Characterizing the communication in the Amazon rainforest:
towards a realistic simulation

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Abstract Many literature papers evaluate solutions in wireless sensor networks by simulation experiments. However, little attention is given to the adequacy of the simulator propagation models to the environment in which such solutions are employed. This can lead to imprecise or inconsistent results in relation to the real world. This paper presents a methodology for adjusting the parameters of these models. In particular, we present experimental results for rainforest environments, which can be the goal of many sensor networks monitoring applications. The impact of the proposed approach is shown by evaluating a localization solution. The results show that this procedure is necessary for a higher fidelity of simulation experiments.

Keywords Propagation · Experimentation · Simulation

1 Introduction

Many research efforts are being dedicated to the wireless sensor networks (WSNs) [3] in the last few years. This is due to their capacity to increase the way we interact with the world. In particular, many environmental monitoring applications are under study [7], and they are being employed in environments with very distinct characteristics which affect the wireless communication. Particularly, many researchers in Brazil and all over the world are concerned about Amazon rainforest monitoring for studies and preservation due to its dimension and biodiversity.

In the initial phase of the WSN’s development many solutions were proposed under conceptual aspects with strong use of experimentation by simulation. These are solutions such as protocols and algorithms for MAC [30], routing [4], localization [9], and many others which depends on the wireless communication and the channel characteristics. Obviously, not all the scenario characteristics and premises considered in these works are observed in practice, as shown by recent work [13], and disparities between simulations and real platforms become clear [11,25].

As a result, many papers have focused on the experimentation of wireless radios within different scenarios [5,6,21,34,41,43]. The main goal is to evaluate how the signal strength and delivery ratio are affected by distance and obstacles. However, simulation is still important for the development of new solutions, because not always there is an infrastructure favorable for practical experimentation, besides the cost and time necessary for this. Thus, we show a methodology to characterize the communication in the Amazon rainforest, which is a very distinct environment from those studied in literature, based on practical experiments.

The main contributions are described as follows. At first we did practical experiments to measure the received power, the link quality and delivery ratio by using two popular platforms for WSN in the Amazon rainforest. This is an important contribution itself, once we do not have similar results in the literature. After that, by using the mean square error (MSE)
minimization we estimated the parameters of the Shadowing propagation model, which is very common in the existing simulation suites. Then, we used a localization system simulation in NS-2\(^1\) based on a multilateration algorithm to compare the obtained results against default parameters normally used in simulations, and other parameters suggested in the literature. We show that fine tuning these parameters is necessary to represent what we can obtain in real experiments.

The remainder of the paper is organized as follows. Some theoretical fundamentals are introduced in Sect. 2. Section 3 presents the related work. In Sect. 4, we characterize the rainforest scenario. Section 5 discusses realistic simulation. Finally, our conclusions and future work are presented in Sect. 6.

2 Background

2.1 Propagation models

The propagation models are traditionally used to predict the average signal strength (power) received at a certain distance of the transmitter, beyond intensity variability of signal in areas near a particular position. The simplest but representative models are the Free Space and Shadowing. Both consider that path loss, the amount of power lost for the medium, is proportional to the distance between transmitter and receiver to the power of N, i.e.,

\[ PL(d) \propto d^N, \]  

where \( d \) is the distance and \( N \) is path loss exponent that indicates the power decay rate with the distance. For the Free Space model, \( N \) is 2, and for the Shadowing model, \( N \) ranges from 1.5 to 6. The higher \( N \) is, the more severe is the environment for communication.

The Free Space model is used only when there is a vision path clearing between the transmitter and receiver. The Shadowing model is used when there are many obstacles for propagation that is difficult to model into equations.

Other models take into account the effects of reflection, diffraction, and scattering [33]. The Two-Ray Ground model [19] deals with ground reflection, useful for predicting signal strength over distances of several kilometers when transmitter and receiver use towers of different heights. Diffraction effect is treated by the Knife-edge Diffraction model [26,33] in cases where there is a hill or mountain between transmitter and receiver. The Radar Cross Section model [35] aims to model the impact of the diffused reflections produced when a radio wave impinges on a non-uniform surface like a building.

