DHMM-based asynchronous finite-time sliding mode control for Markovian jumping Lur’e systems

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Abstract—This paper is concerned with the asynchronous problem for a class of Markovian jumping Lur’e system (MJLSs) via sliding mode control (SMC) in continuous-time domain. Specifically, the discrete hidden Markovian model (DHMM) is employed to describe the nonsynchronization between the controller modes and the MJLSs modes. In particular, considering the nonlinearity of MJLSs, a novel Lur’e-integral-type sliding surface is constructed. In order to ensure the finite-time stability of sliding mode dynamics and the accessibility of the specified sliding surface, the asynchronous Lur’e-type SMC law of the detector mode is presented. Finally, an example of DC motor controller modes and the MJLSs modes. In particular, considering the asynchronous problem between the controller and uncertainty for the asynchronization problem. In [27], a candidate controller and the system mode is asynchronous, the stability of a class of linear systems with mean residence time is studied by using asynchronous switching control. In [26], an asynchronous elastic controller is designed to ensure the system stability but also to accelerate the convergence and final disturbance, the control objective is not only to guarantee stability of the nonlinear switched system subject to time delay and uncertainty for the asynchronization problem. In [27], considering the asynchronous problem between the controller and the system modes, the stability of uncertain time-delay switched nonlinear systems is studied by using asynchronous switching control. However, for nonlinear systems with external disturbance, the control objective is not only to guarantee system stability but also to accelerate the convergence and improve the robustness. Sliding mode control (SMC) has been widely applied in practical engineering in terms of its fast response speed and strong robustness to uncertainties and disturbances. In [28], the $H_{\infty}$ control problem in commodity pricing has been solved by SMC method. Reference [29] develops an adaptive SMC scheme integrating fuzzy control with SMC control for interval type-2 fuzzy systems in the presence of uncertainty parameters, and the issue of SMC for fuzzy singularly perturbed systems with application to electric circuit is investigated in [30]. Furthermore, the extended application to missile system, communication network system and robot control system, has enabled finite-time boundedness to become

I. INTRODUCTION

ONLINEAR systems have received extensive attention and a series of gratifying results are available in the past decades, see, e.g., [2], [3], [4], [5] and many references therein. As an important kind of nonlinear systems, Lur’e system is composed of a linear system and an unknown static nonlinear system with certain constraints through negative feedback [6]. For more topics of Lur’e systems one can refer to [7], [8]. Additionally, many nonlinear dynamic systems in engineering and economic fields, such as manufacturing process, network communication system, power circuit system and economic system, are susceptible to random variations and abrupt changes caused by external environment, internal structure or human intervention in the operation process. Thus, Markovian jumping systems (MJSs) are developed to model these more general nonlinear systems and notable results have been produced [9], [10], [11]. Recently, the stability analyses and control problems of MJSs have received great interests among researchers [12], [13], [14]. However, there are few researches on the Markovian jumping Lur’e systems (MJLSs), which makes this paper have certain research significance.

It should be noted that in the ideal MJLSs, each state of the Markov chain corresponds to an observable physical quantity. But the practical situation is complex, so all the real states observation is notoriously difficult to conduct. In this case, the hidden Markov model (HMM) is introduced to describe such phenomenon. As a statistical analysis model, HMM has became a hot topic over the last few years, and its application including but not limited to speech processing [15], target tracking [16], digital communication [17], biomedical engineering [18] and finance [19]. It’s worth mentioning that the aforementioned HMM includes two random processes, i.e., the random process caused by the Markov chain, and the other process brought from the observed variables. According to the observed values, the HMM can be divided into two types. When the observed value is discrete, it is called the discrete HMM (DHMM). Otherwise, when the observed value is continuous and can be described as a function, such HMM is called continuous HMM (CHMM). However, to the best of our knowledge, most of the existing research are mainly focused on interpreting DHMM in discrete time domain, see [20], [21], [22], and rarely studied in continuous time domain. Therefore, this paper addresses a longstanding open issue on DHMM in continuous time domain.

Based on DHMM, the asynchronous control scheme [23], [24] offers an effective alternative for dynamic system fully utilized the potential information in the presence of time delays and data loss. In [25], when the switching between the candidate controller and the system mode is asynchronous, the stability of a class of linear systems with mean residence time is studied by using asynchronous switching control. In [26], an asynchronous elastic controller is designed to ensure the stability of the nonlinear switched system subject to time delay and uncertainty for the asynchronization problem. In [27], considering the asynchronous problem between the controller and the system modes, the stability of uncertain time-delay switched nonlinear systems is studied by using asynchronous switching control. However, for nonlinear systems with external disturbance, the control objective is not only to guarantee system stability but also to accelerate the convergence and improve the robustness. Sliding mode control (SMC) has been widely applied in practical engineering in terms of its fast response speed and strong robustness to uncertainties and disturbances. In [28], the $H_{\infty}$ control problem in commodity pricing has been solved by SMC method. Reference [29] develops an adaptive SMC scheme integrating fuzzy control with SMC control for interval type-2 fuzzy systems in the presence of uncertainty parameters, and the issue of SMC for fuzzy singularly perturbed systems with application to electric circuit is investigated in [30]. Furthermore, the extended application to missile system, communication network system and robot control system, has enabled finite-time boundedness to become

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considerably embraced [31], [32], [33].

