Intrinsic Detectivity Limits of Organic Near-Infrared Photodetectors

Sam Gielen, Christina Kaiser, Frederik Verstraeten, Jonas Kublitski, Johannes Benduhn, Donato Spoltore, Pieter Verstappen, Wouter Maes, Paul Meredith, Ardalan Armin,* and Koen Vandewal*

Organic photodetectors (OPDs) with a performance comparable to that of conventional inorganic ones have recently been demonstrated for the visible regime. However, near-infrared photodetection has proven to be challenging and, to date, the true potential of organic semiconductors in this spectral range (800–2500 nm) remains largely unexplored. In this work, it is shown that the main factor limiting the specific detectivity ($D^*$) is non-radiative recombination, which is also known to be the main contributor to open-circuit voltage losses. The relation between open-circuit voltage, dark current, and noise current is demonstrated using four bulk-heterojunction devices based on narrow-gap donor polymers. Their maximum achievable $D^*$ is calculated alongside a large set of devices to demonstrate an intrinsic upper limit of $D^*$ as a function of the optical gap. It is concluded that OPDs have the potential to be a useful technology up to 2000 nm, given that high external quantum efficiencies can be maintained at these low photon energies.
showing photodetection up to 1400 nm. Room-temperature values for $D^*$ in this wavelength regime are in the order of 10$^{10}$ Jones and hence one to two orders of magnitude below the performance of the inorganic variants in that wavelength range. Research on organic photodetectors (OPDs) therefore focuses on increasing $D^*$, while simultaneously targeting longer NIR wavelengths and maintaining other performance metrics.

$D^*$ is negatively influenced by high noise levels and low photovoltaic external quantum efficiencies (EQE$_{ph}$).[22,23] Although narrow-gap organic materials absorbing up to 2 μm are known,[24–26] the often observed high dark (and noise) current is one of the major limiting factors, as it strongly increases with a decreasing optical gap.[27,28] While the effect of interfacial layers and traps on the dark and noise current has been elucidated, it is still unclear if under ideal circumstances, organic NIR detectors can reach similar performances as obtained with inorganic photodetectors.[29–31] No (thermodynamic) upper limit for $D^*$ of organic NIR detectors has been rationalized and the longest achievable detection wavelength is still unknown.

In this work, four different NIR-photoactive narrow-gap donor polymers are investigated in a photodiode architecture and for the first time the open-circuit voltage ($V_{OC}$) is shown to be related to the dark current, noise current, noise equivalent power (NEP) and $D^*$. Specifically for these low $E_{CT}$ systems the dark current is not dominated by the shunt current anymore, resulting in a quantitative relation between $V_{OC}$ and dark current. This relation is generalized for previously reported organic NIR detectors and it is shown how, at low optical gaps, $V_{OC}$ limits the ratio between light and dark current ($I_{ph}/I_{0}$). Applying the knowledge on the $V_{OC}$ limits gathered in the organic solar cell field allowed us to calculate an upper limit for the $D^*$ of 10$^{12}$ and 10$^{10}$ Jones at 1500 and 2000 nm, respectively. Moreover, it is found that $D^*$ is predominantly limited by non-radiative losses, effectively decreasing $D^*$ by 3 to 4 orders of magnitude in comparison to the background limited infrared photodetection (BLIP) for perfectly radiative materials.[31] Nonetheless, given the multiple advantages for commercialization, OPDs remain a viable alternative for the NIR regime up to 2 μm.

