Statistical Mechanical Analysis of Low-Density Parity-Check Codes on General Markov Channel

Ryuhei Mori †

Abstract—Low-density parity-check (LDPC) codes on symmetric memoryless channels have been analyzed using statistical physics by several authors. In this paper, statistical mechanical analysis of LDPC codes is performed for asymmetric memoryless channels and general Markov channels. It is shown that the saddle point equations of the replica symmetric solution for a Markov channel are equivalent to the density evolution of the belief propagation on the factor graph representing LDPC codes on the Markov channel. The derivation uses the method of types for Markov chain.

Keywords—Low-density parity-check codes, Markov channel, replica method, method of types, large deviations.

1 Introduction

The replica method is a tool for the evaluation of free energy in statistical physics. Although the mathematical rigorousness of the replica method has not been proved, the replica method has been used not only for problems of statistical physics but for problems in several areas including information theory [1]. The analysis of low-density parity-check (LDPC) codes using the replica method is shown for memoryless symmetric channels by Murayama et al. [2] and Montanari [3]. After some period, LDPC codes on binary-output asymmetric channel is analyzed by using the replica method by Neri et al. [4]. On the other hand, LDPC codes on channel with memory is also of great interest in wireless communication and magnetic recording [5]. LDPC codes on restricted types of finite-state Markov channel are also analyzed by Neri and Skantzos [6]. In this paper, we analyze LDPC codes on general asymmetric memoryless channel and general Markov channel by using the replica method. In the analysis, we use the method of types for Markov chain in order to generalize the method recently obtained in [7].

Despite lack of mathematical rigour, the replica method is known to yield rigorous bounds in some cases [8, 9]. Moreover, a tight bound of MAP threshold, which is derived in an elegant way in [10], is found to empirically coincide with the result of the replica method.

2 Memoryless channel

2.1 LDPC codes and the number of codewords

In this paper, we deal with regular LDPC codes. Generalization for irregular LDPC codes is straightforward as in [11]. We consider factor graph (Tanner graph) including $N$ variable nodes and $N(1-R)$ factor nodes (check nodes). The degrees of variable nodes and factor nodes are $l$ and $r$, respectively, for which the relation $R = 1 - l/r$ holds. Edge connection is randomly chosen uniformly from $(N!)^l$ possible connections. Let $\mathcal{X} = \{0, 1\}$ and $f(x) : \mathcal{X}^n \to \{0, 1\}$ be a factor function taking 1 when the number of 1s in $x$ is even, and taking 0 otherwise. Let $x_{\partial a}$ be the neighborhood of a factor node $a$. The uniform distribution on codewords $x \in \mathcal{X}^N$ of the LDPC code defined by the Tanner graph is given by

$$p(x) := \frac{1}{Z_0} \prod_a f(x_{\partial a})$$

$$Z_0 := \sum_{x \in \mathcal{X}^N} \prod_a f(x_{\partial a}) .$$

Here, $Z_0$ is a constant for the normalization, which in this case is the number of codewords in an LDPC code. Let $E[\cdot]$ be the expectation with respect to the edge connections. In [7], the following lemma is shown for general $f(x)$.

Lemma 1.

$$\lim_{N \to \infty} \frac{1}{N} \log E[Z_0] = \max_{(m_{\tau \to}, m_{\tau \to \tau}) \in \mathcal{R}} \left\{ \log Z_e + \frac{l}{r} \log Z_t - l \log Z_e \right\}$$

(1)

where $\mathcal{R}$ denotes the set of saddle points of the function for which the maximum is taken, and where

$$Z_e := \sum_{x \in \mathcal{X}^l} m_{\tau \to}(x)^l$$

$$Z_t := \sum_{x \in \mathcal{X}^r} f(x) \prod_{i=1}^r m_{\tau \to \tau}(x_i)$$

$$Z_e := \sum_{x \in \mathcal{X}^l} m_{\tau \to}(x) m_{\tau \to \tau}(x).$$

The saddle point equations determining the two probability functions $m_{\tau \to}(x)$, $m_{\tau \to \tau}(x)$, both defined on $\mathcal{X}$, are

$$m_{\tau \to \tau}(x) \propto m_{\tau \to}(x)^{-1}$$

$$m_{\tau \to}(x) \propto \frac{1}{r} \sum_{i=1}^r \sum_{x' \neq x} f(x') \prod_{j=1, j \neq i}^r m_{\tau \to \tau}(x_j).$$

