Universal coding for correlated sources with complementary delivery *†

Akisato Kimura ‡ Tomohiko Uyematsu §
Shigeaki Kuzuoka ¶

February 5, 2008

Abstract

This paper deals with a universal coding problem for a certain kind of multiterminal source coding system that we call the complementary delivery coding system. In this system, messages from two correlated sources are jointly encoded, and each decoder has access to one of the two messages to enable it to reproduce the other message. Both fixed-to-fixed length and fixed-to-variable length lossless coding schemes are considered. Explicit constructions of universal codes and bounds of the error probabilities are clarified via type-theoretical and graph-theoretical analyses.

Keywords: multiterminal source coding, complementary delivery, universal coding, types of sequences, bipartite graphs

1 Introduction

The coding problem for correlated information sources was first described and investigated by Slepian and Wolf [1], and later, various coding problems derived from that work were considered (e.g. Wyner [2], Körner and Marton [3], Sgarro [4]). Meanwhile, the problem of universal coding for these systems was first investigated by Csiszár and Körner [5]. Universal coding problems are not only interesting in their own right but also very important in terms of practical applications. Subsequent work has mainly focused on the Slepian-Wolf coding system [6] [7] [8] since it appears to be difficult to construct universal codes for most of the other coding systems. For example, Muramatsu [9] showed that we

---

*Some of the material in this manuscript has been already published in IEICE Transactions on Fundamentals, Vol.E90-A, No.9, pp.1840-1847, September 2007. Manuscript received December 19, 2006. Manuscript revised March 29, 2007. Final manuscript received April 20, 2007. Several additional results are also included.
†Parts of the material in this paper were presented at the International Symposium on Information Theory, Nice, France June 2007.
‡NTT Communication Science Laboratories, NTT Corporation, 3-1 Morinosato Wakamiya, Atsugi-shi, Kanagawa, 243-0198 Japan. E-mail: research <at> akisato org URL: [http://www.brl.ntt.co.jp/people/akisato/]
§Department of Communications and Integrated Systems, Tokyo Institute of Technology, 2-12-1 Ookayama, Meguro-ku, Tokyo, 152-8550 Japan. E-mail: uyematsu <at> ieee org
¶Department of Computer and Communication Sciences, Wakayama University. 930 Sakaedani, Wakayama-shi, Wakayama, 640-8510 Japan. E-mail: kuzuoka <at> sys wakayama-u ac jp

arXiv:0710.4187v1  [cs.IT]  23 Oct 2007
cannot construct asymptotically optimal fixed-to-variable length (FV) universal code for Wyner-Ziv coding systems \cite{10} in terms of the coding rate.

This paper deals with a universal coding problem for a certain kind of multiterminal source coding system that we call a complementary delivery coding system \cite{11}. Figure 1 shows a block diagram of the complementary delivery coding system. The encoder observes messages emitted from two correlated sources, and delivers these messages to other locations (i.e., decoders). Each decoder has access to one of two messages, and therefore wants to reproduce the other message. Although the previous articles \cite{11} considered lossy configurations, this paper considers a lossless configuration. We show an explicit construction of fixed-to-fixed length universal codes. We also clarify the upper and lower bounds of the error probabilities via type-theoretical and graph-theoretical analyses. Fixed-to-variable universal codes can also be constructed in a similar manner.

This paper is organized as follows: Notations and definitions are provided in Section 2. Previous results for the complementary delivery coding system are shown in Section 3. A proposed coding scheme is described in Section 4. Several coding theorems are clarified in Section 5. Variable-length coding is discussed in Section 6. Finally, concluding remarks are given in Section 7.

2 Preliminaries

2.1 Basic definitions

Let $\mathcal{X}$ be a finite set, $\mathcal{B}$ be a binary set, and $\mathcal{B}^*$ be a set of all finite sequences in the alphabet $\mathcal{B}$. Let $|\mathcal{X}|$ be the cardinality of $\mathcal{X}$ and $\mathcal{I}_M = \{1, 2, \cdots, M\}$. A member of $\mathcal{X}^n$ is written as $x^n = (x_1, x_2, \cdots, x_n)$, and substrings of $x^n$ are written as $x^i_j = (x_i, x_{i+1}, \cdots, x_j)$ for $i \leq j$. When the dimension is clear from the context, vectors will be denoted by boldface letters, i.e., $\mathbf{x} \in \mathcal{X}^n$. $\mathcal{M}(\mathcal{X})$ denotes the set of all probability distributions on $\mathcal{X}$. Also, $\mathcal{M}(\mathcal{X}|P_Y)$ denotes
the set of all probability distributions on $X$ given a distribution $P_Y \in \mathcal{M}(Y)$, namely each member $P_{X|Y}$ of $\mathcal{M}(X|P_Y)$ is characterized by $P_{X|Y} \in \mathcal{M}(X \times Y)$ as $P_{XY} = P_{X|Y}P_Y$. A discrete memoryless source $(X, P_X)$ is an infinite sequence of independent copies of a random variable $X$ taking values in $\mathcal{X}$ with a generic distribution $P_X \in \mathcal{M}(X)$. We will denote a source $(X, P_X)$ by referring to its generic distribution $P_X$ or random variable $X$. For a correlated source $(X,Y)$, $H(X|Y)$ denotes the conditional entropy of $X$ given $Y$. For a generic distribution $P_{X|Y} \in \mathcal{M}(X|P_Y)$, $H(P_{X|Y}|P_Y)$ also denotes the conditional entropy of $X$ given $Y$. $D(P||Q)$ denotes the Kullback-Leibler divergence between two distributions $P$ and $Q$.

In the following, all bases of exponentials and logarithms are set at 2.

