Double SHA-256 Hardware Architecture
With Compact Message Expander for
Bitcoin Mining

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ABSTRACT In the Bitcoin network, computing double SHA-256 values consumes most of the network energy. Therefore, reducing the power consumption and increasing the processing rate for the double SHA-256 algorithm is currently an important research trend. In this paper, we propose a high-data-rate low-power hardware architecture named the compact message expander (CME) double SHA-256. The CME double SHA-256 architecture combines resource sharing and fully unrolled datapath technologies to achieve both a high data rate and low power consumption. Notably, the CME algorithm utilizes the double SHA-256 input data characteristics to further reduce the hardware cost and power consumption. A review of the literature shows that the CME algorithm eliminates at least 9.68% of the 32-bit XOR gates, 16.49% of the 32-bit adders, and 16.79% of the registers required to calculate double SHA-256. We synthesized and laid out the CME double SHA-256 using CMOS 0.18 μm technology. The hardware cost of the synthesized circuit is approximately 13.88% less than that of the conventional approach. The chip layout size is 5.9 mm × 5.9 mm, and the correctness of the circuit was verified on a real hardware platform (ZCU 102). The throughput of the proposed architecture is 61.44 Gbps on an ASIC with Rohm 180nm CMOS standard cell library and 340 Gbps on a FinFET FPGA 16nm Zynq UltraScale+ MPSoC ZCU102.

INDEX TERMS Bitcoin mining, SHA-256, unrolling, ASIC.

I. INTRODUCTION

Bitcoin is the most popular cryptocurrency and was invented by Satoshi Nakamoto in 2008 [1], [2]. Leveraging blockchain technology, Bitcoin uses a distributed public ledger to record all transactions without any third party [3]. Each block added to the public distributed ledger is created by hashing a 1024-bit message, including a version number, a hash of the previous block, a hash of the Merkle root, timestamp, target value, and a nonce. In the 1024-bit message, the nonce must be valid to create a hashing output smaller than the specified target value. Therefore, miners relentlessly seek valid nonces when adding new blocks. The process of finding a valid nonce is called Bitcoin mining [4].

In Bitcoin mining, the double SHA-256 algorithm is used to compute the hash value of the bitcoin block header, which is a 1024-bit message. The use of double SHA-256 protects against the length extension attack [5]. Technically, SHA-256 consists of a message expander (ME) and a message compressor (MC). During the SHA-256 operation, the ME expands the 512-bit input message into 64 chunks of 32-bit data. The MC compresses these 64 32-bit data chunks into a 256-bit hashed output.

Most of the energy consumption required for maintaining the Bitcoin network stems from calculating double SHA-256 values. Therefore, reducing the hardware cost and energy consumption of the SHA-256 circuit is a popular research trend. In [6], the authors optimized the double SHA-256 operation for Bitcoin mining from an algorithmic perspective, but no hardware design was available to evaluate
the power consumption. From a hardware perspective, [7]–[22] proposed solutions to improve SHA-256. For instance, the authors of [7] employed the carry-save adder to improve the computation time of the critical path, which increased the maximum frequency and processing rate, while [8]–[12] used pipeline technology to improve the SHA-256 throughput. A cache memory technique was presented in [13] to reuse data, minimize the critical paths, and reduce the number of memory accesses for SHA-256 processing. The authors of [14] adopted the unfolding technique to reduce the computing latency for SHA-256. The authors of [15] proposed using a 7-3-2 array compressor to reduce the critical path delay for SHA-256. The carry-save adders technique is used in [16] to reduce the latency of additions in the SHA-256 algorithm. The authors of [17] used a combination of techniques such as carry-save-adders and pipelines to increase the performance of SHA-256. Pipeline and unrolled techniques are presented in [18] and [19] to increase the throughput of SHA-256. The authors of [20]–[22] presented a SHA-256 implementation on an FPGA for performance evaluation, with no technique optimization. Despite providing improvements in terms of hardware cost and power consumption, the hardware circuits developed in [7]–[22] have low processing rates because they require several (up to 64) clock cycles to compute a single 256-bit hash value.

To be applicable for Bitcoin mining, a SHA-256 circuit needs not only efficient hardware and power cost but also a high processing rate. To reach a high processing rate, the authors in [23] proposed the fully unrolled SHA-256 datapath for Bitcoin mining hardware. Additionally, the fully unrolled SHA-256 datapath can be designed to run on an application-specific integrated circuit (ASIC) [24], which can reach even higher processing rates. However, because an ASIC implementation of a fully unrolled datapath has high power consumption and hardware costs, [25]–[28] proposed eliminating an 8-round unrolled datapath in the double SHA-256 architecture to reduce the chip area. Furthermore, several technical solutions, such as carry-save adders and optimized message compressor (MC) architectures have been proposed and applied to reduce the hardware and power costs.

In this study, we propose a new approach for reducing the hardware cost and power consumption of high processing rate fully unrolled SHA-256 architecture. We analyze the characteristics of the 1024-bit input data of double SHA-256 and propose compact message expander (CME) algorithms that significantly reduce the hardware cost required to compute the message expander (ME) process of SHA-256. In addition, we propose a CME double SHA-256 accelerator architecture that adopts the proposed CME algorithms to reduce the power consumption. Our architecture generates one 256-bit hash value per clock cycle. We implemented the proposed double SHA-256 accelerator architectures in ASIC CMOS 0.18 \( \mu m \) technology to demonstrate their energy efficiency. The Verilog code and synthesized results of the experiment are publicly available from GitHub.

The remainder of this paper is organized as follows. Section II presents a preliminary study. Section III describes our proposed CME double SHA-256 architecture, and the CME algorithms and hardware circuits are explained in detail. Section IV reports our evaluation in terms of theory, ASIC, and FPGA experiments. Finally, Section V concludes the paper.

