Tilted Microstrip Phased Arrays With Improved Electromagnetic Decoupling for Ultrahigh-Field Magnetic Resonance Imaging

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Abstract: One of the technical challenges in designing a dedicated transceiver radio frequency (RF) array for MR imaging in humans at ultrahigh magnetic fields is how to effectively decouple the resonant elements of the array. In this work, we propose a new approach using tilted microstrip array elements for improving the decoupling performance and potentially parallel imaging capability.

To investigate and validate the proposed design technique, an 8-channel volume array with tilted straight-type microstrip elements was designed, capable for human imaging at the ultrahigh field of 7 Tesla. In this volume transceiver array, its electromagnetic decoupling behavior among resonant elements, RF field penetration to biological samples, and parallel imaging performance were studied through bench tests and in vivo MR imaging experiments.

In this specific tilted element array design, decoupling among array elements changes with the tilted angle of the elements and the best decoupling can be achieved at certain tilted angle. In vivo human knee MR images were acquired using the tilted volume array at 7 Tesla for method validation.

Results of this study demonstrated that the electromagnetic decoupling between array elements and the B1 field strength can be improved by using the tilted element method in microstrip RF coil array designs at the ultrahigh field of 7T.

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Abbreviations: FDTD = finite-difference time-domain, FOV = field of view, g-factor = geometry factor, GRAPPA = generalized autocalibrating parallel partial acquisition, ID = inner diameter, MRI = magnetic resonance imaging, MTL = microstrip transmission line, NEX = number of excitation, OD = outer diameter, RF = radio frequency, SNR = signal-to-noise ratio, TE = echo time, TR = repetition time.

INTRODUCTION

Magnetic resonance imaging (MRI) has become one of the most important non-invasive imaging modalities used in clinical medicine.1–3 The fact that signal to noise ratio (SNR) is proportional to the static magnetic field strength4 has resulted in the rapid development of high and ultrahigh field MRI such as 3T, 7T, and beyond.5–12 With the increase of the Larmor frequency at high fields, radiation losses and interaction between RF coil and imaging subject increase, leading to degraded quality factors, and MR signal intensity. These problems become more pronounced at ultrahigh magnetic fields such as 7T where the proton Larmor frequency reaches the high frequency of 300-MHz range. Therefore, there is a demand for designing sensitive high-frequency radio frequency (RF) coils for efficient MR signal excitation and detection at high and ultrahigh magnetic fields.13,14

The RF coil array was firstly introduced to MR signal acquisition to improve SNR5–5 for large field-of-view (FOV) imaging. In recent years, this technique has been utilized in parallel acquisition16–22 and parallel transmission23–25 to accelerate imaging speed. Transmit/receive (or transceiver) coil arrays are also essential for RF excitation field (referred to as B1þ) shimming by using independent phase and amplitude control of each array element26–30 to address field inhomogeneity problems in high dielectric and conductive biological samples, such as the human body at ultrahigh fields.29,31–35 These techniques require dedicated transceiver arrays with sufficient electromagnetic decoupling among the array elements. However, due to the significantly increased coupling among resonant elements of an array at ultrahigh magnetic fields,34,36–39 it is technically challenging to implement such a dedicated transceiver array. Conventional decoupling methods such as the use of low impedance preamplifier,15 which are usually used in receive-only arrays, are not readily feasible for transceiver array designs. Dedicated decoupling circuits, either capacitive or inductive, have been applied to decouple an n-element array using a 2n-port decoupling interface.40 This method may become complicated and increase electric losses, especially for a massive number of coil elements in an array.