(2)

In the case of regular LDPC codes, it is proved [11] that the maximizer for the right-hand side of (1) is given by $m_{\tau \to}(x) = m_{\tau \to \tau}(x) = 1/2$, for which

$$\lim_{N \to \infty} \frac{1}{N} \log E[Z_0] = R .$$

(3)

Since each independent factor reduces the number of codewords by half, the inequality $(1/N) \log Z_0 \geq R$ always holds. Hence, for regular LDPC codes,

$$\lim_{N \to \infty} \frac{1}{N} \log E[Z_0] = R$$

and for any $\varepsilon > 0$,

$$\frac{1}{N} \log Z_0 \leq R + \varepsilon , \quad \text{with probability } 1 - o(1) .$$

(4)

The result of this paper can be generalized to any irregular LDPC code ensemble satisfying (3).
2.2 Replica analysis
In this subsection, we consider LDPC codes transmitted on an asymmetric memoryless channel. Let \( Y \) be the output alphabet and \( W(y) \) be a probability of an output \( y \) in \( Y \) when \( x \in X \) is transmitted. The a posteriori distribution of a codeword \( x \in X^N \) given an output \( y \in Y^N \) is
\[
p(x | y) := \frac{1}{Z} \prod_a \prod_i f(x_{i^a}) \prod_{i=1}^N W(y_i | x_i).
\]
The aim of this paper is to evaluate
\[
E[\log Z] = \lim_{N \to \infty} E[\log Z]
\]
which gives the expected conditional entropy of codeword given an output \( Y \). The quantity \( E[\log Z] \) is evaluated via the replica method. First, from (1), \( m_0 \) in the definition of \( p(y) \) can be replaced by \( 2^{N\Phi} \). Then, one obtains
\[
Z_n = \sum_{x \in (X^n)^Y} \prod_{k=1}^N \prod_{i=1}^n p(x_{i^k}, x_i) \prod_{i=1}^n W(y_i | x_i)
\]
The replica symmetry (RS) assumption that we use is
\[
\begin{align}
m_{v \to s}(x) &= m_{v \to s}(x) \int \prod_{i=1}^n M_{v \to s}(x_i) d\Phi(M_{v \to s} | x_0) \\
m_{v \to s}(x) &= m_{v \to s}(x) \int \prod_{i=1}^n M_{v \to s}(x_i) d\Phi(M_{v \to s} | x_0)
\end{align}
\]
where \( \Phi \) and \( \Phi \) are distributions of measures on \( X \). In (7), the notation \( x_i \) is used instead of \( x^{(i)} \) for \( i \in \{0, \ldots, n\} \). From (3), \( R \) in the right-hand side of (6) can be replaced by the right-hand side of (1). On the RS assumption (7), we obtain the following lemma.
Lemma 2. For any \( n \in \mathbb{R} \),
\[
\lim_{N \to \infty} \frac{1}{N} \log E[\log Z] = \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \log Z_n + \frac{1}{r} \log Z_t(n) - \log Z_c(n) \right\}
\]
where \( \text{extr}_{x} \{ F(x) \} \) denotes \( F(x^*) \) where \( x^* \) is the saddle point of \( F(x) \), and where
\[
Z_n := \sum_{x \in \mathcal{X}} \prod_{i=1}^n \prod_{j=0}^{r} f\left( x_{i^j} \right) \prod_{i=1}^n W\left( y_i | x_i \right)
\]
From this lemma, the RS solution is obtained straightforwardly (7).

