Alignment procedure for the VIRGO Interferometer: experimental results from the Frascati prototype.

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Abstract

A small fixed-mirror Michelson interferometer has been built in Frascati to experimentally study the alignment method that has been suggested for VIRGO. The experimental results fully confirm the adequacy of the method. The minimum angular misalignment that can be detected in the present set-up is 10 nrad/\sqrt{Hz}.

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Introduction

The search for Gravitational-Waves (GW) with interferometric antennas (VIRGO, LIGO, GEO) aims at the detection of a GW-signal by measuring the relative motion that a GW induces between two widely separated test masses [1, 2, 3]. In all these antennas, the test masses are the mirrors suspended at the ends of two long arms, perpendicular to each other and forming a Michelson interferometer, with Fabry-Perot cavities in the VIRGO and LIGO cases. Inside the detector, IR-light (1.064 µm) is split into two beams which travel down the arms and reflect from the mirrors at the two ends. On their return, the beams interfere on the detection plane and from the illumination changes of the photodiode one can infer the relative changes that have occurred to the interferometer arm lengths. However, all the optical elements of the interferometer undergo position and angle fluctuations that are mostly due to thermal and seismic noise. Any angular tilt leads to a reduction in the sensitivity of the interferometer and in particular causes a variation of the cavity length that simulates a GW-signal.

The specifications required for the angular stability are determined by the sensitivity level one is aiming at for the detection of a GW-signal. These can turn out to be very severe and thus the sensitivities required in monitoring the error signals can become an issue. As a part of the research and development program in support of the VIRGO project, we built a small fixed-mirror interferometer with Fabry-Perot cavities in the two arms. The aim of the work was only to focus on the experimental verification of the theory developed to compute the alignment signals and on the method to extract from them the angular position of each individual mirror.

1 The Fabry-Perot alignment methods

Two are the basic alignment methods that we considered and both of them are based on the phase modulation of the incoming beam. The first was developed at Caltech by D.Z. Anderson et al. [4] and the second at Glasgow by H. Ward et al. [5]. Both of them were originally designed to align one single Fabry-Perot (FP) but no attempt has ever been made to extend this procedure to the case of a complete Michelson Interferometer (MI).

For a brief description of the two methods, let us consider a pure TEM$_{00}$ laser beam entering a Fabry-Perot cavity with a tilting angle $\Delta \alpha$ with respect to the cavity axis. In the limit where this angle is much smaller than the far field divergence of the beam $\alpha_0 = \lambda/\pi w_0$, the cavity sees the input beam as a linear superposition of the TEM$_{00}$ distribution and the first off-axis mode TEM$_{10}$. Similarly, if the TEM$_{00}$ beam enters the cavity with a lateral displacement $\Delta x$ much smaller than the beam waist $w_0$, the cavity sees again the incoming beam profile as a linear superposition of TEM$_{00}$ and TEM$_{10}$.

The difference between these two cases is that the rotations lead to a coupling into the TEM$_{10}$ mode as the translations do, but with a 90° phase shift. This means that a misalignment causes a coupling into the lowest-order off axis mode with a phase that depends upon the type of misalignment. Therefore the transverse field distribution seen by the cavity as a consequence of small walks off of the two terminal mirrors, can be always approximated by a linear combination of these two cavity modes

$$E \simeq C_0 \left( U_0 + CU_1 \right), \quad (1.1)$$

where $E$ is the normalized input field, and $U_0$ and $U_1$ are the usual orthonormal Hermite-Gaussian functions of $x, y$ associated with the TEM$_{00}$ and TEM$_{10}$ modes. $C_0$ and $C$ are the coupling coefficients and in particular:

- for a pure translation $\Delta x$: $C = \Delta x/w_0$;
- for a pure rotation $\Delta \alpha$: $C = i \Delta \alpha/\alpha_0$;

$C_0$ is always real and, since it is always very close to unity, it has been assumed equal to one.

Therefore the basic idea of the methods is to find a way to detect the amplitude and phase of this induced TEM$_{10}$ component.

