Exploiting non-orthogonal multiple access in device-to-device communication

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Article Info

ABSTRACT
This paper proposes an uplink non-orthogonal multiple access (NOMA) system with device-to-device (D2D) communication, enabling NOMA users to communicate with other users/devices using D2D communication to improve the system capacity. In the NOMA-D2D system, two cellular users communicated with the BS using uplink NOMA, and two cellular users simultaneously communicated with the D2D users using downlink NOMA. Closed-form solutions for the ergodic sum capacity of the proposed system are derived analytically. The analytical results are validated via simulations and they are compared with the results obtained from conventional schemes. The comparison shows that, in scenarios where efficient interference cancellation can be achieved, the proposed NOMA system with the D2D model can achieve higher capacity gains than conventional benchmark schemes. When $\rho=20$ dB, NOMA-D2D achieves capacity gains of 192.2% and 157.5% over the conventional OMA and the time-sharing-based NOMA, respectively.

Keywords:
Device-to-device
Ergodic sum capacity
Non-orthogonal multiple access
Successive interference cancellation
Uplink

1. INTRODUCTION
Owing to the recent focus on the design of next-generation wireless communication systems and the pursuit to achieve high spectral efficiency in such systems, non-orthogonal multiple access (NOMA) has attracted significant research interest [1]-[7]. According to the NOMA principle, multiple users can communicate over the same resource block (RB), which can be a time slot, a frequency subcarrier, or a scrambling code depending on the context. This sharing process significantly enhances the spectral efficiency of the system [8]-[13].

Among the available variants of the NOMA principle, power domain NOMA (PD-NOMA) has gained much research interest recently. In PD-NOMA, multiple users (in a user pair) share the same RB. Signal separation is performed in the power domain, by superimposing the signals of the users and superimposed and allocating different power levels to the signals associated to each set of paired users [11], [14], [15]. To recover the signal at the receiver’s end, efficient interference cancellation techniques, such as successive interference cancellation (SIC), are used to the higher channel gain user, while direct signal recovery is applied to the lower channel gain user [10]. It has been reported in the literature that, through efficient user pairing and power allocation, NOMA can achieve higher capacity gains compared to conventional OMA schemes [16].
2. RESEARCH METHOD

This section presents a model of the system being considered and explains the transmission protocol.

a) The uplink NOMA system is investigated, where one of the two NOMA users simultaneously communicates with multiple other users/devices using D2D communication.

b) The ergodic sum capacity of the system is analyzed analytically, validated via simulations, and compared to that of conventional OMA and time-sharing NOMA systems.

c) The proposed model is demonstrated to achieve higher capacity gains than the OMA and time-sharing NOMA schemes, under scenarios where interference cancellation can be ensured.

2.1. System model

Figure 1 depicts a model of the single-cell uplink transmission system consisting of a base station (BS), two cellular users (UE1, UE2), and D2D users. UE1 is located at the center of the cell, and UE2 is located at the edge of the cell. Further, the D2D users are located very close to UE1 and UE2. Each node is equipped with a single antenna. The channel coefficients for the links UE1 → BS, UE2 → BS, UE1 → D1i, and UE2 → D2i are independent Rayleigh flat fading variables with channel coefficients $h_{1b} \sim \mathcal{N}(0, \lambda_{1b} = d^{-\alpha}_{1b})$, $h_{2b} \sim \mathcal{N}(0, \lambda_{2b} = d^{-\alpha}_{2b})$, $h_{1i} \sim \mathcal{N}(0, \lambda_{1i} = d^{-\alpha}_{1i})$ and $h_{2i} \sim \mathcal{N}(0, \lambda_{2i} = d^{-\alpha}_{2i})$ respectively, where $i \in \{1, 2, \ldots, N\}$. $\alpha$ is the path loss exponent, and $d$ denotes distance. Moreover, as $d_{1b} < d_{2b}$ and $d_{1i} < \cdots < d_{1b} < \cdots < d_{1i} < d_{1b}$, it is expected that $|h_{1i}|^2 > |h_{2b}|^2$ and $|h_{1i}|^2 > \cdots > |h_{1i}|^2 > |h_{1b}|^2$. Furthermore, $d_{1i} = d_{2i}$ and $|h_{1i}|^2 = |h_{2i}|^2$ are assumed for simplicity of the argument.