Some models are described for particular environments. The outdoor models, like Okumura–Hata [22] and Walfisch and Bertoni [39], are derived for urban areas usually associated with cellular systems. Foliage or vegetation models estimate the path loss due to propagation though trees and foliage [36]. In this category we have Weissberger’s modified exponential decay model [40], ITU Vegetation model [24], COST235 model [2]. Generically, these models estimate the additional loss \( L_\text{veg} \) by the expression:

\[ L_\text{veg} = \alpha \times f^\beta \times d^\gamma, \]  

where \( f \) is the frequency, \( d \) is the depth of the foliage along the line-of-sight path. Constants \( \alpha, \beta, \) and \( \gamma \) are empirically calculated in each model. In spite of these models considering tree and foliage, they are not suitable for modeling propagation when small and low-power radios are employed inside forests for short distances.

In this work, we use the Shadowing model to estimate path loss and received power in the Amazon rainforest. Next, we describe in detail this model and justify this choice. In addition, we also provide the theory of the Free Space model, once it may be used as base for many models including the Shadowing model.

2.1.1 Free Space propagation model

The Free Space model is used to predict received signal strength when the transmitter and receiver have a clear, unobstructed line-of-sight path between them [33]. The received power by a receiver antenna in dBm is given by [18]

\[ P_r(d) = P_0 - 20 \log(f) - 20 \log(d) + 27.56, \]  

where \( P_0 \) is the power at zero distance from the antenna, \( f \) is the signal frequency in MHz and \( d \) is the distance (in meters) from the antenna.

2.1.2 Shadowing propagation model

Propagation models like Free Space and Two-Ray Ground predict the signal power received as deterministic functions of distance. It represents the communication area as a perfect sphere, however the behavior of a received signal propagation is random and distributed log-normally around the mean. The randomness of the signal behavior is caused by changes in the environment that directly impact the signal power [33].

The Shadowing model is separated in two parts. The first part provides the average of received powers at a certain distance. The second part, models the randomness of the signal at this distance \( d \).

The received or input power is calculated as

\[ P_t(d) = P_t - PL(d). \]
where \( d \) is the distance between the receiver and the transmitter in m and \( P_t \) is the transmission power in dBm. The path loss, \( PL \) in dBm, at a distance \( d \) is calculated as [33]

\[
PL(d) \text{ (dB)} = PL(d_0) + 10N \log\left(\frac{d}{d_0}\right) + X_\sigma, \tag{5}
\]

where \( d_0 \) is the nearby reference distance, \( PL(d_0) \) is the average path loss for \( d_0 \) or an assumption using the Free Space model. \( N \) is the exponent of the path loss that indicates the power decay rate with the distance. \( X_\sigma \) is a zero-mean Gaussian distributed random variable with standard deviation \( \sigma \) (in decibels).

Combining Eqs. 4 and 5, we have

\[
P_t(d) = P_0 - 10N \log\left(\frac{d}{d_0}\right) + X_\sigma, \tag{6}
\]

where

\[
P_0 = P_t - PL(d_0). \tag{7}
\]

The adoption of the Shadowing model in this work is motivated by many reasons. First, it considers attenuation of the signal due to distance, and interferences caused by obstacles which are observed in the measurements. Second, the foliage models are not suitable for communication with both transmitter and receiver inside the forest. Variations of these models are adapted for plantation environments, but they consider trees as measurable or countable objects. In the Amazon rainforest, the trees are very close to each other and there are many other obstacles, such as: roots, branches, lianas, shrubs, and to measure their depth is a difficult task. Third, the Shadowing model is simple, easy, generic, and popular. Many environments were characterized using this model [5, 18, 33, 34] allowing comparison and reuse. Last, the Shadowing model is implemented by simulators like NS-2 and Castalia.²

\footnote{2 http://castalia.npc.nicta.com.au.}

3 Related work

Regarding wireless communication characterization, Zhao and Govindan [43] carried out experiments investigating the packet delivery in three different environments: an indoor office building, a habitat with moderate foliage, and an open parking lot. They showed that by varying the distance between nodes, we can identify regions in which the communication becomes unstable, the so-called “gray areas” or “transitional region”. Although they are trying to characterize wireless communication, they focused on packet delivery.

Zuniga and Krishnamachari [44] provided an extensive study about the “transitional region”. They identified the causes and derived expressions for its width and packet delivery rate. For increasing realism in simulation experiments, the authors recommend the characterization of the interested environment.

The relationship between packet delivery and received signal strength indicator (RSSI) and link quality indicator (LQI) was investigated by Srinivasan and Levis [37]. Their experimental study showed that LQI actually is a better quality indicator than RSSI for a wider signal degradation range. However, the focus of this work was the RSSI itself, not characterization or realistic simulation.