Inspired by the above facts, we investigate the finite-time stabilization problem for a class of MJLSs via SMC scheme under the framework of DHMM which is illustrated in Fig. 1. The main contributions of the proposed scheme can be concluded as:

(1) Asynchronous SMC is used for the first time to solve the finite-time stabilization problem of the MJLSs in continuous time domain under the framework of DHMM.

(2) An asynchronous Lur’e-type SMC law is designed to drive state trajectories onto the specified sliding surface during the prescribed finite-time interval.

(3) Nonlinear terms are added to the integral-type sliding surface in order to reduce the conservatism.

(4) Compared with [34], we avoid the coupling problem caused by mode in the derivation process by selecting the appropriate Lyapunov function in the proof of reachability.

The rest of this paper is organized as follows. The MJLSs with stochastic perturbation is described in Section II followed by the main results in Section III. The effectiveness of the proposed method is illustrated in Section IV. Finally, Section V draws a concluding remark.

II. PROBLEM FORMULATION AND PRELIMINARIES

Given a probability space \((\Omega, \mathcal{F}, \text{Prob})\), where \(\Omega\) is the sample space, \(\mathcal{F}\) represents the algebra of events and \(\text{Prob}\) denotes the probability measure defined on \(\mathcal{F}\). \(\{\theta_\sigma = i, \sigma \geq 0\}\) indicates a Markov chain taking values in a finite set \(\mathcal{N} = \{1, 2, \ldots, N\}\) with a transition probability matrix \(\Pi \triangleq [\pi_{ij}], i, j \in \mathcal{N}\). The transition probability from mode \(i\) at time \(\sigma\) to mode \(j\) at time \(\sigma + \Delta \sigma\) can be described by:

\[
\text{Prob}\{\theta_{\sigma+\Delta \sigma} = j | \theta_{\sigma} = i\} = \begin{cases} 
\pi_{ij}\Delta \sigma + o(\Delta \sigma), & i \neq j \\
1 + \pi_{ii}\Delta \sigma + o(\Delta \sigma), & i = j 
\end{cases}
\]

in which \(\pi_{ij} \in [0, 1], \forall i, j \in \mathcal{N}\) and \(\sum_{j=1}^{N} \pi_{ij} = 1\). The MJLSs with stochastic perturbation can be written as:

\[
\begin{align*}
\dot{x}(\sigma) &= A(\theta_\sigma)x(\sigma) + A_d(\theta_\sigma)x(\sigma - \tau) + B(\theta_\sigma)u(\sigma) \\
&\quad + D(\theta_\sigma)w(\sigma) + F(\theta_\sigma)\varphi(\varphi(\sigma)) \\
\varphi(\sigma) &= M_i x(\sigma) \\
x(\sigma) &= h(\sigma), \sigma \in [\sigma_0 - \tau, \sigma_0]
\end{align*}
\]

where \(x(\sigma) \in \mathbb{R}^n\) is the state vector, \(u(\sigma) \in \mathbb{R}^m\) is the control input, \(w(\sigma) \in \mathbb{R}^p\) is the input disturbance and \(\varphi(\cdot)\) denotes the nonlinear vector.

Some assumptions and the corresponding definitions as well as lemmas are given to facilitate the stability analysis of MJLSs (2).

**Assumption 1** The nonlinear function \(\varphi(\varphi(\sigma)) : \mathbb{R}^p \rightarrow \mathbb{R}^p\) is an additive vector depending on the vector \(\varphi(\sigma)\). Thus, \(\varphi(\cdot)\) satisfies:

\[
\begin{align*}
\mathcal{S}C(\varphi(\cdot), \eta, L) &= \varphi(\eta)^T L \varphi(\eta) - \Omega \eta \leq 0 \\
\varphi(0) &= 0, \forall \eta \in \mathbb{R}^p
\end{align*}
\]

**Assumption 2** Define a finite-time interval \([0, T]\), the unknown input \(w(\sigma)\) satisfies:

\[
\int_0^T w^T(s)w(s)ds \leq \delta, \delta \geq 0.
\]

**Definition 1** [5] Given a time interval \([0, T]\), positive scalars \(\xi, \zeta, \tau\), with \(0 < \xi < \zeta\) and a mode-dependent weighting matrix \(R_i > 0\), MJLSs (2) is finite-time bounded (FTB) if

\[
\mathbb{E}\{x^T(0) R_i x(0)\} \leq \xi \Rightarrow \mathbb{E}\{x^T(\sigma) R_i x(\sigma)\} \leq \zeta, \forall \sigma \in [0, T].
\]

**Definition 2** [35] Given a finite-time interval \([0, T]\), positive scalars \(\xi, \zeta\), with \(0 < \xi < \zeta\) and a mode-dependent weighting matrix \(R_i > 0\), MJLSs (2) is finite-time stabilizable (FTS) with \(w(\sigma) = 0\) if condition (5) is fulfilled.

**Lemma 1:** (Partitioning Strategy)[36] For MJLSs (2) with the specified parameters set \((\tau_1, \tau_2, [0, T], R_i, \delta)\), the closed-loop system is FTB with respect to \((\tau_1, \tau_2, [0, T], R_i, \delta)\), if and only if there exists an auxiliary scalar \(c^*\) satisfying \(\tau_1 < c^* < \tau_2\) such that it is FTB with respect to \((\tau_1, c^*, [0, T], R_i, \delta)\) during the reaching phase and FTB with respect to \((c^*, \tau_2, [\tau_1, T], R_i, \delta)\) during the sliding motion phase.