2.2 Types of sequences

Let us define the type of a sequence $x \in X^n$ as the empirical distribution $Q_x \in \mathcal{M}(X)$ of the sequence $x$, i.e.

$$Q_x(a) = \frac{1}{n} N(a|x) \quad \forall a \in \mathcal{X},$$

where $N(a|x)$ represents the number of occurrences of the letter $a$ in the sequence $x$. Similarly, the joint type $Q_{x,y} \in \mathcal{M}(X \times Y)$ is defined by

$$Q_{x,y}(a,b) = \frac{1}{n} N(a,b|x,y) \quad \forall (a,b) \in \mathcal{X} \times \mathcal{Y}.$$ 

Let $\mathcal{P}_n(X)$ be the set of types of sequences in $X^n$. Similarly, for every type $Q \in \mathcal{P}_n(X)$, let $\mathcal{V}_n(Y|Q)$ be the set of all stochastic matrices $V : \mathcal{X} \to \mathcal{Y}$ such that for some pairs $(x,y) \in \mathcal{X} \times \mathcal{Y}$ of sequences we have $Q_{x,y}(x,y) = Q(x)V(y|x)$.

For every type $Q \in \mathcal{P}_n(X)$ we define

$$T^n_Q \overset{\text{def.}}{=} \{ x \in X^n : Q_x = Q \}.$$ 

Similarly, for every $x \in T^n_Q$ and $V \in \mathcal{V}_n(Y|Q)$ we define

$$T^n_V(x) \overset{\text{def.}}{=} \{ y \in Y^n : Q(x)V(y|x) = Q_{x,y}(x,y), \forall (x,y) \in \mathcal{X} \times \mathcal{Y} \}.$$ 

Hereafter, we call $T^n_V(x)$ a $V$-shell.

Here, let us introduce several important properties of types.

**Lemma 1.** (Type counting lemma [12 Lemma 2.2]) The number of different types of sequences in $X^n$ is less than $(n+1)^{|\mathcal{X}|}$, namely

$$|\mathcal{P}_n(X)| \leq (n+1)^{|\mathcal{X}|}.$$ 

**Lemma 2.** (Sizes of V-shells [12 Lemma 2.5]) For every sequence $x \in X^n$ and every stochastic matrix $V : \mathcal{X} \to \mathcal{Y}$ such that the corresponding V-shell $T_V(x)$ is not empty, we have

$$|T_V(x)| \geq (n+1)^{-|\mathcal{X}|} \exp\{nH(V|P_X)\},$$

$$|T_V(x)| \leq \exp\{nH(V|P_X)\}.$$
Lemma 3. (Probabilities of types [12, Lemma 2.6])

For every type $Q$ of sequences in $X^n$ and every distribution $P_X$ on $X$, we have

\[
P_X(x) = \exp\{ -n(D(Q\|P_X) + H(Q)) \} \quad \forall x \in T_Q,
\]

\[
P_X(T_Q) \geq (n + 1)^{-|X|} \exp\{ -nD(Q\|P_X) \},
\]

\[
P_X(T_Q) \leq \exp\{ -nD(Q\|P_X) \}.
\]

3 Previous results

This section formulates the coding problem investigated in this paper, and shows a fundamental bound of the coding rate. We again note that previous work [11] considered lossy coding, whereas this paper considers lossless coding.

First, we formulate the coding problem of the complementary delivery coding system.

Definition 1. (Fixed-to-fixed complementary delivery (FF-CD) code) [11]

A set $(\varphi_n, \hat{\varphi}_n^{(1)}, \hat{\varphi}_n^{(2)})$ of an encoder and two decoders is an FF-CD code with parameters $(n, M_n, e_n^{(X)}, e_n^{(Y)})$ for the source $(X,Y)$ if and only if

\[
\varphi_n : X^n \times Y^n \rightarrow I_{M_n},
\]

\[
\hat{\varphi}_n^{(1)} : I_{M_n} \times Y^n \rightarrow X^n, \quad \hat{\varphi}_n^{(2)} : I_{M_n} \times X^n \rightarrow Y^n,
\]

\[
e_n^{(X)} = \Pr\{ X^n \neq \hat{X}^n \}, \quad e_n^{(Y)} = \Pr\{ Y^n \neq \hat{Y}^n \},
\]

where

\[
\hat{X}^n \overset{\text{def.}}{=} \hat{\varphi}_n^{(1)}(\varphi_n(X^n, Y^n), Y^n),
\]

\[
\hat{Y}^n \overset{\text{def.}}{=} \hat{\varphi}_n^{(2)}(\varphi_n(X^n, Y^n), X^n).
\]

Definition 2. (FF-CD-achievable rate)

$R$ is an FF-CD-achievable rate of the source $(X,Y)$ if and only if there exists a sequence $(\varphi_n, \hat{\varphi}_n^{(1)}, \hat{\varphi}_n^{(2)})_{n=1}^{\infty}$ of FF-CD codes with parameters $(n, M_n, e_n^{(X)}, e_n^{(Y)})_{n=1}^{\infty}$ for the source $(X,Y)$ such that

\[
\limsup_{n \to \infty} \frac{1}{n} \log M_n \leq R,
\]

\[
\lim_{n \to \infty} e_n^{(X)} = \lim_{n \to \infty} e_n^{(Y)} = 0.
\]

Definition 3. (Inf FF-CD-achievable rate)

\[
R_f(X,Y) = \inf\{ R : R \text{ is an FF-CD-achievable rate of } (X,Y) \}.
\]

Willems, Wolf and Wyner [13, 14] investigated a coding problem where several users are physically separated but communicate with each other via a satellite, and determined the minimum coding rate for the three users when transmitting to and from the satellite. The complementary delivery coding system is a special case of the system described by Willems et al., which considers the case of two users. Therefore, we can immediately obtain the closed form of $R_f(X,Y)$ from the result obtained by Willems et al. 

