A Coordinated Optimal Strategy for Voltage and Reactive Power Control with Adaptive Amplitude Limiter Based on Flexible Excitation System

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Abstract: The flexible excitation system (FES) is a kind of novel excitation system with two channels for damping control. Besides the basic functions of traditional excitation systems, flexible excitation systems can provide reactive power support for the terminal voltage, and the large-capacity FES can improve the voltage stability and power-angle stability of synchronous generator units. However, with the increase in system capacity and the complication of control objectives, the difficulty of controller design will be increased. The randomness and fluctuation of new energy resources such as photovoltaic and wind turbines may cause disturbance and fault to the power system, which requires the coordinated control strategy for the FES to achieve stability in voltage and power angle. In this paper, the basic characteristics of FES are analyzed, and the mathematic model of the single machine infinite bus (SMIB) system based on FES is derived. The coordinated control strategy based on decoupling control of stator and rotor is proposed according to the optimal objectives of voltage stability and power-angle stability, and the linear optimal excitation control (LOEC) is adopted with the adaptive amplitude limiter (AAL) determined by fuzzy rules. The MATLAB/Simulink platform is established and the results verify the superiority of the proposed LOEC + AAL control strategy in large disturbance working conditions, which showed better robustness. The proposed coordinated control strategy provides an effective solution for industrial application and performance improvement of FES.

Keywords: synchronous generator; voltage source converter (VSC); linear optimal excitation control; adaptive amplitude limiter

1. Introduction

The power system in China is one of the largest in the world. With the changes in the structure of the energy resources in recent years, several problems have received attention, including the low-frequency oscillation, the voltage fluctuation, and the deficiency of reactive power support, because of the wide application of distributed power generations such as photovoltaic and wind turbines [1–6]. The changes in the future energy structure will challenge the voltage and reactive power stability of the power grid.

Reactive power compensation is one of the most important methods for power system stability; traditional reactive power compensation devices include capacitors, generator excitation systems, and the static synchronous compensator (STATCOM), etc. The synchronous generator can provide active power to the grid; as for reactive power support, high-power excitation systems, which are considered the most important reactive power compensation devices, are required for large-capacity synchronous generators to meet the demands of stability and rapidity in disturbance conditions. The traditional excitation
The simulation results were obtained in Section 4, and prove the effectiveness of the proposed strategy in voltage and power-angle stability, compared with the traditional adaptive amplitude limiter was proposed to improve the stability during the transient process. The mathematical model of the optimal control problem is derived in Section 2. Considering these points, a coordinated optimal strategy for voltage and reactive power control based on the flexible excitation system was proposed to achieve optimization in different disturbance conditions and reduce difficulties in engineering realization. An adaptive amplitude limiter was proposed to improve the stability during the transient process. The mathematical model of the optimal control problem is derived in Section 2. The coordinated optimal strategy is designed in Section 3 according to the degree of disturbances; the method of adaptive LOEC control, and fuzzy logic control were used. The simulation results were obtained in Section 4, and prove the effectiveness of the proposed strategy in voltage and power-angle stability, compared with the traditional LOEC controller. Section 5 concludes this paper.

2. Mathematical Model

2.1. Mathematical Model of FES

As shown in Figure 1, the physical structure of the FES in this paper consisted of the voltage source converter (VSC), the DC-link capacitor, and the H-bridge chopper. The three-level topology was adopted, which applied to the units with high voltage and large capacity. Compared with the two-level converters, the number of switching tubes was increased in the three-level topology, but the voltage in each single switching tube was reduced, so it had superiority in security and cost. For the VSC, the mathematical model is shown in Equation (1):

$$\begin{align*}
E_a &= L \frac{dI_a}{dt} + RI_a + \frac{1}{2}(2S_{1a} - S_{1b} - S_{1c})u_{dc1} + \frac{1}{2}(-2S_{2a} + S_{2b} + S_{2c})u_{dc2} \\
E_b &= L \frac{dI_b}{dt} + RI_b + \frac{1}{2}(2S_{1b} - S_{1a} - S_{1c})u_{dc1} + \frac{1}{2}(-2S_{2b} + S_{2a} + S_{2c})u_{dc2} \\
E_c &= L \frac{dI_c}{dt} + RI_c + \frac{1}{2}(2S_{1c} - S_{1b} - S_{1a})u_{dc1} + \frac{1}{2}(-2S_{2c} + S_{2b} + S_{2a})u_{dc2} \\
C_1 \frac{dU_{dc1}}{dt} &= S_{1a} I_a + S_{1b} I_b + S_{1c} I_c - i_L \\
C_2 \frac{dU_{dc2}}{dt} &= S_{2a} I_a + S_{2b} I_b + S_{2c} I_c - i_L
\end{align*}$$