2.2. Transmission protocol and received SNRs

2.2.1. Transmission protocol

As shown in Figure 1, the cellular users (UE1, UE2) transmit signals to the BS using the uplink NOMA principle. Moreover, when cellular users transmit signals to the BS, they simultaneously send the low power message signals to the more closely located D2D users using the downlink NOMA principle. Perfect successive interference cancellation (SIC) for NOMA signals is assumed not only at the BS but also for the D2D users. The transmission protocol is presented in detail as follows.

$UE_1$ transmits a superimposed composite signal $S_1 = \sqrt{\alpha_1 P_{x_{1b}}} + \sum_{k=1}^{N} \sqrt{\beta_{1k} P_{x_{1k}}}$ to the BS and to $D1i$. Simultaneously, $UE_2$ transmits a superimposed composite signal $S_2 = \sqrt{\alpha_2 P_{x_{2b}}} + \sum_{k=1}^{N} \sqrt{\beta_{2k} P_{x_{2k}}}$, where the first and second terms represent the $UE_1 \rightarrow$ BS and $UE_2 \rightarrow D1i$ message signals, respectively. The parameters $\alpha_1$, $\alpha_2$, $\beta_{1i}$, and $\beta_{2i}$ are the power allocation coefficients for symbols, where $\alpha_1 > \alpha_2 > \beta_{1i} = \beta_{2i} > \cdots > \beta_{1i} = \beta_{2i}$, $\alpha_1$ and $\alpha_2$ are power allocation factors for the message signals from $UE_1$ and $UE_2$ to the BS, respectively. Moreover, the power allocation factors for the D2D users are $\beta_{1i}$ and $\beta_{2i}$ for $UE_1 \rightarrow D1i$ and $UE_2 \rightarrow D2i$, respectively. Their values are very small because of the following factors: (a) the very small distances between the cellular users and the D2D users, (b) small data rate requirements of the D2D users, and (c) the fact that larger values of $\beta_{1i}$ and $\beta_{2i}$ would cause interference for the cellular users and the BS. Furthermore, it is assumed that $UE_1$ and $UE_2$ cooperate with each other to adjust the power allocation factors for their signals.
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2.2.2. Received SNRs

The term corresponding to the SNR received from a D2D user at the BS can be ignored because its value is negligible. This is because the power allocation factors for D2D users are very small. After receiving the superimposed signals, the BS first decodes the higher power signal $x_{1b}$ from UE1 by treating the lower power signal $x_{2b}$ of UE2 as noise. Subsequently, the BS performs SIC to decode the signal $x_{2b}$. Thus, the SINRs received for the signals $x_{1b}$ and $x_{2b}$ at the BS are respectively denoted by:

\begin{align}
\gamma_{1b} &= \frac{a_1 \rho |h_{1b}|^2}{a_2 \rho |h_{2b}|^2 + 1} \\
\gamma_{2b} &= a_2 \rho |h_{2b}|^2
\end{align}

where $\rho \triangleq \frac{P}{\sigma^2}$.

Meanwhile, when $D_{1i}$ receives the superimposed signal, it performs SIC to obtain the signal $x_{1i}$. The term corresponding to the signal transmitted from $UE_2$ to the D2D user is assumed to be negligible due to channel attenuation. Therefore, the SINR received for the signal $x_{1i}$ at $D_{1i}$ is given by:

$$\gamma_{1i} = \frac{\beta_{1i} \rho |h_{1i}|^2}{\sum_{k=1}^{i-1} \beta_{1k} \rho |h_{1k}|^2 + 1}$$

Likewise, the SINR received for the signal $x_{2i}$ at $D_{2i}$ is given by:

$$\gamma_{2i} = \frac{\beta_{2i} \rho |h_{2i}|^2}{\sum_{k=1}^{i-1} \beta_{2k} \rho |h_{2k}|^2 + 1}$$

3. ERGODIC CAPACITY ANALYSIS

This section investigates the achievable capacity of individual signal, and then computes the overall ergodic sum capacity of the system [23].