4
Theorem 1. (Coding theorem of FF-CD codes)

\[ R_f(X,Y) = \max \{ H(X|Y), H(Y|X) \} = \max \{ H(P_{X|Y}|P_Y), H(P_{Y|X}|P_X) \} \]

4 Code construction

This section shows an explicit construction of universal codes for the complementary delivery coding system defined by Definition 1. The coding scheme is described as follows:

[Encoding]

1. Determine a set \( S_n(R) \) of joint types as

\[
S_n(R) = \left\{ Q_{XY} \in \mathcal{P}_n(\mathcal{X} \times \mathcal{Y}) : \max \{ H(V|Q_X), H(W|Q_Y) \} \leq R, Q_{XY} = Q_XV = Q_YW, V \in \mathcal{V}_n(Y|Q_X), W \in \mathcal{V}_n(X|Q_Y) \right\},
\]

where \( R > 0 \) is a given coding rate. We note that the joint type \( Q_{XY} \) specifies the types \( Q_X, Q_Y, \) and the conditional types \( V \) and \( W \).

2. Create a table (henceforth we call this a coding table, see Figure 2 left) for each joint type \( Q_{XY} \in S_n(R) \). Each row of the coding table corresponds to a sequence \( x \in T^{n}_{Q_X} \), and each column corresponds to a sequence \( y \in T^{n}_{Q_Y} \).

3. Mark cells that correspond to sequence pairs \( (x, y) \in T^{n}_{Q_{XY}} \) (see Figure 2 middle). Codewords will be given only to sequence pairs that correspond to marked cells.

4. Fill the marked cells with \( \exp(nR) \) different symbols such that each symbol occurs at most once in each row and at most once in each column. An example of symbol filling is shown on the right in Figure 2 right.

5. For a given pair of sequences \( (x, y) \in \mathcal{X} \times \mathcal{Y} \) with the joint type \( Q_{XY} \), if \( Q_{XY} \in S_n(R) \), the index assigned to the joint type \( Q_{XY} \) of \( (x, y) \) is the first part of the codeword, and the symbol filling the cell of \( (x, y) \) in the coding table of \( Q_{XY} \) is determined as the second part of the codeword. For the sequence pairs \( (x, y) \) whose joint type \( Q_{XY} \) does not belong to \( S_n(R) \), the corresponding codeword is determined arbitrarily and an encoding error is declared.

[Decoding: \( \hat{\varphi}^{(n)}_{(1)} \)] (Almost the same as for \( \hat{\varphi}^{(n)}_{(2)} \))

1. Find the coding table of the type \( \hat{Q}_{XY} \) that corresponds to the first part of the received codeword. The decoder can find the coding table used in the encoding scheme if no encoding error occurs. In this case, \( \hat{Q}_{XY} \) should be \( Q_{XY} \).

2. Find the cell filled with the second part of the received codeword from the column of the side information sequence \( y \in T^{n}_{Q_Y} \). The sequence \( \tilde{x} \in T^{n}_{Q_X} \) that corresponds to the row of the cell found in this step is reproduced.
First, we show the existence of such coding tables. To this end, we introduce the following two lemmas.

**Lemma 4.** For a given coding table of a joint type \( Q_{XY} \in \mathcal{P}_n(X \times Y) \), the number of marked cells in every row of the coding table, \( N_y(Q_{XY}) \), is a constant value that is less than \( \exp(nR) \), and the number of marked cells in every column of the coding table, \( N_x(Q_{XY}) \), is also a constant value of less than \( \exp(nR) \), both of which depend solely on the joint type \( Q_{XY} \).

**Proof.** Note that the number of marked cells in each row equals the cardinality of the V-shell \( T^n_V(x) \) for the sequence \( x \in T^n_QX \) that corresponds to the row. The cardinality of V-shells \( T^n_V(x) \) is constant for a given joint type \( Q_{XY} \) and any sequences \( x \in T^n_QX \), this cardinality is bounded as follows:

\[
N_y(Q_{XY}) = |T^n_V(x)| \leq \exp\{nH(V|Q_X)\} \quad (\because \text{Lemma 2}) \\
\leq \exp(nR).
\]

In the same way, the number of marked cells in each column equals the cardinality of the V-shell \( T^n_W(y) \) for the sequence \( y \in T^n_QY \) that corresponds to the column, and therefore it can be bounded as

\[
N_y(Q_{XY}) = |T^n_W(y)| \leq \exp(nR).
\]

This concludes the proof of Lemma 4.

**Lemma 5.** For given integers \( m_x, m_y, n_x \) and \( n_y \) that satisfy \( m_x \geq n_x \) and \( m_y \geq n_y \), there exists an \( m_x \times m_y \) table filled with \( \max(n_x, n_y) \) different symbols such that

- at most \( n_y \) cells are filled with a certain symbol for each row (blank cells are possible),
- at most \( n_x \) cells are filled with a certain symbol for each column (blank cells are possible),
- each symbol occurs at most once in each row and at most once in each column.
Proof. The table mentioned in this lemma is equivalent to a bipartite graph such that

- each node in one set corresponds to a row in the table, and each node in the other set corresponds to a column in the table,
- each edge corresponds to a cell in the table, to which a certain symbol is assigned,
- \( \max(n_x, n_y) \) different colors are given to edges, each of which corresponds to a symbol in the table,
- no two edges with the same color share a common node.

Figure 3 shows an example of such a graph. Here, let us introduce the following lemma for bipartite graphs:

Lemma 6. (König [15, 16]) If a graph \( G \) is bipartite, the minimum number of colors necessary for edge coloring of the graph \( G \) equals the maximum degree of \( G \).

Lemma 6 ensures the existence of the above bipartite graph. This concludes the proof of Lemma 5.

From Lemmas 4 and 5 we can easily show the existence of coding tables by setting \( m_x = |T_{QX}| \), \( m_y = |T_{QY}| \), \( n_x = |T_{YX}(y)| \) and \( n_y = |T_{XY}(x)| \) in Lemma 5.