\[ \text{Theorem 3.} \]
\[
\lim_{N \to \infty} \frac{1}{N} E[\log Z] = \lim_{n \to 0} \frac{1}{n} \log E[\log Z]
\]
\[
= \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n m_{v \to s}(x_i) \right\} \right\}
\]
\[
+ \sum_{x \in \mathcal{X}} \sum_{x_{i^j} \in \mathcal{X}} \prod_{j=0}^{r} m_{v \to s}(x_{i^j}) \sum_{y \in \mathcal{Y}} \prod_{i=1}^n W(y | x_i) \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n M_{v \to s}(x_i) \right\} \right\}
\]
\[
\sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \sum_{x_{i^j} \in \mathcal{X}} \prod_{j=0}^{r} m_{v \to s}(x_{i^j}) \sum_{y \in \mathcal{Y}} \prod_{i=1}^n W(y | x_i) \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n M_{v \to s}(x_i) \right\} \right\}
\]
\[
\frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n m_{v \to s}(x_i) \right\}
\]
\[
\sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \sum_{x_{i^j} \in \mathcal{X}} \prod_{j=0}^{r} m_{v \to s}(x_{i^j}) \sum_{y \in \mathcal{Y}} \prod_{i=1}^n W(y | x_i) \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n M_{v \to s}(x_i) \right\} \right\}
\]
\[
\frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n m_{v \to s}(x_i) \right\}
\]
\[
\sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \sum_{x_{i^j} \in \mathcal{X}} \prod_{j=0}^{r} m_{v \to s}(x_{i^j}) \sum_{y \in \mathcal{Y}} \prod_{i=1}^n W(y | x_i) \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n M_{v \to s}(x_i) \right\} \right\}
\]
\[
\frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n m_{v \to s}(x_i) \right\}
\]
\[
\sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \sum_{x_{i^j} \in \mathcal{X}} \prod_{j=0}^{r} m_{v \to s}(x_{i^j}) \sum_{y \in \mathcal{Y}} \prod_{i=1}^n W(y | x_i) \left\{ \text{extr}_{(\Phi, \Phi, m_{v \to s}, m_{v \to s})} \left\{ \frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n M_{v \to s}(x_i) \right\} \right\}
\]
\[
\frac{1}{r} \sum_{x \in \mathcal{X}} \sum_{x_i \in \mathcal{X}} \int f(x) \prod_{i=1}^n m_{v \to s}(x_i) \right\}
\]
where the saddle point equations are (2) and

\[ m_{\rightarrow r}(x) \delta(M_{\rightarrow r} | x) = \frac{1}{r} \sum_{i=1}^{r} \sum_{x_i \in X^r} f(x) \prod_{i=1, i \neq k}^{r} m_{\rightarrow r-i}(x_i) \delta(M_{\rightarrow r-i} | x_i) \]

\[ \Phi(M_{\rightarrow r} | x) = \frac{1}{r} \sum_{y \in Y} W(y | x) \sum_{i=1}^{r} \delta(M_{\rightarrow r-i} | x) \delta(M_{\rightarrow i} | x) \]

As mentioned above, all \( m_{\rightarrow r}(x) \) and \( m_{\rightarrow r-i}(x) \) in Theorem 3 are equivalent to the density evolution of belief propagation for asymmetric memoryless channels [13]. Although the RS assumption is sufficient for symmetric channels [13], for asymmetric channels, 1-step replica symmetry breaking (1RSB) should be considered generally [4]. The equations for 1RSB solutions are obtained straightforwardly from (2) and the 1RSB assumption. However, numerically solving the saddle point equations for 1RSB solutions is much more elaborate. For LDPC codes, there exists a frozen solution which is the minimum of the RS solution with respect to temperature [3, 4]. While the frozen solution can be relatively easily obtained like the RS solution, it is not obvious whether the frozen solution is the correct solution [4].

3 Markov channel

In this section, we consider LDPC codes on the general Markov channel. The space of states is denoted by \( S \). The state-dependent channel is denoted by \( W(y | x, s) \), and the transition probability of states by \( V(s' | y, x, s) \), for \( y \in Y, x \in X, s, s' \in S \). The probability distribution of the initial state \( s \in S \) is denoted by \( V_0(s) \). When the transition probability is independent of \( y \), we call the channel an intersymbol-interference channel. When the transition probability is independent of \( y \) and \( x \), we call the channel a finite-state Markov channel. The a posteriori probability of a codeword \( x \in X^N \) given an output \( y \in Y^N \) is

\[ p(x | y) := \frac{1}{Z} \sum_{x \in X^N} \prod_{u=1}^{r} f(x_u) \prod_{i=1}^{N} W(y_i | x_i, s_i) \cdot V_0(s_1) \prod_{i=1}^{N-1} V(s_{i+1} | y_{i+1}, s_i) \]

\[ Z := \sum_{x \in X^N} \sum_{s \in S^N} \prod_{u=1}^{r} f(x_u) \prod_{i=1}^{N} W(y_i | x_i, s_i) \cdot V_0(s_1) \prod_{i=1}^{N-1} V(s_{i+1} | y_{i+1}, s_i) \]