3.1. Ergodic capacity of $UE_1$

Using (1), the achievable capacity associated to the signal $x_{1b}$, is given by:

$$C_{x_{1b}} = \log_2(1 + \gamma_{1b})$$

Let $X_{1b} = \frac{a_1 \rho |h_{1b}|^2}{a_2 \rho |h_{2b}|^2 + 1}$. As in (6), the cumulative distribution function (CDF) of $X_{1b}$, $F_{X_{1b}}(x)$ can be written as follows.

$$F_{X_{1b}}(x) = 1 - \left( \frac{a_1 \rho A_{1b}}{a_2 \rho A_{1b} + a_2 \rho A_{2b} x} \right) e^{-\frac{x}{a_1 \rho A_{1b}}}$$

Using the identity $\int_0^\infty \log_2(1 + x) f_X(x) \, dx = \frac{1}{\ln 2} \int_0^\infty \frac{1}{1 + x} \, dx$, the ergodic capacity (EC) $C_{x_{1b}}$ of $x_{1b}$ is computed as in (7).
\[
\tilde{C}_{x_{1b}} = \frac{1}{\ln 2} \int_0^\infty \frac{1}{1+x} \left( \frac{\alpha_1 \rho \lambda_{1b}}{\alpha_1 \rho \lambda_{1b} + \alpha_2 \rho \lambda_{2b}} \right) e^{-\frac{x}{\alpha_1 \rho \lambda_{1b}}} dx
\]

Using \( Ei(u) = \int_{-\infty}^u \frac{e^x}{x} dx \), \( u < 0 \), where \( Ei \) represents the exponential integral function, the ergodic capacity \( \tilde{C}_{x_{1b}} \) is computed as follows.

\[
\tilde{C}_{x_{1b}} = -\frac{1}{\ln 2(1-2^\lambda)} \left( e^{\frac{1}{\alpha_1 \rho \lambda_{1b}}} Ei \left( -\frac{1}{\alpha_1 \rho \lambda_{1b}} \right) - e^{\frac{1}{\alpha_2 \rho \lambda_{2b}}} Ei \left( -\frac{1}{\alpha_2 \rho \lambda_{2b}} \right) \right) = -\frac{1}{\ln 2(1-2^\lambda)} \left( e^{\frac{1}{\rho \lambda_{1b}}} Ei \left( -\frac{1}{\rho \lambda_{1b}} \right) - e^{\frac{1}{\rho \lambda_{2b}}} Ei \left( -\frac{1}{\rho \lambda_{2b}} \right) \right)
\]

where \( p = \alpha_1 \rho \lambda_{1b} \), and \( q = \alpha_2 \rho \lambda_{2b} \).

3.2. Ergodic capacity of \( U_{E_2} \)

Using (2), the achievable capacity associated to the signal \( x_{2b} \) is given by:

\[
C_{x_{2b}} = \log_2(1 + y_{x_{2b}})
\]

Let \( X_{2b} = y_{x_{2b}} = \alpha_2 \rho |h_{2b}|^2 \). The CDF of \( X_{2b} \), \( F_{X_{2b}}(x) \) can be written as follows.

\[
F_{X_{2b}}(x) = 1 - e^{-\frac{x}{\alpha_2 \rho \lambda_{2b}}}
\]

Similar to the calculation of \( \tilde{C}_{x_{1b}} \), \( \tilde{C}_{x_{2b}} \) can be computed as follows.

\[
\tilde{C}_{x_{2b}} = \frac{1}{\ln 2} \int_0^\infty \frac{1}{1+y} e^{-\frac{y}{\alpha_2 \rho \lambda_{2b}}} dy
\]

Using \( Ei(u) = \int_{-\infty}^u e^x \frac{1}{x} dx \), \( u < 0 \), \( \tilde{C}_{x_{2b}} \) of \( x_{2b} \) is computed as follows.

\[
\tilde{C}_{x_{2b}} = -\frac{1}{\ln 2} \left( e^{\frac{1}{\alpha_2 \rho \lambda_{2b}}} Ei \left( -\frac{1}{\alpha_2 \rho \lambda_{2b}} \right) \right) = -\frac{1}{\ln 2} \left( e^{\frac{1}{\rho \lambda_{2b}}} Ei \left( -\frac{1}{\rho \lambda_{2b}} \right) \right)
\]

where \( q = \alpha_2 \rho \lambda_{2b} \), exactly as defined earlier in the case of the UE1 ergodic capacity.