Seidel and Rappaport [34] and Andersen et al. [5] have reported many contributions about propagation models, specifically about the Shadowing model and derivatives. Both have conducted a large number of experiments to model and characterize the wireless communications channels in indoor scenarios.

Some studies on propagation models for forest environments have been presented in the literature since the 1960s until today [6, 21, 27, 28, 38]. Some of these studies focus on medium and high frequencies (2–200 MHz) [38], others on UWB (above 3.1 GHz) [6, 28] and others only on theoretical analysis [27, 28]. Except for Tamir’s work [38], the idea is to evaluate the impact of the forest on long-distance telecommunication systems (over 1 km) when transmitter or receiver is outside of the forest. In this work, we evaluate the communication for ZigBee radios inside the forest, operating in 2.4 GHz, for short distances.

In the context of WSN, Ndzi et al. [31] and Mestre et al. [29] carried out experiments in plantations to evaluate vegetation attenuation models (discussed in Sect. 2.1) for precision agriculture.

Gay-Fernandez et al. [21] present a complete study for the deployment of a WSN in a forest in Spain based on ZigBee, including the propagation model analysis. They carried out some propagation experiments and found parameters for different deployment situations. These parameters can be easily adapted to the Shadowing parameters. However, the environment studied, even being a forest, presents features different from the Amazon rainforest which, basically, is denser than the Spanish forest. Thus, their parameters could not be used in applications in our context. One of our main objectives in this work is to characterize the communication in the Amazon rainforest.

Zanca et al. [42] used IFX-Eye node to evaluate the performance of localization algorithms in WSN for an indoor environment. They featured the signal radio propagation by the Shadowing model and carried out some practical localization experiments. They revealed that some localization algorithms achieve worse performance in real testbeds than what predicted by the simulation analysis. Pham et al. [32] also provided results for real deployments of TelosB motes for validating and tuning the MAC protocol implemented in the Castalia simulator. In this paper, we investigate the impact
of the incorrect tuning of propagation parameters in the NS-2 simulator for actual applications.

4 Rainforest scenario characterization

In this section we present the used methodology to characterize the Amazon rainforest scenario. First we carried out practical experiments to assess the behavior of current wireless communication for actual applications. In the second part of characterization, we used the measured received power to estimate the parameters of the Shadowing propagation model for this scenario.

4.1 Experimentation

This part extends our previous work [20] by presenting new results and we tested another platform for comparison. In the following, we show the experimental methodology and results.

4.1.1 Test scenario

The experimentation was conducted in an underexplored rainforest area in which a WSN could be deployed for environmental monitoring. This scenario is defined by a small area inside a preservation area, in which the terrain is flat and we have several natural obstacles to signal propagation, such as a wide variety of trees, bushes and leaves (Fig. 1a). This is a typical dense rainforest with low human intervention. We have experimented with sensor nodes positioned on the ground (Fig. 1b). These nodes are deployed in an ad hoc and unplanned fashion (dropped on the ground) or by using autonomous mobile robots. Our previous work [20] showed the case where the nodes were positioned on a base 1.25 m from the ground. The results were similar with those we present here (in Sect. 4.1.4), but with a higher communication range.

4.1.2 Wireless sensor network platforms

We have chosen the Crossbow’s MicaZ [14] and Iris [15] as the evaluation platforms (Fig. 2), for these belong to a popular sensor platform series that have been used in several prototype applications. MicaZ and Iris represent a evolution compared to the previous representative, the Mica2. The MicaZ platform adopts the CC2420 radio [12] whereas Iris adopts the RF230 radio [8] both compliant with the IEEE 802.15.4 (ZigBee) standard [23]. Table 1 shows the main parameters provided by the manufacturer for both radios.

Both radios provide the RSSI and the LQI. The RSSI indicates the received power at the RF pins of the radio of received packets. The LQI indicates how good is the wireless link and can be affected by spectrum interferences and high noise levels. The received power and LQI will be used as metrics in our experiments. Table 2 presents the manufacturer specifications for both radios. According to the data-sheet of the Iris’s radio [8], the equation which converts RSSI to dBm is only valid for values between 1 and 28. A value of 0 indicates a received power less than $-91$ dBm.