II. PRELIMINARIES

A. DOUBLE SHA-256 ARCHITECTURE FOR BITCOIN MINING

Fig. 1 shows the overview architecture of double SHA-256 applied for Bitcoin mining. The input to the double SHA-256 process is a 1024-bit message, which includes a 32-bit version, a 256-bit hash of the previous block, a 256-bit hash of the Merkle root, a 32-bit timestamp, a 32-bit target, a 32-bit nonce, and 384 bits of padding. The 1024-bit message is split into two 512-bit message parts; then SHA-256\(_1\) calculates a hash value of the first 512-bit message, and SHA-256\(_2\) computes a hash value of the final 512-bit message. Due to the double SHA-256 requirement, the 256-bit hash output from SHA-256\(_2\) must be compressed into the final 256-bit hash by using SHA-256\(_3\). In the Bitcoin mining process, the final 256-bit hash output from SHA-256\(_3\) is compared to the target value. If the final hash is smaller than the target value, the valid 32-bit nonce is specified, and a new Bitcoin block is successfully created. Otherwise, the 32-bit nonce is increased by one and the double SHA-256 circuit recomputes to find a new hash value. This process is repeated until the 256-bit hash of SHA-256\(_3\) meets the target requirement.

Computation inside all three blocks (SHA-256\(_1\), SHA-256\(_2\), and SHA-256\(_3\)) follows the SHA-256 algorithm, which has two processes: a message expander (ME) and a message compressor (MC).

Algorithm 1 shows the ME process, which expands the 512-bit input message into 64 chunks of 32-bit data \( W_j \) (0 \( \leq j \leq 63 \)). In the first 16 rounds, the ME parses the 512-bit message into 16 32-bit data chunks (denoted as \( W_j, j = 0 \) to 15 where \( j \) is the round index). In the final 48 rounds, the ME calculates 48 chunks of 32-bit data \( W_j \) (17 \( \leq j \leq 63 \)). Three 32-bit adders and two logical functions \( \sigma_0(x) \) and \( \sigma_1(x) \) are
Algorithm 1 Message Expander (ME)

- For \( j \) from 0 to 15 {
  \[ W_j = M_j \]
- For \( j \) from 16 to 63 {
  \[ W_j = \sigma_1(W_{j-2}) + W_{j-7} + \sigma_0(W_{j-15}) + W_{j-16} \]
}

needed to compute each \( W_j \) (\( 17 \leq j \leq 63 \)) value. Fig. 2 shows the conventional circuit \( C \) required to calculate \( W_j \) (\( 17 \leq j \leq 63 \)), in which the logical functions \( \sigma_0(x) \) and \( \sigma_1(x) \) are respectively defined as follows:

\[
\sigma_0(x) = S^7(x) \oplus S^{18}(x) \oplus R^3(x) \quad (1)
\]

\[
\sigma_1(x) = S^{17}(x) \oplus S^{19}(x) \oplus R^{10}(x) \quad (2)
\]

Algorithm 2 shows the MC process, which compresses the 64 chunks of \( W_j \) (\( 0 \leq j \leq 63 \)) into a 256-bit hash value. The process involves three main steps: initialization, loop, and add. In the initialization step, eight initial hash values (denoted as \( a, b, c, d, e, f, g, h \)) are assigned to eight initial hashes \( H_1, H_2, \ldots, H_8 \) defined by the SHA-256 algorithm. In the loop step, the internal hash values \( a, b, c, d, e, f, g, h \) are calculated and updated through 64 loops. To compute \( a, b, c, d, e, f, g, h \) in each loop, logical functions such as \( \Sigma_0(x), \Sigma_1(x), Ch(x, y, z), \) and \( Maj(x, y, z) \) are used.

\[
\Sigma_0(x) = S^2(x) \oplus S^{13}(x) \oplus S^{22}(x) \quad (3)
\]

\[
\Sigma_1(x) = S^{6}(x) \oplus S^{11}(x) \oplus S^{25}(x) \quad (4)
\]

\[
Ch(x, y, z) = (x \land y) \oplus (\neg x \land z) \quad (5)
\]

\[
Maj(x, y, z) = (x \land y) \oplus (x \land z) \oplus (y \land z) \quad (6)
\]

In the add step, the final hash is computed by adding the initial hashes \( H_1, H_2, \ldots, H_8 \) to the final internal hashes \( a, b, c, d, e, f, g, h \) resulting from the 64 loops.

Algorithm 2 Message Compressor (MC)

(1) Initialization:

- \( a = H_1; b = H_2; c = H_3; d = H_4; e = H_5; f = H_6; g = H_7; h = H_8 \)

(2) Loop:

  - For \( j \) from 0 to 63 {
    - \( T_1 = h + \Sigma_1(e) + Ch(e, f, g) + K_j + W_j \)
    - \( T_2 = \Sigma_0(a) + Maj(a, b, c) \)
    - \( h = g; g = f; f = e; e = d + T_1; d = c; c = b; b = a; a = T_1 + T_2 \)
  }

(3) Add:

- \( HO_1 = a + H_1; HO_2 = b + H_2; HO_3 = c + H_3; HO_4 = d + H_4; HO_5 = e + H_5; HO_6 = f + H_6; HO_7 = g + H_7; HO_8 = h + H_8 \)

B. THE PROTOTYPE DOUBLE SHA-256 ARCHITECTURE

To be applicable for Bitcoin mining, double SHA-256 hardware should provide a high processing rate. The current optimal solution is to develop and implement a double SHA-256 accelerator in ASIC chips. In [23], the authors proposed an ASIC-based double SHA-256 accelerator that implemented ME and MC processes in a fully unrolled datapath for high processing. Technically, the fully unrolled SHA-256 datapath enables the 64 rounds of ME and MC to run in parallel and be pipelined.

Fig. 3 illustrates a prototype SHA-256 architecture with 64-round unrolled datapaths for the MC and ME processes. The unrolled ME datapath is denoted as Block\( j \) (\( j = 0, \ldots, 63 \)), while the unrolled MC datapath is denoted as Loop\( j \) (\( j = 0, \ldots, 63 \)).

Because the goal of this study is to optimize the ME process, we focus specifically on a hardware implementation for ME. For the first 16 blocks (i.e., Block\( j \) (\( j = 0, \ldots, 15 \))), each ME block requires a 512-bit register (or 16 32-bit registers) to pipeline and store the 16 \( W_j \) (\( j = 0, \ldots, 15 \)) values. For the last 48 blocks, i.e., Block\( j \) (\( j = 16, \ldots, 63 \)), each block needs a 512-bit register (or 16 32-bit registers) and C circuits (Fig. 2) to compute \( W_j \) (\( j = 16, \ldots, 63 \)). As shown in Fig. 1, the double SHA-256 accelerator for Bitcoin mining requires three individual SHA-256 circuits. This means that the accelerator must implement \( 48 \times 3 = 144 \) C circuits (in the 16th to 63th blocks of SHA-256, SHA-2562, and SHA-2563). Thus, it is necessary to both optimize the C circuit and reduce the number of C circuits required for double SHA-256.