5 Coding theorems

We can obtain the following theorem for the universal FF-CD codes constructed in Section 4. The proof is similar to that of the theorem of universal coding for a single source.
Theorem 2. For a given real number $R > 0$, there exists a sequence of universal FF-CD codes with parameters $\{(n, M_n, e_n^{(X)}, e_n^{(Y)})\}_{n=1}^{\infty}$ such that for any integer $n \geq 1$ and any source $(X, Y)$ with a generic distribution $P_{XY} \in M(X \times Y)$

\[
\frac{1}{n} \log M_n \leq R + \frac{1}{n} |X \times Y| \log(n + 1),
\]

\[
e_n^{(X)} + e_n^{(Y)} \leq 2(n + 1)^{|X\times Y|} \times \exp \left\{ -n \min_{Q_{XY} \in \mathcal{S}_n(R)} D(Q_{XY} \| P_{XY}) \right\},
\]

where $\mathcal{S}_n(R) = \mathcal{P}_n(X \times Y) - \mathcal{S}_n(R)$.

Proof. Lemmas 4 and 5 ensure the existence of a coding table for every joint type $Q_{XY} \in \mathcal{P}(X \times Y)$. From the coding scheme, the size of the codeword set is bounded as

\[
M_n \leq |\mathcal{P}_n(X \times Y)| \exp(nR) \\
\leq (n + 1)^{|X\times Y|} \exp(nR), \quad (\because \text{Lemma 1})
\]

which implies the first inequality of Theorem 2. Next, we evaluate decoding error probabilities. Since every sequence pair $(x^n, y^n) \in T^n_{Q_{XY}}$ that satisfies $Q_{XY} \in \mathcal{S}_n(R)$ is reproduced correctly at the decoder, the sum of error probabilities is bounded as

\[
e_n^{(X)} + e_n^{(Y)} \leq 2 \Pr \{ \exists Q_{XY} \in \mathcal{S}_n(R), (X^n, Y^n) \in T^n_{Q_{XY}} \}
\]

\[
\leq 2 \sum_{Q_{XY} \in \mathcal{S}_n(R)} \exp\left\{ -n D(Q_{XY} \| P_{XY}) \right\} \\
(\because \text{Lemma 3})
\]

\[
\leq 2 \sum_{Q_{XY} \in \mathcal{S}_n(R)} \exp\left\{ -n \min_{Q_{XY} \in \mathcal{S}_n(R)} D(Q_{XY} \| P_{XY}) \right\} \\
\leq 2(n + 1)^{|X\times Y|} \times \exp\left\{ -n \min_{Q_{XY} \in \mathcal{S}_n(R)} D(Q_{XY} \| P_{XY}) \right\} \\
(\because \text{Lemma 1})
\]

This completes the proof of Theorem 2. \qed

We can see that for any real value $R \geq R_f(X, Y)$ we have

\[
\min_{Q_{XY} \in \mathcal{S}_n(R)} D(Q_{XY} \| P_{XY}) > 0.
\]

This implies that any real value $R \geq R_f(X, Y)$ is a universal FF-CD achievable rate of $(X, Y)$, namely, there exists a sequence of universal FF-CD codes with parameters $\{(n, M_n, e_n^{(X)}, e_n^{(Y)})\}_{n=1}^{\infty}$ such that

\[
\limsup_{n \to \infty} \frac{1}{n} \log M_n \leq R,
\]
The following converse theorem indicates that the error exponent obtained in Theorem 2 is tight.

**Theorem 3.** Any sequence of FF-CD codes with parameters \( \{(n, M_n, e^{(X)}_n, e^{(Y)}_n)\}_{n=1}^\infty \) for the source \((X, Y)\) must satisfy

\[
e^{(X)}_n + e^{(Y)}_n \geq \frac{1}{2} (n + 1)^{|X \times Y|} \times \exp \left\{ -n \min_{Q_{XY} \in \mathcal{S}_n(R + \epsilon_n)} D(Q_{XY} || P_{XY}) \right\}
\]

for any integer \( n \geq 1 \) and a given coding rate \( R = 1/n \log M_n > 0 \), where

\[
\epsilon_n \overset{\text{def.}}{=} \frac{1}{n} \left( |X \times Y| \log(n + 1) + 1 \right) \rightarrow 0 \quad (n \to \infty).
\]

**Proof.** Note that the number of sequences to be decoded correctly for each decoder are at most \( \exp(nR) \). Here, let us consider a joint type \( Q_{XY} \in \mathcal{S}_n(R + \epsilon_n) \). Lemma 2 and the definition of \( \mathcal{S}_n(R + \epsilon_n) \) imply that for \((x, y) \in T^n_{Q_{XY}}\) we have

\[
\max\{|T^n_V(x)|, |T^n_W(y)|\} \geq (n + 1)^{|X \times Y|} \times \max\{\exp(nH(V|Q_X)), \exp(nH(W|Q_Y))\} \geq (n + 1)^{|X \times Y|} \exp(n(R + \epsilon_n)) = 2 \exp(nR).
\]

Therefore, at least half of the sequences in the V-shell \( T^n_V(x) \) will not be decoded correctly at the decoder \( \varphi^n_{(2)} \), or at least half of sequences in the V-shell \( T^n_W(y) \) will not be decoded correctly at the decoder \( \varphi^n_{(1)} \). Thus, the sum of error probabilities is bounded as

\[
e^{(X)}_n + e^{(Y)}_n \geq \frac{1}{2} \sum_{Q_{XY} \in \mathcal{S}_n(R + \epsilon_n)} \Pr\{ (X^n, Y^n) \in T^n_{Q_{XY}} \} \\
\geq \frac{1}{2} (n + 1)^{|X \times Y|} \times \sum_{Q_{XY} \in \mathcal{S}_n(R + \epsilon_n)} \exp\{-nD(Q_{XY} || P_{XY})\}
\]

\[
\left( \because \text{Lemma 3} \right) \\
\geq \frac{1}{2} (n + 1)^{|X \times Y|} \times \exp \left\{ -n \min_{Q_{XY} \in \mathcal{S}_n(R + \epsilon_n)} D(Q_{XY} || P_{XY}) \right\}
\]

This concludes the proof of Theorem 3. \( \square \)
The following corollary is directly derived from Theorems 2 and 3.