Photodiodes are often photovoltaic-based and are made in some form of diode architecture, operated in the reverse direction. In the ideal case, the total current flowing through the device under illumination is described by the Shockley equation:

$$I = I_D (e^{qV/kT} - 1) - I_{ph}$$

with $q$ being the elementary charge and $k$ the Boltzmann constant. At a temperature $T$, the total current $I$ depends on the dark saturation current $I_D$, the photocurrent $I_{ph}$ and the applied voltage $V$. In the dark, $I = I_D$ and $I_{ph} = 0$, and Equation (1) can be written as:

$$I_D = I_D (e^{qV/kT} - 1)$$

For negative voltages, the diode current from Equation (1) approaches the dark saturation current ($I_D \rightarrow -I_D$). $I_D$ is considered the “thermodynamic equilibrium recombination current”, due to thermally activated band to band transitions as described by Cuevas.[32] However, due to the presence of traps, pinholes, shunts, or charge-carrier injection from the contacts, in a real diode $I_D$ is typically well above $I_0$ and is dominated by the strongest recombination mechanism, which could stem from trap-assisted recombination described by Shockley–Read–Hall statistics.[33] In what follows, it is assumed for simplicity that at sufficiently small bias voltages, where series resistance can be neglected, those additional currents are described by $I_{shunt}$, $I_D$ is therefore expressed as:

$$I_D = I_0 \left( e^{qV/kT} - 1 \right) + I_{shunt}$$

where $I_{shunt} = V/R_{shunt}$ and $R_{shunt}$ is the shunt resistance. In the literature, it was shown that $I_{shunt}$ can be suppressed using interlayers, thick junctions, and an optimized device architecture.[34–36] Moreover, in the case of bulk heterojunctions (BHJs) with a low effective bandgap, $I_D$ can be high in comparison to $I_{shunt}$. Since $I_D$ was previously found to scale with the effective bandgap,[37,38] Importantly, because of the presence of shunts, $I_0$ is experimentally often difficult to access. However, when $R_{shunt}$ is sufficiently high, the ratio of $I_{ph}/I_0$ relates to the $V_{OC}$ of the solar cell. Indeed, an expression for the $V_{OC}$

$$V_{OC} = \frac{kT}{q} \ln \left( \frac{I_{ph}}{I_0} + 1 \right)$$

can be derived by considering that at open circuit ($V = V_{OC}$), the dark current given by Equation (3) equals the photocurrent $I_{ph}$, which is considerably higher than $I_{shunt}$. This establishes a relation between $V_{OC}$ and $I_{ph}/I_0$.

In order to verify the above relations, four narrow-gap polymers are used as donors in combination with PC$_7$BM as acceptor in BHJ photodiodes of the structure glass/ITO/ZnO-PEIE/active layer/MoO$_3$/Ag. The synthesis and performance in NIR photodiodes of three of the used polymers—PTTQAI, PBTQ(OD), and PTTQ(HD)—have been reported before.[16,39] The synthesis route for the novel ultralow gap push–pull copolymer PTTQn(HD) is provided in the Supporting Information (Scheme S1, Supporting Information). The chemical structures and their frontier orbital energy levels, as estimated via cyclic voltammetry (CV, Figure S2, Supporting Information), are shown in Figure 1a–b, respectively.

The donor polymers are characterized by a low LUMO energy level, which results in low optical gaps for the polymers (<1.5 eV, Figure S3, Supporting Information) and effective gaps (charge transfer energy, $E_{CT}$) below 1.12 eV for the BHJs.

Given these effective gaps (between 0.81 and 1.12 eV), low $V_{OC}$ values between 0.12 and 0.44 V are obtained under high intensity radiation (100 mW cm$^{-2}$), in accordance to the previously reported linear relationship between $E_{CT}$ and $V_{OC}$.[40] For such low effective gaps, it is expected that $I_D$ becomes comparable or higher than $I_{shunt}$, which makes Equation (3) and (4) suitable for relating the dark currents at reverse voltages to $V_{OC}$.

A fit to the dark current for the devices based on the four material systems shown in Figure 1 is attempted using Equation (3) and returning $R_{shunt}$ and $I_0$ as fit parameters, as shown in Figure S4, Supporting Information. This approach is indeed only useful when $I_0$ is not much smaller than $I_{shunt}$, a requirement that is only met in narrow-gap systems with very high film quality to minimize the shunt effects. The fit parameters, $R_{shunt}$ and $I_0$, are summarized in Table S1, Supporting Information.
Importantly, at small reverse bias ($\leq -0.1$ V), the contribution of $I_{\text{shunt}}$ can be neglected and hence the experimental $I_D$ equals $I_0$, except for the PTTQn(HD). Excluding PTTQn(HD), we therefore expect close agreement between the $V_{\text{OC}}$ calculated using $I_0$ in Equation (4) and the experimental $V_{\text{OC}}$. Table S1, Supporting Information, shows that both are on average 0.05 V apart, while the largest offset is observed for PBTQ(OD) with 0.09 V.

