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A Signaling-Free Underwater Code Division Multiple Access Scheme

Bo-Min Seo, Junho Cho and Ho-Shin Cho *

School of Electronics Engineering, Kyungpook National University, Daegu 41566, Korea
* Correspondence: hscho@ee.knu.ac.kr; Tel.: +82-53-950-7577

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Abstract: In this paper, we propose an underwater code division multiple access system where each sensor node independently evaluates whether a channel is available or not without control message exchanges with a central data-gathering node named a sink. A sensor node is able to estimate how large power is currently received at a sink in the distance based on the overheard power at the node from neighbors. If the estimated power is below a certain threshold level, the sensor node is allowed to transmit data in a p-persistent manner, where the probability p depends on the available capacity. Simulation results show the traffic estimation works well as demonstrated by a success probability of approximately 100%, and the data throughput improves in most of the offered traffic region because of the removal of the control signaling related to channel allocation.

Keywords: underwater acoustic sensor networks; code division multiple access; multiple access control

1. Introduction

There has been growing demand for underwater acoustic sensor networks for various military and commercial applications including oceanographic data collection, environmental monitoring, resource investigation, disaster prevention, undersea exploration, assisted navigation, distributed tactical surveillance, and mine reconnaissance [1]. To successfully implement these applications, it is important to understand the characteristics of the underwater communication channel and thus design a methodology for communication between the sensor nodes. Previous studies have found that using an acoustic signal is the most suitable approach to transport information in an underwater communications channel [2].

However, acoustic signals have poor characteristics in underwater communication channels, such as a low propagating speed of 1500 m/s, which is five orders of magnitude lower than the terrestrial radio propagation speed, limited available bandwidth, multipath fading, temporary losses of connectivity (shadow zone), and Doppler spread [1,3,4]. Therefore, underwater communication suffers from a high bit error rate and long propagation delays. Enhancements or modifications for an underwater environment in terms of modulation schemes and medium access control (MAC) protocols are required to overcome the harsh underwater channel conditions.

Code division multiple access (CDMA) is a promising underwater MAC scheme and has been intensely studied because of its advantages such as robustness against frequency-selective fading and high frequency-reuse efficiency [5,6]. Thus, the CDMA studies have been mainly focused on spread code assignment, power control, and energy efficiency [7–11]. In [12,13], code assignment and power control schemes for a transmitter-based CDMA system were proposed to achieve high network throughput, low energy consumption, and limited channel access delay in deep underwater communications. In [14], a path-oriented code assignment (POCA)-CDMA-MAC protocol was proposed to resolve the funneling effect in multi-hop underwater acoustic sensor networks where the sink discriminates data arriving from multiple routing-paths simultaneously by assigning each path a distinct spreading sequence.
However, most CDMA systems employ a signaling procedure between the sink and sensor nodes to exchange control messages involved in power control, channel access, etc. Owing to the very limited bandwidth of acoustic channels and long propagation delays, the overhead of the signaling procedure causes severe deterioration in performance in terms of throughput and latency. Although there have been several works for freeing signaling procedure in MAC protocols, they are restricted to ALOHA- and CSMA (Carrier Sense Multiple Access)-family protocols [15–17]. In [18], the ALOHA with carrier sensing (ALOHA-CS) is proposed using a new back-off window to adapt the protocol to variable propagation delay. Similarly, to improve medium access opportunity, CSMA-ALOHA [19] adopts a random sensing duration shorter than the time required for the signal to propagate over the sensing range. However, to the best of our knowledge, the effort to remove the signaling overhead in CDMA has not been made for underwater environments.

In this paper, we propose an underwater CDMA without any signaling procedure related to channel access and power control. Individual sensor nodes evaluate channel availability by estimating the number of nodes currently communicating with the sink, based on individual power measurement. If the estimated number of nodes is less than the system capacity, then the node tries to occupy the channel in a p-persistent manner. The estimation is mathematically driven and evaluated by computer simulation. Numerical results show that the hit probability of this estimation is almost one, and system performance is significantly improved in terms of throughput.

