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Improving the RPC rate capability

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ABSTRACT: This paper has the purpose to study the rate capability of the Resistive Plate Chamber, RPC, starting from the basic physics of this detector. The effect of different working parameters determining the rate capability is analysed in detail, in order to optimize a new family of RPCs for applications to heavy irradiation environments and in particular to the LHC phase 2. A special emphasis is given to the improvement achievable by minimizing the avalanche charge delivered in the gas. The paper shows experimental results of Cosmic Ray tests, performed to study the avalanche features for different gas gap sizes, with particular attention to the overall delivered charge. For this purpose, the paper studies, in parallel to the prompt electronic signal, also the ionic signal which gives the main contribution to the delivered charge. Whenever possible the test results are interpreted on the basis of the RPC detector physics and are intended to extend and reinforce our physical understanding of this detector.

KEYWORDS: Gaseous detectors; Resistive-plate chambers; Timing detectors; Trigger detectors

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1 Introduction

The RPC is a unique gaseous detector due to the absence of any drift time. This means that the full gas volume is critical i.e. the field is strong enough to produce avalanche multiplication as soon as a free electron is generated in any point of the gas. It is therefore an ideal detector for fast decisions and very high time resolutions. Moreover, its simple structure and low cost materials make it suitable for very large area applications. This makes the RPC a very suitable detector for triggering and for time of flight measurements at present and future colliders, in particular the High Luminosity LHC. In the latter case however the required rate capability might be beyond the limits of the RPC generation presently working at LHC [1] and a substantial improvement is required to overcome this limit. At the same time, it has to be stressed that any rate capability improvement requires also a careful analysis of the detector ageing, to avoid that the improved rate causes a shortened detector lifetime. This paper studies adequate methods to improve the RPC rate capability without increasing the ageing.

The RPC rate capability is mainly limited by the current that can flow through the high resistivity electrodes and can be improved by working on a number of interconnected parameters that are discussed below [2].

The current density flowing across the electrodes can change from point to point, on a few millimetre scale, depending on the map of the points where the gas was ionized. In average however, for a uniform irradiation density, a working current $i$ flowing across the electrodes of total resistance $R$ produces a voltage drop

$$V_{el} = V_a - V_{gas} = Ri$$

$V_a$ and $V_{gas}$ being the voltages applied to the detector and to the gas gap respectively. A high rate capability requires to keep the difference $V_{el}$ at a negligible value with respect to $V_{gas}$ even under heavy irradiation. The voltage drop across the electrodes can be expresses as a function of the basic detector parameters with the following relationship

$$V_{el} = \rho t \langle Q \rangle r_u$$

Here $\rho$ is the bulk resistivity of the electrode material; $t$ is the total thickness of both electrodes; $\langle Q \rangle$ is the average charge delivered in the gas for each count and $r_u$ is the counting rate per unit
surface. We define here the “rate capability” as the achieved rate per unit voltage drop across the electrodes:

\[
\text{Rate Capability} = \frac{r}{V_{el}} = \frac{1}{\rho t \langle Q \rangle}
\]

It can be increased by decreasing each of the 3 parameters \(\rho, t, \langle Q \rangle\), with the advantages/disadvantages discussed in the following.

The search for low resistivity materials for the electrode plates is the most natural way of increasing the rate capability. However, a rate improvement obtained only by lowering the electrode resistivity, would imply a corresponding increase of the operating current and of the pollutants generated inside the gas. Moreover, the spontaneous detector noise could increase as well. This focusses a potential ageing problem and requires a new robust ageing test to qualify the RPCs for a high current working mode.

To reduce the electrode thickness \(t\) has in principle a similar effect on the rate capability as to reduce its resistivity, the total electrode resistance being proportional to the product \(\rho t\). However, a further important effect of the electrode thickness is to attenuate the charge induced on the read out electrodes by a factor

\[
A = 1 + \frac{t}{\varepsilon_r g}
\]

[3]. Here \(\varepsilon_r\) and \(g\) are the relative dielectric constant of the electrode material and the gas gap size respectively. A thinner electrode “amplifies” therefore the signal induced on the read out electrodes, thus improving the signal to noise ratio. This effect, which is particularly relevant for very thin gaps, shows that the electrode thickness should be reduced according to the gap size.