1.1 The Anderson method

With this method, one chooses to phase-modulate the input beam at the frequency separation between TEM$_{00}$ and TEM$_{10}$, which in a plano-concave cavity (length $L$, mirror radius $R$) is given by:

$$\Delta \nu = \frac{c}{2\pi L} \arccos \sqrt{1 - L/R} \quad (1.2)$$
By retaining only the resonant terms, namely the fundamental mode at the carrier frequency and the TEM
This equation shows that the intensity component that is in-phase with the modulation (cos Ω
DC-photocurrent is obtained by integrating the first and leading term of eq.(1.5) yielding
where Ω/2π = Δν is the modulation frequency, E₀ is a constant real vector, and J_l(m) is the Bessel function
of order l and phase modulation index m. In this notation, the physical electric field is obtained by taking
the real part of the complex quantities. If the value of the modulation index m is sufficiently small, only the
first three terms in the expansion (1.3) can be retained and the expression of the electric field (1.1) reduces to:
E ≈ E₀(U₀ + CU₁)e^{iωt} \left\{ J₀(m) + 2iJ₁(m) \sin \Omega t \right\}, \quad (1.4)
By retaining only the resonant terms, namely the fundamental mode at the carrier frequency and the TEM₁₀
mode at the sideband frequency (ω + Ω), the resulting transmitted intensity exhibits a spatially dependent
component modulated at the beat frequency Ω as follows (1.6):
I = |T|^2 |I₀ J₀^2 U₀^2 + |C|^2 J₁^2 U₁^2 + 2J₀J₁U₀U₁ \left( \text{Re } |C| \cos \Omega t + \text{Im } |C| \sin \Omega t \right) |, \quad (1.5)
where T is the FP’s complex transmittivity on resonance [7].
Since the Hermite-Gaussian functions are mutually orthogonal when integrated all over the space, the
detection of the entire transmitted beam by a single photodiode results in a DC-signal. The correspondent
DC-photocurrent is obtained by integrating the first and leading term of eq.(1.5) yielding
\begin{align}
I_{dc} = \eta \frac{e \lambda}{hc} |T|^2 I₀ J₀^2(m), \quad (1.6)
\end{align}
where η is the quantum efficiency. On the contrary, a separate detection of each half of the transmitted beam,
followed by electronic subtraction of the two photocurrents, yields a current signal given by
\begin{align}
I_{diff} = 2 \sqrt{\frac{2}{\pi}} I_{dc} J₀ \frac{J₁}{J₀} \left\{ \text{Re } |C| \cos \Omega t + \text{Im } |C| \sin \Omega t \right\}. \quad (1.7)
\end{align}
This equation shows that the intensity component that is in-phase with the modulation (cos Ωt) is proportional
to the translational error and the in-quadrature component (sin Ωt) is proportional to the angular alignment error.
Thus, by demodulating the signal, one can obtain simultaneously and independently both the misalignment
errors. With the same argument, a quadrant photodiode detector permits simultaneous detection of couplings
to both vertical and horizontal off-axis modes.

1.2 The Ward method

This alternative technique was suggested by R.Drever at Caltech and experimentally demonstrated by H.Ward
at the University of Glasgow [5]. It is basically an extension of the method used for the longitudinal locking
(Pound and Drever) [6] and it uses the light which is reflected from the FP. The beam is phase modulated
but the modulation frequency Ω is not to be equal to the frequency difference between fundamental and first
transverse mode as it is in the Anderson method.
Let us suppose that a plano-concave cavity is both laterally and angularly misaligned with respect to the
incoming beam direction of the usual quantities a = Δx/ω₀ and α = Δα/α₀ respectively, with Δx being
measured at the waist position. In the cavity frame the incoming beam is described by
\begin{align}
E'_{in} = E₀e^{iωt} \left[U₀ + (a + iα)U₁ \right] J₀ + 2iJ₁ \sin \Omega t], \quad (1.8)
\end{align}
where U₀ at the carrier frequency is the only resonant term, and all the other terms are assumed to be completely
off-resonance. With the convention that the reflection from a mirror introduces a 90° phase shift, the phase
of a beam reflected from a Fabry-Perot is −90° (+90°) depending on whether the beam is on(off)-resonance.
Assuming a totally reflective end mirror, the expression for the reflected beam in the cavity frame is
\begin{align}
E'_{ref} = -iE₀e^{iωt} \left\{ J₀ \left[U₀ - (a + iα)U₁ \right] - 2iJ₁ \left[U₀ + (a + iα)U₁ \right] \sin \Omega t \right\}, \quad (1.9)
\end{align}
which, transformed into the incoming beam frame, reads as follows
\begin{align}
E_{ref} = -iE₀e^{iωt} \left\{ J₀(U₀ - 2aU₁) - 2iJ₁(U₀ + 2iαU₁) \sin \Omega t \right\}. \quad (1.10)
\end{align}
The idea of the method is to let this beam evolve freely in space and to consider that an additional term intervenes in this process. This term is the phase difference between the real Hermite-Gaussian beam and an ideal plane wave, given by the Guoy phase