Similarly to the previous subsection, let \( E[S] \) denote the expectation with respect to \( y \in Y^N \) obeying

\[ p(y) := \frac{1}{Z_0} \sum_{x \in X^N} \prod_{u=1}^{r} f(x_u) \prod_{i=1}^{N} W(y_i | x_i, s_i) \cdot V_0(s_1) \prod_{i=1}^{N-1} V(s_{i+1} | y_{i+1}, s_i) \]

As in the previous subsection, we consider (5) with \( Z_0 \) replaced by \( 2^{NR} \). Then, one obtains

\[ E[Z^n] = \frac{1}{2^{NR}} \sum_{x \in (X^l)^N} \sum_{s \in (S^l)^N} \left( \prod_{k=0}^{n} f(x_{2k}) \right) \prod_{k=0}^{n} \sum_{y \in Y} W(y | x_{2k+1}, s_{2k+1}) V(s_{2k+1} | y, x_{2k}, s_{2k}) \cdot V_0(s_1) \prod_{i=1}^{N} V(s_{i+1} | y_{i+1}, s_i) \]

Let \( T_n \subseteq S^{n+1} \) be the largest set such that for any \( s_2 \in T_n \)

\[ \sum_{x \in X^l \times Y \times S^2} \sum_{y \in Y} W(y | x_{2j+1}, s_{2j+1}) V(s_{2j+1} | y, x_{2j}, s_{2j}) > 0. \]

In order to use Sanov’s theorem [15] from the method of types for Markov chain [16][18], we assume that the Markov chain on \( T_n \) defined by the transition probabilities

\[ Q_n(s_1 | s_2) := \sum_{x \in X^l \times Y \times S^2} \sum_{y \in Y} W(y | x_{2j+1}, s_{2j+1}) V(s_{2j+1} | y, x_{2j}, s_{2j}) \]

is irreducible for each \( n \in [0, 1, 2, \ldots] \). Then, we obtain the following maximization problem like LDPC codes for memoryless channel [7].

\[ \lim_{N \to \infty} \frac{1}{N} \log E[Z^n] = \sup \left\{ H(X_1, S_1 | X_2, S_2) - l IH(X_1, S_1) + I - H(U_1, \ldots, U_r, T_1, \ldots, T_r) \right\} \]

Here, \( X_1 \) and \( X_2 \) are random variables on \( \mathcal{X}^{n+1} \), \( S_1 \) and \( S_2 \) are random variables on \( T_n \) or equivalently on \( S^{n+1} \). \( U_i \) and \( T_i \) are random variables on \( \mathcal{X}^{n+1} \) and \( S^{n+1} \), respectively, for \( i \in \{ 1, \ldots, r \} \). The notation \( \langle \cdot \rangle \) denotes the expectation with respect to the random variables. The supremum is taken with respect to \( (X_1, S_1, X_2, S_2) \) and \( (U_1, \ldots, U_r, T_1, \ldots, T_r) \) on the following conditions

- \( (X_1, S_1) \) and \( (X_2, S_2) \) have the same distribution
- \( (X_1, S_1) \) and \( (U_k, T_k) \) have the same distribution where \( K \) denotes the uniform random variable on a set \( \{ 1, \ldots, r \} \). By using the variational method [19, 7], we obtain the following lemma.

Lemma 4. For \( n \in [0, 1, 2, \ldots] \),

\[ \lim_{N \to \infty} \frac{1}{N} \log E[Z^n] = \max \left\{ \log Z_n - \log Z_n + \frac{I}{r} \log Z_r - I \log Z_r - R \right\} \]
where $\mathcal{R}$ denotes the set of saddle points of the function for which the maximum is taken, and where

$$
Z_m := \sum_{x_1, s_1, x_2, s_2} \left( \sum_{y \in \mathcal{Y}} \int_{x \in \mathcal{X}} W(y \mid x, s) W(s_2 \mid y, s_2) V(y \mid x_1) \right) \cdot m_{R \rightarrow v}(x, s_1) m_{R \rightarrow v}(x_1).
$$

$$
Z_0 := \sum_{x, s} m_{R \rightarrow v}(x, s) m_{R \rightarrow v}(x).
$$

$$
Z_s := \sum_{x} m_{R \rightarrow v}(x) m_{R \rightarrow v}(x).
$$

The saddle point equations are

$$
m_{R \rightarrow v}(x) = \left( \sum_{y \in \mathcal{Y}} \int_{x \in \mathcal{X}} W(y \mid x, s) m_{R \rightarrow v}(s) m_{R \rightarrow v}(x) \right)^{\lambda} - 1.
$$