C. THE OPTIMIZED DOUBLE SHA-256 ARCHITECTURE

The prototype double SHA-256 accelerator has high power consumption because the fully unrolled datapath results in a large chip area. To reduce the power consumption, [25]–[28] proposed the optimized double SHA-256 accelerator, in which a 64-round unrolled datapath is optimized into a 60-round unrolled datapath.
Fig. 4 shows a schematic diagram of the 60-round unrolled ME datapath used in SHA-256_2 and SHA-256_3. In SHA-256_2, the 60-round unrolled ME datapath includes rounds 4 to 63 (denoted as Block_j (j = 4, . . . , 63)). In SHA-256_3, the 60-round unrolled ME datapath includes rounds 1 to 60 (denoted as Block_j (j = 1, . . . , 60)). Consequently, 8 ME blocks are eliminated compared with the prototype architecture mentioned above.

III. THE PROPOSED CME DOUBLE SHA-256 ARCHITECTURE

A. ARCHITECTURAL OVERVIEW

In Bitcoin mining, the 512 bits of data input to SHA-256_1 does not change frequently because it does not include the 32-bit nonce field. Conversely, the 512 bits of data input to SHA-256_2 are updated frequently because of the changing value of the nonce field. Whenever the output of SHA-256_2 changes, SHA-256_3 also needs to be recomputed. Because the nonce field has 32 bits, each computation of SHA-256_1 requires SHA-256_2 and SHA-256_3 to recompute their values up to 2^{32} times.

Therefore, we propose the CME double SHA-256 accelerator architecture, as shown in Fig. 5. To achieve a high processing rate as well as efficient hardware and power cost, we implement a resource-sharing architecture for SHA-256_1 and a fully unrolled datapath architecture for SHA-256_2 and SHA-256_3. The SHA-256_1 has a single Block_0–63 circuit for calculating W_j (j = 0, . . . , 63) and a single Loop_0–63 circuit for calculating the internal hashes a, b, c, d, e, f, h in 64 clock cycles. Each clock cycle computes one W_j value and updates the internal hash one time.

Similar to the conventional optimized double SHA-256 architecture, our SHA-256_2 has 60-round unrolled datapaths (j = 4, . . . , 63), and our SHA-256_3 has 60-round unrolled datapaths (j = 1, . . . , 60). To reduce the hardware and power costs of SHA-256_2 and SHA-256_3, we propose using CME algorithms and their equivalent hardware circuits. In Fig. 5, the CME for SHA-256_2 is denoted as CME2_j (j = 4, . . . , 63), and the CME for SHA-256_3 is denoted as CME3_j (j = 1, . . . , 60).

Using pipelined and parallel operations, SHA-256_2 and SHA-256_3 can produce an output hash every clock cycle. However, the resource-sharing SHA-256_1 circuit produces one hash value every 64 clock cycles. The low processing rate of the SHA-256_1 circuit does not affect the final processing rate of the CME double SHA-256 accelerator because one SHA-256_1 output value can be used to calculate SHA-256_2 and SHA-256_3 up to 2^{32} times. The final processing rate of the CME-based double SHA-256 is one 256-bit hash value per clock cycle.

In the following subsections, we explain our proposed CME algorithms and the equivalent hardware designs.

B. COMPACT MESSAGE EXPANDER (CME) ALGORITHM

We propose the CME algorithms by analyzing the characteristics of the input data of SHA-256_2 and SHA-256_3.

1) CME FOR SHA-256_2

As seen in Fig. 1, the 512 bits of data input to SHA-256_2 include a 32-bit Merkle root hash, a 32-bit time stamp, a 32-bit target, a 32-bit nonce, and a 384-bit padding+length...
field. It is worth noting that most of the content of the `padding + length` field consists of zeros (refer to Fig. 6a).

Assume that the 512 bits of data are separated into 16 32-bit words \( M_j (j = 0, \ldots, 15) \). The CME operation for SHA-256 is illustrated in Algorithm 3. The algorithm processes the data in 64 loops. During the first 16 loops, \( W_j (j = 0, \ldots, 15) \) are assigned to \( M_j (j = 0, \ldots, 15) \). The values of \( W_j (j = 5, \ldots, 14) \) are all zero because they are equivalent to the zero values of the `padding + length` field. In addition, \( W_4 \) and \( W_{15} \) are constants. During the last 48 loops, the CME calculates \( W_j (j = 16, \ldots, 63) \) by using (7):

\[
W_j = \sigma_1(W_{j-2}) + W_{j-7} + \sigma_0(W_{j-15}) + W_{j-16} \quad (7)
\]

The logical functions \( \sigma_0(x) \) and \( \sigma_1(x) \) are shown in (1) and (2), respectively.

Utilizing the zeros or constant values of \( W_j (j = 4, \ldots, 15) \), we can optimize the calculation of (7). For example, the \( W_{16} \) calculation can be analyzed as follows:

\[
W_{16} = \sigma_1(W_{14}) + W_0 + \sigma_0(W_{1}) + W_0 \\
= 0 + \sigma_0(W_{1}) + W_0 \\
= \sigma_0(W_{1}) + W_0 \\
\]

Note that \( W_{14} = 0 \) and \( W_0 = 0 \). By comparing (7) with (8) for calculating \( W_{16} \), it can be seen that the logical function \( \sigma_1(x) \) and two 32-bit adders have been eliminated.

The computations of \( W_j (j = 17, \ldots, 63) \) are analyzed and optimized similarly. The final results are shown in Algorithm 3.

2) CME for SHA-256

The 512 bits of input data to SHA-256 include the 256-bit hash output from SHA-256 concatenated with a 256-bit `padding + length` field. The value of the first 32 bits of padding is 32'h80000000, while the value of the last 32 bits `padding + length` is 32'h00000100. The remaining values are all zeros (refer to Fig. 6b).