III. MAIN RESULTS

A. Asynchronous Sliding Surface and SMC Law Design

Due to packet loss, transmission delay or random disturbance involved in practice along with the inaccurate model information of the system, there is a need of the asynchronizition between the controller modes and the system modes. In this section, we introduce the DHMM \((\theta_\sigma, \hat{\theta}_\sigma)\) to describe the asynchronous phenomenon described above. More precisely, a stochastic variable \(\theta_\sigma\) can be utilized to estimate \(\theta_\sigma\) with a known II-dependent conditional probability matrix \(\Phi \triangleq [\delta_{il}]\), in which the probability \(\delta_{il}\) is defined as:
\( \text{Prob} \{ \hat{\theta}_m = I | \theta_m = i \} = \delta_{il} \),

where \( 0 \leq \delta_{il} \leq 1 \) and \( l \in \mathcal{M}, \sum_i \delta_{il} = 1 \) for all \( i \in \mathcal{N} \).

Remark 1: It should be mentioned that \( \hat{\theta}_m \) is introduced as a detector to represent the controller mode, and take values in a finite set \( \mathcal{M} = \{1, 2, \cdots, M\} \). Notice that \( \hat{\theta}_m \) depends on \( \theta_m \) according to the given conditional probability.

By preserving the index \( i \) for the modes of the Markov chain \( \theta_m \) and \( l \) for the detector \( \hat{\theta}_m \), we design an Lur'e-type sliding surface function as:

\[
\begin{align*}
\hat{s}(\sigma) &= G_i x(\sigma) - \int_0^\sigma G_i B_i E_i \varphi(M_i x(s))ds \\
&\quad - \int_0^\sigma G_i B_i K_i x(s)ds - \int_0^\sigma G_i F_i \varphi(M_i x(s))ds \quad (7)
\end{align*}
\]

in which \( G_i \in \mathbb{R}^{n \times n} \) is a real matrix such that \( G_i B_i \) is nonsingular.

Then, the asynchronous Lur'e-type SMC law can be achieved by:

\[
\begin{align*}
\hat{u}(\sigma) &= K_i x(\sigma) + E_i \varphi(M_i x(\sigma)) - \beta_i(\sigma) \text{sign}(s(\sigma)) \quad (8)
\end{align*}
\]

where \( K_i \) is the controller gain and \( E_i \) is the nonlinear feedback gain.

B. FTS Analysis of Sliding Mode Dynamics

By applying the partitioning strategy, the FTS problem of MJLSs (2) can be addressed in two phases via asynchronous SMC, i.e., the reaching phase within \([0, \bar{\mathcal{T}}] \) and the sliding motion phase within \([\bar{\mathcal{T}}, \mathcal{T}] \).

When the system states move on \([0, \bar{\mathcal{T}}] \), the system trajectories will not reach the Lur'e-type integral-type sliding surface (7). Substituting the above asynchronous Lur'e-type SMC law (8) into MJLSs (2), the closed-loop MJLSs can be rewritten as:

\[
\begin{align*}
\dot{x}(\sigma) &= A_{ii} x(\sigma) + A_{di} x(\sigma - \tau) + E_i \varphi(M_i x(\sigma)) + D_i w(\sigma) - B_i \beta_i(\sigma) \\
x(\sigma) &= h(\sigma), \quad \sigma \in [\omega_0 - \tau, \omega_0] \quad (9)
\end{align*}
\]

where \( A_{ii} = A_i + B_i K_i, \quad E_{ii} = B_i E_i + F_i, \quad \beta_i(\sigma) = \beta_i(\sigma) \text{sign}(\sigma) \).

Theorem 1: For a given finite time interval \([0, \bar{\mathcal{T}}] \), by implementing the asynchronous Lur'e-type SMC law (8), MJLSs (2) is a FTC, i.e., the closed-loop MJLSs (9) is FTB with respect to \((c_1, c^*, R_1, \delta)\), if there exists a scalar \( \mu > 0 \), positive-definite symmetric matrices \( P_i, Q_i \), and positive-definite matrices \( Q, O_1, O_2, R \), such that the following inequalities hold for all \( i \in \mathcal{N} \) and \( l \in \mathcal{M} \):

\[
\begin{align*}
\Theta_3 & \Theta_4 \Theta_5 P_i B_i - P_i D_i \sum_{i=1}^{M} \delta_{il} \tau A_{di}^2 \\
& \leq Q_i - O_2 \begin{bmatrix} 0 & 0 & 0 \\
0 & 0 & \tau A_i^2 \\
0 & \tau A_i^2 & 0 \\
\mu I & 0 & \tau D_i^2 \\
\mu I & \tau D_i^2 & \tau R_i^2
\end{bmatrix} > 0 \quad (10)
\end{align*}
\]

