Low-frequency and broadband vibration energy harvester driven by mechanical impact based on layer-separated piezoelectric beam

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Abstract  Vibration energy harvesting is to transform the ambient mechanical energy to electricity. How to reduce the resonance frequency and improve the conversion efficiency is very important. In this paper, a layer-separated piezoelectric cantilever beam is proposed for the vibration energy harvester (VEH) for low-frequency and wide-bandwidth operation, which can transform the mechanical impact energy to electric energy. First, the electromechanical coupling equation is obtained by the Euler-Bernoulli beam theory. Based on the average method, the approximate analytical solution is derived and the voltage response is obtained. Furthermore, the physical prototype is fabricated, and the vibration experiment is conducted to validate the theoretical principle. The experimental results show that the maximum power of 0.445 $\mu$W of the layer-separated VEH is about 3.11 times higher than that of the non-impact harvester when the excitation acceleration is 0.2 g. The operating frequency bandwidth can be widened by increasing the stiffness of the fundamental layer and decreasing the gap distance of the system. But the increasing of operating frequency bandwidth comes at the cost of reducing peak voltage. The theoretical simulation and the experimental results demonstrate good agreement which indicates that the proposed impact-driving VEH device has advantages for low-frequency and wide-bandwidth. The high performance provides great prospect to scavenge the vibration energy in environment.

Key words  vibration energy harvester (VEH), layer-separated piezoelectric beam, low-frequency, broad-bandwidth

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1 Introduction

Energy harvesting technology refers to transform the ambient energy, such as solar, heat, sound, wind, and vibration, into electrical energy and store in capacitor or battery in a variety of ways. It is found that the harvesting efficiency of the vibration energy in the environment is as high as 25%–50%. Vibration energy harvesting is a promising technology as it could be used for low-power electronic components, which are widely applied in many engineering fields such as flexible electronics, wearable devices, and wireless sensor system. There are mainly three common electromechanical transformation mechanisms for vibration energy harvesting, i.e., the electromagnetic effect [1–2], the electrostatic effect [3–5], and the piezoelectric effect [6–10]. In recent years, the piezoelectric vibration energy harvesting (PVEH) attracts broad attention because of its advantages of high conversion efficiency and easy integration for microelectromechanical systems.

In nature and engineering fields, there is always lower-frequency environment vibration, while the oscillator frequency of the vibration energy harvester (VEH) is higher. It becomes significant to broaden the response bandwidth and improve the conversion efficiency for the VEH. To achieve this goal, scientists developed many kinds of oscillators based on linear and nonlinear vibration theory, such as bistable models [11–13], multistable models [14–16], and multidegrees models [17–18]. There are also stochastic excitation VEHs [19–21] and nonlinear internal resonance VEHs [7, 13, 22].

In addition to the above-mentioned methods, many researchers proposed impact-driving VEH. Umeda et al. [23] earlier used a free-fall ball to collide a piezoelectric beam to harvest the low-frequency vibration energy. Halim et al. [24] and Halim and Park [25–26] designed several kinds of low-frequency VEHs using stopper-engaged dynamic magnifier. Liu et al. [27] proposed a stopper VEH to increase the energy conversion bandwidth by limiting the vibration amplitude through mechanical stoppers, while the power output was decreased. Li et al. [28] studied a bi-resonant PVEH to extend the operating frequency bandwidth. Yuan et al. [29] focused on the semi-analytical approach based on the harmonic balance method to predict steady-state responses of nonlinear piezoelectric vibration energy harvester, which consists of an oscillator of two magnetic beams. Dechant et al. [30] studied a new kind of low-frequency broadband VEH coupled with two mechanical stoppers, which significantly improves the vibration-to-electricity conversion efficiency by 40%–81%. Two-degree-of-freedom piecewise-linear PVEH was also studied by Liu et al. [31] and Zhao et al. [32], which demonstrates good harvesting performance.

In this paper, a new impact-driving PVEH is proposed for low-frequency and broadband vibration energy harvesting based on a layer-separated piezoelectric beam. Firstly, the generating mechanism of the impact-driving PVEH is introduced in Section 2, where the theoretical model is also constructed. The amplitude-frequency response and the numerical simulation are obtained in Section 3. The vibration experiment is then conducted for the impact-driving VEH and the effects of different parameters on the operating frequency bandwidth and output power of the system are discussed in Section 4. Finally, the conclusion is given.