The experimental $V_{\text{OC}}$ and the ratio $I_{\text{ph}}/I_D$ are therefore expected to be logarithmically related when $I_D$ is evaluated at a sufficiently low reverse voltage. Indeed, as shown in Figure 2a for the four studied material blends, as well as for a large set of reported BHJs (all fullerene-based devices, additional information in Table S2 and Figure S5, Supporting Information), a linear relation between $V_{\text{OC}}$ and $I_{\text{ph}}/I_D$ on a semi-logarithmic scale is observed for $V_{\text{OC}} < 0.25$ V. Since $I_D$ is always larger than $I_0$, the points in Figure 2a are either very close to the intrinsic limit (within the experimental error) or well above, but cannot be significantly below. For $V_{\text{OC}} < 0.25$ V, the experimentally measured dark current is limited to $I_D < 25 \times I_0$.

For higher voltages, $I_{\text{ph}}/I_D$ increasingly deviates from the intrinsic limit, due to the increasing contribution of $I_{\text{shunt}}$. For $V_{\text{OC}} = 1$ V,
$I_D$ is typically 12 orders of magnitude higher than $I_P$. Note that the $I_D$ used in Figure 2a,b was taken at $-0.1 \text{ V}$ for all BHJs to reduce the influence of additional shunt currents. However, if devices are not fully optimized and hence show some leakage, $I_{ph}/I_D$ cannot be close to the intrinsic limit set by the $V_{OC}$.

In agreement with previous literature, a linear relationship between $E_{CT}$ and $I_D$ is observed, but only for $E_{CT} < 0.9 \text{ eV}$, as shown in Figure 2b. This agrees with $V_{OC}$ values up to 0.25 V, since the offset observed between the $V_{OC}$ and the $E_{CT}/q$ in organic BHJs is typically around 0.6–0.7 V. From the trend shown in Figure 2b, the minimum achievable $I_D$ can be estimated for a given material system with known $E_{CT}$. Moreover, it confirms that the fitting procedure of extracting $I_0$ from $I_D$ is reasonable only in the case of low $E_{CT}$ BHJs. From the above it is clear that the lowest achievable dark current, $I_0$, at reverse voltages is set by the $V_{OC}$. Devices with an $E_{CT}$ and optical gap smaller than 0.9 eV ($\approx 1380 \text{ nm}$) approach this limit.

So far, it is shown that for a BHJ with a characteristic $E_{CT}$ and $V_{OC}$, there is a theoretical upper limit for the ratio $I_{ph}/I_D$ that decreases with decreasing $V_{OC}$ ($E_{CT}$). It is clear that efficient photodetection can only be achieved when the ratio $I_{ph}/I_D$ is high. Below, we demonstrate how this relates to upper limits of photodetector performance, which is described by its specific detectivity $D^*$. The experimental $D^*$ is either obtained from the ratio of responsivity ($R$) to noise current ($i_{\text{noise}}$) according to