2. An Overview of the Characteristics of the Underwater Channel and System Model

2.1. The Characteristics of the Underwater Channel

In this paper, we consider Thorp’s underwater channel model [1], the most widely used underwater channel model, which derives the signal attenuation depending on the frequency and the distance. It also derives the underwater ambient noise power spectral density depending on the frequency. Frequency and the distance between transmitter and receiver mainly determine the attenuation of an acoustic wave. Hence, the overall attenuation can be expressed as a function of the frequency and distance, which is given by

\[
A(l, f) = A_0 k^a(f)^l
\]

where \(l\) is a distance between the transmitter and the receiver (km), \(f\) is signal frequency (kHz), \(A_0\) is the normalizing constant, and \(k\) is the spreading factor. The spreading factor has a value between one and two depending on the depth. \(k = 1\) indicates a cylindrical spreading that characterizes shallow water communication, and \(k = 2\) indicates a spherical spreading that characterizes deep water communication. In general, \(k = 1.5\) is often considered as a practical spreading value. Therefore, we assume the spreading factor as 1.5 and the normalizing constant as 30 dB. We note that these values reflect the characteristics of a quiet, deep sea [20]. \(a(f)\) is the absorption coefficient and is expressed in dB/km for \(f\) in kHz:

\[
10 \log a(f) = 0.11 \frac{f^2}{1 + f^2} + 44 \frac{f^2}{4100 + f^2} + 2.75 \times 10^{-4} f^2 + 0.003
\]

This equation is generally valid for frequencies above a few hundred Hz. For lower frequencies, the following formula may be used:

\[
10 \log a(f) = 0.002 + 0.11 \frac{f^2}{1 + f^2} + 0.011 f^2
\]

2.2. System Model

We considered a data gathering CDMA network with a star topology, where the sink is located at the center and sensor nodes in charge of data-sensing and reporting are uniformly distributed over
an ideal circle-shaped coverage. It is assumed that each sensor node acquires the distance to the sink and the unique orthogonal code for CDMA communication through the network initialization stage, in which measurement of round-trip time and orthogonal code allocation take place. Based on Thorp’s underwater channel model, sensor nodes are able to control transmission power with the knowledge of the distance and the frequency, in such a way that regardless of their locations, the received power at the sink is equal to a specific level.

3. Estimating the Number of Nodes in Transmission

As shown in Figure 1, suppose that node X, located at a distance \( x \) from the sink, is examining the channel availability for data transmission while node Y is transmitting. For mathematical description simplicity, using polar coordinates, we locate the sink at the origin \((0, 0)\), node X at \((x, -\frac{\pi}{2})\), and node Y at \((r, \phi)\). Let \( \theta \) and \( l \) denote the angle and distance between node X and Y, respectively. Let \( P_t^Y \) denote the transmission power at node Y and \( P_{rcv}^S \) denote the common received power at sink then,

\[
P_t^Y \cdot \int_B A^{-1}(r, f)df = P_{rcv}^S = S(f) \cdot B
\]

(4)

where \( S(f) \) represent the common received power spectral density at the sink. We assume that \( S(f) \) is flat for \( f \in B \) where \( B \) is a signal bandwidth. Therefore, the transmission power at node Y is given by

\[
P_t^Y = \frac{S(f) \cdot B}{\int_B A^{-1}(r, f)df}
\]

(5)

In a similar manner, the received power at node X when node Y is transmitting is

\[
P_{rcv}^{X \leftarrow Y} = P_t^Y \cdot \int_B A^{-1}(l, f)df
\]

(6)

Because \( P_{rcv}^{X \leftarrow Y} \) is obtained from the specific location of node Y with \( r \) and \( \theta \), over the whole region with radius \( R \), we can obtain the expected value of received power at node X from a node Y:

\[
E[P_{rcv}^{X \leftarrow Y}] = \int_0^R \int_0^{2\pi} P_{rcv}^{X \leftarrow Y} \cdot f_{r,\Theta}(r, \Theta)d\Theta dr
\]

\[
= \int_0^R \int_0^{2\pi} P_{rcv}^{X \leftarrow Y} \cdot f_r(r) \cdot f_{\Theta}(\Theta) d\Theta dr
\]

\[
= \int_0^R \int_0^{2\pi} P_{rcv}^{X \leftarrow Y} \cdot \frac{2r}{R^2} \cdot \frac{1}{2\pi} d\Theta dr
\]

(7)

where \( f_{r,\Theta}(r, \Theta) \), \( f_r(r) \), and \( f_{\Theta}(\Theta) \) represent the joint probability density functions of \( r \) and \( \Theta \), the probability density function of \( r \), and the probability density function of \( \Theta \), respectively. We assume that \( f_{r,\Theta}(r, \Theta) \) is uniformly distributed over the cell with radius R. Because \( E[P_{rcv}^{X \leftarrow Y}] \) corresponds to the value for one node Y, the value is multiplied when multiple nodes are transmitting and accordingly, the number of transmitting nodes is estimated in a way that the measured receiving power at node X becomes closest to the certain multiple times of \( E[P_{rcv}^{X \leftarrow Y}] \) like

\[
N_t = \arg \min_n \left( \left| P_{rcv}^X - nE[P_{rcv}^{X \leftarrow Y}] \right| \right)
\]