2 Mechanical configuration and theoretical modeling

2.1 Mechanical configuration and working principle

As shown in Fig. 1, the proposed impact-driving PVEH is made up of a piezoelectric laminated cantilever beam and a tip mass at the free end, where the laminated beam includes a protective layer, a piezoelectric layer, and a fundamental layer. Here, the particularity of this structure is embodied in the separated layers between the piezoelectric layer and the fundamental layer. The piezoelectric layer is made of polyvinylidene fluoride (PVDF) sheet and used for energy conversion while the fundamental layer and the protective layer are made of the brass sheet and the polyester sheet, respectively. For the effect of the clamping force of the fixture,
a gap distance is formed between the fundamental layer and the piezoelectric layer.

![Diagram](image)

**Fig. 1** The mechanical configuration of the proposed impact-driving PVEH, (a) the 3-dimensional schematic, and (b) the mechanical model with coordinate frame (color online)

When an external vibration with an appropriately large amplitude is imposed on the harvester, the separate piezoelectric layer with the protective layer begins to vibrate and the tip mass moves down over the gap distance producing an impact periodically onto the separate fundamental layer. During the impact, the separate fundamental layer acts as a stopper for its greater stiffness and the separate piezoelectric layer vibrates together with the separate fundamental layer for a short moment, leading to the increase of the stiffness of the separate piezoelectric layer which extends the resonance over a wider range. At the same time, for a smaller stiffness of the piezoelectric layer and the protective layer, the harvester can achieve lower vibration frequency and produce higher output voltage.

### 2.2 Theoretical modeling

In order to investigate the effects and potential benefits of the proposed impact-driving PVEH, the governing equation needs to be established and analyzed. But as the premise, it is necessary to introduce the following assumptions: (a) suppose the separate piezoelectric layer with the protective layer for a cantilever beam, and that the impact between the separate parts occurs only at the free ends of the piezoelectric layer and the fundamental layer, (b) suppose that the section plane of the beam is still perpendicular to the middle plane after transverse bending, (c) neglect the moment of inertia, and (d) only the transverse electric field is considered.

Based on the electromechanical coupling, the following linear constitutive equations could be used to compute the potential energy of the piezoelectric material:

\[
\begin{align*}
T_x &= C_{xx}^E S_x - d_{xx} E_z, \\
D_z &= d_{zz} S_x + \varepsilon_{zz}^S E_z,
\end{align*}
\]

where \(T_x\) and \(S_x\) are the stress and the strain along the length of the beam, respectively. \(D_z\) is the electric displacement, while \(E_z\) is the electric field that develops through the thickness of the piezoelectric layer. \(C_{xx}^E\) is the Young’s modulus of the piezoelectric layer. \(d_{xx}\) denotes the piezoelectric strain coefficient. \(\varepsilon_{zz}^S\) is the laminate permittivity.

The relationship between the nonlinear strain and transversal displacement is given as follows:

\[
S_x = -z \left( v'' + \frac{1}{2} v'^2 v'' \right),
\]

where \(v\) is the transversal displacement. The notation \(\left( \cdot \right)'\) indicates the derivative with respect to \(x\).

The kinetic energy of the system is expressed as follows:

\[
T = \frac{1}{2} \int_0^L \left( \rho_p b h_p + \rho_b b h_s \right) (\dot{v}(x, t) + \dot{z}(t))^2 dx + \frac{1}{2} m (\dot{v}(L, t) + \dot{z}(t))^2,
\]

where \(\rho_p\) and \(\rho_b\) are the density of the protective layer and the piezoelectric layer, respectively.
where \( \rho_p \) and \( \rho_s \) are the densities of the piezoelectric layer and the protective layer, respectively. The notation (·) indicates a time derivative while \( m \) is the tip mass and \( z(t) = A \cos(\omega t) \) is the base motion. \( L \) is the effective length of the piezoelectric beam, and \( b \) is the width of the piezoelectric beam. \( h_p \) and \( h_s \) are the thicknesses of the piezoelectric layer and the protective layer, respectively.