(1)

where the S is the switching function. S = 1 means the upper bridge arm is in the condition mode, otherwise, it is turned off. While the FES transmits reactive power to the terminal, the voltage can be adjusted by changing the control angle, as shown in Equation (2):

$$Q \approx \frac{\delta}{57.3} \frac{3U_s^2}{R} = \frac{U_s^2}{19.1R} \delta$$

(2)
where \( Q \) is the reactive power injection, and \( \Delta \) is the power angle. For the H-bridge, the field winding has three operation modes of \( 0.5U_{dc} \), 0, and \(-0.5U_{dc}\) in the half-voltage chopping condition, while in the full-voltage chopping condition, the three operation modes are \(-U_{dc}, 0, \) and \( U_{dc} \) for intensive excitation. The field voltage is shown in Equation (3) [14]:

\[
U_{fd} = \frac{T_p - T_{sw}}{T_{sw}} u_{dc} = (2D_{Ch} - 1) \cdot u_{dc}
\]

where \( U_{fd} \) is the field voltage, \( T_p \) is the conduction time of IGBT5 and IGBT8, \( T_{sw} \) is the conduction time of IGBT6 and IGBT7, \( T_{sw} \) is the switching period, and \( D_{Ch} \) is the duty ratio of the PWM trigger pulse.

**Figure 1.** Three-level voltage source flexible excitation system.

### 2.2. Mathematical Model of LOEC Control

The major issue in coordinated optimal control is represented by the differences in the inertia time constants because the inertia time constant of the VSC is much lower than the field winding, which is often ignored in other research. In this paper, the state model can be described in Equations (4)–(7):

\[
\Delta \dot{\omega} = \omega_0 \Delta \omega
\]

\[
\Delta \omega = (-\Delta P_e - D \Delta \omega)/M
\]

\[
\Delta E'_q = (\Delta E_{fd} - \Delta E_q)/T'_{d0}
\]

\[
\Delta I_s = -\frac{1}{T} \Delta I_s + \frac{1}{T} \Delta I_{Sref}
\]

where \( \omega \) is the rotate speed of the synchronous generator, \( E_q \) is the q-axis voltage, \( I_s \) represents the reactive current of the VSC, \( T'_{d0} \) is the d-axis transient time constant of the synchronous generator, and the Equations (8)–(10) can be obtained:

\[
\Delta P_e = K_1 \Delta \delta + K_2 \Delta E'_q + K_{10} \Delta I_s
\]

\[
\Delta E'_q = \frac{K_3}{1 + T'_{d0} K_3 \delta} \left( \Delta E_{fd} - K_4 \Delta \delta + K_{11} \Delta I_s \right)
\]

\[
\Delta U_r = K_5 \Delta \delta + K_6 \Delta E'_q + K_{12} \Delta I_s
\]

Thus, the state model is transformed into:
where the parameters $K_1$ to $K_5$, and $K_6$ to $K_{10}$ are [15]:

$$K_1 = \frac{(x_q - x_d')i_{q0}V_b \sin \delta_0}{x_q' \Sigma} + \frac{E'_{q0}V_b \cos \delta_0}{x_q \Sigma}$$ (12)

$$K_2 = i_{q0} \frac{x_q \Sigma}{x_q' \Sigma}$$ (13)

$$K_3 = \frac{x_q' \Sigma}{x_q \Sigma}$$ (14)

$$K_4 = \frac{(x_d - x_d')V_b \sin \delta_0}{x_q' \Sigma}$$ (15)

$$K_5 = \frac{U_{d0}x_q V_b \cos \delta_0}{x_q' \Sigma U_{l0}} - \frac{U_{b0}x_q V_b \sin \delta_0}{x_q' \Sigma U_{l0}}$$ (16)