To reduce radiation losses and increase the RF coil efficiency, the microstrip transmission line (MTL)41,42 approach has been introduced for RF coil design at ultrahigh fields.10–14,34,43 Microstrip coils can readily be operated at ultrahigh Larmor frequencies such as 298 MHz (7 Tesla)10,12 and 400 MHz (9.4 Tesla)13,14,43 for commonly desired coil size. Its simple structure and high frequency operation capability make coil design and construction convenient, especially for...
ultrahigh field volume coils\textsuperscript{12,44} and coil arrays\textsuperscript{34–36,39,45,46} with complicated mechanical structures. Due to the inherent decoupling among microstrip resonators, microstrip transceiver arrays have demonstrated excellent decoupling performance and have been realized at ultrahigh field MRI.\textsuperscript{34,35,45,47–49} However, for RF coil arrays with a massive number of elements, due to the reduced distance between coil elements where the microstrip decoupling conditions are often unable to be met, achieving better than $-18 \text{ dB}$ decoupling between resonant elements is technically challenging in practice.

In this work, an improved decoupling method for microstrip transceiver array designs based on tilted element technique\textsuperscript{50–52} was proposed and investigated. A specific microstrip transceiver array with 1.3 cm thick Teflon substrate, 1.3 cm wide microstrip resonator and 16 cm length, operating at 300 MHz, was designed and constructed for this investigation at the ultrahigh field of 7T. Bench tests on a network analyzer were made to measure the B$_1$ strength variation with the tilted angle, and compare the mutual coupling between a tilted array and non-tilted array. Numerical simulation by using Finite-Difference Time-Domain (FDTD) method was used to compare the B$_1$ field distribution between 2 microstrip resonators with $0^\circ$ and $30^\circ$ (optimized for decoupling) tilted angles. In vivo MR images were acquired using the transceiver array from healthy human volunteers. To investigate parallel imaging performance of the proposed tilted arrays, GRAPPA accelerated parallel imaging\textsuperscript{18} was used for image reconstruction at various acceleration rates. The geometry factor (g-factor)\textsuperscript{17} for 1D accelerated imaging were also calculated and evaluated.

\section*{METHODS}

\subsection*{Design of Coil Array With Titled Straight-Type Microstrip Elements}

An 8-channel transceiver volume array with tilted straight-type microstrip elements was built. Each element was a capacitive terminated half-wavelength ($\lambda/2$) microstrip resonator with 1.3 cm width and 16 cm length as shown in Figure 1B. Termination capacitors were DLC2R7 (Dalian Dalicap Co., Ltd., Dalian, China) with a nominal capacitance of 2.7 pF and a variable capacitor (NMAP25HV, Voltronics Co., Salisbury, MD) ranging from 1 to 25 pF. The strip conductor was built on the surface of a Teflon substrate with 1.3 cm thickness and permittivity of 2.1. On the bottom of the substrate, there was a piece of adhesive copper tape acting as the ground plane. The resonant frequency of such a capacitive terminated $\lambda/2$ resonator can be estimated by solving the following equation\textsuperscript{12}: 

\begin{equation}
    f_{re} = \frac{(2\pi f_{re} Z_0)^2 C_1 C_2 - 1}{2\pi Z_0(C_1 + C_2)} \tan \left(\frac{2\pi \sqrt{\varepsilon_{eff}}}{c} f_{re}\right)
\end{equation}

where $C_1$ and $C_2$ are the capacitance of the termination capacitors of the microstrip resonator shown in Figure 1B. $Z_0$ is the characteristic impedance of the microstrip resonator. $l$ and $c$ denote the length of the microstrip conductor and the velocity of light in free space, respectively. $\varepsilon_{eff}$ is the effective permittivity of the microstrip resonator.

A total of 8 elements were built on a supporting frame with 21 cm ID and 28 cm OD and the prototype of this volume array is shown in Figure 2. Thus each element can be tilted to any angle to become a tilted microstrip array. The cross section views of this tilted array and a regular array are shown in Figure 3. The tilted angle $\theta$ was defined tangentially to a circle crossing all coil elements as shown in Figure 3B. This structure also facilitated comparison between regular array and tilted array with different tilted angle. The impedance matching of each element was achieved by varying the trimmer capacitor (NMAP25HV, Voltronics Co.), which was connected in series to feed port of the microstrip resonator.

\begin{figure}[h]
\centering
\includegraphics[width=0.5\textwidth]{figure1.png}
\caption{Diagrams of (A) typical open-circuited MTL resonator, (B) capacitor terminated MTL resonator.}
\end{figure}

\begin{figure}[h]
\centering
\includegraphics[width=0.5\textwidth]{figure2.png}
\caption{Prototype of the 8-channel transceiver volume array with tilted straight-type microstrip elements: views from left side (A) and right side (B).}
\end{figure}
Bench Measurement and MR Test

MR imaging study with healthy human subjects on 7T MRI was obtained from the Committee on Human Research (CHR) at the University of California San Francisco (i.e., UCSF’s IRB). Subjects were provided informed written consent prior to participating in the study.