We divide the 512-bit input data into 16 32-bit words \( M_j (j = 0, \ldots, 15) \). The CME operation for SHA-256 is illustrated in Algorithm 4. It processes the data in 64 loops. In the first 16 loops, \( W_j (j = 0, \ldots, 15) \) are assigned to \( M_j (j = 0, \ldots, 15) \). The values of \( W_j (j = 9, \ldots, 14) \) are all zero because they are equivalent to the zero values of the `padding + length` field. In addition, \( W_8 \) and \( W_{15} \) are constants.

In the last 48 loops, CME calculates \( W_j (j = 16, \ldots, 63) \) using (7).

Utilizing the zero or constant characteristics of \( W_j (j = 8, \ldots, 15) \), we optimize the calculation of (7) for calculating \( W_j (j = 16, \ldots, 63) \). The final results are shown in Algorithm 4.

Utilizing Algorithms 3 and 4, we can significantly reduce the number of 32-bit adders and the number of logical func-

Algorithm 3 Compact Message Expander in SHA-256

1. For \( j \) from 0 to 3 {
   \( W_j = M_j \)
2. \( W_4 = 32'h80000000 \)
3. For \( j \) from 5 to 14 {
   \( W_j = 32'h00000000 \)
4. \( W_{15} = 32'h0000280 \)
5. \( W_{16} = \sigma_0(W_1) + W_0 \)
6. For \( j \) from 17 to 19 {
   \( W_j = \sigma_1(W_{j-2}) + \sigma_0(W_{j-15}) + W_{j-16} \)
7. \( W_{20} = \sigma_1(W_{18}) + W_4 \)
8. \( W_{21} = \sigma_1(W_{19}) \)
9. For \( j \) from 22 to 29 {
   \( W_j = \sigma_1(W_{j-2}) + W_{j-7} \)
10. \( W_{30} = \sigma_1(W_{28}) + W_{23} + \sigma_0(W_{15}) \)
11. For \( j \) from 31 to 63 {
   \( W_j = \sigma_1(W_{j-2}) + W_{j-7} + \sigma_0(W_{j-15}) + W_{j-16} \)

Algorithm 4 Compact Message Expander in SHA-256

1. For \( j \) from 0 to 7 {
   \( W_j = M_j^{(i)} \)
2. \( W_8 = 32'h80000000 \)
3. For \( j \) from 9 to 14 {
   \( W_j = 32'h00000000 \)
4. \( W_{15} = 32'h0000100 \)
5. \( W_{16} = \sigma_0(W_1) + W_0 \)
6. For \( j \) from 17 to 21 {
   \( W_j = \sigma_1(W_{j-2}) + \sigma_0(W_{j-15}) + W_{j-16} \)
7. For \( j \) from 22 to 23 {
   \( W_j = \sigma_1(W_{j-2}) + W_{j-7} + \sigma_0(W_{j-15}) + W_{j-16} \)
8. \( W_{24} = \sigma_1(W_{22}) + W_{17} + W_8 \)
9. For \( j \) from 25 to 29 {
   \( W_j = \sigma_1(W_{j-2}) + W_{j-7} \)
10. \( W_{30} = \sigma_1(W_{28}) + W_{23} + \sigma_0(W_{15}) \)
11. For \( j \) from 31 to 63 {
   \( W_j = \sigma_1(W_{j-2}) + W_{j-7} + \sigma_0(W_{j-15}) + W_{j-16} \)
tions $\sigma_0(x)$ and $\sigma_1(x)$ required to calculate $W_{16}$ to $W_{63}$ in SHA-256$_2$ and SHA-256$_3$.

**C. CME HARDWARE CIRCUITS**

From Algorithm 3 and 4, we propose four types of shortened computation (SC) circuits as shown in Fig. 7. Compared with the traditional $C$ circuit shown in Fig. 2, the proposed SC$_1$ eliminates two 32-bit adders and the logical function $\sigma_1(x)$; SC$_2$ eliminates one 32-bit adder; SC$_3$ eliminates two 32-bit adders and the logical function $\sigma_1(x)$; and SC$_4$ eliminates one 32-bit adder and the logical function $\sigma_0(x)$. Note that eliminating either $\sigma_0(x)$ or $\sigma_1(x)$ also eliminates two 32-bit rotations, one 32-bit shift, and two 32-bit XOR circuits.

Based on the $C$ circuit shown in Fig. 2 and the four types of SC circuits shown in Fig. 7, we develop hardware architectures for the CME processes of SHA-256$_2$ and SHA-256$_3$ as shown in Fig. 8 and Fig. 10, respectively.

The proposed CME circuit for SHA-256$_2$ (Fig. 8) is divided into three phases. Phase 1 includes CME$_{24}$ to CME$_{219}$. Each operation requires a 128-bit register (or four 32-bit registers) to store and pipeline $W_0$ to $W_3$. In phase 1, instead of using the conventional $C$ circuit in Fig. 2, the SC$_1$ and SC$_2$ circuits in Fig. 7 are implemented to reduce hardware costs. Phase 2 includes CME$_{220}$ to CME$_{299}$, for which the SC$_2$ and SC$_3$ circuits are appropriately implemented (refer to algorithm 3). Phase 3 includes CME$_{31}$ to CME$_{63}$, and the $C$ circuit is implemented in all the blocks of this phase.

The three phases are classified based on the characteristics of the datapath bit width. In phase 1, the datapath bit-width is constant (128 bits). The 384-bits of $W_{4}$ to $W_{15}$ are fixed constants. Hence, phase 1 do not need to store and pipeline $W_4$ to $W_{15}$. In phase 2, $W_{20}$ to $W_{30}$ must be stored and pipelined. Thus, the datapath bit-width in phase 2 is appropriately increased from 160 bits to 480 bits. In phase 3, the datapath bit-width of CME$_{31}$ to CME$_{57}$ is 512 bits without optimization. To eliminate unnecessary values of $W_j$ in subsequent blocks, the datapath bit-width of CME$_{57}$ to CME$_{63}$ appropriately reduces from 480 bits to 32 bits. To understand the reason for the datapath bit-width adjustment, we show the detailed data flow and computational circuit of the CME$_2$ process in Fig. 9. In this figure, the number represents the $j$ index of $W_j$. For example, we need four 32-bit
FIGURE 10. Block diagram of the 60-round unrolled datapath CME3 process for SHA-256.

