons, Eq. (C9). The integral over $\omega_q$ to the leading order in the small parameter $g = \delta/\Delta \ll 1$ can be evaluated using Eq. (B23) with $x = \epsilon - \omega_q$. The result is

$$
f_e^{\text{opt}} (\epsilon) \approx \frac{h \Omega}{\Delta} \frac{\cos^2 (\pi \epsilon)}{\pi^2 |\rho(\epsilon)|^2} \zeta, \quad \epsilon > 0,
$$

(C15)

where

$$
\zeta = \sum_{a=-\infty}^{0} \left\{ \frac{1}{(\epsilon - a)^2 - \frac{1}{4}} + \frac{1}{(\epsilon + 1 - a)^2 - \frac{1}{4}} \right\} = \frac{1}{(\epsilon - \frac{1}{2})^2} + \sum_{a=1}^{\infty} \frac{2}{(\epsilon + a)^2 - \frac{1}{4}}.
$$

The main contribution to the factor $\zeta$ comes from $\epsilon \sim 0.5$ and $\epsilon = 0$. Thus we obtain

$$
f_e^{\text{opt}} (\epsilon) \approx \frac{h \Omega}{\Delta} \frac{\cos^2 (\pi \epsilon) \nu^{\text{opt}} (\epsilon)}{\pi^2 (\epsilon^2 - \frac{1}{4})^2}, \quad \epsilon > 0.
$$

(C16)

Here we introduce the frozen density of states (DOS) of the cavity working under optimal conditions, $\nu^{\text{opt}} (\epsilon) = 1/(|\rho(\epsilon)|^2 \Delta)$, see Eqs. (43), (15) and (35). Note, if the cavity is driven by the pulsed potential and the working conditions are optimal, the frozen DOS, $\nu^{\text{opt}} (\epsilon)$, is time-independent. Using Eq. (B10) we find

$$
\nu^{\text{opt}} (\epsilon) = \sum_{a=-\infty}^{\infty} \frac{\delta/\pi}{\Delta^2 (\epsilon + 0.5 - a)^2 + \delta^2},
$$

(C17)

where we re-introduce the level width $\delta = g \Delta$. The DOS peaks at $\epsilon_a = a - 0.5$. However, at this values the factor $\cos (\pi \epsilon)$ in Eq. (C16) vanishes. Therefore, at positive $\epsilon$ the leading contribution arises form the peak around $\epsilon_1 = 0.5$ where the zero in the denominator compensate the zero in the numerator. Thus we arrive at the following equation for the distribution function of non-adiabatically emitted electrons ($\epsilon > 0$),

$$
f_e^{\text{opt}} (\epsilon) \approx \frac{h \Omega}{\Delta} \frac{\cos^2 (\pi \epsilon) \nu^{\text{opt}} (\epsilon)}{\pi^2 (\epsilon^2 - \frac{1}{4})^2 \Delta^2 (\epsilon - 0.5)^2 + \delta^2},
$$

(C18)

For practical purposes this equation can be simplified even further,

$$
f_e^{\text{opt}} (\epsilon) \approx \frac{h \Omega \cos^2 (\pi \epsilon) \nu^{\text{opt}} (\epsilon)}{\pi^2 (\epsilon^2 - \frac{1}{4})^2 \Delta^2 (\epsilon - 0.5)^2 + \delta^2}, \quad 0 < \epsilon < 1,
$$

(C19)

where we restrict $\epsilon = (E - \mu)/\Delta$ from above (by one level spacing) to keep the same mean energy and its variance calculated with the help of either Eq. (C18) or Eq. (C19).

Using Eq. (C14) in Eq. (C10), we calculate the distribution function for holes, $f_h^{\text{opt}} (\epsilon) = f_e^{\text{opt}} (-\epsilon)$. 

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