According to IEEE 802.15.4 standard definitions [23], a LQI measurement of a received packet should be an integer
Table 1 Transceivers main parameters (data-sheet [8,12,15,14])

| Parameter          | MicaZ        | Iris         |
|--------------------|--------------|--------------|
| Transceiver        | CC2420       | RF230        |
| Radio frequency (GHz) | 2.4          | 2.4          |
| Bandwidth (kbps)   | 250          | 250          |
| RF power (dBm)     | –24 to 0     | –17.2 to 3   |
| Receiver sensitivity (dBm) | –94          | –101         |
| Outdoor comm. range (m) | 75–100       | >300         |
| Indoor comm. range (m) | 20–30        | >50          |

Table 2 RSSI specifications (data-sheet [8,12])

| Parameter          | CC2420 | RF230 |
|--------------------|--------|-------|
| Dynamic range (dB) | 100    | 81    |
| Resolution (dB)    | 1      | 3     |
| Register length (bits) | 8      | 5     |
| Values range\(^a\) | –60 to 40 | 0 to 28 |
| Power convers. equ.\(^b\) | –45 + RSSI | –91 + 3 (RSSI – 1) |

\(^a\) Values are registered in signed 2’s complement.
\(^b\) Power in dBm.

value between 0 and 255. However, the CC2420 provides only an average correlation value (also known as a chip correlation indicator—CCI) of the eight first symbols of the received packet which can be combined with the RSSI to generate a LQI measurement by software. This correlation value varies between 50 and 110. Normally, only the average correlation value is used as LQI, and no additional processing is done [1,37]. For RF230, LQI values vary between 0 and 255 and is highly associated to an expected packet error rate (PER). A value of 255 indicates no frame error, whereas low values represent many erroneous received frames.

4.1.3 Test procedure

In every test round, the source node sends 20 packets to the sink node at a rate of 10 packets per s. We use the 20 transmitted packets to compute the delivery ratio.\(^3\) The transmission rate of 10 packets per s allows us to collect more data in less experimentation time without congestion in the network. By hardware limitations, we configured the output transmission power on 0 dBm for MicaZ and 0.5 dBm for Iris node. In addition, the sink node assesses the received power (obtained from the RSSI) and LQI for every packet. These three compose the set of metrics evaluated on the experiment. For each platform, we vary the distance between source and sink nodes, and in order to ensure the statistical relevance, we perform 33 rounds for every different platform.

\(^3\) The percentage of packets received regarding the total packets sent by the source node.

4.1.4 Experimental results

Figure 3a presents the relationship between the delivery ratio and the distance between source and sink nodes. We show all 33 values for each distance. Complementarily, we plot the average delivery ratio for every distance. For MicaZ nodes, the largest distance we could reach was only 6 m with about 30% of delivery ratio (average). Iris nodes could reach 13.5 m with a high level of delivery ratio which can be explained by the “transitional region”, discussed in Sect. 3. In this region, the communication becomes unstable due to the low signal-to-noise ratio (SNR) and presence of obstacles in such a configuration that contributes to the signal instead of degrading it. This phenomena is also observed at 10 m. Indeed, this region ranges from 7 to 13.5 m. Further than 13.5 m, no communication was observed.

Iris nodes reach larger distances than MicaZ because of the higher radio sensibility level (presented in Table 1). For Iris node, significant obstacle\(^4\) produced local attenuations at 1.80 and 2.25 m.

For both platforms, we observe a great variation on the delivery ratio when the nodes’ distance was closer to the maximum distance we had attained, just like we observed in our previous work [20] and also reported by the literature [37,43] (discussed in Sect. 3).

Figure 4a presents the received power values when we increase the distance between source and sink nodes. Once more, we show all 33 values for each scenario and the average values as well. Confirming the theory presented in Sect. 2.1.2, the received power decreases with the distance until it reaches the radio sensibility (presented in Table 1).

\(^4\) This obstacle is clearly seen in Fig. 1b, and is very common on the rainforest.
For Iris node, all the received power averages for distances greater than 7 m were $-94$ dBm. This saturation occurred because, at these distances, all the RSSI readings were zero indicating a received power lower than $-91$ dBm, according to the radio’s data-sheet [8]. In fact, we do not know the actual value for these distances, therefore, we plot on the graph the value $-94$ dBm which is the power calculated by the equation described in Table 2 when RSSI = 0. For many applications, such as localization systems, received power measurements are very important and are usually calculated using RSSI readings. Actually, using the Iris Mote, we only have relevant RSSI measurements (in range 1–28) when the distance is up to 6 m in the Amazon rainforest environment. For further distances we only have a “poor signal” indication.