\[ \phi_{n,m}(z) = (n + m + 1) \phi(z), \quad \tan \phi(z) = \frac{\lambda z}{\pi w_0^2}. \]  

(1.11)

The indexes \( n, m \) refer to the \((n, m)\)-th order mode and \( z \) is the propagation coordinate whose origin, \( z = 0 \), is at the location of the beam waist. This means that different modes evolve differently and the two components \( U_0 \) and \( U_1 \) acquire the phase difference \( \phi(z) \). The eq. (1.10) becomes:

\[ E_{ref}(z) = -i E_0 e^{i(\omega t + \phi)} \left\{ J_0(U_0 - 2aU_1 \cos \phi) - 4aJ_1 \cos \phi \sin \Omega t \right\} - 2i \left\{ aJ_0 U_1 \sin \phi + J_1(U_0 - 2aU_1 \sin \phi) \sin \Omega t \right\}. \]  

(1.12)

The intensity associated with this field depends upon the position where the detector is located. If the current difference between the two halves of a photodetector is taken at a given \( z \)-position, and the signal is demodulated at the frequency \( \Omega \), the dominant component that will be detected is given by:

\[ I_{diff} \propto -J_0J_1U_0U_1(a \sin \phi + \alpha \cos \phi) \sin \Omega t. \]  

(1.13)

Since the beam has a waist on \( M_1 \), \( \sin \phi = 0 \) right behind \( M_1 \) and in this region the method is sensitive only to tilts. However, if the beam is let evolve, the angle \( \phi \) goes to \( \pi/2 \) and so, at a very large distance from the waist, \( I_{diff} \) becomes sensitive only to displacements. A more practical way to change \( \phi \) is to employ an appropriate telescope, as discussed in detail in ref. [5].

2 Beam evolution inside a complex interferometer

Let us consider a TEM\(_{00}\) phase modulated beam \( \Psi_{in} \)

\[ \Psi_{in} = U_0 e^{i\omega t} (J_0 + 2iJ_1 \sin \Omega t) \]  

(2.1)

impinging on a misaligned interferometer of a VIRGO-like structure (see fig.1) with the beam waist on \( M_0 \). In the following the ”main frame” will always refer to this input beam.

In order to compute the expected alignment signals, we have developed a mathematical procedure that, under the following approximations and assumptions, leads to a fully analytical solution:

1) the beam mismatches, due to the difference in length of the Michelson arms, and to the plane-plane configuration of the recycling cavity (VIRGO case), have been neglected

2) all the misalignment angles are assumed to be much smaller than the angular divergence of the beam

3) second or higher order terms in the misalignment angles have been neglected in the expressions of the beam amplitudes

4) since, under the above approximations, no horizontal-vertical coupling can be envisaged, the analysis has been limited to one plane only

5) since in Virgo the angular noises that need to be corrected are confined in the region below 1 Hz, the calculation has been performed in a static condition, i.e. with all the optical elements being kept at their misaligned positions.