$$K_6 = \frac{U_{b0}x_q}{x_q' \Sigma U_{l0}}$$ (17)

$$K_{10} = \frac{x_S}{U_{d0}^2} \frac{x_d'^2 V_b^2 \sin \delta_0}{x_d'^2 \Sigma^2} - \frac{x_S}{U_{l0}} \frac{V_b^2 (x_q - x_d')}{x_d'^2 \Sigma^2} \left( \frac{x_d'}{x_q} + \frac{x_q}{x_d'} \right) \sin 2\delta_0$$ (18)

$$K_{11} = \frac{(x_d - x_d')x_S U_{l0}}{x_q' \Sigma U_{l0}^2}$$ (19)

$$K_{12} = \frac{x_S}{V_{d0}^2} \left( \frac{x_S U_{d0}^2}{x_q' \Sigma^2} + \frac{x_d'^2 U_{l0}^2}{x_q' \Sigma^2} \right)$$ (20)

Considering the time constant of the VSC, the Philips–Heffron model is shown in Figure 2.

![Figure 2. Philips–Heffron model based on flexible excitation system.](image-url)
According to Equation (21):
\[
\begin{bmatrix}
\Delta P_e \\
\Delta \omega \\
\Delta U_t \\
\Delta I_S \\
\end{bmatrix} =
\begin{bmatrix}
K_1 & 0 & K_{10} \\
0 & 1 & 0 & 0 \\
K_5 & 0 & K_6 & K_{12} \\
0 & 0 & 0 & 1 \\
\end{bmatrix}
\begin{bmatrix}
\Delta \delta \\
\Delta \omega \\
\Delta E_q \\
\Delta I_S \\
\end{bmatrix} +
\begin{bmatrix}
0 & 0 \\
0 & 0 \\
0 & 0 \\
0 & 0 \\
\end{bmatrix}
\begin{bmatrix}
\Delta E_{fd} \\
\Delta I_{gref} \\
\end{bmatrix}
\tag{21}
\]

where the \( P_e \) is the electromagnetic power of the synchronous generator that can be represented by active power \( P_{eq} \), so the Equation (21) can be reorganized into the new form:
\[
z = T_1 x + T_2 u, z_1 = z - T_2 u = z = T_1 x, \dot{z}_1 = \dot{z} = T_1 \dot{x}.
\]

The Riccati equation can be obtained in Equation (22):
\[
P \bar{A} + \bar{A}^T P - P \bar{B} \bar{R}^{-1} \bar{B}^T P = -Q
\]
\tag{22}

where: \( \bar{A} = T_1 A T_1^{-1}, \bar{B} = T_1 B, \bar{Q} = (T_1^{-1})^T Q T_1^{-1} \).

Thus, the optimal coordinated controller is:
\[
u = -(I - \bar{K} T_u)^{-1} \bar{K} z.
\]

As shown in Equation (23):
\[
\begin{bmatrix}
\Delta E_{fd} \\
\Delta I_{gref} \\
\end{bmatrix} =
\begin{bmatrix}
k_{Ep} & k_{E\omega} & k_{EU} & k_{ES} \\
k_{Ip} & k_{I\omega} & k_{IU} & k_{IS} \\
\end{bmatrix}
\begin{bmatrix}
\Delta P_e \\
\Delta \omega \\
\Delta U_t \\
\Delta I_S \\
\end{bmatrix}
\tag{23}
\]

In summary, the control variables were the field voltage \( \Delta E_{fd} \) and the reference reactive current \( \Delta I_{gref} \), which are determined by the electromagnetic power \( \Delta P_e \), the rotate speed \( \omega \), the terminal voltage \( \Delta U_t \), and the VSC current \( \Delta I_S \).