Bench test of the microstrip volume array with different tilted angle $\theta$ was made on an Agilent E5070B network analyzer. The transmission parameter $S_{21}$ was used to measure the decoupling performance between the adjacent elements of the array. The decoupling performance, which varies with the tilted angle $\theta$, then can be plotted as a function of $\theta$. The mutual coupling between each pair of the elements in the coil array was also measured using the network analyzer, and the mutual coupling matrix was generated to show the decoupling performance of the array. MR imaging experiments with the tilted microstrip array was performed on a General Electric (GE) 7 Tesla whole body MR scanner. This scanner is equipped with 2 quadrature transmit channels and 2 T/R switchers. To test the transceiver arrays on this scanner, scan series were conducted by connecting 2 coil elements into the transmit channels each time, and combining all sub-images offline. Axial plane images of the human knee from a healthy volunteer were acquired with this array by using gradient echo sequences. The acquisition parameters used were $TE = 7 \text{ ms}$, $TR = 100 \text{ ms}$, matrix size $= 256 \times 256$, receiver bandwidth $= 15.6 \text{ kHz}$, field of view (FOV) $= 14 \text{ cm}$, slice thickness $= 3 \text{ mm}$, number of excitation (NEX) $= 1$.

Evaluation of Parallel Imaging Performance

To demonstrate the performance of the tilted array in parallel accelerated imaging, GRAPPA accelerated imaging was used for human knee image reconstruction. The acceleration was applied to phase encoding direction ($y$-axis) and 32 Auto-Calibration Signal (ACS) lines were acquired in the central region of the $k$-space to estimate the missing lines. GRAPPA reconstructions with acceleration factor from 2 to 8 were performed and were compared with the reference image reconstructed from the sum of squares method from full $k$-space data.

In addition, the noise correlation matrix was calculated to demonstrate the noise correlation among array elements. The spatial varying geometry factors (g-factors) were calculated using:

$$g_r = \frac{\sqrt{\left\langle S^H \psi^{-1} S \right\rangle}}{\sqrt{\left\langle S^H \psi^{-1} S \right\rangle}}$$  \hspace{1cm} (2)

where $\rho$ denote the index of voxel in the field of view, and $S$ is the reduced Fourier encoding. $\psi$ is the noise correlation matrix between channels. G-factor strongly depends on the voxel position and has a direct impact on SNR:

$$\text{SNR}_{\text{red}} = \frac{\text{SNR}_{\text{full}}}{g_r^\rho R}$$  \hspace{1cm} (3)

where $R$ is the reduction factor (or acceleration factor). This relationship means the SNR of the accelerated image is inversely proportional to the g-factor. The larger the g-factor, the worse the parallel imaging performance is. The average and maximum g-factors were plotted in term of acceleration factor to demonstrate the performance of the tilted array for accelerated parallel imaging.

RESULTS

Bench Measurements of Tilted Array With Straight-Type Microstrip Elements

Bench measurements of the tilted volume array with straight-type microstrip elements at different tilted angle $\theta$ from $0^\circ$ to $40^\circ$ are plotted in Figure 4. When $\theta = 0^\circ$ this array is actually a regular microstrip array. It is shown that with the variation of the tilted angle, the decoupling between array elements changes obviously. For this specific microstrip array, the tilted angle of $\sim 30^\circ$ can be regarded as an inflexion point and at this tilted angle the decoupling between elements reaches
optimal value of $-22 \, \text{dB}$. Compared with regular microstrip volume array with the same structure, the decoupling can be greatly improved without using any extra decoupling techniques.