The potential energy distributed along the length of the beam is given as follows:

\[
U = \frac{1}{2} \int_{V_p} (T_x S_x - D_z E_z) dV_p + \frac{1}{2} \int_{V_s} E_s S_x^2 dV_s,
\]

where \( E_s \) is the Young’s modulus of the protective layer. \( V_p \) and \( V_s \) are the volumes of the piezoelectric layer and the protective layer, respectively. The electric field \( E_z \) is expressed as follows by regarding the electrodes on the lower and upper surfaces of the piezoelectric layer as a single electrode pair:

\[
E_z = -\frac{V}{h_p},
\]

where \( V \) indicates the voltage of the system.

To obtain the ordinary differential equations, Galerkin’s method with a trial function is used,

\[
v(x, t) = \sum_{n=1}^{m} \phi_n(x) \cdot r_n(t),
\]

where \( n \) indicates the mode shape number. To simplify the derivations, the first mode shape of the cantilever beam is used \((n = 1)\),

\[
\phi_1(x) = C_4 \left(\cosh(\lambda_1 x) - \frac{\cosh(\lambda_1 L) + \cos(\lambda_1 L)}{\sinh(\lambda_1 L) + \sin(\lambda_1 L)}(\sinh(\lambda_1 x) - \sin(\lambda_1 x))\right),
\]

where \( C_4 \) is determined by the following equation:

\[
(\rho_p bh_p + \rho_s bh_s) \int_0^L \varphi_1^2(x) dx + m \varphi_1^2(L) = 1.
\]

The frequency \( \lambda_1 \) can be expressed by

\[
\lambda_1^2 = \omega_1 \sqrt{\frac{\rho_p bh_p + \rho_s bh_s}{C_{xx} I_p + E_s I_s}}.
\]

\( I_p \) and \( I_s \) are the moments of inertia of the fundamental layer and the piezoelectric layer, respectively, and can be described as follows:

\[
I_p = \frac{1}{3} bh_p(3h_p^2 - 3h_p h_n + h_n^3), \quad I_s = \frac{1}{3} bh_s(3h_s^2 + 3h_p h_n + h_n^3),
\]

where \( h_n \) is the position of the neutral axis and can be written as \( h_n = (C_{xx}^E h_p^2 - E_s h_n^2)/2(C_{xx}^E h_p + E_s h_s) \). The first natural frequency of the piezoelectric cantilever beam \( \omega_1 \) is determined by the boundary conditions of the cantilever beam as follows:

\[
\begin{align*}
\varphi_1(x) &= 0 \quad \text{at} \quad x = 0, \\
\frac{d\varphi_1(x)}{dx} &= 0 \quad \text{at} \quad x = 0, \\
(C_{xx}^E I_p + E_s I_s) \frac{d^2 \varphi_1(x)}{dx^2} &= 0 \quad \text{at} \quad x = L, \\
(C_{xx}^E I_p + E_s I_s) \frac{d^3 \varphi_1(x)}{dx^3} &= -m \omega_1^2 \varphi_1(x) \quad \text{at} \quad x = L.
\end{align*}
\]
The motion equation and the voltage equation of the system can be obtained by employing Lagrange’s equations for electromechanical systems,

\[
\begin{align*}
\frac{d}{dt}\left(\frac{\partial T}{\partial \dot{r}_1}\right) - \frac{\partial T}{\partial r_1} + \frac{\partial U}{\partial r_1} &= F_1, \\
\frac{d}{dt}\left(\frac{\partial T}{\partial \dot{V}}\right) - \frac{\partial T}{\partial V} + \frac{\partial U}{\partial V} &= Q,
\end{align*}
\]

where \(F_1\) is a generalized force and can be expressed as follows:

\[
F_1 = \begin{cases} 
-c_1 \dot{r}_1, & r_1 \geq -d/\varphi_1(L), \\
-(c_1 + c_2) \dot{r}_1 - k_2(\varphi_1(L)r_1 + d), & r_1 < -d/\varphi_1(L).
\end{cases}
\]

\(Q\) is the generalized charge and can be obtained by

\[
V = R_L \dot{Q},
\]

where \(R_L\) is the load resistance. \(c_1\) and \(c_2\) are the dampings of the separate piezoelectric layer with the protective layer and the separate fundamental layer, respectively. \(d\) is the gap distance, while \(k_2\) is the stiffness of the separate fundamental layer, and can be expressed as follows:

\[
k_2 = \frac{3E_bI_b}{L^3},
\]

where \(E_b\) and \(I_b = bh^3/12\) are the Young’s modulus and the moment of inertia of the separate fundamental layer, respectively.