3. The Adaptive LOEC with Amplitude Limiter

3.1. The Adaptive LOEC

The adaptive LOEC strategy of the VSC was achieved by controlling the weight matrix \( R \) (or weight matrix \( Q \)) in real-time. According to Equation (24), where the \( P(t) \) is obtained by calculating the Ricatti function in Equation (22), \( R(t) \) is a diagonal matrix:
\[
u = -R^{-1}(t) \bar{B}^T(t) P(t) x'(t)
\tag{24}
\]

According to Equation (25), the variation of \( R \) is determined by the limiter \( U_{\text{max}} \). So, the element value of \( R \) will be smaller while the limiter is unlocked and the \( U_{\text{max}} \) becomes larger:
\[
R = \frac{1}{U_{\text{max}}} R_0
\tag{25}
\]

where \( k \) is a constant coefficient and usually \( k = 1 \). In this case, the conventional linear-quadratic optimal controller cannot explain the phenomenon of limited control. \( R_0 \) is the initial value of \( R \).

According to the minimum principle:
\[
H[x,u,\lambda] = \frac{1}{2} [x^T(t)Q(t)x(t) + u^T(t)R(t)u(t)] + \lambda^T(t)[A(t)x(t) + B(t)u(t)]
\tag{26}
\]

Where the quadratic matrix \( Q \) and \( R \) is obtained by pole assignment, \( H \) is the Hamiltonian function, the adjoint equations are:
\[
\dot{\lambda}(t) = -\frac{dH}{dx} = -Q(t)x(t) - A^T(t)\lambda(t)
\]
\[
\dot{x}(t) = \frac{dH}{dx} = A(t)x(t) + B(t)u(t) = A(t)x(t) - B(t)R^{-1}(t)B^T(t)\lambda(t)
\tag{27}
\]
So, the minimum principle is to realize the minimization of Hamiltonian function:

$$H[x^*, \lambda^*, u^*, t] = \min_{u \in U} H[x^*, \lambda^*, u, t]$$  \hfill (28)

3.2. The Amplitude Limiter

The amplitude limiter was unlocked, determined by Equations (29) and (30):

\begin{align*}
\Delta t &< \Delta t_{\text{max}} \\
\Delta P_{\text{eo}} &< \Delta P_{\text{eo,max}} \\
\Delta V_t &< \Delta V_{\text{max}} \\

u(1) &= U_{\text{max}} \\
u(2) &= -R^{-1}(t)B^T(t)P(t)x^*(t) \\
u(3) &= U_{\text{min}}
\end{align*}

where $\Delta t$ is the time of voltage instability, $\Delta P_{\text{eo}}$ is the deviation of active power $P_{\text{eo}}$, $\Delta V_t$ represents the deviation of terminal voltage, and $u(2)$ represents the control variables $E_{\text{fd}}$ and $Q$ within the interval determined by $U_{\text{max}}$ and $U_{\text{min}}$.

Where the mode selection from $u(1)$ to $u(3)$ are determined by Figure 3. When the voltage instability or power-angle instability occurs, the system state is definitive and the control variables $u$ can be selected by Equation (30).

![Figure 3. Mode selection of the amplitude limiter.](image)

3.3. The Fuzzy Controller

The fuzzy controller for amplitude limiter was designed in Table 1 and Figure 4, according to the fuzzy rules and fuzzy relationship function [16], where MN (Min), L (Low), SL (Small low), M (Middle), SH (Small high), H (High), and MX (Max) are the rules for determining the values of inputs and the output. The inputs of the fuzzy controller were active power deviation $\Delta P_{\text{eo}}$ and the voltage deviation $\Delta V_t$, and the output was the limit value $U_{\text{max}}$. While the system instability occurs, the voltage deviation $\Delta V_t$ and active power deviation $\Delta P_{\text{eo}}$ determined the limiter output $U_{\text{max}}$ according to Equation (25), and the variation of the weight matrix $R$ was determined at the same time.
Switch was designed in Equation (31). Where \( \Delta Pe \) is the signal for mode switching, while \( \Delta Pe = 0 \), the LOEC + AAL was adopted, otherwise the mode was LOEC. In Equation (31), \( \Delta Pe_0 > \Delta Pe_1 \) and \( \Delta Vt_0 > \Delta Vt_1 \): the Schmitt trigger.

\[
\text{Switch}(Pe) = \begin{cases} 
0 & \Delta Pe > Pe_0 \\
1 & \Delta Pe < Pe_1
\end{cases} \\
\text{Switch}(Vt) = \begin{cases} 
0 & \Delta Vt > \Delta Vt_0 \\
1 & \Delta Vt < \Delta Vt_1
\end{cases} \\
\text{Switch}(t) = \begin{cases} 
0 & \Delta t > t_0 \\
1 & \Delta t < t_0
\end{cases} \\
\text{Switch} = \text{Switch}(Pe) \& \text{Switch}(Vt) \& \text{Switch}(t)
\]