The approach of reducing the average charge \(\langle Q \rangle\) has the advantage to improve the achievable rate with minimum or null worsening of other parameters such as the detector ageing. The most ambitious goal, in this perspective, is to increase the rate at constant current, thus making any further ageing test unnecessary. This approach requires: i) to optimize the gas gap size in order to find the best compromise between detection efficiency and delivered charge; ii) to transfer, from the gas to the front end electronics, a relevant part of the amplification needed to get a detectable avalanche signal. The circuit required for this purpose must be very sensitive, with an excellent signal to noise ratio. Moreover, it requires also a careful optimization of the chamber structure as a Faraday cage, to achieve a high level suppression of the noise originated both by the detector itself and by external sources. The limit of this approach is just the achievable signal to noise ratio.

This paper has the purpose to investigate how the RPC rate capability can be improved, following a pure “low charge” approach, based on a systematic study of the performance vs the gap size, under the constraints imposed by the front end electronics gain and signal to noise ratio. This approach is particularly suitable for the RPCs built with electrodes of phenolic High Pressure Laminate (HPL), whose resistivity, of about \(\sim 3 \cdot 10^{10}\) Ohm \(\times\) cm, [1] is already sufficiently low, with respect to other materials used so far. Electrodes of this material have been used for the largest and most relevant RPC detectors in operation now or in the recent past: Atlas, CMS, Alice Trigger, Opera, Argo-YBJ.
2 Study of the signals generated by RPCs of different gap sizes

We report here a Cosmic Ray laboratory test performed on three RPCs with different gap sizes, 0.5 mm, 1 mm and 2 mm respectively, but identical for any other aspect. The gas volumes, of active area $52.6 \times 7.6 \, \text{cm}^2$, are built with the same materials as those operating in Atlas [4] and in particular with the same electrodes of phenolic High Pressure Laminate, 1.8 mm thick. The working gas is a mixture of Tetra-Fluor-Ethane (commercial name Suva 134 a), iso-Butane and Sulphur Hexafluoride: $\text{C}_2\text{H}_2\text{F}_4/\text{i-C}_4\text{H}_{10}/\text{SF}_6 = 94.5/5.0/0.5$, and is also almost identical to the standard Atlas and CMS gas. The chamber lay out, specifically conceived for this test, is shown in figure 1.

The Cosmic Rays crossing the detector under test are selected by the coincidence of other three RPCs, focussing a chamber area of $50 \times 3.0 \, \text{cm}^2$ at the centre of the tested RPC. This area is covered by a single read out strip shaped as shown in figure 2, which behaves like a 50 Ohm transmission line terminated at both ends with its impedance, compatible with the standard 50 Ohm coaxial cables.

The prompt signal charge mentioned in this work is defined as the charge induced on the read out copper strip by the avalanche electrons drifting in the gas. It is corrected taking into account that the signal detected by the scope at one end of strip brings only half of the primary charge, the other half being absorbed by the resistor matching the line at the opposite end.

![Figure 1. Sketch of the RPCs under test.](image1)

![Figure 2. The shape of the read out strip. This shape allows to cover a 2.7 cm width with relatively high impedance line, about 50 Ohm, well matched with the standard coaxial cables used to send the signals to the oscilloscope.](image2)

The experimental set up includes also: the high and low voltage systems, the gas mixing station, a few logic units to elaborate the trigger signal and the Data Acquisition. For each trigger signal, the following waveforms are recorded on two different oscilloscopes:
Figure 3. Waveforms generated by (from left to right): 2 mm gap, 1 mm gap, 0.5 mm gap respectively. The prompt signals due to the electron drift on top and the slow signals due to the ions drift on the bottom are shown for each trigger. Note the horizontal scale, in nano and micro-second respectively.

a) the prompt signal, induced on the readout strip by the electrons drifting inside the gas gap. This signal is recorded with a sampling rate of 10 Gsample/s and 1 GHz analog band.

b) the slow signal, induced on the graphite electrode by the ions drift. It is read out on a 10 kOhm resistor introduced in the HV circuit as shown in figure 1 and recorded with 5 Msample/s and 1 GHz analog band.