Under the condition of (16), we can obtain the following inequality by substituting (18) into (17)

\[
\begin{align*}
3V_1(x(\sigma)) &\leq \sum_{i=1}^{M} \delta_{ii} 2x^2(\sigma)P_i \dot{x}(\sigma) + x^2(\sigma) \left( \sum_{i=1}^{N} \pi_{ij} P_j x(\sigma) \right) + x^2(\sigma)Q_i x(\sigma) + x^2(\sigma)Q_i x(\sigma) + \tau x^2(\sigma)Q_i x(\sigma) \\
&+ \tau x^2(\sigma)R_i x(\sigma) - \int_{\sigma - \tau}^{\sigma} \dot{x}^2(s) R_i x(s)ds + x^2(t)O_1 x(\sigma) \\
&+ x^2(\sigma - \tau)O_2 x(\sigma - \tau) - \int_{\sigma - \tau}^{\sigma} \dot{x}^2(s) W_i x(\sigma) ds \\
&+ 2 \tau x^2(\sigma) W_i x(\sigma) - 2 \tau x^2(t) W_i x(\sigma) \\
&- 2 \tau x^2(\sigma) W_i \int_{\sigma - \tau}^{\sigma} \dot{x}(s)ds + \tau x^2(\sigma) W_i R_i W_i x(\sigma). \quad (19)
\end{align*}
\]

Under Assumption 1, for any given \( \Omega \), there exists \( \Delta \) such that

\[
\begin{align*}
SC[\varphi(\cdot), M_i x(\sigma), \Delta] = \varphi^2(M_i x(\sigma)) \Delta^2 M_i x(\sigma) - \Omega M_i x(\sigma) \leq 0.
\end{align*}
\]
Recalling to (19) and (20), we know that
\[ \mathcal{S}V_1(x(\sigma)) = \sum_{i=1}^{M} \delta_{li} 2x^T(\sigma)P_i\dot{x}(\sigma) \]
\[ + x^T(\sigma) \left( \sum_{j=1}^{N} \pi_{ij} P_j x(\sigma) + x^T(\sigma)Q_i x(\sigma) \right) \]
\[ - x^2(\sigma - \tau)Q_i x(\sigma - \tau) + \tau x^2(\sigma)W_i R_i^{-1} W_i^T x(\sigma) \]
\[ + \tau x^2(\sigma)Q_i x(\sigma) + \tau x^2(\sigma)R_i x(\sigma) + x^2(\sigma)O_i x(\sigma) \]
\[ + x^2(\sigma - \tau)O_2 x(\sigma - \tau) - 2\sigma \mathcal{S}[\psi(\cdot), M_i x(\sigma), \Delta] \]
\[ + 2x^2(\sigma)W_i x(\sigma - \tau) + 2x^2(\sigma)W_i x(\sigma - \tau - \sigma) \]
\[ - \int_{\tau}^{\sigma} [W_i^T x(\sigma) + R_i x(\sigma)] R_i^{-1} [W_i^T x(\sigma) + R_i x(\sigma)] d\sigma \]
\[ = \chi^2(\sigma) Z_1 x(\sigma) + \tau \dot{x}(\sigma) R_i x(\sigma) \]
\[ - \int_{\tau}^{\sigma} [W_i^T x(\sigma) + R_i x(\sigma)] R_i^{-1} [W_i^T x(\sigma) + R_i x(\sigma)] d\sigma \tag{21} \]

where
\[ Z_1 = \begin{bmatrix} \Theta_1 & P_i A_i & -W_i & -P_i B_i & P_i D_i \\ O_2 & -Q_i & 0 & 0 & 0 \\ \ast & \ast & -2\Delta & 0 & 0 \\ \ast & \ast & \ast & 0 & 0 \\ \ast & \ast & \ast & \ast & 0 \end{bmatrix}. \]

\[ \chi(\sigma) \Delta = \left[ x^2(\sigma) x^2(\sigma - \tau) \varphi^2(M_i x(\sigma))^2(\tau) w^2(\sigma) \right]^T, \]

\[ \Theta_1 = \sum_{i=1}^{M} \delta_{li} (A_i^T P_i + P_i A_i) + \sum_{j=1}^{N} \pi_{ij} P_j + Q_i + \tau Q \]
\[ + W_i + W_i^T + \tau W_i R_i^{-1} W_i^T + O_i, \]

\[ \Theta_2 = \sum_{i=1}^{M} \delta_{li} P_i E_i + (\Delta \Omega M_i)^T. \]

Next, we define an auxiliary variable \( \mathcal{F} \) as:
\[ \mathcal{F} = \mu V_1(x(\sigma)) + \mu \beta^2(\sigma) \beta(\sigma) + \mu w^2(\sigma) \]
\[ - \mathcal{S}V_1(x(\sigma)). \tag{22} \]

Further defining the auxiliary variable \( \mathcal{F} \) as follows:
\[ \mathcal{F} = \mathcal{S}V_1(x(\sigma)) - \mu V_1(x(\sigma)) \]
\[ - \int_{\tau}^{\sigma} [W_i^T x(\sigma) + R_i x(\sigma)] R_i^{-1} [W_i^T x(\sigma) + R_i x(\sigma)] d\sigma \]
\[ \geq \mu V_1(x(\sigma)) + \mu \beta^2(\sigma) \beta(\sigma) - \chi^2(\sigma) Z_2 \chi(\sigma) \]
\[ - \tau \dot{x}(\sigma) R_i x(\sigma) \]
\[ \chi^2(\sigma) Z_2 \chi(\sigma) \tag{23} \]

where
\[ Z_2 = \begin{bmatrix} \Theta_3 & \Theta_4 & \Theta_5 & P_i B_i & -P_i D_i & M \sum_{i=1}^{M} \delta_{li} \tau A_i^T \\ \ast & Q_i & -O_2 & 0 & 0 & 0 \\ \ast & \ast & 2\Delta & 0 & 0 & 0 \\ \ast & \ast & \ast & \ast & \ast & \ast \\ \ast & \ast & \ast & \ast & \ast & \ast \end{bmatrix}, \]

\[ \Theta_3 = \mu \dot{P}_i - \Theta_1, \quad \Theta_4 = W_i - P_i A_i, \quad \Theta_5 = -\sum_{i=1}^{M} \delta_{li} P_i E_i - (\Delta \Omega M_i)^T. \]