Under the above mentioned conditions, all the beams running along or emerging from the interferometer, can be expressed as a linear combination of TEM\(_{00}\) and TEM\(_{10}\). Hence, each frequency component of the beam \( \Psi_1 \) (see fig.1) can be written in the following form

\[ \Psi_1 = A e^{i\theta} \left[ U_0 + (a + i \alpha) U_1 \right] \]  

(2.2)
where $A, \theta, a, \alpha$ are the four quantities to be determined. Once this expression is given, all the other beams follow immediately, since the distances between the optical elements and the propagation phases are all known. After the beam $\Psi_1$ has been transported to the two FPs, the reflected beams $\Psi_{r1}$ and $\Psi_{r2}$ are calculated according to a procedure similar to the one of sec.(1.2). These two reflected beams are then propagated back toward the beam splitter and their recombination $\Psi_2$ hits the recycling mirror $M_0$ which is supposed to be misaligned of the angle $\theta_0$. The expression for the reflected beam in the main frame, takes the form

$$\Psi_{2r} = A^* e^{i \theta^*} \left[ U_0 + (a^* + i \alpha^*) U_1 \right] $$

(2.3)

where the quantities $A^*, \theta^*, a^*, \alpha^*$ contain $A, \theta, a, \alpha$, the propagation phases, and the angles of all the mirrors. Finally, by imposing that the sum $\Psi_{2r} + t_0 \Psi_{in} = \Psi_1$, one obtains:

$$t_0 U_0 J_i + A^* e^{i \theta^*} \left[ U_0 + (a^* + i \alpha^*) U_1 \right] = A e^{i \theta} \left[ U_0 + (a + i \alpha) U_1 \right]$$

(2.4)

By equalizing real and imaginary parts of $U_0$, $U_1$, eq.(2.4) can be broken down into a system of four linear equations in the four unknown $A, \theta, a, \alpha$. The solution of this system shows that $A$ and $\theta$ do not depend on the misalignment angles, while $a$ and $\alpha$ are given by:

$$a = \sum_{j=0}^{4} b_j \theta_j, \quad \alpha = \sum_{j=0}^{4} c_j \theta_j$$

(2.5)

where the $b_j, c_j$ coefficients depend on the geometrical and optical characteristics of the interferometer (propagation phases, mirrors transmittivities and reflectivities).

Once this procedure has been repeated for all the three frequency components, we have the full description of the beam inside the recycling cavity that can be propagated toward any position of interest, inside or outside the interferometer. Given a quadrant photodiode placed on an output beam, four signals can be detected: up-down and left-right differences demodulated in-phase and in-quadrature. In general, a signal $S^i$ from the $i$-th quadrant photodiode will be of the form:

$$S^i = \sum_{k=0}^{4} C_k^i \theta_k$$

(2.6)
This means that each signal will receive contributions from any misalignment and one has to find a way to extract the information relative to each of the five mirrors. The most common way of solving a problem of this kind is the $\chi^2$ procedure: having $n$ signals (with $n \geq 5$), the best estimate of the angles $\theta_j$ is found searching for the set of values leading to minimum $\chi^2$. In this case, given the simple structure of eq. (2.6), this set of $\theta_j$-values can be obtained analytically, together with their uncertainties $\Delta\theta_j$, due to the presence of noise on the signals (shot-noise and electronic noise).

The reconstruction of the mirror angles can be optimized by choosing a sufficient number of quadrant photodiodes and by studying the dependence of their signal on the Guoy phase.

### 3 The experimental apparatus

The Frascati prototype is a small-scale, fixed-mirror Michelson interferometer in air, sufficient for us to concentrate on optical alignment and stability studies. The mechanical and optical apparatus lies on top of a TMC-optical table ($1.5 \times 2.0$ m), which is isolated from seismic ground motion with an air over-pressure spring system that couples it to the ground floor. A sound absorbing enclosure protects the whole assembly against the ambient acoustical noise. The interferometer is constructed with commercial mirror mounts and the space between the cavity mirrors is protected by plexiglass tubes to minimize the fluctuations in the refraction index induced by air circulation.