FIGURE 11. Detailed computational circuit of the CME3 process for SHA-256.

registers (equivalent to 128 bits) to store $W_0$ to $W_3$ in blocks CME2<sub>4</sub> to CME2<sub>15</sub>. As another example, CME2<sub>32</sub> needs sixteen 32-bit registers ($16 \times 32 = 512$ bits) to pipeline store 16 values of $W_j$ ($j=16, 17, \ldots, 31$), which are required for the calculation of its following blocks.

Similarly, the proposed CME circuit for SHA-256 has three phases (Fig. 10). Phase 1 includes CME3<sub>1</sub> to CME3<sub>23</sub>. Because of the zero and constant property of input data $W_8$ to $W_{15}$, all blocks of phase 1 have the same datapath of 256 bits only (which is required to pipeline store eight 32-bit values $W_0$ to $W_7$). A large number of registers are thus eliminated. In this phase, circuits SC<sub>1</sub>, SC<sub>2</sub>, and C are appropriately implemented (refer to algorithm 4). Phase 2 includes blocks from CME3<sub>24</sub> to CME3<sub>30</sub>. Circuits SC<sub>4</sub>, SC<sub>3</sub>, and SC<sub>2</sub> are appropriately implemented (refer to algorithm 4). Phase 3 includes blocks from CME3<sub>31</sub> to CME3<sub>60</sub>. We do not implement blocks from CME3<sub>61</sub> to CME3<sub>63</sub> because we can detect early whether the final hash is smaller than the target value without waiting for results from CME3<sub>61</sub> to CME3<sub>63</sub>. Circuit C is implemented in all blocks.

Three phases are classified based on the characteristics of the datapath bit-width. In phase 1, the datapath bit-width is constant (256 bits). The 256-bits of $W_8$ to $W_{15}$ are fixed constants and do not need to be stored and pipelined in phase 1. In phase 2, $W_{24}$ to $W_{30}$ must be stored and pipelined. Therefore, the datapath bit-width of CME3<sub>24</sub> to CME3<sub>30</sub> is appropriately increased from 288 bits to 480 bits. In phase 3, the datapath bit-width of CME3<sub>51</sub> to CME3<sub>53</sub> is 512 bits without optimization. The datapath bit-width of CME3<sub>34</sub> to CME3<sub>60</sub> is reduced from 480 bits to 32 bits. To prove that the datapath bit-width adjustment is appropriate, we show the detailed data flow and the computational circuit of the CME3 process in Fig. 11. In this figure, the number represents the $j$-th index of $W_j$. For example, each block from CME3<sub>0</sub> to CME3<sub>15</sub> requires eight 32-bit registers (equivalent to $8 \times 32 = 256$ bits) to store $W_0$ to $W_7$. These values are required to calculate the blocks from CME3<sub>16</sub> to CME3<sub>22</sub>. As another example, block CME3<sub>59</sub> requires five 32-bit registers ($5 \times 32 = 160$ bits) to store $W_{44}$, $W_{45}$, $W_{53}$, $W_{58}$, and $W_{59}$, which are required for the CME3<sub>60</sub> calculation.

IV. EVALUATION

In this section, we evaluate the efficiency of the CME method when it is applied in the CME double SHA-256 accelerator. We evaluate the performance from three aspects: theory, ASIC, and FPGA experimental results.

A. THEORETICAL REVIEW

For comparison purposes, we developed three hardware circuits, all of which follow the architecture proposed in Fig. 5. The three circuits differ only in how they implement the ME processes of SHA-256<sub>2</sub> and SHA-256<sub>3</sub>. The first circuit (named Prototype double SHA-256) was proposed in [23] and mentioned in section II-B. The second circuit (named
TABLE 1. Theory comparison: hardware-resource required for ME process.

| Stage   | Resource              | Prototype | Optimized | Proposed |
|---------|-----------------------|-----------|-----------|----------|
| SHA-256₂ | 32-bit adders         | 144       | 144       | 117      |
|         | 32-bit XOR gates      | 192       | 192       | 170      |
|         | 32-bit Rotation       | 240       | 240       | 217      |
|         | 32-bit Shift          | 48        | 48        | 38       |
|         | 32-bit registers      | 1009      | 801       | 651      |
| SHA-256₃ | 32-bit adders         | 144       | 135       | 116      |
|         | 32-bit XOR gates      | 192       | 180       | 166      |
|         | 32-bit Rotation       | 240       | 225       | 210      |
|         | 32-bit Shift          | 48        | 45        | 39       |
|         | 32-bit registers      | 1009      | 825       | 697      |
| SHA-256₄ + | 32-bit adders         | 288       | 279       | 233      |
|         | 32-bit XOR gates      | 384       | 372       | 336      |
|         | 32-bit Rotation       | 480       | 465       | 427      |
|         | 32-bit Shift          | 96        | 93        | 77       |
|         | 32-bit registers      | 2018      | 1620      | 1348     |

Optimized double SHA-256 was proposed in [25]–[27], and [28], and is mentioned in Section II-C. The last circuit is our proposed CME double SHA-256.

Table 1 shows the theoretical hardware resources required by the three architectures in terms of the number of adders, XOR gates, rotations, shifts, and registers. In Table 1, SHA-256₂ and SHA-256₃ are the evaluation targets because they are the most hardware-intensive parts.

Compared to the prototype and optimized architectures, the proposed architecture respectively decreases the total number of 32-bit adders by approximately 19.1% and 16.49%, the total number of 32-bit XOR gates by approximately 12.5% and 9.68%, and the total number of 32-bit rotation operations, by approximately 11% and 8.17%.

In addition, the proposed architecture reduces the total number of 32-bit shift operations by approximately 19.8% and 17.2% compared to the prototype and optimized architectures, respectively.

Notably, the proposed architecture eliminates 33.2% and 16.79% of the total number of registers compared to the prototype and optimized architectures, respectively.

B. ASIC EXPERIMENT

1) AREA AND POWER APPROACH

To ensure a fair comparison, the three double SHA-256 circuits were coded in Verilog and synthesized in an ASIC using the Synopsys Design Compiler with the Rohm 0.18µm CMOS standard cell library [29]. Table 2 shows the synthesized area of the three architectures. Note that the total area is the sum of the combinational and non-combinational area (registers), as well as other types of circuits, including buff/Inv, wires, etc. The total area of the proposed

CME double SHA-256 is smaller by 17.6% and 13.9% compared to the prototype and optimized architectures, respectively.