From (10), one can infer \( \Phi > 0 \) and the following inequality holds:
\[ \mathcal{E} \geq \mu V_{12}(x(\sigma)) + \mu V_{13}(x(\sigma)) \]
\[ + \int_{\tau}^{\sigma} [W_i^T x(\sigma) + R_i x(\sigma)] R_i^{-1} [W_i^T x(\sigma) + R_i x(\sigma)] d\sigma > 0. \tag{24} \]

From (22) and (24), it is easily checked that
\[ \mathcal{S}V_1(x(\sigma)) < \mu V_1(x(\sigma)) + \mu \beta^2(\sigma) \beta(\sigma) + \mu w^2(\sigma) w(\sigma). \tag{25} \]

Next, we perform a series of equivalent transformations on (25). By multiplying the left and right sides of the (25) with \( e^{-\mu \sigma} \), and then integrate the result from 0 to \( \sigma \) with \( \sigma \in [0, T_1] \) on both sides, we obtain
\[ \mathcal{E} \{ e^{-\mu \sigma} V_1(x(\sigma)) \} < V_1(0) + \mu \int_{0}^{\sigma} e^{-\mu \sigma} \beta(\sigma) \beta(\sigma) \]
\[ + \mu \int_{0}^{\sigma} e^{-\mu \sigma} w^2(\sigma) w(\sigma) d\sigma. \tag{26} \]

Define \( \tilde{P}_i = R_i^{-1} P_i R_i^{-1}, \tilde{Q}_i = R_i^{-1} Q_i R_i^{-1} \), we have
\[ \mathcal{E} \{ V_1(x(\sigma)) \} < e^{\mu T} \int_{0}^{\sigma} \tau x^2(s)x(s) d\sigma \]
\[ + 4\mu \gamma^2 \mathcal{F} + 4\mu \alpha_2^2 \int_{0}^{\sigma} \tau x(s) x(s - \tau) d\sigma + \mu \delta \]
\[ + 4\mu \alpha_2^2 \delta^2 \mathcal{F} \tag{27} \]

where \( \beta^2(\sigma) \beta(\sigma) \leq 4 \gamma^2 + 4 \alpha_2^2 x^2(\sigma) x(\sigma) + 4 \alpha_2^2 x^2(\sigma - \tau) x(\sigma - \tau) + 4 \alpha_2^2 w^2(\sigma) w(\sigma). \)

More specifically, it follows:
\[ \mathcal{E} \{ V_1(x(\sigma)) \} \geq \tau x^2(s) O_2 x(s - \tau) d\sigma \]
\[ + \int_{0}^{\sigma} \tau x(s - \tau) O_2 x(s - \tau) d\sigma. \tag{28} \]

Based on (13) and (14), it yields:
\[ x^2(\sigma) R_i x(\sigma) < \int_{t_1}^{t} \tau x(s)x(s) + 4\mu \gamma^2 \mathcal{F} + 4\mu \alpha_2^2 \delta^2 \mathcal{F} \]
\[ \frac{1}{\lambda^2 - e^{-\mu \sigma}}. \tag{29} \]

Hence, it is easily to find that \( x^2(\sigma) R_i x(\sigma) < \iota^* \) for all \( \sigma \in [T_1, T_2] \), i.e., the sliding mode phase. According to \( \dot{s}(\sigma) = 0 \), the equivalent control law can be written as:
\[ u_{eq} = K_i x(\sigma) + E_i \varphi(M_i x(\sigma)) - (G_i B_i)^{-1} G_i A_i x(\sigma) \]
\[ - (G_i B_i)^{-1} G_i A_i d_i x(\sigma - \tau) - (G_i B_i)^{-1} G_i D_i w(\sigma). \tag{30} \]

By substituting the equivalent control law (30) into the MJLSs (2), the closed-loop MJLSs become
\[ \begin{cases} \dot{x}(\sigma) = A_i x(\sigma) + \Gamma A_i d_i x(\sigma - \tau) + E_i \varphi_i (M_i x(\sigma)) \\
+ G_i D_i w(\sigma) \end{cases} \tag{31} \]
\[ x(\sigma) = h(\sigma), \sigma \in [\sigma_0 - \tau, \sigma_0] \]
in which \( \tilde{A}_i = \bar{A}_i + B_i K_i, \quad \bar{A}_i = I - B_i (G_i R_i)^{-1} G_i \).