The topology of FES and the structure of the coordinated controller used the LOEC in the small disturbance interval (the interval 1), and adopted the fuzzy controller to adaptively adjust the weight matrix and the limiter output when the large disturbance occurs, as shown in Figure 5, where the \( U_t^* \) is the reference value of the terminal voltage.

| Reference Value | Active Power Deviation \( \Delta Pe \) |
|-----------------|----------------------------------------|
| Voltage deviation \( \Delta Vt \) |
| L | SL | M | SH | H |
| L | MN | L | L | SL | SL |
| SL | L | L | SL | SL | M |
| M | SL | SL | M | M | SH |
| SH | M | M | SH | SH | H |
| H | SH | SH | H | H | MX |

Table 1. Fuzzy Control Rules for Amplitude Limiter.

![Fuzzy control membership function](image)

Figure 4. Fuzzy control membership function. (a) three-dimensional view (b) relationship of \( \Delta Pe \) and \( \Delta Ut \) (c) relationship of \( U_{max} \) and \( \Delta Ut \) (d) relationship of \( U_{max} \) and \( \Delta Pe \).
The control diagram of the proposed strategy is shown in Figure 6. The control system adopted two fuzzy controllers to give the limiter output $U_{\text{max}}$ and $U_{\text{max2}}$, and these fuzzy controllers could also be incorporated if the fuzzy rules were the same, for simplicity.

![Figure 5. Coordinated control strategy based on decoupling control of stator and rotor.](image)

![Figure 6. The control diagram of LOEC + AAL.](image)

4. Simulation Verification

4.1. The Simulation Platform

A simulation model was built under the MATLAB/SIMULINK. The structure of the simulation platform is shown in Figure 7, and the coordinated control system was established according to Figure 5. The structure of the LOEC + AAL controller in the simulation platform is shown in Figure 8, which used the same fuzzy rules and input quantization factors $K_{pe}/K_{v}$ in two fuzzy controllers for simplification; the output quantization factors $K_{out1}/K_{out2}$ were different. The parameters of SMIB system are shown in Table 2.
4.2. The Simulation Results

According to the analysis and parameters above, the effectiveness of the proposed strategy was compared with the traditional LOEC control strategy in this part. The contrast tests included different disturbances in the power system that could cause large disturbances on voltage or power-angle stability. Three typical disturbances were tested in the simulation verification. The three-phase short-circuit fault was divided into ”proximal fault” and ”remote fault”. The proximal fault occurred at a short distance from the synchronous generator; the short-circuit impedance was small and the terminal voltage fell to around 0.2 p.u. The remote fault occurred at a long distance from the synchronous generator; the short-circuit impedance was large and the terminal voltage fell to around 0.8 p.u. The disconnection fault occurred at one of the double transmission lines, which caused disturbance to the power angle.

Table 2. Parameters of simulation experiment.

| Parameter                                                | Value         |
|----------------------------------------------------------|---------------|
| Rated terminal voltage                                   | 10.5 kV       |
| Rated system voltage                                     | 220 kV        |
| Voltage rate of main transformer                          | 220 kV/10.5 kV|
| Voltage rate of excitation transformer                    | 10.5 kV/380 V |
| Rated frequency                                           | 50 Hz         |
| Rated capacity of VSC                                     | 6.47 Mvar     |
| Rated capacity of synchronous generator                   | 64.7 MVA      |
| Rated capacity of main transformer                        | 150 MVA       |
| Reactance of main transformer                             | 0.1 pu        |
| Reactance of power line                                   | 0.23 pu       |
| d-axis transient reactance Xd’ of synchronous generator   | 0.314 pu      |
| d-axis transient time constant Tdo’ of synchronous generator | 7.31 s       |

Figure 7. Structure of simulation platform.