$$D^* = \frac{R \sqrt{\Delta f}}{i_{\text{noise}}} \quad (5a)$$

or directly from the NEP via

$$D^* = \frac{\sqrt{\Delta f}}{\text{NEP}} \quad (5b)$$

Note that $D^*$ is always normalized to the square root of the area ($A$) multiplied by the bandwidth ($\Delta f$). To obtain $D^*$ according to Equation (5a), $R$ is first calculated from the measured EQE$_{pv}$ (Figure S6 and S7, Supporting Information) at $-2$, $-1$, and 0 V according to $R = \text{EQE}_{pv} \lambda/1240$. Secondly, $i_{\text{noise}}$ is measured at different bias using a spectrum analyzer (spectra shown in Figure S8, Supporting Information). In this way, a spectrum of $D^*$ is obtained as shown in Figure 3b for PBTQ(OD) and in Figure S9, Supporting Information, for the other materials investigated in this study. However, the above assumes that the current response of the device will decrease linearly with light power until it vanishes in the current noise. To verify this assumption, a second method is used to obtain $D^*$ directly from the NEP (Equation (5b)) at a single wavelength (520 nm) by gradually decreasing the light power and measuring the respective current. A graphical representation is shown in Figure S10, Supporting Information, for two photodiodes demonstrating a linear dynamic range over 9 orders of magnitude for the PBTQ(OD) based device. Both methods are found to be in good agreement, since $D^*$ values obtained from Equation (5a) and (5b) differ only by a factor of 1.6 and 2.6 for PBTQ(OD) and PTTBAI, respectively (Figure S11, Supporting Information).

With an accurate determination of the experimental $D^*$ at hand, we can now consider upper limits of $D^*$. To do so, $i_{\text{noise}}$ is decomposed into different noise sources, such as shot noise ($i_{\text{shot}}$), thermal noise ($i_{\text{thermal}}$), and flicker noise ($i_{\text{flicker}}$). Shot noise is due to the randomness of the emission of electrons and hence linearly proportional to the current magnitude,

$$i_{\text{shot}}^2 = 2qI_{\text{ph}} \Delta f$$

Thermal noise depends on the total parallel resistance $R_p$, including $R_{\text{shunt}}$ (more information is provided in the Supporting Information) and effective diode resistance at zero bias voltage according to

$$i_{\text{thermal}}^2 = \frac{4kT}{R_p} \Delta f$$

While shot and thermal noise are considered white noise, flicker noise is frequency dependent in the low frequency regimes, as shown in Figure S8, Supporting Information. To minimize the influence of the flicker noise, an experimental frequency range between 1 and 3 kHz was chosen, and therefore the total noise current can be described as

$$i_{\text{noise}} = \sqrt{i_{\text{shot}}^2 + i_{\text{thermal}}^2}$$

![Figure 3](image-url)  
Figure 3. a) Calculated and experimental noise currents for a photodiode based on PBTQ(OD). The experimental noise current (black marker) is at least twice the noise current calculated from the measured dark current (blue marker) or from the dark saturation current (orange marker). Below $-0.1 \text{ V}$, the total calculated noise current is dominated by shot noise (red dashed line), and above $-0.1 \text{ V}$ by thermal noise (green dashed line). b) The specific detectivity of a photodiode with PBTQ(OD) calculated using different levels of noise current (same marker notation as for Figure 3a). For the calculation, the responsivity and noise current at $-2 \text{ V}$ were considered.
Figure 3a shows the experimental $i_{\text{noise}}$ alongside the calculated noise current using Equation (8) for the photodiode based on PBTQ(OD) (see Figure S12, Supporting Information, for the other donor materials). Note that the thermal noise becomes dominant at lower magnitudes of the reverse voltage, here for $|V| < 0.1$ V. Moreover, the experimental $i_{\text{noise}}$ is expected to be above the calculated one since Equation (8) does not account for all noise sources (which are known to exist but not well described) and experimental limitations imposed by the amplifier noise. Consequently, the experimental $D^*$ is expected to be lower than the calculated one according to Equation (5a).

First, we use Equation (6) and assume that $I = I_D$ to calculate $i_{\text{noise}}$, which is subsequently used in Equation (8) to obtain $D^*$. $I_D$ is taken from the dark current–voltage ($I–V$) characteristics as shown in Figure 1c. This procedure of obtaining $D^*$ is often encountered in literature in the absence of experimental noise current spectra as well as in neglect of $i_{\text{thermal}}$. As a result, $D^*$ is often overestimated, as shown in Figure 3b.