**Theorem 2:** For a given finite-time interval \([\mathcal{T}_1, \mathcal{T}]\), MJLSs (2) is FTB with the equivalent control law (33), i.e., the closed-loop MJLSs (34) is FTB respect to \((\tau^*, t_2, R_1, \delta)\), for any \( i \in \mathcal{N} \) and \( l \in \mathcal{M} \), if there exists a scalar \( \mu > 0 \), such that for all positive-definite symmetric matrix \( P_i \) and positive-definite symmetric matrix \( O_3 \) satisfies

\[
\begin{bmatrix}
\Theta_{12} & \sum_{i=1}^{N} \delta_{il} P_i A_{di} & \sum_{i=1}^{N} \delta_{il} \bar{P}_i E_{li} & P_i \Gamma D_i
\end{bmatrix}
\begin{bmatrix}
* & -O_3 & 0 & 0 \\
* & * & -2\Delta & 0 \\
* & * & * & -\mu I
\end{bmatrix}
< 0
\]

(32)

Then, under condition (32), the following inequality can be guaranteed from (38):

\[
\mathcal{J} V_2(x(\sigma)) - \mu V_2(x(\sigma)) < \mu w^T(\sigma) w(\sigma).
\]

(39)

Repeating the equivalent transformation of the (25) to the (39): multiplying both sides of (39) with \( e^{-\mu \sigma} \), and then integrate the result from \( \mathcal{T}_1 \) to \( \sigma \) with \( \sigma \in [\mathcal{T}_1, \mathcal{T}] \) on both sides. We can get:

\[
\mathbf{E}\{e^{-\mu \sigma} V_2(x(\sigma))\} < V_2(\mathcal{T}_1) + \mu \int_{\mathcal{T}_1}^{\sigma} e^{-\mu s} w^T(s) w(s) \mathbf{d}s.
\]

(40)

From Theorem 2, we immediately have \( x^T(\mathcal{T}_1) R_i x(\mathcal{T}_1) < \tau^* \). Then defining \( \hat{P}_i = \hat{R}_i^{-2} P_i \hat{R}_i^{- \frac{1}{2}} \), one can observe

\[
\mathbf{E}\{V_2(x(\sigma))\} < e^{\mu \tau^*} [\tau^* \bar{A}_i + \mu \bar{\delta}].
\]

(41)

On the other hand, it follows:

\[
\mathbf{E}\{V_2(x(\sigma))\} \geq x^T(\mathcal{T}_1) P_i x(\mathcal{T}_1).
\]

(42)

Based on (41) and (42), it yields:

\[
x^T(\sigma) R_i x(\sigma) < \tau^* \bar{A}_i + \mu \bar{\delta}.
\]

(43)

Accordingly, when conditions (33) and (34) are satisfied with a scalar \( c^* \), for all \( \sigma \in [\mathcal{T}_1, \mathcal{T}] \), we can deduce \( x^T(\sigma) R_i x(\sigma) < t_2 \) by (43).

**Theorem 3:** Given a finite-time interval \([0, \mathcal{T}]\), MJLSs (2) is FTB with respect to \((i_1, \tau^*, t_2, R_1, \delta)\), if there exists scalars \( \mu > 0, \sigma_1 > 0, \sigma_2 > 0 \) such that the following inequalities hold for all \( i \in \mathcal{N} \) and \( \mathcal{M} \):

\[
\begin{bmatrix}
L(X_i, Y_i) & A_2 & A_3 & -B_i & D_i \\
* & -\mathcal{F}_{i1} & 0 & 0 & 0 \\
* & * & -2\Delta & 0 & 0 \\
* & * & * & -\mu I & 0 \\
* & * & * & * & -\mu I \\
* & * & * & * & * \\
* & * & * & * & * \\
* & * & * & * & * \\
\end{bmatrix}
< 0
\]

(44)
\[
\begin{bmatrix}
  L_1(X_i, Y_i) & \Gamma A_{di} X_i \\
  \sum_{i=1}^M \delta_{ii} E_{ii} + X_i(\Delta \Omega M_i)^	op
\end{bmatrix}
\begin{bmatrix}
  M(X_i) \\
  0
\end{bmatrix} < 0
\]

(45)

\[
\begin{bmatrix}
  -t^2 e^{-\mu T} + 4\mu y^2 T + 4\mu a_2^2 \delta T \sqrt{\gamma_1} \\
  \sigma_1 - \tau \sigma_2
\end{bmatrix}
\begin{bmatrix}
  t^* \\
  -\sigma_{2t^*}
\end{bmatrix} < 0
\]

(46)

\[
\begin{bmatrix}
  -Z_1 e^{-\mu T} & 2\sigma_1 Z_1 \\
  0
\end{bmatrix} < 0
\]

(48)

\[
\begin{bmatrix}
  -Z_2 e^{-\mu T} & 2\sigma_2 Z_2 \\
  0
\end{bmatrix} < 0
\]

(49)

\[
\sigma_1 R_i^{-1} < X_i < R_i^{-1}
\]

(50)

\[
\left\lVert (\mathcal{G}_i)^{\top} \mathcal{G}_i \right\rVert < \gamma_i
\]

(54)

The control objective in this paper is to design an asynchronous Lur'e-type SMC law (8) that ensures system trajectories reach the predetermined sliding surface within a finite time interval.

**Theorem 4:** Consider MILLSs (2), the reachability of the specified sliding surface (7) can be guaranteed in a finite time $\mathcal{T}$ with $0 < \mathcal{T} < \mathcal{T}$ by the designed asynchronous Lur'e-type SMC law (8). The robust term $\beta_1(\varpi)$ is chosen as

\[
\beta_1(\varpi) = \gamma_1 + \alpha_1 \| x(\varpi) \| + \alpha_2 \| x(\varpi - \tau) \| + \alpha_3 \delta,
\]

(53)

in which $\alpha_1 = \| (G_iB_i)^{-1} G_i A_i \|$, $\alpha_3 = \| (G_i(B_i)^{-1} G_i A_i) \|$, $\alpha_2 = \| (G_i B_i)^{-1} G_i A_i \|$, and the adjust scalar $\gamma_i$ is determined by:

\[
\gamma_i > \frac{\| G_i x(0) \|}{\mathcal{T} \sqrt{\min \{ (G_i B_i)^{-1} G_i A_i \}}}
\]