Substituting Eqs. (3), (4), (13), and (14) into Eq. (12) yields the ordinary differential equation of the beam

\[
\begin{align*}
\ddot{r}_1 + 2\zeta_1\omega_1 \dot{r}_1 + \omega_1^2 r_1 + f(\dot{r}_1, r_1) - \theta_1 V - \theta_2 r_1^2 V + \kappa \ddot{z} + \kappa_1 r_1^2 + \kappa_2 r_1^2 &= 0, \\
C_p \dot{V} + \frac{V}{R_L} + \theta_1 \dot{r}_1 + \theta_2 r_1^2 \dot{r}_1 &= 0,
\end{align*}
\]

where

\[
f(\dot{r}_1, r_1) = \begin{cases} 
0, & r_1 \geq -\delta, \\
2\zeta_2 \omega_2 \dot{r}_1 + \varphi L k_2(r_1 + \delta), & r_1 < -\delta,
\end{cases}
\]

\[
\omega_2 = \sqrt{\frac{k_2}{(\rho_0 Lbh_0)}}, \quad \zeta_1 = \frac{c_1}{2\omega_1}, \quad \zeta_2 = \frac{c_2}{2\omega_2}, \quad \theta_1 = \frac{1}{2}bd_\alpha(2h_n - h_p) \int_0^L \varphi''_1 dx, \\
\theta_2 = \frac{3}{4}bd_\alpha(2h_n - h_p) \int_0^L \varphi''_1 \varphi''_2 dx, \quad \kappa = (\rho_0 A_p + \rho_s A_n) \int_0^L \varphi_1 dx + m\varphi_1(L), \\
\kappa_1 = 2(C_{xx} I_p + E_s I_s) \int_0^L \varphi''_1 \varphi''_2 dx, \quad \kappa_2 = \frac{3}{4}(C_{xx} I_p + E_s I_s) \int_0^L \varphi''_1 \varphi''_2 dx, \\
C_p = \frac{Lbh_0^3}{h_p}, \quad \delta = d/\varphi_1(L), \quad \varphi_L = \varphi_1(L).
\]

3 Theoretical analysis

3.1 Frequency response equation

When the ratio of deflection to length is less than 1/3 under the appropriate external excitation, the energy harvesting system can be considered as a piecewise-linear system. Therefore, the dynamic equation of the piecewise-linear system can be obtained by ignoring the second,
the third, and the fifth terms of the dynamic equation (16), and can be analyzed by the average method\cite{26,32},

\[
\begin{cases}
\dot{r}_1 + b_1 r_1 = F, \\
\dot{V} + b_2 \dot{r}_1 + b_3 V = 0,
\end{cases}
\]

where

\[
F = -\kappa \ddot{z} - 2\zeta_1 \omega_1 \dot{r}_1 - f(\dot{r}_1, r_1) + \theta_1 V,
\]

\[
b_1 = \omega_1^2, \quad b_2 = \frac{\theta_1}{C_p}, \quad b_3 = \frac{1}{C_p R_L}.
\]

The corresponding derived linear system of Eq. (19) is obtained by ignoring the parameter \(F\), and its solution can be obtained as follows:

\[
\begin{cases}
r_1 = B \cos \psi, \\
\dot{r}_1 = -B \omega_1 \sin \psi, \\
V = -\frac{B b_2 \omega_1}{\omega_1^2 + b_3^2} (\omega_1 \cos \psi - b_3 \sin \psi),
\end{cases}
\]

where \(B\) and \(\psi = \omega t + \theta\) represent the amplitude and the phase angle of the displacement \(r_1\), respectively. \(\theta\) represents the difference between the phase angles of the displacement and the excitation. For the nonlinear equation (19), \(B\) and \(\theta\) are variables relative to time. Therefore, the nonlinear system can be analyzed by analyzing the amplitude and the phase difference of the displacement \(r_1\).