Fig. 12 summarizes the energy consumption of the three architectures obtained from the ASIC synthesis results. In terms of cell internal power, the proposed double SHA-256 circuit consumes 133 mW, which is a reduction of 15.82% and 11.92% compared to the prototype and optimized architectures, respectively. In terms of net switching power, the proposed CME double SHA-256 circuit consumes 95 mW, which constitutes reductions of 12.04% and 9.52% compared to prototype and optimized architectures, respectively. These energy consumption reductions are due to the smaller hardware circuit, which matches our expectations.

Based on the timing report of ASIC synthesis, the maximum frequency of the three architectures is 60 MHz. This means that the architectures achieve throughput of 1024 bits × 60 MHz = 61.44 Gbps.

In addition, we successfully laid out the proposed CME double SHA-256 circuit in ASIC technology with the Rohm 0.18µm CMOS standard cell library. Fig. 13 shows the chip layout, and Fig. 14 shows the chip energy distribution map. The size of the chip layout is 5.9 mm × 5.9 mm.

TABLE 2. Practical comparison: The ASIC synthesized area of three double SHA-256 architectures.

| Design                  | Combinational Area (µm²) | Non-combinational Area (µm²) | Total Area (µm²) |
|-------------------------|--------------------------|------------------------------|------------------|
| Prototype               | 5,741.967                | 6,716.928                    | 13,823.243       |
| Optimized               | 5,480.520                | 6,428.772                    | 11,909.292       |
| Proposed               | 4,876.523                | 5,409.863                    | 10,286.386       |

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2) PROCESSING RATE AND HARDWARE EFFICIENCY APPROACH

In this experiment, we prove that the ASIC design of our proposed CME double SHA-256 architecture outperforms previous works in terms of processing rate and hardware efficiency. To ensure a fair comparison, we also synthesized our architecture in ASIC TSMC 0.18\(\mu\)m technology using the CMOS standard cell library. We then compare our results with the previous works in [15], [16], and [17].

The comparison is shown in Table 3. It is worth noting that the designs of [15], [16], and [17] are single SHA-256 circuits. To be applied to Bitcoin mining, these circuits must repeat their calculations three times to generate a double SHA-256 hash value from the 1024-bit input message. The number of cycles required to compute the double SHA-256 (denoted by \(C_d\)) is thus triple the number of cycles required to compute a single SHA-256 (denoted by \(C_s\)); refer to (9).

\[
C_d = 3 \times C_s \quad (9)
\]

Then, we calculate the processing rate for double SHA-256 \((R_d)\) by using (10). The BlockSize is 1024 bits.

\[
R_d = \frac{\text{BlockSize} \times \text{Frequency}}{C_d} \quad (10)
\]

From the \(R_d\) and area results, the hardware efficiency for double SHA-256 (denoted by \(E_d\)) is computed by (11).

\[
E_d = \frac{R_d}{\text{Area}} \quad (11)
\]

Table 3 summarizes the synthesized area results, the calculated processing rate, and the hardware efficiency. The processing rate \((R_d)\) and hardware efficiency \((E_d)\) of our proposed architecture are significantly improved compared to those of the works in [15], [17], and [16]. The numerical results are as follows.

In terms of processing rate \((R_d)\), our CME double SHA-256 architecture is faster than the designs proposed in [15], [16], and [17] by 86, 60, and 47 times, respectively.

In terms of hardware efficiency \((E_d)\), our CME double SHA-256 architecture improves the efficiency by 102\%, 23\%, and 194\% compared to the designs in [15], [16], and [17], respectively.

3) FPGA SYNTHESIS RESULTS

To ensure a fair comparison with other existing SHA-256 architectures, such as [18]–[21], [22], and [23], we synthesized the proposed CME double SHA-256 circuit on four Xilinx FPGA boards, including Kintex UltraScale (XCKU5P-ffva676-3-e), Virtex 7(XC7VX1140T-FLG1926-2), Artix 7 (XC7A200T-FBG484-1), and Zynq UltraScale+ ZCU102 (XCUZ9EG-FFVB1156-2-e).

The results are shown in Table 4. It is worth noting that the existing architectures in [18]–[22] and [23] are single SHA-256 architectures that must repeat the computation three times to generate a double SHA-256 hash value for Bitcoin mining. Thus, the number of clock cycles required

### Table 3. A comparison of ASIC synthesis results.

| Technology | Design  | Frequency (MHz) | No. of Cycles | Throughput \(R_d\) (Gbps) | Area (\(\mu m^2\)) | Hardware Efficiency \(E_d\) (Gbps/\(\mu m^2\)) |
|------------|---------|-----------------|---------------|------------------------|-------------------|---------------------------------------------|
| SMIC 0.18\(\mu m\) | [15]    | 208             | 65            | 195                    | 211,955           | 5.16                                        |
| UMC 0.18\(\mu m\)  | [16]    | 302             | 66            | 198                    | 185,256           | 8.44                                        |
| TSMC 0.18\(\mu m\) | [17]    | 380             | 65            | 195                    | 562,704           | 3.54                                        |
| Prop.       |         | 92              |               | 1                      | 9,038,148         | 10.42                                       |

FIGURE 13. Post-layout circuit of the proposed CME double SHA-256 accelerator.

FIGURE 14. Energy distribution map of the post-layout CME double SHA-256 circuit.
to compute a double SHA-256 is tripled. We focus on evaluating the hardware efficiency (Mbps/LUT) of the single and double SHA-256 architectures in this subsection. In general, the proposed CME double SHA-256 outperforms the existing SHA-256 architectures in terms of hardware efficiency. The numerical results are as follows.

On the Kintex UltraScale FPGA, the hardware efficiency (Mbps/LUT) of the proposed architecture is enhanced by 1,047% (9.52 vs. 0.83), 488% (9.52 vs. 1.62), and 7% (9.52 vs. 8.92) compared to the hardware efficiencies of the architectures in [18], [19], and [23], respectively.

On the Virtex 7 FPGA, the hardware efficiency (Mbps/LUT) of the proposed architecture is enhanced by 7% (5.86 vs. 5.5), 762% (5.86 vs. 0.68), and 7% (5.86 vs. 5.48) compared to the hardware efficiencies of the architectures in [20], [21], and [23], respectively.