In addition, the delivery ratio is used by many routing algorithms as estimation of link quality for choosing routes.

In order to allow the comparison of the received power with the link quality, Fig. 4b correlates the received power values with the delivery ratio. We observe that when the signal strength is greater than a threshold ($-90$ dBm for MicaZ and $-94$ dBm for Iris—or a RSSI indication of zero) we probably will have high delivery ratios. However, for smaller values, the received power quickly decreases and presents great variations. This fact is also observed in other scenarios reported in different papers, such as Srinivasan and Levis [37] discussed in Sect. 3. Thus, we can conclude that RSSI may be used in all cases as a simple threshold for the link quality.

By respecting the same format used to assess the received power, Fig. 5a shows the LQI variation. As discussed in Sect. 4.1.2, LQI for MicaZ varies between 50 and 110, whereas for Iris, the values are between 0 and 255—following the IEEE 802.15.4 standard specifications. As expected, LQI
behavior is similar to the RSSI, i.e., it decreases according to the distance between nodes. Once again, we observed great variations in the LQI when the distance was close to the maximum distance.

Comparing LQI variation with delivery ratio and received power variation, we observe that, for Iris nodes, the LQI is more related to the delivery ratio than received power (Figs. 3, 5a). But, for MicaZ, LQI is more related to the received power (Figs. 4a, 5a). Figure 5b supports this observation. Although the delivery ratio also presents great variation, for Iris nodes, this indicator is more representative than the received power, since such a variation is smoother for the LQI. For MicaZ nodes, we have a more abrupt variation with the decay of the indication around the value 75. As a link quality indication, the LQI should be highly correlated with the delivery ratio instead of the received power. Thus, we can conclude that Iris’s LQI, although still far from ideal, is better than MicaZ’s. Moreover, the usage only of the average correlation value (CCI) as LQI is not recommended, requiring additional calculation as suggests the manufacturer of MicaZ’s radio.

All these experiments reinforce the difficulty of using received power and LQI to represent the link quality and distance estimation. These factors are relevant as they have been considered by solutions for data routing and node localization that usually adopt imprecise theoretical values.

4.2 Shadowing model parameters estimation

After practical experiments, we had a database with collected pairs of average of received power and distance \((p_i, d_i)\). Using this database, we estimated the parameters of the Shadowing propagation model (described in Sect. 2.1.2) for MicaZ and Iris nodes. Normally, the \(N\) coefficient from Eq. 6 varies between 1.5 and 6 and can be obtained by minimizing the MSE described as follows

\[
\text{MSE}(N) = \frac{1}{k} \sum_{i=1}^{k} (p_i - \hat{p}_i)^2
\]  

(8)

where \(p_i\) is a received power measured for a given distance \(d_i\), \(\hat{p}_i\) represents the estimated received power using a propagation model. For the Shadowing model, \(\hat{p}_i\) can be written as

\[
\hat{p}_i = P_0 - 10N \log \left( \frac{d_i}{d_0} \right).
\]  

(9)

Note that Eq. 9 differs from Eq. 6 by the term \(X_\sigma\) which does not appear in Eq. 9. It means that we only consider the deterministic part of the Shadowing model to estimate the received power. \(d_0\) was the shortest distance in the propagation experiment. \(P_0\) was the average measured power at \(d_0\).

According to Rappaport [33], the standard deviation \(\sigma\) from Eq. 6 can be calculated using the minimum mean square error (MMSE) found by:

\[
\sigma = \sqrt{\text{MSE}}.
\]  

(10)

The estimated parameters were compared in Table 3 with parameters for others environments. The first set of parameters are the default values used in NS-2 [17]. The NS-2 is one of the most used simulators in WSNs for scale scenarios. Sets 2 and 3 represent the parameters for a Spanish forest with two different configurations [21]. Finally, the last two sets represent the estimated values in our practical experiments for MicaZ and Iris nodes. We have also added the MSE for MicaZ’s and Iris’ collected measurements of each set.

We can notice that the MSE of sets for MicaZ and Iris were close compared to the others sets. The sets 1, 2 and 3 were at least ten times greater than the Amazon Forest’s sets for the collected data. Acceptable set of parameters (those with low MSE) generally has both \(N\) and \(P_0\) parameters approximate to the actual values for the same \(d_0\). This can be observed on the first four sets, where we had the same value of \(d_0\) but an unbalance on the \(N\) and \(P_0\) parameters. For sets 4 and 5, the \(d_0\) was different, but the others parameters were correctly balanced.