**Corollary 1.** For a given real number $R > 0$, there exists a sequence of universal FF-CD codes with parameters $\{(n, M_n, e^{(X)}_n, e^{(Y)}_n)\}_{n=1}^{\infty}$ such that for any source $(X,Y)$

\[
\limsup_{n \to \infty} \frac{1}{n} \log M_n \leq R,
\]

\[
\lim_{n \to \infty} \frac{1}{n} \log (e^{(X)}_n + e^{(Y)}_n) = \min_{Q_{XY} \in \mathcal{S}(R)} D(Q_{XY} \parallel P_{XY}),
\]

where

\[
\mathcal{S}(R) = \{Q_{XY} \in \mathcal{M}(X \times Y) : \\
\qquad \max\{H(V|Q_X), H(W|Q_Y)\} \leq R, \quad Q_{XY} = Q_XV = Q_YW, \quad V \in \mathcal{M}(Y|Q_X), W \in \mathcal{M}(X|Q_Y)\},
\]

and $\mathcal{S}(R) = \mathcal{M}(X \times Y) - \mathcal{S}(R)$.

In a similar manner, we can investigate the probability such that the original sequence pair is correctly reproduced. The following theorem shows the lower bound of the probability that can be attained by the proposed coding scheme.

**Theorem 4.** For a given real number $R > 0$, there exists a universal lossless f-FCD code $\{(\varphi_n, \hat{\varphi}^{(1)}_n, \hat{\varphi}^{(2)}_n)\}_{n=1}^{\infty}$ such that for any integer $n \geq 1$ and any DMS $(X,Y)$

\[
\frac{1}{n} \log M_n \leq R + \epsilon_n,
\]

\[
1 - (e^{(1)}_n + e^{(2)}_n) \geq \exp \left\{ -n \left( \epsilon_n + \min_{Q_{XY} \in \mathcal{T}_n(R)} D(Q_{XY} \parallel P_{XY}) \right) \right\}.
\]

**Proof.** The first inequality is derived in the same way as the proof of Theorem 2. Next, we evaluate the probability such that the original sequence pair is correctly reproduced. Since every sequence pair $(x^n, y^n) \in T^n_{Q_{XY}}$ that satisfies $Q_{XY} \in \mathcal{T}_n(R)$ is reproduced correctly at the decoder, the sum of the probabilities is bounded as

\[
1 - (e^{(1)}_n + e^{(2)}_n) \geq \Pr \{ (X^n, Y^n) \in T^n_{Q_{XY}} : Q_{XY} \in \mathcal{T}_n(R) \}
\]

\[
\geq \sum_{Q_{XY} \in \mathcal{T}^n(R)} (n + 1)^{-|X \times Y|} \exp \left\{ -n D(Q_{XY} \parallel P_{XY}) \right\}
\]

\[
\geq (n + 1)^{-|X \times Y|} \exp \left\{ -n \min_{Q_{XY} \in \mathcal{T}_n(R)} D(Q_{XY} \parallel P_{XY}) \right\}
\]

\[
= \exp \left\{ -n \left( \epsilon_n + \min_{Q_{XY} \in \mathcal{T}_n(R)} D(Q_{XY} \parallel P_{XY}) \right) \right\},
\]

where Eq. (1) comes from Lemma 3. This completes the proof of Theorem 4. □
Thus, the sum of the probabilities is obtained as

\[1 - (\epsilon_n^{(1)} + \epsilon_n^{(2)})\]

where Eq. (3) comes from Lemma 2. Therefore, the probability

\[P_{\text{c}}(Q_{XY}) \leq \exp \left\{ n \left( -\epsilon_n + \min_{Q_{XY} \in \mathcal{P}_{n}(X \times Y)} |\max(1/H(\hat{V}|Q_X), H(W|Q_Y)) - (R + \epsilon_n)|^+ + D(Q_{XY}||P_{XY}) \right) \right\}\]

for any integer \(n \geq 1\) and a given coding rate \(R = 1/n \log M_n > 0\).

**Proof.** Note that the number of sequences to be decoded correctly for each decoder are at most \(\exp(nR)\). Here, let us consider a joint type \(Q_{XY} \in \mathcal{P}_n(X \times Y)\) such that \(Q_{XY} = Q_X V = Q_Y W, V \in \mathcal{M}(Y|Q_X)\) and \(W \in \mathcal{M}(X|Q_Y)\). The ratio \(r_c(Q_{XY})\) of sequences in the sequence set \(T_{Q_{XY}}\) such that the sequences are correctly reproduced is at most

\[r_c(Q_{XY}) \leq \min \left\{ \min \left( \frac{\exp(nR)}{|T_X^n(x)|}, \frac{\exp(nR)}{|T_Y^n(y)|} \right), 1 \right\}\]

\[\leq \min \left[ \exp(nR) \cdot (n + 1)^{|X \times Y|}, \exp\left\{ -n \max(1/H(\hat{V}|Q_X), H(W|Q_Y)) \right\}, 1 \right]\]

\[= \exp \left\{ -n \left| \max(1/H(\hat{V}|Q_X), H(W|Q_Y)) - (R + \epsilon_n) \right|^+ \right\}\]

where Eq. (3) comes from Lemma 2. Therefore, the probability \(P_{\text{c}}(Q_{XY})\) such that the original sequence pair with type \(Q_{XY}\) is correctly reproduced is bounded as

\[P_{\text{c}}(Q_{XY}) \leq r_c(Q_{XY}) \Pr \{(X^n, Y^n) \in T_{Q_{XY}}^n\}

\leq \exp \left\{ -n \left| \max(1/H(\hat{V}|Q_X), H(W|Q_Y)) - (R + \epsilon_n) \right|^+ + D(Q_{XY}||P_{XY}) \right\}.\]