According to the first and the second formulas of Eq. (22), we can get

\[
\frac{dB}{dt} \cos \psi - B \frac{d\theta}{dt} \sin \psi = B(\omega - \omega_1) \sin \psi.
\]

By substituting the linear solution of Eq. (22) into the first formula of the nonlinear equation (19), we can obtain

\[
\frac{dB}{dt} \sin \psi + B \frac{d\theta}{dt} \cos \psi = -\frac{F}{\omega_1} - B(\omega - \omega_1) \cos \psi.
\]

The parameter \(F\) is also replaced by the linear solution of Eq. (22), and can be expressed as

\[
F = \kappa A \omega^2 \cos \omega t + 2B \zeta_1 \omega_1^2 \sin \psi - f(B, \psi) - \frac{B \theta_1 b_2 \omega_1}{\omega_1^2 + b_3^2} (\omega_1 \cos \psi - b_3 \sin \psi),
\]

where the nonlinear term is transformed with the first order Fourier transformation, and can be expressed as follows:

\[
f(B, \psi) = \frac{2}{\pi} \varphi_L k_2 B (\psi_0 \cos \psi_0 - \sin \psi_0) + \frac{1}{\pi} B h_0 (\varphi_L k_2 \cos \psi - 2\omega_1 \omega_2 \zeta_2 \sin \psi),
\]

where \(\psi_0 = \arccos(\delta/B)\) indicates the displacement of the system corresponding to the phase angle of the whole periodic motion when the separate piezoelectric layer collides with the separate fundamental layer, and \(h_0 = \psi_0 - \sin \psi_0 \cos \psi_0\).

The expressions of \(dB/dt\) and \(d\theta/dt\) can be obtained by Eqs. (23) and (24),

\[
\begin{cases}
\frac{dB}{dt} = -\frac{F}{\omega_1} \sin \psi, \\
\frac{d\theta}{dt} = -\frac{F}{\omega_1 B} \cos \psi + \omega_1 - \omega.
\end{cases}
\]
We assume that the average values of phase angle $\theta$ and amplitude $B$ are constant in a complete period. So the expressions of $dB/dt$ and $d\theta/dt$ can be expressed by the average method,

\[
\begin{align*}
\frac{dB}{dt} &= \frac{1}{2\pi} \int_0^{2\pi} \left( -\frac{F}{\omega_1} \sin \psi \right) d\psi = \frac{\kappa A\omega^2}{2\omega_1} \sin \theta - \eta_e(B)B, \\
\frac{d\theta}{dt} &= \frac{1}{2\pi} \int_0^{2\pi} \left( -\frac{F}{\omega_1 B} \cos \psi + \omega_1 - \omega \right) d\psi = -\frac{\kappa A\omega^2}{2B\omega_1} \cos \theta + \omega_e(B) - \omega,
\end{align*}
\]

(28)

where parameters $\omega_e(B)$ and $\eta_e(B)$ represent the equivalent natural frequency and the attenuation coefficient of the amplitude function $B$, respectively,

\[
\begin{align*}
\omega_e(B) &= \omega_1 + \frac{\theta_1 b_2 \omega_1}{2(\omega_1^2 + b_3^2)} + \frac{\varphi_L k_2 h_0}{2\pi \omega_1}, \\
\eta_e(B) &= \zeta_1 \omega_1 + \frac{\theta_1 b_2 b_3}{2(\omega_1^2 + b_3^2)} + \frac{\omega_2 \zeta_2 h_0}{\pi}.
\end{align*}
\]

(29)

As discussed above, the approximate values of $dB/dt$ and $d\theta/dt$ can be assumed to be zero. Therefore, Eq. (28) can be transformed as

\[
\begin{align*}
\kappa A\omega^2 \sin \theta &= 2\omega_1 \eta_e(B)B, \\
\kappa A\omega^2 \cos \theta &= 2\omega_1 (\omega_e(B) - \omega)B.
\end{align*}
\]

(30)

(31)

Since the separate piezoelectric layer vibrates in the resonance region, the parameter $2\omega_1$ in Eq. (30) can be approximately replaced by $2\omega$. The parameter $2\omega_1$ in Eq. (31) can be approximately replaced by $\omega_e(B) + \omega$. Then, the expression of the amplitude-frequency response of amplitude $B$ relative to the excitation frequency can be obtained as follows:

\[
B = \sqrt{\frac{\kappa A\omega^2}{(\omega_e^2(B) - \omega^2)^2 + (2\omega \eta_e(B))^2}}.
\]

(32)