On the Artix 7 FPGA, the hardware efficiency of the proposed architecture is enhanced by 332% (2.94 vs. 0.68) and 7% (2.94 vs. 2.76) compared to the hardware efficiencies of the architectures in [22] and [23], respectively.

On the Zynq UltraScale+ ZCU102 FPGA, the hardware efficiency of the proposed architecture is enhanced by 7% (8.32 vs. 7.8) compared to the hardware efficiency of the architecture in [23].

C. FPGA EXPERIMENT

1) FUNCTIONAL VERIFICATION ON A REAL SOC HARDWARE PLATFORM

To prove that the circuit operates correctly not only in the software simulation tool but also on real hardware, we built a System on Chip (SoC) platform to execute the proposed CME double SHA-256 circuit. The SoC platform overview is shown in Fig. 15.
The platform includes two primary components: a host PC and a Zynq UltraScale+ ZCU102 evaluation board. The host PC exchanges data with the ZCU102 board via JTAG and UART cables. The ZCU102 board includes an ARMv8 microprocessor, a programmable logic (PL), and a clock generator. Our developed circuit, CME double SHA-256, is embedded in the PL of ZCU102. The PL also has block ram (BRAM) and an integrated logic analyzer (ILA). We used the BRAM to store the valid nonce value for Bitcoin mining and ILA to monitor the outputs of the CME double SHA-256 circuit. The maximum operating frequency of the ZCU102 board is 333 MHz.

The host PC consists of a Vivado, a Software Development Kit (SDK), and a Bitcoin Mining Verification (BMV) program. Vivado is a software suite for SoC development. We use the Vivado suite to design and load the SoC-based system onto the Zynq UltraScale+ ZCU102 board. Moreover, the Vivado helps to export the outputs of the CME double SHA-256 circuit in the ZCU102 into an ILA result file for verification by the BMV program. The SDK is intended for the development of embedded software applications for SoC systems. We use the SDK to embed the real block information from the Bitcoin blockchain network onto our SoC-based system. The BMV is a C-code program that verifies the correctness of the embedded CME double SHA-256 circuit. The BMV executes a double SHA-256 on the host PC and compares the results with the outputs of the CME double SHA-256 circuit.

The abovementioned SoC system has been used to thoroughly verify the correctness of the CME double SHA-256 circuit at different operating frequencies, such as 333 MHz (maximum frequency) and 200 MHz. All the cases result in 100% accuracy, which proves that the proposed CME double SHA-256 architecture works correctly in a real hardware platform. The maximum processing rate of the circuit on the ZCU102 board is 333 MHash/s (or 333 MHz × 1024 bit/CLK = 340.992 Gbps).

Fig. 16 shows an image of the SoC evaluation platform, which includes a host PC (Toshiba Satellite B652 / G Core i5 3320M 2.6GHz / 4GB) and the UltraScale+ ZCU102 evaluation board.

2) PROCESSING-RATE EVALUATION ON A REAL HARDWARE PLATFORM

In this subsection, we evaluate the processing rate and power consumption of the proposed CME double SHA-256 on real hardware platform ZCU102 to prove that our architecture outperforms other high-performance platforms, including CPUs, GPUs, and the existing SHA-256 architectures.

Table 5 shows the execution time of the double SHA-256 algorithm on several hardware platforms, including a CPU, GPU, and FPGA. To compute the same number of hashes (e.g., 500,000 hashes) the proposed architecture running on the FPGA ZCU102 requires only 1.5 ms, while the CPU i7-6950X, CPU XEON 6144, and GPU Tesla V100 require 770 ms, 740 ms, and 140 ms, respectively.

| Device | Device | Hash rate (MHash/s) | Power (W) |
|--------|--------|---------------------|-----------|
| CPU Core i7-6950X@3.30GHz (10 cores), Memory128GB | CPU Intel i7-2600K@3.33GHz, 4 Cores | 1.9 | - |
| GPU NVIDIA GeForce GTX 550 TI | GPU NVIDIA GeForce GTX 570 | 20.4 | - |
| CPU Intel Core i7-990x@3.46GHz, 6 Cores | CPU Intel Core i7-950@3.06GHz | 33 | - |
| GPU NVIDIA GeForce GTX 570 | GPU ATi Radeon HD 5770 | 155 | - |
| ASIC Block Erupter Sapphire | FPGA Zynq UltraScale+ ZCU102 (333MHz) | 333 | 2.55 |

This work shows the execution time of the double SHA-256 on several hardware platforms, including CPUs, GPUs, and the existing SHA-256 architectures.
which means that the proposed architecture reduces the execution time by 513 times, 493 times, and 93 times, respectively.

Table 6 summarizes the hash rate and power consumption from several studies that reported double SHA-256 results. As the table shows, the hash rate of our proposed architecture running on an FPGA is significantly higher than those of the works in [30] and [31]. Although [32] was executed on an ASIC and our architecture was executed on an FPGA, our architecture still achieves the same hash rate but consumes less power.

V. CONCLUSION

Bitcoin mining is an important process in keeping the Bitcoin network secure; however, it consumes massive amounts of energy. To reduce the power consumption and increase the processing rate of the Bitcoin mining process, we proposed a CME double SHA-256 hardware circuit in this paper. The architecture includes three SHA-256 circuits in which the first circuit (SHA-256\(_1\)) is a resource-sharing architecture while the last two circuits (SHA-256\(_2\) and SHA-256\(_3\)) are fully unraveled datapath architectures. The combination of these two types of architecture results in a high processing rate but low hardware costs. Specifically, we propose several compact message expander (CME) algorithms and associated hardware architectures to further reduce the power consumption and hardware costs. Our proposed circuit generates one 256-bit hash value per clock cycle. We thoroughly verified and evaluated the proposed circuit on both ASIC and FPGA platforms. The experimental results showed that the proposed circuit outperforms other high-performance CPU and GPU platforms for computing double SHA-256 values. The proposed circuit also outperforms existing works with specific hardware circuits for computing the double SHA-256 values. The double SHA-256 circuit was laid out on the ASIC with Rohm 0.18 \(\mu\)m CMOS standard cell library, resulting in a chip size of 5.9 mm \(\times\) 5.9 mm and the throughput of 61.44 Gbps. The circuit is also proven to work correctly in a real hardware platform (ZCU102), achieving a processing rate of 340.992 Gbps.