(8)

where \( P_{rcv}^X \) denotes the measured receiving power at node X.
4. An Example of Traffic Estimation

Figure 2 illustrates a topological scenario as an example for which the proposed scheme is applied. The reference node $X$ indicates who is trying to estimate the number of transmitting nodes and is located at $(2, 1.5\pi)$ in polar coordinates, and the other five nodes are randomly located. Let us say that $S(f) \cdot B$ is 81 dB re 1 $\mu$Pa (i.e., the power associated with the reference sound pressure level of 1 $\mu$Pa). We note that this value is chosen based on the range of the power spectral density of ambient underwater noise, which is from 60 to 100 dB re 1 $\mu$Pa/Hz [4]. Then, the expectation $E[p_{rcv}^{x\rightarrow y}]$ is calculated to be 67 dB re 1 $\mu$Pa using Equation (7). Meanwhile, the measured receiving power at node $X$, $P_{rcv}^{X}$, which depends on the node distribution, is ≈74 dB re 1 $\mu$Pa. Then, the reference node $X$ finds the number of nodes, $n$, such that the difference between $n \cdot E[p_{rcv}^{x\rightarrow y}]$ and $P_{rcv}^{X}$ is minimized, as described in Equation (8). The procedure to find $N_e$ is summarized in Table 1, where $N_e$ is found to be 5, which is equal to the number of transmitting nodes in the example of Figure 2, and thus the estimation is correct.

Figure 1. Nodes’ locations.

Figure 2. A topological scenario as an example for traffic estimation.
Table 1. Parametric values generated during traffic estimation for the example of Figure 2.

| n | 1   | 2   | 3   | 4   | 5   | 6   |
|---|-----|-----|-----|-----|-----|-----|
| $E[p_{\text{rcv}}^X \leftarrow Y]$ (dB re 1 $\mu$pa) |     |     |     |     |     | 67  |
| $n \cdot E[p_{\text{rcv}}^X \leftarrow Y]$ (dB re 1 $\mu$pa) | 67  | 70  | 71  | 73  | 74  | 75  |
| $p_{\text{rcv}}^X \leftarrow Y - n \cdot E[p_{\text{rcv}}^X \leftarrow Y]$ (dB re 1 $\mu$pa) | 73  | 72  | 70  | 68  | 62  | (minimum) 65 |
| $N_e$ |     |     |     |     |     | 5   |

5. Probabilistic Channel Access

We assumed that the system has a predefined CDMA system capacity—the number of nodes that are allowed to transmit simultaneously—which we denote as $C$. The nodes examining the channel availability estimate the number of currently transmitting nodes $N_e$ (obtained by Equation (8)). If $N_e$ is less than $C$, then the examining nodes are allowed to join the transmission. However, if they try immediate joining, it may cause overflow—total number of transmitting nodes including the newly joining nodes exceeds the capacity $C$. Therefore, a probabilistic channel access in a $p$-persistent manner is employed to avoid such overflow. The channel access probability is proportional to the availability of the channel and is given by

$$p_{\text{tr}} = \begin{cases} \frac{C-N_e}{C}, & C > N_e \\ 0, & \text{otherwise} \end{cases}$$

where $p_{\text{tr}} = 0$ means the transmission is delayed by a random back-off and restarts from the beginning, i.e., channel estimation.

6. Performance Evaluation

As comparing schemes, we considered whether or not a signaling procedure is present, and whether or not it is CDMA-based, and we chose from three schemes: $p$-persistent CSMA, MACA-U [22], and CDMA, using a four-way handshake of RIPT protocol [21]. Table 2 categorizes the schemes. In the CDMA of RIPT protocol, the sink broadcasts ready-to-receive (RTR) packets to invite sensor nodes that are willing to transmit data packets. Having gathered the responses (RTP: Ready-to-participate) from the sensor nodes, the sink lists the sensor nodes that are served within CDMA capacity and replies to them with an ORDER packet. Reading the ORDER packets, the sensor nodes that are included in the list transmit their data packets. The operation of CDMA of RIPT protocol is illustrated in Figure 3.