#### 3.1 The interferometer layout

The layout is shown in fig. 2. Light from a frequency stabilized Newport NL-1 He-Ne laser (0.5 mW at 632.8 nm) is phase modulated and focused with two positive lenses ($L_0$, $L_1$) into a beam waist of $5.9 \cdot 10^{-2}$ cm at the position of the recycling mirror ($M_0$), 1.94 m away from the laser output.

![Figure 2: Complete lay-out of the whole interferometer](image)

The laser is isolated from spurious light feed-backs with a system of two Faraday rotators and two polarizing beam splitters. If only one of these two systems is used, the back reflection reduces to a fraction of $2.0 \cdot 10^{-5}$ of the incident power but becomes undetectable with a cascade of two. A third polarizing beam splitter ($P_2$)
is added before the entrance of the interferometer to ensure linear polarization in the horizontal plane for both the beams directed to and reflected from the interferometer. A quadrant photodiode (QS) is placed on the beam transmitted from the mirror \( T_2 \): low frequency movements of the beam directed to the interferometer, due mainly to angular jitters of the laser output, are so detected and corrected, acting in a feedback loop on two piezos mounted on the mirror \( T_0 \).

To find room on the optical table, the recycling cavity is folded over a zig-zag sequence of two flat mirrors \((T_3, T_4)\), a positive lens \((L_2)\) and a curved mirror \((T_5)\) with a total optical path of 2.68 m. The beam waist at the position of the two flat mirrors \((M_1, M_3)\) is \(7.2 \cdot 10^{-2}\) cm and is mode matched to the geometry of the two following FP-cavities. The FP mirrors \((M_{1,2} \text{ and } M_{3,4})\) and the recycling mirror \((M_0)\) are mounted on Burleigh PZ-91 (piezo-mounting) to allow for fine longitudinal (2 nm/V) and angular adjustments (80 nrad/V horizontal, 92 nrad/V vertical). The mirror characteristics have been measured and the results are summarized in table I.

### Table I

| Mirror | Transmittivity \( t^2 \) (%) | Curvature Radius (m) |
|--------|-----------------------------|---------------------|
| \( M_0 \) | 40.8 ± 0.3 | \( \infty \) |
| \( M_1, M_3 \) | 10.1 ± 0.2 | \( \infty \) |
| \( M_2, M_4 \) | 0.2 ± 0.01 | 10 |
| BS | 50.3 ± 0.7 | \( \infty \) |
| \( T_3 \) | \((6.0 \pm 1.0) \cdot 10^{-2}\) | \( \infty \) |

The piezo mounts have a frequency response that relates the applied voltage to the motion effectively executed by the mirror. This response is described by a transfer function which, in general, depends upon the exciting frequency and can only be determined with an experimental measurement. In the low frequency interval \((\leq 10 \text{ Hz})\), where most of the mechanical noise is concentrated, the transfer function is definitely well-behaved. However, structures are visible in the region between 200 Hz and 500 Hz and are associated with the mechanical structure of the mirror holder. They have been efficiently damped by adding an extra load to the mount.

### 3.2 Fabry-Perot’s cavities

Both the Fabry-Perot cavities are 70.6 cm long and the separation between the longitudinal modes is \(c/2L = 212.3 \text{ MHz}\). From eq. (1.3), the transverse-longitudinal separation is 18.2 MHz. The line-shape profile has been measured with a photodiode located behind the terminal mirrors and reproduces the expected expression in Ref. [7]. The measurement of the width \( \Gamma \) of the resonance yields the finesse

\[
\mathcal{F} = \frac{c}{2L} = \pi \frac{\sqrt{r_1 r_2}}{1 - r_1 r_2} = 57.5 \pm 2.0
\]

and allows for a direct measurement of the product \( r_1 r_2 = 0.947 \pm 0.002\), in good accordance with the reflectivity values reported in table I. The value of the power enhancement (the power stored into the cavity per unit incident power), the reflectivities and the transmittivities of the FPs have been measured at resonance and the results are shown in table II together with the same values expected on the base of the mirror transmittivities quoted in table I. The departure from one of the sum \( T^2 + R^2 \) measures the total amount of the cavity losses. They turn out to be 7.5 % and 9.4 % for FP\(_1\) and FP\(_2\) and imply coating losses of 0.11 % and 0.15 %, both well within specs for standard commercial mirrors.