Thus, the sum of the probabilities is obtained as

\[1 - (\epsilon_n^{(1)} + \epsilon_n^{(2)}) \leq \sum_{Q_{XY} \in \mathcal{P}_n(X \times Y)} P_{\text{c}}(Q_{XY})\]

\[\leq \exp \left\{ -n \left| \max(1/H(\hat{V}|Q_X), H(W|Q_Y)) - (R + \epsilon_n) \right|^+ + D(Q_{XY}||P_{XY}) \right\}\]

\[\leq \exp \left\{ -n \min_{Q_{XY} \in \mathcal{P}_n(X \times Y)} \left( \left| \max(1/H(\hat{V}|Q_X), H(W|Q_Y)) - (R + \epsilon_n) \right|^+ + D(Q_{XY}||P_{XY}) \right) \right\}.

This completes the proof of Theorem 5. \qed
We can see that for any real value \( R \geq R_f(X, Y) \) we have
\[
\min_{Q_{XY} \in \mathcal{M}(X \times Y)} \left[ \left| \max \{ H(V|Q_X), H(W|Q_Y) \} - R \right|^+ + D(Q_{XY}\|P_{XY}) \right] \\
= \left| \max \{ H(P_V|X), H(P_W|Y) \} - R \right|^+ \\
= 0.
\]
On the other hand, for any real value \( R < R_f(X, Y) \) we have
\[
\min_{Q_{XY} \in \mathcal{T}(R)} \left[ \left| \max \{ H(V|Q_X), H(W|Q_Y) \} - R \right|^+ + D(Q_{XY}\|P_{XY}) \right] \\
\geq 0.
\]
This implies that the error exponent obtained in Theorem 4 might not be tight.

6 Variable-length coding

This section discusses variable-length coding for the complementary delivery coding system, and shows an explicit construction of universal variable-length codes. The coding scheme is similar to that of fixed-length codes, and also utilizes the coding tables defined in Section 4.

6.1 Formulation

**Definition 4.** (Fixed-to-variable complementary delivery (FV-CD) code) A set \( (\varphi^n, \widehat{\varphi}_n^{(1)}, \widehat{\varphi}_n^{(2)}) \) of an encoder and two decoders is an FV-CD code for the source \( (X, Y) \) if and only if
\[
\varphi^n : X^n \times Y^n \to B^n \\
\widehat{\varphi}_n^{(1)} : \varphi^n(X^n, Y^n) \times Y^n \to X^n, \\
\widehat{\varphi}_n^{(2)} : \varphi^n(X^n, Y^n) \times X^n \to Y^n, \\
e_n(X) = \Pr \{ X^n \neq \hat{X}^n \} = 0, \\
e_n(Y) = \Pr \{ Y^n \neq \hat{Y}^n \} = 0,
\]
where
\[
\hat{X}^n \overset{\text{def}}{=} \widehat{\varphi}_n^{(1)}(\varphi^n(X^n, Y^n), Y^n), \\
\hat{Y}^n \overset{\text{def}}{=} \widehat{\varphi}_n^{(2)}(\varphi^n(X^n, Y^n), X^n),
\]
and the image of \( \varphi^n \) is a prefix set.

**Definition 5.** (FV-CD-achievable rate) \( R \) is an FV-CD-achievable rate of the source \( (X, Y) \) if and only if there exists a sequence of FV-CD codes \( \{ (\varphi^n, \widehat{\varphi}_n^{(1)}, \widehat{\varphi}_n^{(2)}) \} \) for the source \( (X, Y) \) such that
\[
\limsup_{n \to \infty} \frac{1}{n} \mathbb{E} [l(\varphi^n(X^n, Y^n))] \leq R,
\]
where \( l(\cdot) : B^n \to \mathbb{R} \) is a length function.
Definition 6. (Inf FV-CD-achievable rate)

\[ R_v(X, Y) = \inf \{ R : R \text{ is an FV-CD-achievable rate of } (X, Y) \} \]

6.2 Code construction

We can construct universal FV-CD codes in a similar manner to universal FF-CD codes. Note that the coding rate depends on the type of sequence pair to be encoded, whereas the coding rate is fixed beforehand for fixed-length coding. The coding scheme is described as follows:

[Encoding]
1. Create a coding table for each joint type \( Q_{XY} \in \mathcal{P}_n(X \times Y) \) in the same way as Step 2 of Section 4.
2. Mark cells that correspond to sequence pairs \((x, y) \in T_{Q_{XY}}\).
3. Fill the marked cells on the coding table with different max \( \{ |T^x_v(x)|, |T^y_w(y)| \} \) symbols such that each symbol occurs at most once in each row and at most once in each column, where \( x \in T^n_{Q_X}, y \in T^n_{Q_Y} \).
4. For a given pair of sequences \((x, y) \in X^n \times Y^n\), the number (index) assigned to the joint type \( Q_{XY} \) of \((x, y)\) is the first part of the codeword, and the symbol filling the cell of \((x, y)\) in the coding table of \( Q_{XY} \) is determined as the second part of the codeword.

[Decoding]
Decoding can be accomplished in almost the same way as the fixed-length coding. Note that the decoder can always find the coding table used in the encoding scheme.

6.3 Coding theorems

We begin by showing a theorem for (non-universal) variable-length coding, which indicates that the inf coding rate of variable-length coding is the same as that of fixed-length coding.

Theorem 6. (Coding theorem of FV-CD code)

\[ R_v(X, Y) = R_f(X, Y) = \max \{ H(X|Y), H(Y|X) \} \]

Proof. See Appendix.

