Experimental Verification of a Two-Degree-of-Freedom Resonant Actuator Using Its Detent Force

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This study aims to present a new two-degree-of-freedom resonant actuator and its motion control method. Accordingly, the design method for this resonant actuator that resonates using its detent force without mechanical spring is proposed. In this actuator, a centering force is generated according to the moving distance, regardless of the moving direction on the X-Y plane. Using this feature, various resonance motions on the X-Y plane can be realized. This study also proves that the reciprocating motion, elliptical motion, and scroll motion can be controlled. The results of finite element analysis support the characteristics of the proposed actuator, which were compared to the results of measurements on a prototype. Finally, the effectiveness of motion control was validated through the prototype.

Keywords: multi-degree-of-freedom, linear resonant actuator, detent, spring-less, motion control

1. Introduction

Recently, various actuators for application to household appliances and industrial machines have been actively studied. Linear resonant actuators (LRAs) are an actuator that reciprocates with high efficiency using m-k resonance. LRAs are mainly used for compressors, electric shavers, electric toothbrushes, and artificial hearts (1)-(5). On the other hand, a multiple-degree-of-freedom (M-DOF) resonant actuator (RA) which extends DOF of LRA has also been studied. This M-DOF RA can be used for applications requiring various motions such as a haptic device (6)-(7). However, since general M-DOF RA can change the effective air-gap, precise motion cannot be realized due to variations in the motor parameters. Furthermore, as general M-DOF RA uses mechanical spring for resonant operation, the number of parts increases and the overall system becomes bulky and complicated. Also, the M-DOF RA is difficult to realize full resonant operation in a variety of motions due to differences in vertical stiffness and lateral stiffness of the mechanical springs. For example, the two-DOF RA can be designed to have the same resonance frequency in the X and Y directions, but the resonance frequency for some angles may be different. The phase difference between the current and the velocity of the mover is made due to the change of resonance frequency depending on the moving direction. The phase difference eventually makes it difficult to control the motion. Moreover, since M-DOF RA controls two or more coils independently, interference between mutual inductances is large and the estimation error of displacement may accordingly be large. However, designing the axial gap type M-DOF RA with a star or delta connection of three-phase symmetric coils without using a mechanical spring can solve the problems mentioned (8)-(9). Nd-Fe-B permanent magnet was used in the cited actuator. However, it was difficult to design the support structure due to the strong attraction force generated in the Z-direction. Thus, we redesigned it by changing as a ferrite permanent magnet.

In this study, we redefine that the detent characteristic is similar to a type of mechanical springs. In addition, the thrust characteristic by phase current is presented. They are verified through finite element analysis (FEA). To control various motion, we propose a simplified motion that can express all of the motion in X-Y plane. As a result, they can be represented as only three types of variables. We also present the prototype of the proposed actuator. All of the static parameters including detent, thrust constant, inductance, and resistance are measured and compared to the results of FEA. Finally, we confirm that the various motion with two-DOF is can be realized by experimental result.

2. Characteristics of the proposed actuator

2.1 Proposed actuator

Fig. 1 demonstrates the basic structure of the proposed two-DOF RA. The actuator consists of stator, coil, permanent magnet, and moving core. The three-phase<br>
The coils are connected by a star-connection and wound counterclockwise when viewed from the top. Magnets placed on U-V-W phase with 120° interval are fixed to the back-iron. In addition, three magnets are magnetized in the positive direction of the Z-axis while the others are magnetized in the negative direction of the Z-axis. Since the mover is constrained only in the Z-direction, various resonant motion can be realized on the X-Y plane.

2.2 Detent characteristics

Fig. 2 indicates the position of each phase magnet when the mover is placed at the position of a distance r and an angle θ. To define the detent force reacted in each phase magnet, the force vector in each phase frame can be defined as:

\[
F_r = \begin{bmatrix} Ar \cos \theta \\ Br \sin \theta \end{bmatrix} \quad \text{(1)}
\]

\[
F_r = \begin{bmatrix} Ar \cos \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) \\ Br \sin \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) \end{bmatrix} \quad \text{(2)}
\]

\[
F_r = \begin{bmatrix} Ar \cos \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) \\ Br \sin \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) \end{bmatrix} \quad \text{(3)}
\]

where \( F_r \) means the reaction force vector of magnets placed in U-phase, \( F_r \) denotes the reaction force vector in V-phase, \( F_r \) is the reaction force vector in W-phase, \( A \) is a gain in the direction of each phase axis, and \( B \) stands for a gain in the normal direction of each phase axis. Since \( A \) and \( B \) are determined by material and the shape of magnets, Eqs.(1-3) are the generalized expression.