To compare the isolation between each pair of elements in the tilted and non-tilted arrays, the mutual coupling matrices of the coil arrays with $0^\circ$ to $30^\circ$ tilted angles were measured and are shown in Figure 5. The tilted array achieved better than $-20 \, \text{dB}$ mutual coupling while the conventional non-tilted array achieved better than $-12 \, \text{dB}$ mutual coupling, demonstrating an improvement of at least $8 \, \text{dB}$ in decoupling performance by using the tilted element strategy.

From the bench test, it is also observed that the $B_1$ field was strengthened at the central region of the coil array when increasing the tilted angle in the range of $0^\circ$ to $90^\circ$. As shown in Figure 6, the $B_1$ field strength was increased by $28\%$ when the tilted angle was changed from $0^\circ$ to $30^\circ$. Simulation comparison between 2 elements with $0^\circ$ to $30^\circ$ tilted angles in Figure 7 illustrates the change of the $B_1$ field distribution in a human brain imaging area. With element tilting, the imaging coverage of coil elements is enlarged compared with that of non-tilted elements.

**FIGURE 4.** Decoupling between array elements of the volume transceiver array with tilted straight-type microstrip resonators measured in unloaded case. The black curve describes the transmission parameter $S_{21}$ measurement varies with the tilted angle: between $0^\circ$ and $30^\circ$ the decoupling increases with the tilted angle, indicating that the self-cancellation from the ground plane can be compensated by tilting the elements.

**FIGURE 5.** Mutual coupling matrices among all the coil elements of the microstrip array with $0^\circ$ angle (left) and optimized $30^\circ$ angle (right). The array with $0^\circ$ angle can achieve better than $-12 \, \text{dB}$ mutual coupling, while the array with $30^\circ$ tilted angle can achieve better than $-20 \, \text{dB}$ mutual coupling.

**FIGURE 6.** $B_1$ field strength of the microstrip array increases with tilted angle from $0^\circ$ to $40^\circ$. The field strength is improved by approximately $28\%$ at $30^\circ$ tilted angle which is the optimized angle for achieving best decoupling for this specific MTL coil array design: Teflon substrate with $1.3 \, \text{cm}$ thickness, microstrip resonator with $1.3 \, \text{cm}$ width and $16 \, \text{cm}$ length.

**Parallel Imaging Performance**

Figure 8 shows the reconstructed images from in vivo human knee MR imaging. The first image is the reference image, which was reconstructed from full $k$-space data using a sum of squares method. The images from the second to the eighth were reconstructed using GRAPPA accelerated imaging method at acceleration factors of 2 to 8. It is clearly shown that with the increase of acceleration factor, the imaging speed was enhanced while SNR decreased. When the acceleration factor was smaller than 4, there was no obvious distortion to the image and the image quality was acceptable. For higher acceleration factor up to 8, obvious distortion can be observed in the reconstructed image. Tradeoffs must be made between the imaging speed and the SNR in parallel imaging. In Figure 9, the average and maximum g-factor is plotted as a function of acceleration factor, which demonstrates the parallel imaging performance at different acceleration factors.

**CONCLUSIONS**

In this study, a transceiver array design technique using tilted elements was investigated for microstrip coils numerically and experimentally. By using the tilted element strategy, mutual inductance between the coil elements could be diminished and thus element-decoupling performance of microstrip coil array could be improved. Bench test measurement was performed to compare the microstrip array with tilted elements and non-tilted elements in terms of RF field strength and the mutual coupling between each pair of elements, demonstrating an improvement.
in the RF field penetration and the decoupling performance of the proposed tilted element array. Numerical simulations using FDTD method were also conducted to compare the $B_1^+$ field distribution in the 2 different cases (ie, tilted and non-tilted), showing that the tilted array could obtain larger imaging coverage for this specific design. It might be interesting to investigate the tilted element method for microstrip arrays with a large number of array elements, where the decoupling conditions are often unable to be satisfied due to the reduced distance between array elements, resulting in insufficient decoupling for imaging acquisitions with reasonable quality.

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