In contrast, when $i_{\text{noise}}$ is calculated according to Equation (6) assuming $I = I_D$, the resulting upper limit of $D^*$ for the photodiode with PBTQ(OD) is only 3.2 times higher than the experimentally determined $D^*$ across the spectrum. Herein, $I_D$ was obtained from the fittings to the dark $I–V$ characteristics, as described earlier. Hence, for this material system, it is clear that the device architecture is well optimized, since it is working close to the limit of $D^*$ imposed by the active layer blend and its $V_{OC}$.

The above analysis shows that at a given responsivity $R$, the maximum $D^*$ is reached when $I_D = I_0$, of which the lower limit is set by the $V_{OC}$. In accordance to Equation (5a), $D^*$ can be further increased if $R$ can be increased. In the Supporting Information, an expression is derived, using the relationship between $I_{ph}$ and $R$, in which $D^*$ is proportional to $I_{ph}/\sqrt{I_0}$. As $I_0$ is limited by the $V_{OC}$, known factors preventing the $V_{OC}$ being closer to the optical gap in organic BHJs will also affect the maximum achievable $D^*$. This is further explored in the next paragraph.

From previous work, it is known that $V_{OC}$ losses are intrinsic to organic BHJs due to the radiative and non-radiative recombination of charge carriers.[40,44] The dark saturation current comprises a radiative contribution $I_{0,R}$ and a non-radiative contribution $I_{0,NR}$. $I_{0,R}$ can be calculated via

$$I_{0,R} = q \int_{0}^{\infty} EQE_{\text{pv}}(E) \phi_{BB}(E) dE \tag{9}$$

where $\phi_{BB}(E)$ is the temperature dependent black body spectrum. In order to calculate upper limits for $D^*$, $I_{0,R}$ is calculated using Equation (9) for a perfect photodiode with a step-like EQE$_{\text{pv}}$ that has its maximum value of 1 for energies above the $E_{CT}$ and 0 below. By taking only the shot noise component of $I_{0,R}$ into account, $D^*$ is calculated via Equation (5a). The respective $D^*$ spectrum belongs to a so-called background limited infrared photodetector (BLIP; derivation shown in the Supporting Information), where the only source of current noise corresponds to the shot noise of the photocurrent induced by thermal radiation.[12] Figure 4b shows the background limited $D^*$ at 300 K for different optical gap energies corresponding to wavelengths up to 2 µm. By lowering the optical gap, $D^*$ decreases down to $5 \times 10^7$ Jones at 2 µm, which marks the thermodynamic limit of $D^*$ for any type of photodetector without photo-multiplicative gain (EQE$_{\text{pv}} \leq 1$). For organic BHJs, the EQE$_{\text{pv}}$ is below 1 and has a strong spectral dependence as well as sub-gap features, which will increase $I_{0,R}$ in comparison to the step-like EQE$_{\text{pv}}$ example and hence decrease $D^*$. Furthermore, it is known that for organic BHJs, the total dark saturation current is dominated by the non-radiative component $I_{0,NR}$. A realistic upper limit for the achievable $D^*$ for organic NIR

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**Figure 4.** a) $\Delta V_{OC,NR}$ as a function of $E_{CT}$ for BHJs studied in this work (hollow symbols) and literature-reported BHJs (black squares).[26] The red line represents a model based lower limit,[45] the green line shows a previously reported[43] empirical lower limit to $\Delta V_{OC,NR}$ and the blue line shows an average value for $\Delta V_{OC,NR}$ as observed in this study. b) Specific detectivity as a function of wavelength. The black lines represent the $D^*$ for BHJs studied in this work using the same symbols as in Figure 4a. The upper limit of the $D^*$ is calculated for different contributions of non-radiative recombination currents (represented as EQE$_{el}$ values according to Equation (10)) to the dark current ranging from zero (solid grey line: BLIP at 300 K) to very high (EQE$_{el} = 10^{-8}$). The colored lines represent the upper limit of the $D^*$ determined from the models used in Figure 4a. The $D^*$ of commercial photodetectors (orange lines) are added for comparison.[47]
OPDs can therefore be obtained by considering \( I_{0,NR} \) and a box-shaped EQE\(_{pv}\) that is 1 for energies above \( E_{CT} \) as will be shown below.