The above reaction force vectors are presented in each phase axis and their normal direction. They can be represented by converting to the X-Y orthogonal frame as follows:

\[
\begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = R(\theta) \begin{bmatrix} Ar \cos \theta \\ Br \sin \theta \end{bmatrix}
\]

\[
\begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = R\left( \frac{2}{3} \pi \right) F_r
\]

\[
= \begin{bmatrix} \frac{1}{2} Ar \cos \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) & \frac{\sqrt{3}}{2} Br \sin \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) \\ \frac{\sqrt{3}}{2} Ar \cos \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) & -\frac{1}{2} Br \sin \left( \frac{\theta - \frac{2}{3} \pi}{3} \right) \end{bmatrix}
\]

\[
\begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = R\left( \frac{4}{3} \pi \right) F_r
\]

\[
= \begin{bmatrix} \frac{1}{2} Ar \cos \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) & \frac{\sqrt{3}}{2} Br \sin \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) \\ \frac{\sqrt{3}}{2} Ar \cos \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) & -\frac{1}{2} Br \sin \left( \frac{\theta - \frac{4}{3} \pi}{3} \right) \end{bmatrix}
\]

where \( F_{dx} \) and \( F_{dy} \) present detent force of U-phase magnet in X-Y orthogonal frame, \( F_{dx} \) and \( F_{dy} \) means the detent force of V-phase magnet, \( F_{dx} \) and \( F_{dy} \) are the detent force of W-phase, and \( R \) is Park’s transformation matrix.

Since the detent force of the mover is equal to the sum of detent force in each phase magnet, it can be obtained as:

\[
F_{dx} = F_{dx} + F_{dx} + F_{dx} = \frac{3}{2} (A + B) r \cos \theta \quad \text{(7)}
\]

\[
F_{dy} = F_{dy} + F_{dy} + F_{dy} = \frac{3}{2} (A + B) r \sin \theta \quad \text{(8)}
\]

where \( F_{dx} \) denotes the detent force of the mover in X-direction and \( F_{dy} \) means the detent force in Y-direction.

In addition, when the mover is placed at the position of the distance \( r \) and the angle \( \theta \), the detent force in \( \theta \) direction and its normal direction can be obtained as follows:

\[
F_{d\theta} = R(\theta) \begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = \begin{bmatrix} \frac{3}{2} (A + B) r \\ 0 \end{bmatrix}
\]

\[
\begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = R(\theta) \begin{bmatrix} F_{dx} \\ F_{dy} \end{bmatrix} = \begin{bmatrix} \frac{3}{2} (A + B) r \\ 0 \end{bmatrix}
\]

where \( F_{d\theta} \) denotes the vector presented as components of \( \theta \)-direction and its normal direction.

From Eq. (9), since the component of normal direction keeps the value of zero, we know that the mover always received the force in the origin direction wherever it is positioned. This characteristic is similar to the case that several mechanical springs are mounted radially in the X-Y plane.

2.3 Thrust characteristics

As the proposed actuator uses a star-connection, assuming that three-phase currents are in equilibrium as follows:

\[
i_u + i_v + i_w = 0 \quad \text{(10)}
\]

where \( i \) denotes current, and subscript means each phase.

Furthermore, as the force constant \( K \) of each phase is all the same, the sum of the torque \( T \) generated at each phase becomes 0 as follows:

\[
T = l(F_u + F_v + F_w) = 1K f (i_u + i_v + i_w) = 0 \quad \text{(11)}
\]

where \( l \) is the distance between the center of the stator and the tooth, and \( F \) means the force generated in each phase.
Fig. 3. Switch states and force direction. (a) force direction in each switch state (b) force direction on the two phases.

Fig. 4. Definition of elliptical motion in the reference frame.