$$
m_{R \rightarrow v}(x) = \sum_{i=1}^{n} \sum_{l=1, l \neq i}^{n} \int_{x \in \mathcal{X}} W(y \mid x, s) V(y \mid x, s) \cdot m_{R \rightarrow v}(x_1).
$$

$$
m_{R \rightarrow v}(x, s) = \sum_{x_1, s_1} \left( \sum_{y \in \mathcal{Y}} \int_{x \in \mathcal{X}} W(y \mid x, s) V(y \mid x, s) \right) \cdot m_{R \rightarrow v}(x_1).
$$

$$
m_{R \rightarrow v}(x, s) = \sum_{x_1, s_1} \left( \sum_{y \in \mathcal{Y}} \int_{x \in \mathcal{X}} W(y \mid x, s) V(y \mid x, s) \right) \cdot m_{R \rightarrow v}(x).
$$

Here, the domains of sums $x, x_1, x_2 \in \mathcal{X}^{n+1}, s, s_1, s_2 \in \mathcal{S}^{n+1}$ are omitted.

Although a proof of this lemma is omitted for lack of space, the derivation is similar to the memoryless case in [7].

**Lemma 5.** For $n = 0$, $m_{R \rightarrow v}(x, s)$ is uniform. Hence, the stationary condition is given by (2) and

$$
m_{R \rightarrow v}(x, s) \propto \sum_{y \in \mathcal{Y}} \int_{s_1} W(y \mid x, s_2) W(s_1 \mid y, s_2) V(y \mid x_1) \cdot m_{R \rightarrow v}(x_1).
$$

Proof: By considering (3), one can confirm that $X_1$ and $S_1$ (also $X_2$ and $S_2$) should be independent for $n = 0$. As a consequence, it turns out that $m_{R \rightarrow v}(x, s)$ is also an independent measure (This part relates to the proof of Lemma[4] Hence, the proof is omitted for lack of space). Then, $\sum_{s \in \mathcal{S}} m_{R \rightarrow v}(x, s)$ should be uniform, which is the unique right eigenvector corresponding to the eigenvalue 1 of the stochastic matrix of irreducible Markov chain $Q_0$. Then, one straightforwardly obtains that $\sum_{s \in \mathcal{S}} m_{R \rightarrow v}(x, s)$ is also uniform. $\square$

For finite-state Markov channels, (2) implies that $m_{R \rightarrow v}(x, s)$ is the stationary distribution of the Markov chain $V(s_1 \mid s_2)$. The RS assumption for this problem is

$$
m_{R \rightarrow v}(x) = \int_{s \in \mathcal{S}} W(x \mid s) m_{R \rightarrow v}(s) m_{R \rightarrow v}(x).
$$

Similarly to the previous subsection, we obtain the RS solution as follows, which is the main result of this paper.

**Theorem 6.** The RS solution is given by (10). The saddle point equations are (2), (9) and (13).
Although the saddle point equations in [6] are obtained by simplifying Theorem 6 for finite-state Markov channels, they are omitted in this paper for lack of space. Similarly to the memoryless case, we have to deal with RSB effect generally. Fortunately, the frozen solution does not generally exist. The necessary and sufficient condition of existence of the frozen solution is that for any \( y \in \mathcal{Y} \), when variables \( x_2, s_2 \) are fixed, at most one value \( s_1 \) exists such that

\[
W(y \mid x_2, s_2) V(s_1 \mid y, x_2, s_2) > 0
\]

and when \( s_1 \) is fixed at most one pair of values \((x_2, s_2)\) satisfying (11) exists. This condition is called hard constraint in [14]. When a state depends only on the previous state or input, we can obtain a simpler expression of RS solution than Theorem 6, and the necessary and sufficient condition for the existence of the frozen solution becomes weaker.