In both cases the signals from the RPC under test are directly fed into the oscilloscope and the recorded waveforms for the three gaps are shown in figure 3.

The comparison of the fast and slow signal rise times, for all three gaps, shows a constant ratio around $5 \times 10^3$, consistent with the drift velocities of electrons and ions respectively.

The comparison of waveforms from different gaps shows that the rise times of both prompt and ionic signals scale proportionally to the gap size, suggesting that the drift velocities of electrons and ions are both field independent.

The prompt signals are recorded under two different conditions: a) the signal is directly fed into the oscilloscope, as shown above; b) the signal is previously amplified and integrated with a fast charge amplifier [5] and subsequently sent to the oscilloscope. These two kinds of signal are referred, in the following, as non-amplified and amplified respectively. The ionic signal, recorded together with both amplified and not amplified prompt pulses, is used to correlate them. For each working condition, the recorded waveforms are analyzed to measure all relevant parameters, such as signal charge and amplitude, signal duration and timing.

The features of the charge amplifier used in the test, are reported in table 1.

The amplifier calibration plot, shown in figure 4, gives the output amplitude measured for different input charges. The points shown in this plot are obtained by binning the ionic charge in bins of 0.4 pC and correlating, in each bin, the average output amplitude of the amplified signal with the average input charge of the not amplified signal. The plot shows that the amplifier output saturates at 800 mV and, for small signals, the ratio of the output-amplitude/input-charge is about $200 \text{ mV}/0.1 \text{ pC} = 2 \text{ mV}/\text{fC}$. The supplied front end voltage is 5 V.
Table 1. Main features of the charge amplifier used in the test.

| Feature            | Value          |
|--------------------|----------------|
| Sensitivity        | 2 mV/fC        |
| Noise              | 0.3 fC RMS     |
| Latch capability   | 100 ps         |
| Bandwidth          | 50 MHz         |
| Power consumption  | 6 mW           |
| $V_{th}$           | > 15 mV        |
| $Q_{th}$           | > 6 fC         |

The detection efficiency can be measured using both the prompt and the ionic signal, with appropriate threshold criteria to discriminate the signal from the noise. These criteria are listed below:

a) The amplitude of the non-amplified prompt signal exceeds a fixed threshold of 1.5 mV (corresponding to a charge about 150 fC for a 2 mm gap). This threshold is equivalent to the one of the RPCs working in Atlas.

b) The amplitude of the amplified prompt signal exceeds a threshold of 40 mV. According to the calibration plot of figure 4 this corresponds to a 20 fC charge threshold for the input signal.

c) For the ionic signal instead, the threshold in amplitude is normalized to the output noise: $V_{thresh} = 5 \sigma_{noise}$, where $\sigma_{noise}$ is the noise standard deviation.

d) It has to be stressed that with the criterion b), when we operate at very high plateau voltages, the higher intrinsic noise of the oscilloscope scale could give some false efficiencies at the plateau. On the contrary, below 90% efficiency, where we evaluate the amplification effect on the operating voltage, this problem does not exist.

The detection efficiencies achievable with the prompt signal, for the three different gaps, are shown in figure 5 vs the applied field. This figure allows to compare the efficiencies achievable with both amplified and non-amplified signals.

The comparison shows that:

1. The operating field strongly increases for decreasing gap size.

2. The amplified efficiency plots, compared to the not amplified ones, are systematically shifted towards a lower field. For the 2 mm gap, e.g., the voltage shift is $\Delta V = -350$ V. The avalanche charge is consequently lowered by a factor of about 3 (see figure 6) at fixed efficiency and the rate capability should be improved accordingly.

3. The plateau efficiency of the 0.5 mm gap is substantially lower, about $-15\%$, with respect to the 1 mm gap, indicating an insufficient primary ionization.

4. The plateau efficiency comparison among the 3 gaps has to take into account a systematic difference in the construction method: the 2 mm and 0.5 mm gaps have a line of spacers in the middle which introduce a dead area of 3.9%. In this case avalanches generated at the
Figure 4. Front end amplifier calibration plot.