The following direct theorem for universal coding indicates that the coding scheme presented in the previous subsection can achieve the inf achievable rate clarified in Theorem 6.

Theorem 7. There exists a sequence of universal FV-CD codes \( \{ (\varphi^n, \bar{\varphi}^{n(1)}, \bar{\varphi}^{n(2)}) \}_{n=1}^\infty \) such that for any integer \( n \geq 1 \) and any source \((X, Y)\), the overflow probability, namely the probability that the length of a codeword exceeds a given real number \( R > 0 \), is bounded as

\[ \overline{\rho}_n(R) \overset{\text{def.}}{=} \Pr \{ l(\varphi^n(X^n, Y^n)) > n(R + \epsilon_n) \} \]
\[ \leq (n + 1)^{|X \times Y|} \times \exp \left\{ -n \min_{Q_{XY} \in S_n(R)} D(Q_{XY} || P_{XY}) \right\}, \]

where \( \epsilon_n \) is defined in Theorem 3. This implies that there exists a sequence of universal FV-CD codes \( \{(\varphi^n, \widehat{\varphi}^{(1)}_n, \widehat{\varphi}^{(2)}_n)\}_{n=1}^\infty \) that satisfies

\[ \limsup_{n \to \infty} \frac{1}{n} l(\varphi^n(X^n, Y^n)) \leq R_v(X, Y) \text{ a.s.} \quad (4) \]

**Proof.** The overflow probability can be obtained in the same way as an upper-bound of the error probability of FF-CD codes, which has been shown in the proof of Theorem 2. Thus, from Theorem 6 we have

\[ \sum_{n=1}^\infty \Pr \left\{ \frac{1}{n} l(\varphi^n(X^n, Y^n)) > R_v(X, Y) + \delta \right\} < \infty \]

for a given \( \delta > 0 \). From Borel-Cantelli’s lemma [17, Lemma 4.6.3], we immediately obtain Eq.(4). This completes the proof of Theorem 7.

The following converse theorem for variable-length coding indicates that the exponent of the overflow probability obtained in Theorem 7 is tight. This can be easily obtained in almost the same way as Theorem 3.

**Theorem 8.** Any sequence of FV-CD codes \( \{(\varphi^n, \widehat{\varphi}^{(1)}_n, \widehat{\varphi}^{(2)}_n)\}_{n=1}^\infty \) for the source \((X, Y)\) must satisfy

\[ \rho_n(R) \geq (n + 1)^{-|X \times Y|} \times \exp \left\{ -n \min_{Q_{XY} \in S_n(R + \epsilon_n)} D(Q_{XY} || P_{XY}) \right\}. \]

for a given real number \( R > 0 \) and any integer \( n \geq 1 \), where \( \epsilon_n \) is defined in Theorem 3.

The following corollary is directly derived from Theorems 7 and 8.

**Corollary 2.** There exists a sequence of universal FV-CD codes \( \{(\varphi^n, \widehat{\varphi}^{(1)}_n, \widehat{\varphi}^{(2)}_n)\}_{n=1}^\infty \) such that for any source \((X, Y)\)

\[ \limsup_{n \to \infty} \frac{1}{n} l(\varphi^n(X^n, Y^n)) \leq R_v(X, Y) \text{ a.s.} \]

\[ \lim_{n \to \infty} -\frac{1}{n} \log \rho_n(R) = \min_{Q_{XY} \in S(R)} D(Q_{XY} || P_{XY}) \]

Next, we investigate the underflow probability, namely the probability that the length of a codeword falls below a given real number \( R > 0 \). For this purpose, we present the following two theorems. The proofs are almost the same as those of Theorems 4 and 5.
Theorem 9. There exists a universal lossless v-FCD code \(\{(\varphi_n, \tilde{\varphi}_n^{(1)}, \tilde{\varphi}_n^{(2)})\}_{n=1}^{\infty}\) such that for any integer \(n \geq 1\) and any DMS \((X, Y)\), the underflow probability \(\rho_n(R)\) is bounded as

\[
\rho_n(R) \overset{\text{def.}}{=} \Pr \{l(\varphi_n(X^n, Y^n)) < nR\} \leq \exp\left\{ -n \left( \epsilon_n + \min_{Q_{XY} \in \mathcal{T}_n(R - \epsilon_n(2))} D(Q_{XY} \| P_{XY}) \right) \right\}.
\]

This implies that there exists a universal lossless v-FCD code \(\{(\varphi_n, \tilde{\varphi}_n^{(1)}, \tilde{\varphi}_n^{(2)})\}_{n=1}^{\infty}\) that satisfies

\[
\liminf_{n \to \infty} \frac{1}{n} l(\varphi_n(X^n, Y^n)) \geq R_v(X, Y) \quad \text{a.s.} \quad (5)
\]

Theorem 10. Any lossless v-FCD code \(\{(\varphi_n, \tilde{\varphi}_n^{(1)}, \tilde{\varphi}_n^{(2)})\}_{n=1}^{\infty}\) for the DMS \((X, Y)\) must satisfy

\[
\rho_n(R) \leq \exp\left\{ -n \left( -\epsilon_n(1) + \min_{Q_{XY} \in \mathcal{T}_n(R + \epsilon_n)} \max(H(V|Q_X), H(W|Q_Y)) - (R + \epsilon_n)^+ + D(Q_{XY} \| P_{XY}) \right) \right\}
\]

for a given real number \(R > 0\) and any integer \(n \geq 1\).

7 Concluding remarks

We investigated a universal coding problem for the complementary delivery coding system. First, we presented an explicit construction of universal fixed-length codes, which was based on a graph-theoretical technique. We clarified that the error exponent achieved by the proposed coding scheme is asymptotically optimal. Next, we applied the coding scheme to construction of universal variable-length codes. We clarified that there exists a universal code such that the codeword length converges to the minimum achievable rate almost surely, and that the exponent of the overflow probability achieved by the proposed coding scheme is optimal. This paper dealt with only the lossless configuration, and therefore constructing universal lossy codes for the complementary delivery coding system still remains as an open problem.