4 Numerical calculation

While Theorem 6 gives the expected conditional entropy on general Markov channel, e.g., trapdoor channel, generally we have to deal with densities of messages similarly to the memoryless case [11], making implementation of the density evolution involved. In [20], generalized erasure channel is defined as a channel where density evolution can be described by one parameter. As the simplest generalized erasure channel, the dicode erasure channel (DEC) is introduced. For \( \mathcal{X} = \mathcal{S} = \{0, 1\} \) and \( \mathcal{Y} = \{-1, 0, 1, \ast\} \), the DEC(\( \varepsilon \)) is defined as

\[
W(y \mid x, s) = \begin{cases} 1 - \varepsilon, & y = x - s \\ \varepsilon, & y = * \end{cases}
\]

for \( \varepsilon \in [0, 1] \). For the DEC, \( m_{l \rightarrow s}(s) = 1/2 \). We modify Theorem 6 for the case where the state only depends on the previous input, on the basis of which the RS solution for the DEC(\( \varepsilon \)) is obtained without dealing with densities as

\[
\lim_{N \to \infty} \frac{1}{N} \mathbb{E} [\log Z] = \min_{\varepsilon \in [0, 1]} \left\{ 2(\varepsilon \log \varepsilon + (1 - \varepsilon) \log (1 - \varepsilon)) - \left[ \varepsilon R_{\varepsilon \rightarrow s} + (1 - \varepsilon) \varepsilon L_{\varepsilon \rightarrow s} \right] \log 2 - \varepsilon \log \varepsilon - (1 - \varepsilon) \log (1 - \varepsilon) + \left[ \varepsilon R_{\varepsilon \rightarrow s} + (1 - \varepsilon) \varepsilon L_{\varepsilon \rightarrow s} + \varepsilon L_{\varepsilon \rightarrow s} \right] \log 2 \right\}
\]

where the saddle point equations are

\[
\begin{align*}
\varepsilon_{s \rightarrow s} &= 1 - (1 - \varepsilon_{s \rightarrow s})^{-1} \\
\varepsilon_{s \rightarrow t} &= \varepsilon_{t \rightarrow t}^{-1} \varepsilon_{s \rightarrow v} \left( \varepsilon + \varepsilon_{s \rightarrow s} \right) \\
\varepsilon_{\varepsilon \rightarrow s} &= \varepsilon + (1 - \varepsilon \varepsilon_{\varepsilon \rightarrow s}) \\
\varepsilon_{\varepsilon \rightarrow s} &= \varepsilon_{s \rightarrow t}^{-1} \varepsilon_{t \rightarrow t} \varepsilon_{s \rightarrow v} \left( \varepsilon + \varepsilon_{t \rightarrow t}^{-1} \varepsilon_{s \rightarrow s} \right)
\end{align*}
\]

(12)
This is equivalent to the density evolution of the joint iterative decoding, which we call the belief propagation (BP) decoding, on the DEC \( \epsilon \) [20]. For the DEC \( \epsilon \), the expected conditional entropy is

\[
\lim_{N \to \infty} \frac{1}{N} \mathbb{E}[H(X | Y)] = \left( \lim_{N \to \infty} \frac{1}{N} \mathbb{E}[\log Z] \right) = -\epsilon \log \epsilon - (1 - \epsilon) \log(1 - \epsilon)
\]

\[
= -\epsilon \theta_{r \to s} + \epsilon \theta_{L \to s} + \frac{r - 1}{r} \theta_{e_{\infty} \to e_{s}} (1 - \epsilon) \epsilon (1 - \epsilon)
\]

where \( \theta_{r \to s}, \theta_{L \to s} \) and \( \theta_{e_{\infty} \to e_{s}} \) satisfy (12).

In Fig. 1, the results for (2, 4), (3, 6), (4, 8), (5, 10) and (6, 12) regular ensembles are shown. When \( \epsilon \) is small, there exists only a trivial saddle point, which always gives 0 expected conditional entropy. A non-trivial saddle point appears above BP threshold. However, similarly to the memoryless case, except for (2, 4)-regular ensembles, the non-trivial saddle point yields negative expected conditional entropy for some regions as in Fig. 1. In the region, while BP decoding fails, MAP decoding succeeds since the trivial saddle point should be chosen for the non-negative conditional entropy above the MAP threshold [21].

The method in this paper can be generalized to other models, e.g., IRA/ARA LDPC codes [21]. CDMA on a channel with memory, compressed sensing of a Markov source, etc.

Figure 1: The expected conditional entropy of LDPC codes on the DEC calculated by the replica method. The results for (2, 4), (3, 6), (4, 8), (5, 10) and (6, 12) regular ensembles are plotted.

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