In the radiative limit \( (I_0 = I_{0,R}) \), the upper limit of the \( V_{OC} \) \((V_{OC,R})\) is obtained via Equation (4). The difference between \( V_{OC,R} \) and the experimental \( V_{OC} \) equals the non-radiative voltage losses \( (\Delta V_{OC,NR} = V_{OC,R} - V_{OC}) \). Figure 4a shows \( \Delta V_{OC,NR} \) for a large set of BHJs with different \( E_{CT} \) including the devices studied here. Clearly, \( \Delta V_{OC,NR} \) depends on \( E_{CT} \). An increase of \( \Delta V_{OC,NR} \) by 300 mV is observed when \( E_{CT} \) decreases from 1.6 to 0.8 eV. As a next step, \( I_{0,NR} \) is obtained from \( \Delta V_{OC,NR} \) via

\[
\Delta V_{OC,NR} = \frac{kT}{q} \ln \left( \frac{\text{EQE}_R}{\text{EQE}} \right)
\]

where \( \text{EQE}_R = \frac{I_{0,R} + I_{0,NR}}{I_{0,R}} \). \[45,46\] Table 1 summarizes the device parameters related to non-radiative losses for the devices studied in this work. Since \( I_{0,NR} \gg I_{0,R} \), the dark saturation current will be dominated by the non-radiative recombination current and it can be assumed that \( I_0 = I_{0,NR} \).

Subsequently, \( D^* \) is calculated under consideration of non-radiative recombination currents that are unavoidable in organic BHJs. First, \( I_{0,NR} \) is considered to be independent of \( E_{CT} \) and thus constant across the spectrum and corresponding to realistic EQE\(_R\) values between \( 10^{-3} \) to \( 10^{-9} \). As shown in Figure 4b, \( D^* \) decreases by one order of magnitude across the spectrum for a decrease of EQE\(_R\) by two orders of magnitude (corresponding to an increase of \( I_{0,NR} \) by two orders).

Recent reports have identified high frequency molecular vibrations as the source of the rather large non-radiative decay rates in organic BHJs,\[40\] explaining the previously experimentally observed dependence of \( I_{0,NR} \) on \( E_{CT} \). Below, we distinguish three different cases for \( \Delta V_{OC,NR} \) as a function of \( E_{CT} \): i) a model-based lower limit for \( \Delta V_{OC,NR} \) reported by Azzouzi et al. (in which high and low frequency vibrational modes, 150 and 50 meV, respectively, are considered to be low for organic semiconductors, hence providing a lower limit for the non-radiative voltage losses),\[45\] ii) an empirical lower limit for \( \Delta V_{OC,NR} \) reported by Benduhn et al.,\[46\] and iii) an average value for \( \Delta V_{OC,NR} \) as a function of \( E_{CT} \) as observed in this study. While case (i) predicts an upper limit for \( D^* \) to about \( 10^{10} \) Jones at 2 \( \mu \)m, the empirical and average cases (ii) and (iii) predict that OPDs based on materials with similar recombination properties as the ones found in current BHJs will typically not reach more than \( 10^9 \) Jones at 2 \( \mu \)m. At the same time, we expect \( D^* \) to reach a respectable \( 10^{12} \) Jones at 1500 nm if the EQE\(_{pv}\) can approach 1. This value is comparable to commercial InGaAs detectors and even surpasses commercial Ge detectors (Figure 4b).\[47\] Importantly, the upper limit of \( D^* \) decreases significantly faster for models where \( I_{0,NR} \) is dependent on \( E_{CT} \) than if it is not. Hence, further red-shifting the absorption to the mid-infrared (2–10 \( \mu \)m) in OPDs is not expected to be useful, as it seems impossible to compete with inorganic alternatives. Nonetheless, OPDs remain a viable alternative for the NIR regime up to 2 \( \mu \)m, given that EQE\(_{pv}\) values approaching 1 can be achieved at such low gaps with a steep EQE\(_{pv}\) tail onset.