3.3. Ergodic capacity of \( D_{11} \)

Using (3), the achievable capacity associated to the signal \( x_{11} \) is given by:

\[
C_{x_{11}} = \log_2(1 + y_{11})
\]

Let \( Y_{11} = y_{11} = \beta_{11} \rho |h_{11}|^2 \). The CDF of \( Y_{11} \), \( F_{Y_{11}}(y) \) can be written as follows.

\[
F_{Y_{11}}(y) = 1 - e^{-\frac{y}{\beta_{11} \rho \lambda_{11}}}
\]

Similar to (7) and (8), the ergodic capacity \( \tilde{C}_{x_{11}} \) of \( x_{11} \) can be computed as follows.

\[
\tilde{C}_{x_{11}} = \frac{1}{\ln 2} \int_0^\infty \frac{1}{1+y} e^{-\frac{y}{\beta_{11} \rho \lambda_{11}}} dy = -\frac{1}{\ln 2} \left( e^{\frac{1}{\beta_{11} \rho \lambda_{11}}} Ei \left( -\frac{1}{\beta_{11} \rho \lambda_{11}} \right) \right)
\]

where \( r = \beta_{11} \rho \lambda_{11} \).
3.4. Ergodic capacity of $D_{12}$

Using (4), the achievable capacity associated to the signal $x_{12}$ is given by:

$$C_{x_{12}} = \log_2(1 + \gamma_{12})$$  \hspace{1cm} (16)

Let $Y_{12} = \frac{\beta_{12} \rho |h_{12}|^2}{\beta_{11} \rho |h_{11}|^2 + 1}$ can be written as

$$F_{Y_{12}}(y) = 1 - e^{-\left(\frac{\gamma_{12}}{\beta_{11}}\rho \lambda_{12}\right)^{1/\rho}}$$  \hspace{1cm} (17)

According to [25], it is quite difficult to obtain a closed-form solution for functions such as $F_{Y_{12}}(y)$. Hence, by applying a high SNR approximation ($\rho \rightarrow \infty$), $Y_{12}$ can be approximated as $\gamma_{12} = \frac{\beta_{12}}{\beta_{11}}$. The ergodic capacity of $D_{12}$ is

$$C_{\text{appr}}^{x_{12}} = \log_2(1 + \frac{\beta_{12}}{\beta_{11}})$$  \hspace{1cm} (18)

3.5. Conventional multiple access

For comparison with proposed scheme, an OMA scheme and a time-sharing based NOMA are devised as benchmarks. First, an OMA scheme assumes that the frequency is properly divided and transmitted. Let $B$ ($0 \leq B \leq 1$) denote the bandwidth used. In the model of the system assumed in this paper, $\frac{1}{6}B$ is allocated to each cellular user and each D2D user. Secondly, time-sharing based NOMA assumes that time is divided into three phases. In the first phase, the cellular users transmit signals to the BS simultaneously. In the second phase, UE1 transmits signals to $D_{11}$ and $D_{12}$. In the third phase, UE2 transmits signals to $D_{21}$ and $D_{22}$.