Figure 5. Plots of the detection efficiencies vs the applied electric field, for the amplified (red points) and not amplified (black points) signals. The plots refer to 2 mm, 1 mm and 0.5 mm gas gaps respectively.

Figure 6. Prompt signal amplitude vs electric field for the three gaps.
spacer edge, shaped as shown in figure 1, induce a very small charge on the read out strip, that can only be detected through the amplifier and give a slightly higher efficiency for the amplified plot. The 1 mm gap on the contrary has 2 lines of spacers at the sides so that the tested area is spacer free and the measured plateau inefficiency is the intrinsic one.

5. Finally, it has to be noted that at the plateau voltage, the amplified efficiency might be slightly over evaluated due to the noise problem mentioned above.

Figure 6 shows the plot of the prompt signal amplitude vs the applied field for the three gaps. The corresponding prompt charge is plotted in figure 7. It has to be stressed that this is the prompt charge delivered in the gas, which is twice the signal charge detected by the oscilloscope, as explained above. The comparison of the plots shows that just above the plateau knee of the non-amplified plot, the amplitudes are approximately the same for all three gaps, whereas the charges scale proportionally with the gap size.

![Figure 6](image)

**Figure 6.** Prompt signal amplitude vs the electric field for the three gaps. The corresponding prompt charge is plotted in Figure 7.

**Figure 7.** Prompt signal charge vs the electric field for the three gaps. Typical values of the signal charge and amplitude at the plateau knee, for the three gaps, are reported in Table 2, together with the output amplitude of the front-end integrator.

| Gap size (mm) | Amplified Efficiency @ plateau knee | Op. Voltage for amplified signal (V) | Average input prompt charge (pC) | Average input prompt amplitude (mV) | Average output amplified amplitude (mV) |
|---------------|-------------------------------------|-------------------------------------|---------------------------------|-----------------------------------|----------------------------------------|
| 2             | 90%                                 | 9660                                | 0.4                             | 1.6                               | 640                                    |
| 1             | 90%                                 | 6010                                | 0.2                             | 2.0                               | 440                                    |
| 0.5           | 77%                                 | 3780                                | 0.11                            | 2.5                               | 260                                    |

The overall result of these tests is that thinner gas gaps reduce the delivered charge linearly with the gap size and improve the rate capability accordingly. The detection efficiency however tends
to decrease due to the smaller number of primary ionizations produced in the gap. This number can be evaluated assuming a model in which the efficiency is given by the ionizations occurring in the “effective gap”, which is a convenient fraction of the gas gap on the cathode side. Its width increases with the applied voltage. The inefficiency, $1 - \varepsilon$, is the “zero probability” of a Poisson distribution with $\langle n \rangle$ ionizations in average [6].

$$1 - \varepsilon = e^{-\langle n \rangle} \rightarrow \langle n \rangle = \ln \left( \frac{1}{1 - \varepsilon} \right).$$

The value of $\langle n \rangle$, which is proportional to the effective gap size, is plotted in figure 8 vs the applied voltage, for the 1 mm gas gap. At 6560 V in the plateau, e.g. the intrinsic inefficiency is 1.5% with 4.2 primary ionizations in average.

![Figure 8](image)

**Figure 8.** Number of primary ionizations contributing to the detection efficiency vs the applied voltage. The plot refers to the 1 mm gap.

The same model can be applied to the 0.5 mm gap. In this case the 89% intrinsic plateau efficiency, after correction for the spacers dead area, is consistent with $\langle n \rangle = 2.2$, in agreement, within the approximations, with the value obtained for the 1 mm gap, twice larger.

In the case of 2 mm gap, the 3.9% spacers dead area do not allow to evaluate the intrinsic inefficiency with the needed accuracy.

The ionic charge, shown in figure 9, is normally much larger than the prompt charge [3].

