Improved Torque and Efficiency of Induction Motors
by Changing Rotor Structure of Permanent Magnet Assistance Synchronous Reluctance Motors

Bui Minh Dinh, Bui Duc Hung*, Trieu Viet Linh and Dang Quoc Vuong
School of Electrical and Electronic Engineering, Hanoi University of Science and Technology, Vietnam.
* Corresponding author. Email: hung.buidac@hust.edu.vn

ABSTRACT

Many authors have recently studied line start permanent magnet assistance synchronous reluctance. This paper presents a method to improve the electromagnetic torque and efficiency of induction motors of 7.5 kW-4P (which consists of 36 stator slots and 40 rotor bars) by changing the design of the permanent magnet assistance synchronous reluctance rotors. This means that permanent magnets will be inserted into the squirrel cage induction motors (induction motors). The electromagnetic torque and efficiency of an induction motor are analyzed and compared with that of the line-start synchronous reluctance motor via a finite element analysis. In addition, the influence of position and length of permanent magnets on the electromagnetic torque and efficiency of the line-start synchronous reluctance motor is also considered and simulated. The model of this motor Linh is finally designed with four U layered- magnet rotors to verify the developed method. The development of the method is also validated by the practical induction motor.

1. Introduction

The electric energy consumption of electric motors has been increasing in industrial applications up to 40% [1], [2]. For a low efficiency of the electric motor, environmental problems such as emission of greenhouse gases are created. In order to overcome these problems, the minimum energy performance standards (MEPS) are presented by changing the efficiency of motors. In practice, induction motors (IMs) account for 70% of industrial motors because of their low manufacturing cost and simple structure [3]. But, these motors appear copper losses leading to a limitation of the improvement in their efficiency [4]-[5]. Thus, in order to replace these motors, several types of motors are studied to improve the efficiency such as line-start synchronous reluctance motors (LS-SynRMs). For this motor, it can be operated at a synchronous speed without exhibiting secondary copper loss [5].

2. Finite Element approach

As presented in the previous part, so as to compare the efficiency of LS-SynRMs, the IM of and LS-SynRM with same power of 7.5 kW-4P are designed as shown in Figure 1.

It should be noted that the main parameters of stators of two of these motors are the same, but the rotors are different. The number of squirrel cage bars of the reference motors is similar. Table 1 gives the diameter of stator and rotor, the air-gap length, the numbers of slot and poles, the air-gap length and stack length.
The geometric parameters of motor has been already calculated by estimation equations given in [1], [2]. An analytical model has been presented by many steps of computation to determine basic parameters. Based on a torque volume density (TVD) from 65 to 80 kNm/m^3 [5], it is assumed that the rotor diameter (D) is equal to the rotor length (L), the electromagnetic torque (T) of the LS-SynRM can be thus defined as [6]-[8]

\[ T = \frac{\pi}{4} D^2 L, TRV, \]  

(1)

where \( L_s \) is the length of core (m) and D is the out diameter (m).

In general, the LS-SynRM is designed in the same way of the IM. The main parameters (such as the rotor diameter, outer diameter, stator slot, motor length, and airgap length) can be determined as actual factors with the requirement of desired inputs.

3. Design of LS-SynRMs

The permanent magnet (PM) is embedded in rotor configuration of the LS-SynRM to create sufficiently the magnetic voltage for magnetic circuit. In practice, the possible configurations are classified by the shape and position of PMs mounted on the rotor as given in Table 1, such as the number
of poles, slots and stack lengths. The rotor topology of the proposed motor is presented in Figure 2. It can be also considered as the 4 layered-shape magnet arrangement.