(55)

Proof: Consider the following Lyapunov-Krasovskii functional as:

\[
V_3(x(\varpi)) = \frac{1}{2} s^T(\varpi)s(\varpi).
\]

Then, it follows from (7) that

\[
\dot{V}_3(x(\varpi)) = s^T(\varpi) \dot{s}(\varpi)
\]

\[
= s^T(\varpi) G_i [A_i x(\varpi) + A_{di} x(\varpi - \tau) + B_i u(\varpi)]
\]

\[
+ F_i \psi(M_1 x(\varpi) + D_i w(\varpi) - B_i K_i x(\varpi))
\]

\[
- B_i E_i \psi(M_1 x(\varpi)) - F_i \psi(M_1 x(\varpi))
\]

\[
= [G_i B_i s^T(\varpi)] [G_i B_i s^T(\varpi)] + \psi(\varpi)
\]

\[
+ (G_i B_i)^{-1} G_i A_{di} x(\varpi - \tau) + (G_i B_i)^{-1} G_i D_i w(\varpi)
\]

\[
- K_i x(\varpi) - E_i \psi(M_1 x(\varpi))
\]

\[
= [G_i B_i s^T(\varpi)] [G_i B_i s^T(\varpi)] + K_i x(\varpi)
\]

\[
+ (G_i B_i)^{-1} G_i A_{di} x(\varpi - \tau) + E_i \psi(M_1 x(\varpi))
\]

\[
- \beta_1(\varpi) \text{sign}(s(\varpi))
\]

\[
+ (G_i B_i)^{-1} G_i D_i w(\varpi) - K_i x(\varpi) - E_i \psi(M_1 x(\varpi))
\]

\[
\leq \| B^T G_i s^T(\varpi) \| \| (G_i B_i)^{-1} G_i A_i \| \| x(\varpi) \|
\]

\[
+ \| (G_i B_i)^{-1} G_i A_{di} \| \| x(\varpi - \tau) \| - \beta_1(\varpi) \text{sign}(s(\varpi))
\]

Denote $X_i = P_i^{-1}, Y_i = K_i X_i, \mathcal{F}_{ii} = X_i Q_i X_i, \mathcal{F}_{2i} = X_i Q_i X_i, \mathcal{F}_{3i} = X_i W_i X_i, \mathcal{F}_{4i} = X_i O_3 X_i, Z_1 = O_1^{-1}, Z_2 = O_2^{-1}$, $\mathcal{F}_{3i} = X_i O_3 X_i$.

Then, the congruence transformation of (10) can be performed by a block-diagonal matrix $\text{diag}(P_i^{-1}, P_i^{-1}, I, I, I, I, P_i^{-1})$. Based on Schur's complement properties, we can get (44). Similarly, we can obtain (45) by taking (32) through a congruence transformation with a block-diagonal matrix $\text{diag}(P_i^{-1}, P_i^{-1}, I, I, I)$.

Subsequently, in view of Schur's complement, it is easily to achieve LMIs (46)-(49) under conditions (11)-(13) and (33). This completes the proof.

Remark 3: It is easily checked that $0 \leq (X_i - Q_i) Q_i^{-1} (X_i - Q_i)$ is a symmetric matrix, which implies $-X_i Q_i^{-1} X_i^T \leq -X_i - X_i^T + Q_i$. Following this method, one immediately obtains $-\tau P_i^{-1} R_i^{-1} P_i^{-1} \leq -X_i X_i^T - R$.
+ \|(G_i B_i)^{-1} G_i D_i inf\| w(\sigma)\|\|
\leq -\gamma_i \|(B_i^T G_i^T s(\sigma))\|
\leq -\bar{\gamma} V_i^2(x(\sigma)), \quad (56)

where \(\bar{\gamma} = \sqrt{2} \gamma_i \{\lambda_{min}(G_i B_i B_i^T G_i^T)\} > 0\). As shown in (56), there exists a positive scalar \(\mathcal{T}_1\):

\[\mathcal{T}_1 \leq \frac{2V_1^2(0)}{\bar{\gamma}_i}. \quad (57)\]

such that \(V_3(x(\sigma)) = 0\) when \(\sigma \geq \mathcal{T}_1\), and we have \(s(\sigma) = 0\). By (55), it follows that

\[V_3(0) \leq \frac{\|G_i x(0)\|}{\gamma_i \{\lambda_{min}(G_i B_i B_i^T G_i^T)\}}. \quad (58)\]

Substituting (58) into (57), it yields:

\[\mathcal{T}_1 \leq \frac{\|G_i x(0)\|}{\gamma_i \{\lambda_{min}(G_i B_i B_i^T G_i^T)\}}. \quad (59)\]

Under condition of (54), the following inequality can be derived by (59)

\[\mathcal{T}_1 \leq \mathcal{T}. \quad (60)\]

Thus, we can also find a time \(\mathcal{T}_1\) within \([0, \mathcal{T}]\) to guarantee the accessibility of the sliding mode surface. This completes the proof.