A further comparison parameter for different gaps is the ratio $Q/q$ of the ionic ($Q$) to the prompt ($q$) charge. Figure 10 shows this ratio vs the ionic charge $Q$ for the three gap sizes. The three plots have very similar shapes, with a maximum for small $Q$, in the range of very few pC, and an asymptotic value for large $Q$ values. The maxima slightly grow for increasing gap, suggesting that, at low voltages, the prompt charge collection is more efficient for thinner gaps. The asymptotic values are the same, about 12, for all three gaps.

## 3 Combined effect of thinner gaps and higher sensitivity front end electronics

The purpose of all tests reported so far is to minimize, for each detected signal, the total charge delivered in the gas, i.e. the sum of the prompt+ionic contributions. This charge, indeed multiplied
by the counting rate, gives the working current flowing across the high resistivity electrodes. So far we studied, in an independent way, the effects of thinner gas gaps and of the new front end electronics with improved sensitivity. Both effects contribute to reduce the delivered charge and therefore to improve the rate capability. However, when the two effects are combined, the overall gain depends on their correlation and could range, in principle, between two extreme cases:

i) the two effects are uncorrelated (the most favourable case) and the overall gain is the product of the two independent gains;

ii) the gain due to front end electronics embodies the one of the thinner gap, which does not give any further advantage (the most unfavourable case).

This section shows that the real case is in between. A comprehensive presentation of the delivered charge, under all tested conditions, is summarized in table 3, which shows the results of the three gap sizes with and without the front end amplifier.
Table 3. Average value of the total charge delivered by the 3 gaps at the efficiency plateau knee. The columns 3 and 4 refer to the output signal directly fed into the oscilloscope. In the columns 5 and 6 the signal was previously amplified by the front end circuit described in table 1.

| Gap size (mm) | Efficiency @ plateau knee | Operating Voltage for not amplified signal (V) | Average total charge for not amplified signal (pC) | Operating Voltage for amplified signal (V) | Average total charge for amplified signal (pC) |
|---------------|---------------------------|-----------------------------------------------|--------------------------------------------------|-------------------------------------------|-----------------------------------------------|
| 2             | 90%                       | 10040                                         | 19.6                                             | 9660                                      | 6.5                                           |
| 1             | 90%                       | 6260                                          | 9.3                                              | 6010                                      | 4.9                                           |
| 0.5           | 77%                       | 4000                                          | 4.3                                              | 3780                                      | 3.05                                          |

Table 3 allows to factorize the effects of the gas gap and the front end amplification. It shows that:

1. When the gap is reduced from 2 mm to 0.5 mm in the absence of amplification the delivered charge decreases by a factor $\frac{19.6}{4.3} = 4.6$.

2. When the signal is amplified at constant gap size of 2 mm the delivered charge decreases by a factor $\frac{19.6}{6.5} = 3.0$.

3. The combined effect in the absence of correlation would give a factor $4.6 \times 3 = 13.8$.

4. In the real case, due to the correlation, the overall effect, in reducing the delivered charge, is a factor $\frac{19.6}{3.05} = 6.4$.

The advantage of thin gaps, in addition to what can be obtained with an improved front end sensitivity, is due to the fact that the higher achievable field produces a faster avalanche saturation, thus reducing the large amplitude spread of the signals generated inside the gas. In the exponential avalanche growth indeed, the signal charge/amplitude variability is due to: a) the primary cluster Poisson statistics, b) the number of electrons in each cluster and c) the different weight of these clusters which decreases exponentially with the distance from the anode. The cluster saturation substantially reduces the variability due to b) and c).

4 Conclusions

We performed a systematic study of the parameters determining the charge delivered in the RPC gas for each output signal. Reducing this charge indeed is crucial to improve the RPC rate capability. We have shown that the output signal duration is proportional to the gap and therefore, for thinner gaps, the delivered charge is reduced accordingly. The combined effect of reducing the gap size from 2 mm to 0.5 mm and improving the front end electronics, thanks to a new charge amplifier with $2 \text{mV/fC}$ sensitivity and $0.3 \text{fC rms}$ noise, is a delivered charge a factor 6.4 smaller and a rate capability correspondingly higher. We have also demonstrated that the operating field strongly increases for decreasing gap size and that the ratio of the ionic to prompt charge is asymptotically the same for all gaps, with a numerical value of about 12.
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