![Figure 2. LS-SynRM with 4 layered magnets.](image)

In practice, in order to obtain the maximum reluctance and electromagnetic torque, the number of PMs has to be limited. The PM arrangement is related to the efficient operation in the U-shape. The shapes of prototype model are very complicated to insert the PM inside the rotor. It is also very difficult to compare and combine the effectiveness of the PM position with all different sizes [9] - [11]. In general, the rib gives a fixed value based on the relation of manufacturing limitations. By coupling the Matlab program to CAD automatically program, the change in the number of flux barriers ($K_w$) can be expressed as

$$K_w = \frac{\sum W_{\text{air}}}{\sum W_{\text{iron}}},$$

(2)

where $\sum W_{\text{air}}$ is the total flux barrier width and $\sum W_{\text{iron}}$ is the width of total rotor iron sheets. The total volumes of the PMs are pre-computed. It should be noted that the total weight of copper winding and magnet segment is the same. For one polar, the rotor structures consist of five and six pieces of magnets. Thus, it is very complicated. The geometric parameters of IP and LS-PMA-SYNRM are shown in Table 2.

| Parameter                        | IM (kg) | LS-PMA-SynRM (kg) |
|----------------------------------|---------|-------------------|
| Stator Lam (Back Iron)           | 8.26    | 8.26              |
| Stator Lam (Tooth)               | 5.245   | 5.245             |
| Stator Lamination [Total]        | 13.5    | 13.5              |
| Armature EWdg [Front]            | 1.027   | 1.027             |
| Armature Winding [Active]        | 4.138   | 4.138             |
| Armature EWdg [Rear]             | 1.027   | 1.027             |
| Armature Winding [Total]         | 6.193   | 6.193             |
| Wire Ins. [Total]                | 0.2172  | 0.2172            |
| Rotor Lam (Back Iron)            | 5.248   | 5.37              |
| Rot Inter Lam (Back Iron)        | 2.25E-05| 2.25E-05          |
| IPM Magnet Pole                  | 4.723   | 4.833             |
| Rotor Lamination [Total]         | 10.5    | 10.74             |
| Magnet                           | -       | 0.5               |
| Total                            | 34.55   | 33.98             |

The back electromagnetic force (EMF) of 4U shape is the sinusoidal waveform. It gets a small value of 120V because the flux barriers are quite big and the magnet is considered as an assistance. The EMF of LS-PMA-SynRM is pointed out in Figure 3.
The main data of two models are given in Table 3, which consists of the average and ripple torque. The power losses, cogging torque and efficiency are then computed and simulated. It can be seen that the total losses due to the core loss and magnet eddy-current loss with four-segmented magnets are the lowest. Hence, this is a basic to select this model as the prototype machine.

![Figure 3. The back EMF of two models (SynRM and PMA-SynRM).](image)

| Parameter                             | LS-SynRM | IM     | Unit   |
|---------------------------------------|----------|--------|--------|
| Shaft Torque                          | 55.667   | 44.662 | Nm     |
| Input Power                           | 9273.3   | 7597.5 | Watts  |
| Output Power                          | 8744.1   | 7015.5 | Watts  |
| Total Losses (on load)                | 529.15   | 581.98 | Watts  |
| System Efficiency                     | 94.294   | 92.34  | %      |
| Armature DC Copper Loss (on load)     | 340      | 382.7  | Watts  |
| Magnet Loss (on load)                 | 178.8    | 188.5  | Watts  |
| Stator iron Loss [total] (on load)    | 10.33    | 10.78  | Watts  |
| Harmonic Distortion Line-Line Terminal Voltage | 3.089   | -      | Volts  |
| Phase Terminal Voltage (rms)          | 289.1    | 298.5  | Volts  |
| Harmonic Distortion Phase Terminal Voltage | 11.27   | -      | %      |
| Back EMF Line-Line Voltage (peak)     | 114      | -      | %      |
| Back EMF Line-Line Voltage (rms)      | 82.29    | -      | Volts  |
| Back EMF Phase Voltage (peak)         | 57.39    | -      | Volts  |

3.1. Performance Design

In this part, the model of PMA-SynRM motor is designed and compared with the IM model. The total torque and efficiency are important performance for those motors. The electromagnetic torque \( T \) of this motor is formed by the reluctance torque and magnetic torque. For the PM component, it is produced via the interaction between magnetic field at the airgap and the armature reaction magnetic field. For the reluctance component, it is replaced on the asymmetry between the magnetic circuit of \( d \)-axis and \( q \)-axis. Thus, the term \( T \) can be determined:

\[
T = \frac{3}{2} p \left[ \lambda_{\text{pm}} \cdot i_d + (L_d - L_q) \cdot i_d \cdot i_q \right],
\]

where \( p \) is the number of pole pairs, \( \lambda_{\text{pm}} \) is the linkage flux generated by the PM field, \( i_d, i_q \) are respectively direct and quadrature currents, and \( L_d, L_q \) are respectively direct and quadrature axis inductance. These parameters are calculated due to the rotor magnet barrier and magnet pole V angle.