According to Eq. (22), the relationship between the voltage amplitude $V$ and the displacement amplitude $B$ can be obtained as follows:

\[
V = \frac{b_2 \omega_1}{\sqrt{\omega_1^2 + b_3^2}} B.
\]

(33)

3.2 Analytical results

In this section, the parameters of the proposed energy harvester are set as Table 1. In order to get the open circuit voltage of the system, the load resistance $R_L$ is set as 20 MΩ, which is far greater than its internal impedance. Substituting the parameters in Table 1 into Eq. (33), the open circuit voltage of the proposed energy harvester can be obtained. The results are shown by the red curve in Fig. 2, which illustrates the variation of the open circuit voltage with the external excitation frequency. At the same time, the open circuit voltage versus the external excitation frequency of the original harvester is shown by the blue curve. The entire red curve can be divided into three stages when the frequency increases from 0 Hz to 300 Hz. Stage I starts from Point A to Point B, where the open circuit voltage grows with the excitation frequency. Stage II starts from Point B to Point C, where the end of the separate piezoelectric layer with the tip mass reaches the gap distance $d$ and starts to impact the separate fundamental layer. During the impact (Stage II), the stiffness and damping of the separate piezoelectric with the protective layer are increased due to that of the separate fundamental layer, and the amplitude-frequency response curve of output voltage presents a multi-solution phenomenon,
in which the red solid line shows the stable solution while the green dashed line represents the unstable solution. As a result, the equivalent resonance frequency is enlarged. Therefore, the operating frequency bandwidth of the harvester is broadened. Stage III starts when the excitation frequency reaches Point \( C \), where the hard spring characteristic happens and the open circuit voltage of the system drops dramatically to Point \( D \). Then, the open circuit voltage decreases gradually when the excitation frequency changes from Point \( D \) to Point \( E \).

### Table 1  The parameters of the proposed energy harvester

| Position          | Parameter | Value |
|-------------------|-----------|-------|
| **Beam**          | Length \( L \) | 25 mm |
|                   | Width \( b \) | 10 mm |
|                   | Tip mass \( m \) | 0.0015 kg |
|                   | Damping \( c_1 \) | 0.439 kg \( \cdot \) s \(^{-1} \) |
|                   | Young’s modulus \( E_{xx} \) | 2.5 GPa |
|                   | Piezoelectric strain coefficient \( d_{xx} \) | –18 pC/N |
| **Piezoelectric layer** | Thickness \( h_p \) | 0.3 mm |
|                   | Laminate permittivity \( \varepsilon_{zz} \) | 2.18 nF/m |
|                   | Capacitance of piezoelectric layer \( C_p \) | 1.82 nF |
|                   | Density \( \rho_p \) | 2000 kg/m\(^3\) |
| **Protective layer** | Young’s modulus \( E_s \) | 3 GPa |
|                   | Thickness \( h_s \) | 0.1 mm |
|                   | Density \( \rho_s \) | 1000 kg/m\(^3\) |
| **Fundamental layer** | Young’s modulus \( E_b \) | 90 GPa |
|                   | Density \( \rho_b \) | 8900 kg/m\(^3\) |
|                   | Damping \( c_2 \) | 3.328 kg \( \cdot \) s \(^{-1} \) |

As demonstrated by the blue curve, the open circuit voltage grows with the increase of the excitation frequency, and then decreases gradually. From the comparison of the two curves, we can see that the resonance frequency of the separate energy harvester is much smaller than that of the original energy harvester. It is well known that under the same external excitation acceleration, a system will have large amplitude when the external excitation frequency is small. Therefore, the separate energy harvester can harvest low-frequency vibration energy and output larger voltage. At the same time, since the separate parts collide with each other during the vibration process, the operating frequency bandwidth of the system can be widened. Obviously, the separate energy harvester, as a miniature device, has great advantages in collecting low-frequency vibration energy.