Table 2. Categorization of proposed and comparing scheme.

|                  | Signaling-Free | Signaling-Based |
|------------------|----------------|-----------------|
| CDMA             | Proposed scheme | CDMA of RIPT [21] |
| Non-CDMA         | $p$-persistent CSMA | MACA-U [22] |
A data packet is generated at each node to follow the Poisson process at the rate of $\lambda$ (packet/sec). To cope with traffic congestion, all schemes are assumed to employ the binary exponential random back-off specified in the IEEE 802.11 standard [23]. Packet losses are only caused by collisions, not by physical channel deterioration such as noise and multipath fading. Table 3 summarizes the system parameters used in the computer simulation. The data rate in Table 3 corresponds to the multiplexed one, i.e., in the case of CDMA with a capacity of four, each channel has a data rate of 2400 bps. We averaged the results of 1000 times iteration for each parameter to derive the result.

During packet generation, the sensor node measures receiving power based on the estimated number of nodes currently transmitting packets to the sink. Because the transmission power of a sensor node is controlled not to extend far beyond the sink, in the case where the estimating node is located at the opposite side from the transmitting nodes, the received power may not be adequate for correct estimation. However, even with such a difficult condition in the simulation, the successful traffic estimation ratio is maintained above 95% for almost all cases of offered load as shown in Figure 4.
Figure 4. Success rate of traffic estimation.

Figure 5 shows the throughput varying the offered load per node. Throughput is defined as the number of data bits successfully received at the sink per second from all sensor nodes. The throughput of the proposed system is shown to be much higher than the comparing schemes for almost cases of the offered load. It proves that the channel access without signaling procedure works well, significantly reducing both the signaling overhead and the associated delay. An issue that we need to resolve is that as the offered load gets heavier, the throughput of the proposed scheme degrades rapidly, and even falls below CDMA of RIPT and MACA-U at over 0.25 offered load. This might be an extremely heavy traffic condition related to various monitoring-related applications [25]. The reason for this is that while the signaling-based schemes like CDMA of RIPT and MACA-U can prevent throughput from decreasing below a certain level using a channel reservation, as traffic load increases the proposed system, this shows excessive caution for transmission and keeps delaying it because the estimated traffic mostly exceeds capacity. Therefore, the probability of transmission $p_{tr}$ needs to be adjusted for the special case of heavy traffic. This is not dealt with in this paper but remains as a potential future work.

Figure 5. Throughput for proposed scheme with other MAC protocols.

Regarding the fairness for channel occupation, in signaling-assisted protocols such as MACA-U and CDMA of RIPT, the nodes closer to the sink are more probably occupying the channel because the channel reservation control message such as RTS may arrive earlier at the sink than from the nodes far away from the sink. In contrast, the proposed scheme equally gives the opportunity for channel
occupation independently of the nodes’ location because any node finding an available channel is allowed to occupy the channel.

Figure 6 shows the throughput of sensor nodes in relation to the location from the sink. As the node gets farther from the sink, the traffic estimation may be incorrect because the received power from the other nodes is probably not strong enough. Especially when other nodes are mostly located at the opposite side beyond the sink, the node is probably blind to their existence and accordingly, underestimates the number of nodes. Such a wrong estimation causes collisions at the sink and thus the throughput decreases. Meanwhile, the node closer to the sink is able to catch all the power transmitted from any location inside the cell and thus is able to estimate the traffic more precisely.

![Figure 6. Throughput of the node at different distances from the sink.](image)

7. Conclusions

In this paper, we proposed a CDMA scheme that removes signaling overhead associated with channel access and thereby increases throughput along with reducing delay. Sensor nodes are able to correctly estimate the number of nodes in transmission only by means of received power measurement. Computer simulation shows that the ratio of successful traffic estimation is above 95% for varied offered load conditions. With reference to throughput, the proposed scheme outperforms the schemes against which it was compared for almost all cases of offered load. We note that the proposed scheme can be implemented and evaluated in a real underwater environment. Underwater nodes consist of a commercial underwater modem and a DSP board where we can program the proposed MAC protocol algorithm. Then, each node is deployed in the underwater environment to build an underwater star topology network. After deployment, each node executes the MAC protocol operation and maintains a log of the events. Using the MAC protocol event logs, the performance of the proposed scheme can be evaluated. In a future study, the persistent transmission probability $p_T$ will be researched to make the proposed scheme more sustainable regardless of traffic conditions.

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