### IV. Illustrative Example

In this section, a DC motor circuit [37] is used to evaluated the effectiveness of the proposed methodology, which is shown in Fig. 2. The dynamics is described by:

\[
\begin{align*}
\frac{d}{dt} & (\sigma) = A_i \sigma + B_i u(\sigma) + D_i w(\sigma) + F_i \psi(s(\sigma)) \\
\phi & = \lambda_1 x(\sigma) + \lambda_2 y(\sigma)
\end{align*}
\]

(63)

By (55), it follows that

\[\mathcal{T}_1 \leq \mathcal{T}. \quad (57)\]

Thus, we can also find a time \(\mathcal{T}_1\) within \([0, \mathcal{T}]\) to guarantee the accessibility of the sliding mode surface. This completes the proof.

The parameters are given in Table II.

Considering the time-delays, disturbances and nonlinearities involved, we have the following system

\[
\begin{align*}
\dot{x}(\sigma) & = A_i x(\sigma) + B_i u(\sigma) \\
\lambda_1 x(\sigma) + \lambda_2 y(\sigma) & = \lambda_1 x(\sigma)
\end{align*}
\]

(63)

Then, we assume that system (63) and the designed controller contain two modes respectively, i.e., \(N = 2\) and \(M = 2\). In addition, we select the initial conditions as \(x_1(0) = [0.4; 0.1]\) and \(\mathcal{T} = 0.1\). The other relevant parameters are given in Table III.

Besides, the random nonlinearity function \(\psi(\cdot)\) is taken as \(\psi(M_1 x(\sigma)) = 0.5 \Omega M_1 x(\sigma) (0.1 + 0.2 \cos(M_1 x(\sigma)))\) with \(\Omega = 2.6\). Then, by solving LMIs (44) – (52) in Theorem 3, and the circularly optimizing parameter \(\epsilon'\) is chosen as \(\epsilon'_{min} = 1.9683\), we have:

\[
\begin{align*}
K_1 & = [-183.2166 - 243.1673], \\
K_2 & = [-182.4272 - 263.5429], \\
E_1 & = [0.1233 0.0444], \\
E_2 & = [0.1233 0.0444].
\end{align*}
\]
Then the asynchronous Lur'e-type SMC law in (8) can be calculated as:

\[
u(\omega) = \begin{cases} 
= [-183.2166 - 243.1673]x(\omega) + [0.1233 \\
0.0444] \varphi(M_1x(\omega)) - \beta_1(\omega) \text{sign}(s(\omega)) \\
i = 1, l = 1 \\
= [-182.4272 - 263.5429]x(\omega) + [0.1233 \\
0.0444] \varphi(M_1x(\omega)) - \beta_2(\omega) \text{sign}(s(\omega)) \\
i = 2, l = 2 \\
= [-183.2166 - 243.1673]x(\omega) + [0.1233 \\
0.0444] \varphi(M_1x(\omega)) - \beta_1(\omega) \text{sign}(s(\omega)) \\
i = 1, l = 1 \\
= [-182.4272 - 263.5429]x(\omega) + [0.1233 \\
0.0444] \varphi(M_1x(\omega)) - \beta_2(\omega) \text{sign}(s(\omega)) \\
i = 2, l = 2 
\end{cases}
\]

The simulation results are shown in Figs. 3-7, where Fig. 3 depicts a possible sequence of the system/controller modes.

Thus, the sliding surface in (7) is:

\[
s(\omega) = \begin{cases} 
= [0, 2]x(\omega) \\
- \int_0^t [-1.10680.5774] \varphi(M_1x(s)) ds \\
- \int_0^t [-732.8664 - 972.6690] x(s) ds i = 1, l = 1 \\
= [0, 2]x(\omega) \\
- \int_0^t [-1.10680.5774] \varphi(M_1x(s)) ds \\
- \int_0^t [-729.7 - 1054.2] x(s) ds i = 1, l = 2 \\
= [0, 2]x(\omega) \\
- \int_0^t [1.0932 - 0.6226] \varphi(M_1x(s)) ds \\
- \int_0^t [-732.8664 - 972.6690] x(s) ds i = 2, l = 1 \\
= [0, 2]x(\omega) \\
- \int_0^t [1.0932 - 0.6226] \varphi(M_1x(s)) ds \\
- \int_0^t [-729.7 - 1054.2] x(s) ds i = 2, l = 2 
\end{cases}
\]

Select the adjustable parameters \(\gamma_1 = 0.0018\) and \(\gamma_2 = 0.0015\). Then the asynchronous Lur'e-type SMC law in (8) can be
The responses of the sliding variable $s(\sigma)$ and the asynchronous SMC input $u(\sigma)$ are depicted in Fig. 4 and Fig. 5, respectively. From Figs. 6-7, we can see that MJLSs (2) is FTS during the finite-time interval with the designed asynchronous controller.

V. CONCLUSION

This paper investigates the asynchronous SMC problem of a class of MJLSs. Firstly, the asynchronous problem of system modes and controller modes is modeled by HMM. Then, asynchronous sliding mode controller with Lur’e nonlinear information is designed to guarantee the reachability of sliding surface. Finally, according to the partitioning strategy, sufficient conditions are given to ensure the finite-time stabilization of the MJLSs. Future work will be directed at integrating the CHMM theory with SMC and fuzzy control.

Funding The work is supported by the National Natural Science Foundation of China (Basic Science Center Program: 61988101), National Natural Science Foundation of China (61725301), International (Regional) Cooperation and Exchange Project(61720106008).

Data availability statement The data that support the findings of this study are available from the corresponding author upon reasonable request.

Conflict of interest The authors declare that they have no conflicts of interest.

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