The gap distance \( d \) and the stiffness of the separate fundamental layer \( k_2 \) are two important parameters for the power generation of the system. According to Eq. (15), the value of \( k_2 \) depends on the thickness of the fundamental layer \( h_b \), and the values of \( k_2 \) are 388.8 N/m, 921.6 N/m, 1800.0 N/m, and 3100.4 N/m when the values of \( h_b \) are 0.3 mm, 0.4 mm, 0.5 mm, and 0.6 mm, respectively. In order to find out the effect of the change of \( k_2 \) on the output voltage, the external excitation acceleration \( a = A\omega^2 \) is set to 0.2 \( g \), and the gap distance \( d \) is set to 0.5 mm. Figure 3 depicts the curve of open circuit voltage versus the frequency when \( k_2 \) takes different values (388.8 N/m, 921.6 N/m, 1800.0 N/m, and 3100.4 N/m). As shown in the figure, with the increase value of \( k_2 \), the peak voltage of the system will decrease while the operating frequency bandwidth becomes wider. The reason is that a greater \( k_2 \) will increase the values of the equivalent stiffness and the equivalent damping when the separate parts collide with each other. Therefore, as discussed above, the values of the operating frequency bandwidth will be broadened. The operating frequency bandwidth are 3.6 Hz, 5.8 Hz, and 8.9 Hz when the values of \( k_2 \) are 388.8 N/m, 921.6 N/m, and 1800.0 N/m, respectively. Meanwhile, the greater equivalent damping reduces the amplitude of the separate piezoelectric layer, resulting in a reduction of the open circuit voltage.
In order to find out the influence of the parameter $d$ on the generation law of the system, the external excitation acceleration $a$ is set to $0.2 \text{ g}$, and the impact stiffness $k_2$ is set to $1800.0 \text{ N/m}$. Figure 4 shows the curve of the open circuit voltage versus the frequency when $d$ takes different values ($0.2 \text{ mm}, 0.5 \text{ mm}, 0.8 \text{ mm}, 1.0 \text{ mm},$ and $1.5 \text{ mm}$). As it can be seen from the figure, the operating frequency bandwidth is widened while the peak voltage is reduced with the gap distance decreases. The separate piezoelectric layer impacts the fundamental layer early when $d$ is small, bringing in an increase of the stiffness and the damping at an early time. Meanwhile, a small gap distance can enlarge the impact frequency range. Thus, the peak voltage will be decreased and the operating frequency bandwidth will be broadened.

**Fig. 4** The analytical results for the open circuit voltage with different gap distance $d$ (color online)

4 Experimental research

4.1 Experimental setup

The fabricated prototype is shown in Fig. 5. The protective layer and the fundamental layer are bonded on the two surfaces of the polarized piezoelectric layer. The effective length and the width of the beam are $25 \text{ mm}$ and $10 \text{ mm}$, respectively. The thickness of the piezoelectric layer and the protective layer are $0.3 \text{ mm}$ and $0.1 \text{ mm}$, respectively. The piezoelectric beams with the piezoelectric layers of $0.3 \text{ mm}, 0.4 \text{ mm},$ and $0.5 \text{ mm}$ are used in the experiment. A tip mass is fixed on the upper surface of the end of the beam. In the experiment, the piezoelectric layer is separated from the fundamental layer according to the need, and ensured that the gap at the
end of the separate parts to be 0.5 mm by the clamping force from the fixture. The fixture is fixed on the vibration exciter.

The experimental set-ups are shown in Fig. 6. It includes a signal generator, a power amplifier, an electric vibration exciter, a laser displacement sensor, a signal acquisition, a digital oscilloscope, multimeter, and a signal analysis software. Figure 7 shows the flow chart of signal in the experimental process. The signals are sent to the power amplifier by the signal generator. The power amplifier conveys the processed signal to the vibration exciter to control vibration of the piezoelectric energy harvester. Keeping the external excitation acceleration as 0.2 g, the excitation frequency is gradually increased from 0 Hz to 300 Hz by adjusting the signal generator. The output voltage is recorded by a digital oscilloscope directly. The deflection of piezoelectric beam under different excitation accelerations is measured by the laser displacement sensor to ensure that the movement of the beam is not a large deformation movement.
4.2 Experimental results

Figure 8 shows the comparison of the open circuit voltage of experimental results with theoretical results. Here, the red dotted curve shows the theoretical result and the blue triangular curve is the experimental result of the proposed harvester with $k_2 = 388.8\,\text{N/m}$ and $d = 0.5\,\text{mm}$ while the pink solid curve shows the theoretical result and the black dot curve is the experimental result of the original harvester. As can be seen from the figure, the peak voltage of the proposed harvester is higher than that of the original harvester, and the excitation frequency at peak voltage of the proposed harvester is much less than that of the original harvester. That is to say, the proposed harvester is more beneficial to harvest low-frequency vibration energy and acquire high output voltage. The operating frequency bandwidth in the experimental curve is 7.9 Hz more than the theoretical one, while the peak open circuit voltage is 2.12 V less than the theoretical one. The relative errors are calculated as 54.4% and 19.7%, respectively. Those errors are mostly caused by modeling, installation, and measurement. Although some errors occurred in the experimental results, the waveform curve of the open circuit voltage and the performance of broadening the operating frequency bandwidth and improving the peak voltage agreed well with the theoretical results. Therefore, the theoretical model and the analytical solution are proved suitable and can be further optimized.