Acknowledgements

The authors would like to thank Dr. Jun Muramatsu of NTT Communication Science Laboratories, and Prof. Ryutaroh Matsumoto of Tokyo Institute of Technology for their valuable discussions and helpful comments, which led to improvements in this work. The authors also thank Drs. Yoshinobu Tonomura, Hiromi Nakaiwa, Tatoto Takeuchi, Shoji Makino and Junji Yamato of NTT Communication Science Laboratories for their help.
A Proof of Theorem 6

A.1 Direct part

Proof. We can apply a sequence of achievable FF-CD codes (fixed-length codes). The encoder $\varphi^n$ assigns the same codeword as that of the fixed-length code to a sequence pair $(x, y) \in X^n \times Y^n$ that is correctly reproduced by the fixed-length code. Otherwise, the encoder sends the sequence pair itself as a codeword.

The above FV-CD code can always reproduce the original sequence pair at the decoders, and it attains the desired coding rate.

A.2 Converse part

Proof. We can prove the converse part in a similar manner to that for fixed-length coding. Let a sequence $\{ (\varphi^n, \hat{\varphi}^{(1)}_n, \hat{\varphi}^{(2)}_n) \}_{n=1}^\infty$ of FV-CD codes be given that satisfies the conditions of Definitions 4 and 5. From Definition 5, for any $\delta > 0$ there exists an integer $n_1 = n_1(\delta)$ and then for all $n \geq n_1(\delta)$, we can obtain

$$\frac{1}{n} E[l(\varphi^n(X^n, Y^n))] \leq R + \delta. \quad (6)$$

Here, let us define $A_n = \varphi^n(X^n, Y^n)$. Since the decoder $\hat{\varphi}^{(1)}_n$ can always reproduce the original sequence $X^n$ from the received codeword $A_n$ and side information $Y^n$, we can see that

$$H(X^n | A_n Y^n) = 0. \quad (7)$$

Similarly, we can obtain

$$H(Y^n | A_n X^n) = 0.$$ 

Substituting $A_n$ into Eq.(6), we have

$$n(R + \delta) \geq E[l(A_n)] \geq H(A_n) \quad (\because A_n \text{ is a prefix set}) \geq H(A_n | Y^n) \geq I(X^n; A_n | Y^n) = H(X^n | Y^n). \quad (\because \text{Eq.}(7))$$

Since we can select an arbitrarily small $\delta > 0$ for a sufficient large $n$, we can obtain

$$R \geq \frac{1}{n} H(X^n | Y^n) = H(X | Y).$$

In the same way, we also obtain

$$R \geq H(Y | X).$$
References

[1] D. Slepian and J.K. Wolf, “Noiseless coding of correlated information sources,” IEEE Trans. Inf. Theory, Vol.19, No.4, pp.471–480, July 1973.

[2] A.D. Wyner, “On source coding with side information at the decoder,” IEEE Trans. Inf. Theory, Vol.21, No.3, pp.294–300, May 1975.

[3] J. Körner and K. Marton, “Images of a set via two channels and their role in multi-user communication,” IEEE Trans. Inf. Theory, Vol.23, No.6, pp.751–761, November 1975.

[4] A. Sgarro, “Source coding with side information at several decoders,” IEEE Trans. Inf. Theory, Vol.23, No.2, pp.179–182, March 1977.

[5] I. Csiszár and J. Körner, “Towards a general theory of source networks,” IEEE Trans. Inf. Theory, Vol.26, No.2, pp.155–165, March 1980.

[6] I. Csiszár, “Linear codes for source and source networks: Error exponents, universal coding,” IEEE Trans. Inf. Theory, Vol.28, No.4, pp.585–592, July 1982.

[7] Y. Oohama and T.S. Han, “Universal coding for the Slepian-Wolf data compression system and the strong converse theorem,” IEEE Trans. Inf. Theory, Vol.40, No.6, pp.1908–1919, November 1994.

[8] T. Uyematsu, “An algebraic construction of codes for Slepian-Wolf source networks,” IEEE Trans. Inf. Theory, Vol.47, No.7, pp.3082–3088, November 2001.

[9] J. Muramatsu, Universal data compression algorithms for stationary Ergodic sources based on the complexity of sequences, Ph.D. thesis, Nagoya University, March 1998.

[10] A.D. Wyner and J. Ziv, “The rate-distortion function for source coding with side information at the decoder,” IEEE Trans. Inf. Theory, Vol.22, No.1, pp.1–10, January 1976.

[11] A. Kimura and T. Uyematsu, “Multiterminal source coding with complementary delivery,” Proc. International Symposium on Information Theory and its Applications (ISITA), pp.189–194, October 2006.

[12] I. Csiszár and J. Körner, Information theory: Coding theorems for discrete memoryless systems, Academic Press, New York, 1981.

[13] F.M.J. Willems, J.K. Wolf, and A.D. Wyner, “Communicating via a processing broadcast satellite,” Proc. of the 1989 IEEE/CAM Information Theory Workshop, June 1989.

[14] A.D. Wyner, J.K. Wolf, and F.M.J. Willems, “Communicating via a processing broadcast satellite,” IEEE Trans. Inf. Theory, Vol.48, No.6, pp.1243–1249, June 2002.

[15] D. König, “Graphok és alkalmazásuk a determinánsok és a halmazok elméletére,” Mathematikai és Természettudományi Értesítő, Vol.34, pp.104–119, 1916. (in Hungarian).
[16] N.L. Biggs, E.K. Lloyd, and R.J. Wilson, Graph Theory, Oxford University Press, 1976.

[17] R.M. Gray, Probability, Random Processes, Ergodic Properties, Springer-Verlag, New York, 1988.