Meanwhile, the Park’s and Clarke’s transformation matrix $T$ can be defined with magnitude invariance system as follows:

$$T(\theta) = \frac{2}{3} \begin{bmatrix} \cos \theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3}) \\ -\sin \theta & -\sin(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \end{bmatrix}$$

(12)

By considering the sum of the current force on each phase, the thrust of the mover can be defined as:

$$\begin{bmatrix} F_{ex} \\ F_{ey} \end{bmatrix} = T \left( \frac{\pi}{2} \right) \begin{bmatrix} F_u \\ F_v \\ F_w \end{bmatrix} = \frac{2}{3} K_f \begin{bmatrix} \frac{\sqrt{3}}{2} i_v - \frac{\sqrt{3}}{2} i_w \\ \frac{\sqrt{3}}{2} i_v + \frac{\sqrt{3}}{2} i_w \\ -i_u + \frac{\sqrt{3}}{2} i_v + \frac{\sqrt{3}}{2} i_w \end{bmatrix}$$

(13)

where $F_{ex}$ denotes the thrust in the $X$-direction, $F_{ey}$ means the thrust in the $Y$-direction.

### 2.4 Current mode and force direction

By using a three-phase voltage source inverter, current mode can be generated as shown in Fig. 3(a). As shown in Fig. 3(b), the direction of the force according to the current mode is determined by the sum of the vector components of the force generated in each phase. In addition, force components of various directions can be created by a composite vector of two modes.

#### 3. Model dynamics

##### 3.1 Electrical dynamics

The electrical dynamics of the two-DOF RA with three-phase coils can be expressed as follows:

$$\begin{align*}
\begin{bmatrix} v_u \\ v_v \\ v_w \end{bmatrix} &= \begin{bmatrix} R_u & 0 & 0 \\ 0 & R_v & 0 \\ 0 & 0 & R_w \end{bmatrix} \begin{bmatrix} i_u \\ i_v \\ i_w \end{bmatrix} + \begin{bmatrix} e_u \\ e_v \\ e_w \end{bmatrix} \\
\frac{d}{dt} \begin{bmatrix} L_{uu} & M_{uv} & M_{uw} \\ M_{vu} & L_{vv} & M_{vw} \\ M_{wu} & M_{uw} & L_{ww} \end{bmatrix} \begin{bmatrix} i_u \\ i_v \\ i_w \end{bmatrix} &= \begin{bmatrix} e_u \\ e_v \\ e_w \end{bmatrix}
\end{align*}$$

(14)

where $v$ and $i$ denote the phase voltage and the phase current, $R$ is the coil resistance, $L$ is the self-inductance, $M$ is the mutual inductance, $e$ is the back electromotive force of each phase.

Since each phase coil is arranged symmetrically with the same number of turns, below conditions are satisfied:

$$R_u = R_v = R_w = R$$

(15)

$$L_{uu} = L_{vv} = L_{ww}$$

(16)

$$M_{uv} = M_{wu} = M$$

(17)

From Eqs.(15-17), electrical dynamics of the proposed actuator can be defined as:

$$\begin{align*}
\begin{bmatrix} v_u \\ v_v \\ v_w \end{bmatrix} &= \begin{bmatrix} R + \rho L & 0 & 0 \\ 0 & R + \rho L & 0 \\ 0 & 0 & R + \rho L \end{bmatrix} \begin{bmatrix} i_u \\ i_v \\ i_w \end{bmatrix} + \begin{bmatrix} e_u \\ e_v \\ e_w \end{bmatrix}
\end{align*}$$

(18)

where $\rho = \frac{d}{dt}$, $L = L_{uu}M$. It is a relational expression of leakage inductance, self-inductance, and mutual inductance. As a result, the electrical dynamics of the proposed actuator is the same as that of a typical surface permanent magnet synchronous motor (SPMSM).

##### 3.2 Mechanical dynamics

The mechanical equation of two-DOF RA without mechanical spring can be expressed as follows:

$$\begin{align*}
\begin{bmatrix} \dot{x} \\ \dot{\chi} \\ \dot{y} \end{bmatrix} &= \begin{bmatrix} 0 & 1 & 0 & 0 \\ -c & 0 & 0 & 0 \\ -\frac{m}{m} & 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x \\ \chi \\ y \end{bmatrix} + \frac{1}{m} \begin{bmatrix} 0 \\ F_{ex} - F_{dx} - F_{dt} \\ F_{cy} - F_{dy} - F_{bt} \end{bmatrix}
\end{align*}$$

(19)

where $\chi$ is the displacement in the $X$-direction, $y$ is the displacement in the $Y$-direction, $m$ denotes the mass of the mover, $c$ means a viscosity coefficient considering the support structure of the mover, $F_{ex,ey}$ is the force generated by the current, $F_{dx,dy}$ means the centering force by the detent, $F_{bt}$ is the force of the load.