Figure 6 presents the collected pairs \((p_i, d_i)\) for both platforms. Complementarily, we plot the mean curves (discarding the random term \(X_\sigma\) from Eq. 2.1.2) for the Shadowing propagation model using the sets of parameters from Table 3. For Iris node, the values greater than 6 m were not considered for the regression and MSE calculation.

Although there is the difference of values in Table 3 between sets 4 and 5, the curves in Fig. 6 for these sets are roughly similar which can be verified by the low MSE values. This indicates that the parameters are describing the environment instead of motes. If we keep the same basic
characteristics (frequency and transmission power), we will have similar results for the environment. Besides, the model focuses on path loss (according to Eq. 5) allowing even the representation of different transmission powers. Actually, the unique parameter related to the device is \( \sigma \), once a particular radio could use better components that are more immune to variations/noise. Further, designers of the radio can also provide extra hardware to deal with this problem decreasing the sigma parameter. We also can see that the mean curves for the sets 1, 2 and 3 are far from the curves for our regression lines. This is in agreement with the high values of MSE showed in Table 3. Clearly, we can observe that sets 1, 2 and 3 do not represent the characteristics of the Amazon rainforest.

4.3 Tuning propagation model parameters

In this section we provide the procedure for characterizing new environments and for adjusting the propagation model parameters in the NS-2 simulation tool. The goal is to find parameters for a given environment even though we do not understand all the details of its behavior. In other words, our concern is about the statistical behavior of the communication/channel, not the particularities of the forest. The steps are:

1. **Signal strength experimentation** First, we need to collect RSSI readings in the region varying the transmitter–receiver distance (see Sect. 4.1.3). The more readings you collect, the more accurate and statistically relevant will be your results. We suggest at least 30 samples per distance. A good idea is to find the maximum communication range, and then, take as many as possible measurements within this range. It is important to take note of the transmission power \( P_t \).

2. **Parameter estimation** By using the steps described in Section 4.2, we estimate the Shadowing propagation model parameters \( N \) and \( \sigma \). We need to consider the transmission power \( P_t \) used in the experimentation.

3. **Simulator parameter adjusting** In the NS-2 simulator, we set the earlier found shadowing propagation model parameters \( N \), \( \sigma \), and \( d_0 \) in the TCL configuration file.\(^5\) The parameter \( d_0 \) is the shortest distance between transmitter and receiver in the signal strength experiment.

5. **Localization experiments**

Many solutions for wireless communication are proposed in the literature and validated in simulation tools considering erroneous or unrealistic assumptions. In this section, we show the impact of the incorrect tuning of the propagation parameters in a simulation tool. In the following, we describe the experiment.

After estimating the propagation model parameters, we carried out two experiments, both based on a localization application: (1) a practical experiment; and (2) a simulation experiment. This application was chosen due to its relevance in monitoring and event detection systems using WSNs [9, 10]. Hereafter, we use only the Iris platform in our analysis.

The practical experiment was implemented based on the localization system for indoor environment proposed by BluePass [16]. Four fixed nodes were placed at the corners of a square area of \( 4 \times 4 \) m\(^2 \) in the Amazon rainforest. Another node, called mobile node, was placed inside this square area. Periodically, the fixed nodes send data to mobile node. The mobile node gets the RSSI values and stores the received power in a list. For every new received packet, we calculate a new average of input powers. So, we estimate the distances between fixed nodes and mobile node, using the following equation:

\[
\hat{d}_i = 10^{(P_0 + 10N \log(d_0) - p_i)/(10N)},
\]

where \( \hat{d}_i \) is the estimated distance from mobile node to a fixed node \( i \). \( P_0 \), \( N \) and \( d_0 \) are the Shadowing propagation model parameters. \( p_i \) is the current average received power related to a packets received from fixed node \( i \). Using these distances we estimate the mobile node position, called \( ppos \), using multilateration. For every new received packet, we reestimate