Figure 8. Structure of LOEC + AAL controller.
4.2.1. Three Phase Proximal Fault

The single-phase proximal fault occurred at 1 s and continued for 0.1 s. Figure 9 shows the contrasted results of terminal voltage, active power, field voltage, and the reactive power injection of the VSC. The LOEC + AAL could provide higher $E_{fd}$ and $Q$ while the proximal fault occurred, and the excitation transformer could resist the impact within a short period, which brought better stability. As can be seen in Figure 9, the voltage returned to interval 1 within 1.5 s and the power angle returned to interval 1 within 2 s under the LOEC + AAL, whereas it needed more than 3.5 s under the LOEC.

![Figure 9](image-url)

**Figure 9.** Comparison of coordinated control and traditional control under the single phase proximal fault. (a) terminal voltage (b) active power (c) field voltage (d) reactive power injection.

4.2.2. Three Phases Remote Fault

The three-phase remote fault occurred at 0.5 s and continued for 0.1 s. Figure 10 shows the contrasted results, which were similar to the results in Figure 9. In the transient period, the LOEC + AAL strategy increased the limit of field voltage and reactive power in a short time, which effectively stabilized the fluctuation in the power system and improved transient stability.

![Figure 10](image-url)

**Figure 10.** Comparison of coordinated control and traditional control under the three phase remote fault. (a) terminal voltage (b) active power (c) field voltage (d) reactive power injection.
4.2.3. The Disconnection Fault

The disconnection fault occurred at 0.5 s and continued for 0.1 s. Figure 11 shows the contrasted results. The disconnection fault resulted in changes in system parameters, especially the system impedance, as the line of \(X_{L1}\) or \(X_{L2}\) was cut off, and, in this case, the LOEC + AAL controller could better improve the voltage and power stability than traditional LOEC. The system returned to the stable value within 1 s under the LOEC + AAL, which is a great improvement compared with the LOEC.

![Figure 11](image1)

**Figure 11.** Comparison of coordinated control and traditional control under the disconnection fault. (a) terminal voltage (b) active power (c) field voltage (d) reactive power injection.

In summary, the LOEC + AAL coordinated controller stabilized the system within 2 s under large disturbances, which is a great improvement compared with the conventional LOEC strategy, according to Figures 9–11. As the control objective was to keep the stability of the terminal voltage and the power angle, it was suitable to increase the field voltage \(E_{fd}\) and the reactive power injection \(Q\) in a short period, which could be quickly stabilized after 3 s, and the safety of the excitation transformer could be guaranteed.

4.2.4. The Fuzzy Amplitude Limiter

The reference value of the fuzzy amplitude limiter is shown in Figures 12 and 13. Taking the reference value of \(U_{max1}\) (the output of the limiter to restrict the variation of field voltage) for example, the fuzzy amplitude limiter changed \(U_{max1}\) according to different kinds of disturbances, in Figure 12. The weighted matrix \(R1\) also changed in real-time corresponding to \(U_{max1}\), in Figure 13. In these three typical fault cases, when the system was detected to enter the large disturbance interval (the interval 2) in which the voltage or the power angle exceeds the reference values of stability, the limiter opened and adjusted the duration according to the length of the fault. However, if the duration of the fault was too long (longer than 4 s), the limiter closed mandatorily to protect the safety of the excitation transformer. In the studied cases, the duration of transient periods was no longer than 2 s under the fuzzy amplitude limiter control.
Figure 12. The fuzzy amplitude limiter for $U_{max1}$.

Figure 13. The fuzzy amplitude limiter for $R1$.

5. Conclusions

The flexible excitation system can provide rapid reactive power injection for the synchronous generator, improve the stability of the terminal voltage as well as the power-angle stability, ameliorate the damping characteristics, and guarantee the robustness of the power system. In this paper, a kind of FES was introduced; the mathematical model of the coordinated control problem was figured out for designing the LOEC + AAL controller. The LOEC + AAL control strategy, which combines the methods of linear optimal excitation control, the adaptive amplitude limiter control, and the fuzzy logic control, was proposed for improving the stability of the power system and releasing the capability of the excitation transformer. The simulation results prove the effectiveness of the LOEC + AAL strategy, compared with the conventional LOEC algorithm. As the FES has been connected to the grid in field operation, the novel control algorithm proposed in this paper can be applied in engineering practice. In future works, the parameters of the LOEC + AAL controller will be calculated and tested in detail for industrial applications.

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