In addition, the resonant frequency of the two-DOF RA can be defined as follows:

$$f_n = \frac{1}{2\pi} \sqrt{\frac{k}{m}} = \frac{1}{2\pi} \sqrt{\frac{1}{m} \frac{F_{dt}}{|x|}} = \frac{1}{2\pi} \sqrt{\frac{1}{m} \frac{F_{bt}}{|y|}}$$

(20)

where $k$ is the stiffness of the mechanical spring. However, since
the proposed actuator has no mechanical spring, the resonant

frequency of the proposed actuator is determined by the ratio
between the amplitude of detent force and displacement.

4. Motion estimation

The electrical dynamics of the proposed actuator in X-Y
orthogonal reference frame can be defined as:

$$\begin{bmatrix}
    v_x \\
    v_y
\end{bmatrix} =
\begin{bmatrix}
    R + \rho L & 0 \\
    0 & R + \rho L
\end{bmatrix}
\begin{bmatrix}
    i_x \\
    i_y
\end{bmatrix} +
\begin{bmatrix}
    K_e \dot{y} \\
    -\dot{x}
\end{bmatrix}
$$

where $K_e$ is the back-EMF constant. From the electrical dynamics,
position and angle information can be estimated as:

$$\begin{bmatrix}
    x \\
    y
\end{bmatrix} = \frac{1}{K_e}
\begin{bmatrix}
    \int (-v_y + Ri_y) dt + Li_y \\
    \int (v_x - Ri_x) dt - Li_x
\end{bmatrix}
$$

$$\begin{bmatrix}
    x_A \\
    y_A
\end{bmatrix} =
\begin{bmatrix}
    \cos \theta & \sin \theta \\
    -\sin \theta & \cos \theta
\end{bmatrix}
\begin{bmatrix}
    x \\
    y
\end{bmatrix}
$$

where $x_A$ and $y_A$ are the represented position in the orthogonal frame
rotated by an angle $\theta$. In addition, stroke of motion can be obtained
as follows:

$$\begin{bmatrix}
    \alpha \\
    \beta
\end{bmatrix} =
\begin{bmatrix}
    \max(x_A) - \min(x_A) \\
    \max(y_A) - \min(y_A)
\end{bmatrix}
$$

where $\alpha$ and $\beta$ are the stroke of motion. Fig. 4 shows a type of
elliptical motion. If elliptical motion can be controlled, scrolling
motion and reciprocating motion can be performed by changing the
ratio between $\alpha$ and $\beta$.

5. Verification through FEA

FEA is performed to confirm the trend of parameter
characteristics such as the direction of the detent and the thrust in
all of the mover position. Dense meshes are only applied in the
surface of the teeth-tip and permanent magnet due to the reduction of
computation time. In addition, 441 cases of mover positions are
analyzed by positioning the mover within a range of 4 mm in X-Y
plane. As an analysis tool, JMAG software is used.

Fig. 5 presents the detent characteristics of the proposed actuator.
The degree of the brightness means the norm of the detent, arrows
are pointing at a direction of detent force. We confirm that contours
are drawn with regular circles in X-Y plane and all of the arrows are
pointing at the origin. As a result, detent can be used as stiffness for
resonance with two-DOF.

The result of the thrust characteristics is shown in Fig. 6. In Fig.
6(a), each phase current is set with $i_v = 0 \text{ A}, i_c = 0.58 \text{ A}, i_v = -0.58 \text{ A}$. All of the arrows mean the directions of the thrust and they are
pointing at X-direction. Fig. 6(b) shows the result of thrust in Y-
direction. To generate the thrust in Y-direction, each phase current
is set with $i_v = -0.67 \text{ A}, i_c = 0.33 \text{ A}, i_v = 0.33 \text{ A}$. All of the arrows
keep in Y-direction.

6. Experimental verification

Fig. 7 present the prototype of the proposed two-DOF RA. In
order to support the mover and limit the stroke, stator guide, mover
stopper, spacer, ball stopper, ball-bearing were added. The
specification of the prototype is listed in Table 1.