### Table II

| Mirror | Reflectivity \( R^2 \) (%) | Transmittivity \( T^2 \) (%) | Power Enh. |
|--------|----------------------------|-----------------------------|------------|
| FP\(_1\) | 85.8 ± 3.0 | 6.7 ± 0.2 | 32.8 ± 1.0 |
| FP\(_2\) | 84.3 ± 3.0 | 6.3 ± 0.2 | 31.1 ± 1.0 |
| expected | 92.6 | 7.4 | 36.2 |
Both FP cavities have been longitudinally locked with a “dithering” technique. The end mirrors of FP1 and FP2 are set into longitudinal oscillation with an amplitude of \( \sim 1.3 \cdot 10^{-5} \lambda \) at a frequency of 7.0 kHz and 4.3 kHz respectively. By means of two analog dividers, the detectors PM_{1,2} have been normalized to PM_{3} which is proportional to the recycling cavity inner power. The resulting ratios are then sent to two lock-in amplifiers, with reference 7.0 kHz and 4.3 kHz, whose outputs constitute the error signals of the feedback loops. These signals are then integrated, amplified and, finally, applied to the piezos of the two FP’s terminal mirrors M_{2} and M_{4}. In the “feedback on” condition, the power fluctuations in the two cavities never exceed a small percentage of the stored power.

### 3.3 The recycling cavity

The recycling cavity comprises the sequence of elements that run from M_{0} to M_{1,3}. It is arranged in a plano-plano geometry where the two flat terminal mirrors are connected by the two focusing elements L_{2} and T_{5}. The distances from the beam splitter to the initial FP mirrors have been made slightly different \((M_{BS} - M_{1} = 11.9 \text{ cm}, M_{BS} - M_{3} = 9.4 \text{ cm})\), in analogy to what they will be in VIRGO, to increase the signal on the dark fringe detector \([8]\). The total cavity loss \(\alpha_{rc}\), inclusive of the lens L_{2} and the three mirrors T_{3}, T_{4}, T_{5}, has been obtained by a direct measurement of the power lost in one trip from L_{2} to T_{5}. Unfortunately measurements of this kind are affected by large errors because they always require to take differences between almost equal numbers. In our case we have obtained \(\alpha_{rc} = (1.92 \pm 1.0)\%\).

The recycling cavity can be looked at as a FP cavity where the initial mirror is M_{0} and the terminal mirror is the ensemble of the beam splitter and the two FPs. This ensemble can be considered as an equivalent mirror with an effective reflectivity defined as:

\[
 r_{eq}^2 = \frac{1}{2} (R_{FP1}^2 + R_{FP2}^2) (1 - \alpha_{rc})^2 (1 - \alpha_{fr}) = (81.5 \pm 2.6)\% \quad (3.2)
\]

where \(\alpha_{fr}\) accounts for the power that is inevitably lost in the incomplete light extinction on the fringe detector. This power loss has been evaluated to be of the order of 1 %. The enhancement factor for this cavity is expected to be

\[
 F_{pe} = \frac{\text{stored power}}{\text{incident power}} = \frac{T_{0}^2}{(1 - r_{eq}r_{0})^2} = 4.4 \pm 0.4 \quad (3.3)
\]

and it has been obtained by measuring the power transmitted from T_{3} (or equivalently by FP_{1,2}) with and without the insertion of M_{0}. The experimental result has been determined to be 4.8 \pm 0.3 in good agreement with eq.(3.3).

Two fast photodiodes were used in two feedback loops to keep the recycling cavity on resonance (FP_{1}) and to maintain the dark fringe condition (FP_{2}). The demodulated signal from FP_{1} was used to control the M_{0} position, while the feedback from FP_{2} acted on both the FP_{2} mirrors, thus constraining the dark fringe without perturbing the status of FP_{2}.