| Set | Environment       | \( N \) | \( \sigma \) (dBm) | \( d_0 \) (m) | \( P_0 \) (dBm) | MSE\(_{\text{MicaZ}}\) | MSE\(_{\text{Iris}}\) |
|-----|-------------------|--------|-------------------|-------------|--------------|-----------------|-----------------|
| 1   | NS-2 default      | 2.0    | 4.0               | 1.0         | -40.2        | 1,254           | 1,214           |
| 2   | Spanish forest 1  | 2.55   | 7.29              | 1.0         | -55.9        | 297             | 346             |
| 3   | Spanish forest 2  | 3.43   | 6.04              | 1.0         | -43.4        | 667             | 874             |
| 4   | Amazon rainforest | 2.14   | 1.86              | 1.0         | -75.2        | 3               | 32              |
| 5   | Amazon rainforest | 3.21   | 5.23              | 0.1         | -40.4        | 14              | 27              |

\[^5\]http://www.isi.edu/nsnam/ns/doc/node221.html.
We repeated this experiment for several mobile node positions and for every set of parameters from Table 3.

The simulation experiment was conducted in a NS-2 simulator. We configured the simulator with the same area, nodes position and other characteristics as we did in the practical experiment. We did the following changes with the NS-2 code:

- Setting of the $P_0$ parameter on the Shadowing model.
- Calculation of the Shadowing model power, at a certain distance, according to Eq. 6 in dBm.
- Implementation of the Iris rules for reading RSSI/received power values.

The estimated mobile node position in the simulation experiment was called $spos$. Thus, we calculate the localization error $\varepsilon$ according to the following equation:

$$\varepsilon = \text{dist}(mpos, epos),$$

where $mpos$ is the mobile node position and $epos$ is the estimated position which can be a $ppos$ or $spos$. The function $\text{dist}(\cdot)$ is the Euclidean distance between two points. Figure 7 summarizes the results. $\varepsilon_p$ indicates the localization error in practical experiment and $\varepsilon_s$ the error in simulation. $\varepsilon_{sp}$ will be introduced later.

Figure 8 presents average localization error for practical and simulation experiment when we increase the number of packets used to calculate the average power. These graphs compare the error for the parameters from Table 3, excluding the NS-2 default parameters that presented errors in order of kilometers. From this point on, the set of parameters 5 will be referred to as estimated parameters, and the sets 2 and 3 as literature parameters.

In the practical experiment, our estimated parameters presented the smallest error, less than 2 m whereas the others were around 35 m. We observe a stability of the error values and this is important because some localization systems try to reduce the errors accumulating and averaging readings. Figure 8a shows that this procedure do not increase significantly the method precision.

In the simulation experiment, we had a different behavior for the literature parameters. The maximum error measured was lower than 5 m and this was reduced to a level lower than 0.5 m. This difference occurred because in practical experiments we had actual power readings whereas, in the simulation, we had artificial readings based on the Shadowing propagation model. So, for each received packet, the distance was estimated (by Eq. 11) using the same parameters used to create the artificial readings (by Eq. 6). Basically, in simulation, the errors come only from the $X_\sigma$ term.
We can consider a good simulator when its results are near to the practical results. Thus, to evaluate the simulation precision we should analyze the error between the estimated node position in simulation and estimated position in a practical experiment. This error will be called simulation precision error—\( \varepsilon_{sp} \) (shown in Fig. 7) and can be calculated as
\[
\varepsilon_{sp} = \text{dist}(spos, ppos).
\]

Figure 9 shows the \( \varepsilon_{sp} \) when we increase the number of packets. We observe a low error in the experiments with the estimated parameters indicating that estimations in simulation were near to estimations in practical experiments. This is an indication that we have a realistic simulation using the estimated parameters. The errors for the literature parameters were almost ten times greater than the length of one side of the square region used in the experiment.

6 Conclusion

This work presented experimental data for rainforest wireless communication, and the adequacy of the parameters of the Shadowing propagation model common in simulation suites. This adequacy was achieved from the measurements of received powers and a regression by using the minimization of the MSE technique. Default parameters, some found in literature and the new ones, were compared in simulations by considering a localization solution. The results showed the importance of proper environment characterization to obtain accurate results in relation to the real world.

All the described procedures can be very expensive to do every time we need a new development. However, the data and parameters achieved in this paper can be easily included in different simulators when a similar environment is the goal.

As future work we intend to evaluate at the simulation level the influence of the rainforest propagation model on several wireless protocols. The goal is to study their feasibility under the approximated conditions of these real scenarios, which is very much worse than the conditions these solutions were originally evaluated. Also, we plan to perform similar experimentation in flooded forest regions, which are very common in the rainforest.

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