The electromagnetic torque of reluctance and permanent magnetic parts are depicted in Figure 5. It presents that the 4U shape of PMA-SynRM motor gets a higher average torque of 10 N.m.
Figure 4. Flux density of reluctance (top) and Permanent Magnet Assistance (bottom).

Figure 5. Comparison of ripple torque between 4U and 4V.

Figure 6. Temperature values of LS-PMa-SynRM
For the U shape of LS PMA-SynRM machine, the peak torque and power characteristics have been checked at 1500 rpm. The average torque is 52 Nm and the maximum efficiency is 94% for the phase current density of 5 A/mm². The iron and copper losses have been applied to the thermal simulation. Temperature results are presented in Figure 6 and Figure 7. The maximum temperature of winding is 90.7°C, which is lower than isolation class of H (180°C), and 88°C for the temperature of magnet. The detail values of other parts are given in Table 4.

**Figure 7. Temperature values of IM.**

| No | Component                          | LS-SynRM (T°C) | IM (T°C) |
|----|------------------------------------|----------------|----------|
| 1  | T [Ambient]                        | 40             | 40       |
| 2  | T [Housing - Active]               | 78.717         | 80.987   |
| 3  | T [Stator Lam (back iron)]        | 84.833         | 83.33    |
| 4  | T [Stator Surface]                | 88.403         | 86.24    |
| 5  | T [Rotor Surface]                 | 88.65          | 95.829   |
| 6  | T [Airgap Banding]               | 88.651         | 95.992   |
| 7  | T [Magnet]                         | 88.197         | --       |
| 8  | T [Airgap Banding]               | 88.651         | 95.972   |
| 9  | T [Rotor Lamination]              | 87.674         | 95.502   |
| 10 | T [Shaft - Center]                | 87.375         | 88.422   |
| 13 | T [Active Winding Minimum]       | 87.352         | 90.363   |

### 4. Conclusions

The model of four-layered LS PMA-SynRM motors with U-shape has been successfully proposed. It has been indicated that volume of magnets of the proposed structure is the lowest due to the closer air gap, high torque and power density. In addition, this type of shape gets the lower ripple torque and core loss, because the harmonics of air-gap density are reduced by the increase of magnet layers. In order to check the presented model, a detailed design of a three-phase hybrid rotor PM machine with 36 slots and 4 poles is proposed to verify the torque, power, and efficiency performances. The harmonic orders of the back electromotive force have been shown via the finite element analysis. The obtained results of temperatures from the thermal simulation have been also pointed out to validate the overheat capacity.

### Acknowledgments

This research was supported by Institute for Control Engineering and Automation- ICEA for High Processing Speed Computer and CAD software to complete this paper.
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Dr. Bui Minh Dinh is currently working as a lecturer at Department of Electrical Engineering, School of Electrical and Electronic Engineering, Hanoi University of Science and Technology. He obtained the PhD degree in the Department of Electrical Engineering, TU university, in 2014.

Dr. Bui Duc Hung is currently working as a team leader of electrical machines’s group, and also a lecturer of Department of Electrical Engineering, School of Electrical Engineering, Hanoi University of Science and Technology. He obtained the PhD degree in the Department of Electrical Engineering, Hanoi University of Science and Technology, in 2000.

Dr. Trieu Viet Linh is currently working as a lecturer of electrical machines’s group, and also a lecturer of Department of Electrical Engineering, School of Electrical Engineering, Hanoi University of Science and Technology. He obtained the PhD degree in the Department of Electrical Engineering, Hanoi University of Science and Technology, in 1999.

Assoc. Prof. Dang Quoc Vuong is currently a deputy director of Training Center of Electrical and Electronic Engineering, and also a lecturer of Department of Electrical Engineering, School of Electrical Engineering, Hanoi University of Science and Technology. He obtained the PhD degree in the Electrical Engineering and Computer Science Department of the University of Liège, Belgium, in 2013.