### 4 Experimental Results

According to the analytical simulation described in sec. 2 (for more details see ref. [3]) small mirror misalignments do not prevent the field amplitude from reaching an equilibrium condition that retains the information on each individual mirror position. This means that the demodulated signals from a quadrant photodiode placed on any of the beams leaving the interferometer are sensitive to the misalignment status of all mirrors, either in their in-phase or in-quadrature components, or in both. Furthermore, since each of these components can be analytically expressed in terms of the individual mirror misalignments, with a sufficient number of measurements the alignment status of the whole interferometer can be fully determined. To better achieve this, the modulation frequency of the laser beam has to coincide with the TEM_{00} − TEM_{10} frequency separation of the FPs (18.2 MHz in our case), since in this way the beams transmitted from the FPs have the maximum sensitivity to misalignments.

In the experimental test we conducted, we set into oscillation M_{0}, M_{1}, M_{2}, M_{3}, M_{4} one at a time, at the fixed frequency of 3 Hz and with an amplitude of approximately 200 nrad. The signals seen by Q_{0}, Q_{1} and Q_{2}...
were each time demodulated at $0^\circ$ and $90^\circ$ in order to construct the two quantities

$$A(M_i) = \sqrt{I(M_i)^2 + Q(M_i)^2}, \quad \varphi(M_i) = \arctan \frac{I(M_i)}{Q(M_i)}.$$  \hspace{1cm} (4.1)

where $I(M_i)$ and $Q(M_i)$ are the in-phase and in-quadrature signals generated by the $i$-th mirror.

Since the angle/voltage calibrations are the same for all the piezos in the system, it is quite appropriate to present the results normalized to one of them. Tables III, IV, V present the comparison between the experimental and expected values of these ratios for $Q_0$, $Q_1$ and $Q_2$, respectively.

Table III: Beam reflected from $M_0$ for a fixed value of $\Phi_G/\pi$: comparison between the experimental and theoretical values for the quantity $R_{ij} = A(M_i)/A(M_j)$ and $\Delta \varphi_{ij} = |\varphi(M_i) - \varphi(M_j)|$.

| $\Phi_G/\pi = 0.1063$ | $R_{12}$ | $\Delta \varphi_{12}$ (deg) | $R_{20}$ | $\Delta \varphi_{20}$ (deg) | $R_{30}$ | $\Delta \varphi_{30}$ (deg) |
|------------------------|-----------|-----------------------------|-----------|-----------------------------|-----------|-----------------------------|
| Expt.                  | 1.45 ± 0.08 | 174.3 ± 3.0                | 1.33 ± 0.06 | 1.2 ± 1.6           | 1.33 ± 0.07 | 179.3 ± 3.1               |
| Theor.                 | 1.22       | 179.0                      | 1.26      | 1.1                      | 1.11      | 179.1                      |

Table IV: Beam transmitted from FP$_1$.

| $\Phi_G/\pi = 0.0415$ | $R_{02}$ | $\Delta \varphi_{02}$ (deg) | $R_{12}$ | $\Delta \varphi_{12}$ (deg) | $R_{42}$ | $\Delta \varphi_{42}$ (deg) |
|------------------------|-----------|-----------------------------|-----------|-----------------------------|-----------|-----------------------------|
| Expt.                  | 0.40 ± 0.02 | 45.1 ± 3.0             | 0.83 ± 0.03 | 173.8 ± 2.2       | 0.8 ± 0.1 | 25.2 ± 6.5                |
| Theor.                 | 0.395      | 49.6                      | 0.835     | 174.7               | 0.7       | 23.0                      |

Table V: Beam transmitted from FP$_2$ for two values of $\Phi_G/\pi$.