Figure 9 shows the experimental open circuit voltage for different $k_2$ when $d = 0.5\,\text{mm}$. The red curve and the blue curve show the open circuit voltage of the proposed harvester when the values of $k_2$ are 921.6 N/m and 1800.0 N/m, respectively. The operating frequency bandwidth in the experimental curve is 8.6 Hz more than the theoretical one while the peak open circuit voltage is 1.94 V less than the theoretical one, when the value of $k_2$ is 921.6 N/m. The relative errors are calculated as 17.2% and 19.5%, respectively. The operating frequency bandwidth in the experimental curve is 10.7 Hz more than the theoretical one while the peak open circuit voltage is 1.71 V less than the theoretical one when the value of $k_2$ is 1800.0 N/m. The relative errors are calculated as 16.8% and 20.4%, respectively. From the analysis above, we know that the operating frequency bandwidth increases gradually with the increase of the value of $k_2$ while the relative error of the operating frequency bandwidth of the experimental and theoretical results decrease gradually.

In order to measure the load voltage and the output power of the system, an adjustable load resistance with the variation ranging from 0 MΩ to 12 MΩ which can cover the optimum value is connected to two electrodes of the harvester, and the load voltage is measured by adjusting the value of the resistance. The load voltages and the output powers of the original harvester and the separate harvester with $k_2 = 388.8\,\text{N/m}$ and $d = 0.5\,\text{mm}$ versus the resistance are shown in Fig.10. As can be seen from the figure, the maximum output power 0.445 µW of the separate harvester is obtained when the load resistance is 2.5 MΩ at 36 Hz while the maximum power 0.143 µW of the original harvester is obtained when the load resistance is 0.34 MΩ at 265 Hz. The maximum load voltage and the maximum power of the separate harvester are about 4.78 and 3.11 times higher than those of the original harvester, respectively. The impedance of
piezoelectric layer can be calculated by $R_s = 1/(2\pi fC_p)$. When the impedance of piezoelectric layer $R_s$ equals the load resistance $R_L$, the output power reaches the maximum.

![Graph](image1)

**Fig. 8** The comparison of the open circuit voltage of experimental results with theoretical results (color online)

![Graph](image2)

**Fig. 9** The experimental results for the open circuit voltage with different $k_2$ (color online)

![Graph](image3)

**Fig. 10** The load resistance voltage and the output power of the impact-driving PVEH and the non-impact PVEH (color online)

5 Conclusions

In this paper, an impact-driving piezoelectric vibration energy harvester is proposed for a lower resonance frequency and a higher output power based on a layer-separated piezoelectric laminated cantilever beam. The working mechanism and the theoretical model are introduced in detail. When the ratio of the deflection to length is less than $1/3$ under the appropriate external excitation, the separate piezoelectric harvester can be considered as a piecewise-linear system by ignoring higher order nonlinear terms of the governing equation. The approximate analytical solution is obtained based on the average method. The numerical simulation and vibration test are also conducted and all the results demonstrate the advantages of the vibration energy harvesting for the lower resonance frequency and the higher output power. The operating frequency bandwidth can be widened by increasing the stiffness of the fundamental layer and decreasing the gap distance of the system. But the increasing of operating frequency bandwidth comes at the cost of reducing peak voltage. The maximum power of 0.445 μW of the separate harvester is about 3.11 times higher than that of the original harvester when the
excitation acceleration is $0.2\, g$. Although there are little quantitative errors between the analytical and experimental results, the qualitative results are well consistent with each other. In view of the close agreement of the variation trend between the theoretical values and experimental results, the proposed impact-driving PVEH could be applied to further micro-electromechanical system (MEMS) fabrication process and optimized design.

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