In a recent study, Wu et al. reported on an upper limit for \( D^* \) that is around a 100 times higher at 2 \( \mu \)m than calculated here.\[48\] Their model considers as a limit to \( D^* \) the effect of tail-state broadening on noise related to thermally activated charge generation and recombination over the effective bandgap. Here, we follow a different approach and considered the non-radiative dark saturation current as a more important limiting factor. This dark mechanism has been shown to be present in both non-fullerene acceptor (NFA) containing and NFA free BHJs, and results, as compared to Ref. [48], in a stronger increase in the noise current toward longer wavelengths.\[40\]

To increase \( D^* \) to the theoretical maximum for NIR organic photodiodes, as indicated by the red curve in Figure 4b, one should find a way to decrease \( \Delta V_{OC,NR} \) while maintaining the NIR absorption. A reasonable pathway could be the use of NFA materials as these are known to yield lower non-radiative \( V_{OC} \) losses,\[49,50\] a hurdle to overcome is to find a well-performing combination of a NIR-photoactive donor and NFA, as only a selected group of donor materials tend to work well with the best NFAs produced to date.\[51\]

In this work, the intrinsic limit of the specific detectivity of NIR organic photodiodes was determined. For narrow-gap donor polymers combined with fullerene acceptors, the open-circuit voltage was related to the dark current, noise current, and specific detectivity. For devices with open-circuit voltages up to 0.25 V under 100 mW cm\(^{-2}\) illumination, corresponding to effective and optical gaps lower than 0.9 eV, it is found that the experimental dark current approaches its theoretical limit. Above 0.25 V, the deviation increases with increasing open-circuit voltage. Moreover, it is observed that the upper limit for the \( D^* \) of NIR OPDs is predominantly limited by non-radiative losses, effectively decreasing \( D^* \) by 3 to 4 orders of magnitude in comparison to the BLIP for perfectly radiative materials. This allowed to calculate an intrinsic upper limit for \( D^* \) of \( 10^{12} \) and \( 10^{10} \) Jones at 1500 and 2000 nm, respectively. The applicability of OPDs to the NIR region above 2 \( \mu \)m seems therefore

**Table 1. Device performance parameters for the photodiodes with the studied donor materials.**

| Material     | \( V_{OC} \) [V] | \( I_{0,R} \) [A]\(^{0}\) | EQE\(_R\) \(^{0}\) | \( \Delta V_{OC,NR} \) [V]\(^{1}\) | \( E_{CT} \) [eV]\(^{0}\) |
|--------------|-----------------|-----------------|-----------------|-----------------|-----------------|
| PTBBA        | 0.40            | 5.59 \times 10^{-16} | 6.54 \times 10^{-8} | 0.42            | 1.07            |
| PBTQ(OH)     | 0.44            | 8.54 \times 10^{-17} | 1.18 \times 10^{-8} | 0.47            | 1.12            |
| PTTQ(HD)     | 0.22            | 4.60 \times 10^{-15} | 4.10 \times 10^{-9} | 0.49            | 0.96            |
| PTTQn(HD)    | 0.12            | 3.65 \times 10^{-15} | 3.86 \times 10^{-9} | 0.49            | 0.81            |

\(^{0}\)Calculated from Equation (9); \(^{1}\)Calculated from Equation 59, Supporting Information (more detailed description in Supporting Information); \(^{2}\)Calculated from Equation (10) (more detailed description in Supporting Information); \(^{3}\)EQE\(_R\) values obtained from the intercept of the reduced EQE\(_{pv}\) and EL spectrum (Figure S13, Supporting Information).
negligible and future research should focus on reducing voltage losses, in which NFAs will likely play an important role.

**Supporting Information**
Supporting Information is available from the Wiley Online Library or from the author.

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**Conflict of Interest**
The authors declare no conflict of interest.

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