| $\Phi_G/\pi$ | $\Phi_G/\pi = 0.0965$ | $\Phi_G/\pi = 0.4144$ |
|--------------|------------------------|------------------------|
| Expt.        | Theor.                 | Expt.                  | Theor.                 |
| $R_{02}$     | 0.52 ± 0.02 | 0.57          | 0.56 ± 0.01 | 0.55          |
| $\Delta \varphi_{02}$ (deg) | 10.1 ± 2.0 | 26.7      | 15.5 ± 1.4 | 24.2        |
| $R_{12}$     | 0.77 ± 0.02 | 0.91          | 0.9 ± 0.02 | 0.895        |
| $\Delta \varphi_{12}$ (deg) | 183.8 ± 1.4 | 179.2     | 176.7 ± 1.0 | 179.2      |
| $R_{32}$     | 1.28 ± 0.03 | 1.2           | 0.94 ± 0.02 | 1.17         |
| $\Delta \varphi_{32}$ (deg) | 146.6 ± 1.27 | 152.4     | 137.6 ± 1.0 | 149.2      |
| $R_{42}$     | 1.2 ± 0.03  | 1.43          | 1.18 ± 0.01 | 1.39         |
| $\Delta \varphi_{42}$ (deg) | 31.9 ± 1.1  | 22.4         | 35.3 ± 0.6  | 24.6        |

As shown in the tables, we have selected different values for the Guoy phases, either by choosing the appropriate position of the quadrant photodiode, or by inserting a lens between it and the output mirror, at
whose position we define the zero of the Guoy phase. The quoted errors are only computed from the measured noise fluctuations. In all the cases the agreement appears extremely satisfactory.

The effect of the evolution of the Guoy phase has been directly tested on the beam reflected from $M_0$. As indicated at the end of section 1.2, we have kept $Q_0$ fixed at about 60 cm from $M_0$, and placed a lens with $f = +20$ cm at different distances from $Q_0$, thus obtaining values for the total Guoy phase in the range $(0.11 \div 0.94) \pi$ rad. The behaviours of both $A(M_2)/A(M_0)$ and $\varphi(M_2) - \varphi(M_0)$, expressed in terms of the calculated Guoy phase $\Phi_G$, are shown in fig. 3.

The expected curve for $A(M_2)/A(M_0)$ peaks quite sharply at $\Phi_G/\pi = 0.8$ and the experimental data follow this shape with a reasonable agreement. The smallness of the $A(M_0)$ value at the peak position makes the error bar on the ratio unusually large. The same observation holds for $\varphi(M_2) - \varphi(M_0)$ as well. By looking at the two figures, one could argue that there is a possible small systematic shift, of approximately 9 degrees, between theory and experiment. While there are a lot of possible explanations for that, one has to keep in mind that the theoretical curves have been computed using the measured values for distances, transmittivities etc., and thus they can be affected by systematic uncertainties as well.

The validity of this theoretical approach can be further appreciated in the following way. Starting from a coarse pre-alignment condition, one can read the residual error signals given by the three photodiodes and elaborate the corrections to be applied to each individual mirror for the fine alignment of the system. The sequence of the power levels that the dark fringe detector reads, when the corrections are applied one after the other, is presented in fig. 4. With this correction procedure the light extinction on the dark fringe improves of more than a factor 3, reaching a value of 0.4 % of the total power stored into the cavity.

In conclusion, we have demonstrated that even with a very low average power in the photodiodes, $(1 \div 30) \mu$W, a small phase modulation index (0.25 % of the total power in each sideband) and a longitudinal locking scheme that could be largely improved, it is possible to accurately align a complex interferometer. Indeed, under these circumstances, the average noise level on the demodulated signals from the quadrant photodiodes is equivalent to about $10 \text{ urad}/\sqrt{\text{Hz}}$ or $4.0 \cdot 10^{-5} \alpha_0/\sqrt{\text{Hz}}$. Therefore, since this angle is the minimum angle that can be detected in our set-up, this quoted value represents the lowest limit to our present alignment.
Figure 4: Fringe loss (the ratio between the power on the fringe detector and the power in the recycling cavity) when the calculated corrections are applied in sequence to $M_{0,1,2,3,4}$.

capability. Finally, since the VIRGO sensitivity to GW-detection requires the angular noise to be reduced below $1 \mu\text{rad}/\sqrt{\text{Hz}}$ [10], we claim that our present results already satisfy this requirement.

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