3.6. Ergodic capacities of $D_{21}$ and $D_{22}$

Similar to the calculations of $C_{\text{appr}}^{x_{11}}$ and $C_{\text{appr}}^{x_{12}}$, the ergodic capacities of $C_{x_{21}}$ and $C_{x_{22}}$ can be computed as:

$$C_{x_{21}} = -\frac{1}{\ln 2} \left( e^{\frac{1}{\beta_{21}} \rho \lambda_{21}} E(1) - \frac{1}{\beta_{21}} \rho \lambda_{21} \right)$$  \hspace{1cm} (19)

$$C_{\text{appr}}^{x_{22}} = \log_2(1 + \frac{\beta_{22}}{\beta_{21}})$$  \hspace{1cm} (20)

3.7. Ergodic sum capacity

Using (8), (12), (15), and (20), the ergodic sum capacity of the considered NOMA-D2D system model can be calculated by (21).

$$C_{\text{sum}} \sim C_{x_{1b}} + C_{x_{2b}} + C_{x_{11}} + C_{x_{12}} + C_{x_{21}} + C_{\text{appr}}^{x_{22}}$$  \hspace{1cm} (21)

where the first two terms correspond to the signals transmitted by the cellular users to the BS, and the other terms correspond to the D2D communication.

4. RESULTS AND DISCUSSION

This section presents the simulation and the analytical results of the considered NOMA-D2D system model. A close agreement between the simulation and the analytical results validates the mathematical analysis. Moreover, the ergodic sum capacity of the NOMA-D2D system model being considered is compared with that of the conventional OMA scheme and the time-sharing-based NOMA. It is shown that, under efficient interference cancellation, the NOMA-D2D system demonstrates higher capacity gains than the conventional benchmark schemes. The simulation is performed using MATLAB R2016b.

Considering the simulation setup, various system parameters are detailed as follows. The normalized distances of cellular users UE1 and UE2 from the BS are maintained as $d_{1b} = 0.3$ and $d_{2b} = 1$, respectively. Moreover, the D2D devices $D_{11}$, $D_{12}$, $D_{21}$, and $D_{22}$ are at distances $d_{11} = d_{21} = 0.05$, and $d_{12} = d_{22} = 0.1$ away from UE1 and UE2 respectively. The path loss factor is $\nu = 4$. Furthermore, the power allocation coefficients are set as $\alpha_1 = 0.64, \alpha_2 = 0.24, \beta_{11} = \beta_{21} = 0.01, and \beta_{21} = \beta_{22} = 0.05$. 

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Figure 2 provides a comparison of the ergodic sum capacities of the NOMA-D2D system and the conventional OMA and the time-sharing-based NOMA with respect to the transmitted SNRs. The analytical and simulated results of NOMA-D2D system model are presented. A close concordance between the results validates the presented mathematical analysis. As shown, the ergodic sum capacity of NOMA-D2D shows a remarkable performance gain over that of the OMA scheme under perfect SIC assumptions. For instance, when $\rho = 20$ dB, NOMA-D2D achieves capacity gains of 192.2% and 157.5% over the conventional OMA and the time-sharing-based NOMA, respectively.

Figure 3 presents the ergodic capacity of each user under the scenario of perfect interference cancellation. Both analytical and simulated results are plotted, and a close agreement between the results validates the mathematical analysis. As is evident from the results, the capacity of $D_{11}$ is the highest at the all-transmitted SNR values. This is because $D_{11}$ is located very close to UE1. It can be argued that the channel corresponding to $D_{11}$ is in very good condition. The results for scenarios with imperfect interference cancellation are not presented for individual users. However, it is well known that, if imperfections were to occur in the interference cancellation and SIC, the capacities of UE2 and $D_{11}$ would further decrease.

5. CONCLUSION

In this paper, an uplink communication system that integrates NOMA with D2D communication has been investigated. In the NOMA-D2D system, two cellular users communicated with the BS using uplink NOMA, and two cellular users simultaneously communicated with the D2D users using downlink NOMA.
Comprehensive analytical derivations were presented to compute the ergodic sum capacity of the model of the investigated system. The mathematical analysis was validated through simulations, where a close agreement between the analytical and simulated results was observed. Capacity comparisons have been presented. The proposed NOMA-D2D system outperformed the conventional OMA and the time-sharing-based NOMA in the perfect interference cancellation scenario. Future works may focus on the generalization and downlink of the NOMA-D2D. The multi-cell case also can be considered.

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