6.1 Detent measurement Fig. 8 shows the setup environment for measuring the static characteristics of the
prototype. It consists of a load cell that can move in one direction
through a micrometer and the indicator that measures the output
signal of the load cell.
Fig. 9 illustrates the results of detent characteristics. Fig. 9(a) is the result in the X-direction, and Fig. 9(b) is the result in the Y-direction. The lines mean the result of FEA, and dots denote the result of measurement. The slopes of the measured values of the detent in the X and Y directions are 540 N/m and 543 N/m, indicating an error within 0.6%. On the other hand, the result of FEA is 576 N/m, which is 6% higher than the measured value. This is due to the fact that the air-gap length may be increased by assembly tolerances. As a result, the resonant frequency is lower than the expected value, but it is possible to realize the two-DOF resonant motion.

Table 1. Model specification.

| Parameter                  | Value |
|----------------------------|-------|
| Mass of mover (g)          | 66.1  |
| Effective stiffness of detent (N/m) | 540   |
| Force constant (N/A)       | 1.67  |
| Back-EMF constant (Vs/m)   | 1.67  |
| Phase resistance (Ω)       | 4.1   |
| Phase inductance (mH)      | 15.7  |
| Resonant frequency (Hz)    | 14.9  |

Fig. 7. Prototype of the proposed two-DOF RA.

Fig. 8. Setup for measuring the static characteristics.

6.2 Force constant measurement Fig. 10 denotes the result of force constant characteristics. To generate thrust in the Y direction, the phase current is energized as switching-states (0,1,1) in Fig. 3. The measured value is about 2% smaller than the result of FEA. This is also due to the fact that the air-gap length is increased. As a result, force constant is about 1.67 N/A in the moving distance within 2 mm.

6.3 Inductance and resistance measurement From
(16), to estimate the motion of the mover, we have to know the information about inductance and resistance. However, by operating frequency, their impedance may be changed due to the component of eddy-current and hysteresis loss. Fig. 11 denotes the result of inductance and resistance in the frequency domain. Since we know that the operating frequency is about 15 Hz, we decide the value of inductance and resistance for motion estimation as 15.7 mH and 4.1 Ω.

6.4 Motion control Fig. 12 describes the setup environment for measuring the dynamic characteristics of the prototype. It consists of the prototype, position sensor, oscilloscope, and three-phase inverter. Since the inverter has four-channel of digital to analog converter (DAC), measured value and estimated value can be compared under the oscilloscope. The operation condition is presented in Table 2.

| Parameter                      | Value   |
|-------------------------------|---------|
| Operating frequency (Hz)      | 15.0    |
| Carrier frequency (kHz)       | 4.0     |
| DC-link voltage (V)           | 24      |
| Target rotation angle (degree)| 45      |
| Target stroke (mm)            | 2.0     |
| Ratio of reciprocating motion ($\beta/\alpha$) | 0       |
| Ratio of elliptical motion ($\beta/\alpha$) | 0.5     |
| Ratio of scrolling motion ($\beta/\alpha$) | 1.0     |

Fig. 13 shows the result of changing the motion. Each channel of the oscilloscope denotes DAC outputs of $x$, $y$, $\alpha$, and $\beta$. We confirmed that reciprocating motion, elliptical motion, scrolling motion can be realized by controlling the ratio of $\alpha$ and $\beta$.

Fig. 14 illustrates the result of the comparison between the measured displacement and the estimated displacement. Fig. 14(a) indicates the result of displacement trajectory for reciprocating motion. As indicated in the figure, reciprocating motion in a direction can be controlled as well. Fig. 14(b) denotes the result of elliptical motion. It can be confirmed that all of the command motion, measured motion, and estimated motion match successfully. Finally, the result of scrolling motion is shown in Fig. 14(c). As a result, we proved that all of the motion can be estimated and controlled as well.

Fig. 15 shows the result of the comparison between the measured stroke and the estimated stroke in the elliptical motion. By using the measured parameters in the stroke estimation process, a good result of estimation accuracy can be obtained.
In this study, the authors have proposed an axial gap type two-DOF RA which resonates without a mechanical spring. The proposed two-DOF RA showed that constant detent force is generated with respect to the moving direction from theoretical and experimental results. Furthermore, a method of estimating the motion of an actuator without a position sensor has been proposed. We have built the prototype to experimentally verify the effectiveness of the proposed method. The static characteristics of the prototype closed to the result of theoretic and FEA. In addition, we designed an estimator and controller in the microcomputer (MICOM) of the inverter. Finally, it has been confirmed that various resonant motion can be precisely estimated and controlled as well.

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