Blockchain is not only the Bitcoin network. Blockchain technology is outgrowing in its potential to be applied in many fields of life, such as smart health care, autonomous cars, and supply chains. Other blockchain networks may employ not only SHA-256 but also other cryptography hash functions, such as SHA-512 or SHA-3. Therefore, developing a flexible and programmable accelerator that can compute several hash functions is a future need. By developing a low-cost low-power-consumption blockchain accelerator, we help to enhance the security and decentralized features of the blockchain network. Therefore, we believe that developing a blockchain accelerator that can compute multiple cryptography hash functions at low cost and with low power consumption will be an important research trend in the near future.

APPENDIX

The Verilog code and the synthesized results of the prototype, optimized, and proposed architectures can be found at https://github.com/archlab-naist/Double-CME-SHA256/

REFERENCES

[1] P. D. DeVries, “An analysis of cryptocurrency, bitcoin, and the future,” Int. J. Bus. Manage. Commerce, vol. 1, no. 2, pp. 1–9, 2016.
[2] S. Nakamoto, Bitcoin: A Peer-to-Peer Electronic Cash System. Manoub, Nov. 2019.
[3] M. Rahouti, K. Xiong, and N. Ghani, “Bitcoin concepts, threats, and machine-learning security solutions,” IEEE Access, vol. 6, pp. 67189–67205, 2018.
[4] H. Vranken, “Sustainability of bitcoin and blockchains,” Current Opinion Environ. Sustainability, vol. 28, pp. 1–9, Oct. 2017.
[5] J. Taskinsoy, “Bitcoin and turkey: A good match or a perfect storm,” SSRN Electron. J., Oct. 2019, doi: 10.2139/ssrn.3477849.
[6] N. T. Courtois, M. Grajek, and R. Naik, “Optimizing SHA256 in bitcoin mining,” in Proc. Int. Conf. Cryptogr. Secur. Syst., Berlin, Germany: Springer, 2014, pp. 1–36.
[7] X. Zhang and H. Hu, “Optimization of hash function implementation for bitcoin mining,” in Proc. 3rd Int. Conf. Mechatronics Eng. Inf. Technol. (ICMEEIT), 2019.
[8] M. D. Rote, V. N., and D. Selvakumar, “High performance SHA-2 core using the round pipelined technique,” in Proc. IEEE Int. Conf. Electron., Comput. Commun. Technol. (CONNECT), Jul. 2015, pp. 1–6.
[9] G. Dadda, M. Macchetti, and J. Owen, “ASIC design of a high speed ASIC unit for the hash function SHA-256 (384, 512),” in Proc. Design, Autom. Test Eur. Conf. Exhib., Feb. 2004, pp. 70–75.
[10] P. Madhi and R. Chaudhari, “An optimized pipelined architecture of SHA-256 hash function,” in Proc. 7th Int. Symp. Embedded Comput. Syst. Design (ISED), Dec. 2017, pp. 1–4.
[11] X. Zhang, R. Wu, M. Wang, and L. Wang, “A high-performance parallel computation hardware architecture in SHA-256 hash,” in Proc. 21st Int. Conf. Adv. Commun. Technol. (ICACT), Feb. 2019, pp. 52–55.
[12] S. B. Suhaili and T. Watanabe, “Design of high-throughput SHA-256 hash function based on FPGA,” in Proc. 6th Int. Conf. Electr. Informat. (IEEEI), Nov. 2017, pp. 1–6.
[13] R. Garcia, I. Algedro-Badillo, M. Morales-Sandoval, C. Feregrino-Uribe, and R. Cumplido, “A compact FPGA-based processor for the secure hash algorithm SHA-256,” Comput. Electr. Eng., vol. 40, no. 1, pp. 194–202, Jan. 2014.
[14] H. Michail, G. Athanasiou, A. Kritikakou, C. Goutis, A. Gregoriades, and V. Papadopoulou, “Ultra high speed SHA-256 hashing cryptographic module for ipsec hardware/software codesign,” in Proc. Int. Conf. Secur. Cryptogr. (SECYR), Jul. 2010, pp. 1–5.
[15] L. Bai and S. Li, “VLSI implementation of high-speed SHA-256,” in Proc. IEEE Int. Conf. Adv. Commun. Technol., vol. 1, no. 2, pp. 1–9, 2016.
[16] S. Tillich, M. Feldhofer, M. Kirschbaum, P. Thomas, S. Jörn-Marc, and S. Alexander, “High-speed hardware implementations of BLAKE, blue midnight wish, CubeHash, ECHO, Fugue, Grøstl, Hamsi, JH, Keccak, Luffa, Shabal, SHAvite-3, SIMD, and Skein,” IACR Cryptol. ePrint Arch., vol. 310, 2009.
[17] F. Opritou, S. L. Jurj, and M. Vladutiu, “Technological solutions for throughput improvement of a secure hash algorithm-256 engine,” in Proc. IEEE 23rd Int. Symp. Design Technol. Electron. Package. (SITIME), Oct. 2017, pp. 159–164.
[18] R. Martino and A. Cilardo, “A flexible framework for exploring, evaluating, and comparing SHA-2 designs,” IEEE Access, vol. 7, pp. 72443–72456, 2019.
[19] R. Martino and A. Cilardo, “A configurable implementation of the SHA-256 hash function,” in Proc. Int. Conf. WPTP Parallel, Grid, Cloud Internet Comput., Cham, Switzerland: Springer, Nov. 2019, pp. 558–567.
[20] F. Kahri, B. Bouallegue, M. Machhout, and R. Tourki, “An FPGA implementation and comparison of the SHA-256 and blake-256,” in Proc. 14th Int. Conf. Sci. Techn. Autom. Control Comput. Eng. STA, Dec. 2013, pp. 152–157, doi: 10.1109/STACON.2013.6783122.
[21] R. Florin and R. Ionut, “FPGA based architecture for securing IoT with blockchain,” in Proc. Int. Conf. Speech Technol. Hum.-Comput. Dialogue (SpED), Oct. 2019, pp. 1–8, doi: 10.1109/SPED.2019.8906595.
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