A single electron transistor (SET) is a high-sensitivity electrometer that measures a small fraction of the elementary charge, $e$, on a small island \([1]\). If the island, or quantum dot atom, shows well-defined eigenstates in a zero-dimensional confinement potential, quantum mechanical effects found in atomic physics can be reproduced with tunable parameters \([2]\). The ability to control single-particle quantum states will be helpful for developing quantum logic gates \([3, 4]\). So the dynamical behavior of charge states need to be studied for those interested. Most of the experiments have been done by taking the dc response from an ac modulation. However, direct measurement of the ac response of a single-particle state has not been done due to technical difficulties. Recently, a radio-frequency (rf) operated SET technique for following the fast response of the charge has been proposed and demonstrated by R. J. Schoelkopf et al. \([5]\). A bandwidth greater than 100 MHz would be useful for studying single-electron dynamics such as single-electron tunneling oscillations \([6]\) and coherent charge oscillation \([7]\), as well as for various sensors \([8]\).

In this letter, we describe an application of the rf-SET for detecting individual emission and capture events of a trap in a semiconductor. A modified rf-SET technique that measures the transmission of rf signals through a resonator, is also used for a quantum dot fabricated in an AlGaAs/GaAs heterostructure \([9]\). The charge noise of the quantum dot is studied both for low-frequency \(1/f\) noise and for random telegraph signals (RTSs) originating from a trap near the dot. For a specific RTS, the statistics of the electron capture and emission are given by the Poisson process.

The transmission-type rf-SET technique is shown schematically in Fig. 1(a). The SET, or quantum dot, is fabricated in an AlGaAs/GaAs modulation doped heterostructure using focused ion beam implantation and patterning of fine Schottky gates \([10]\). The two gate voltages, $V_L$ and $V_R$, control the two tunneling barriers independently, and effectively lift the electrostatic potential of the dot. The SET is placed in an LC resonator (two inductors of 2\(L\) and one capacitor of $C$). Other lumped elements allow measurement of the dc current and rf-transmission simultaneously. When the rf carrier signal $V_c e^{i \omega t}$ is supplied at the resonant frequency, $\omega = \frac{1}{\sqrt{L C}}$, to the resonator, a transmitted signal $V_t e^{i \omega t}$ appears at the other end of the resonator. The transmitted signal is given by $R(q)$, of the SET \([11]\), or the charge on the island, $q$, i.e., $V_t/V_c = -(1 - 4q^2 Z_0/R(q))$ for $R(q) \gg 4q^2 Z_0$ and $Q \gg 1$, where $Z_0 = 50 \, \Omega$ is the cable impedance and $Q$ is the quality factor of the resonator. The rf excitation voltage across the SET, $v_{exc}$, is given by $v_{exc} = Q V_c$.

Compared with the reflection-type rf-SET originally demonstrated in Ref. 6, the transmission-type rf-SET has some advantages for convenient measurements. Since the incident and the transmitted signals are separated, a directional coupler necessary for the reflection measurement is not required (although two inductors are necessary for the transmission measurement). The transmission signal always shows a clear and sharp resonance signal when the frequency is swept, while the reflection signal does not particularly if the sample is highly resistive or Coulomb blockaded. Once the frequency is adjusted properly, both transmitted and reflected signals behave similarly if the sample resistance is high, $R(q) \gg 4q^2 Z_0$. High-sensitivity \([12]\) and high-frequency (ideally upto $\omega/Q$) operation of the rf-SET could also be applied to the transmission-type rf-SET.

Figure 1(b) shows typical Coulomb blockade oscillations measured by dc current (upper trace) and rf-transmission (lower trace). The traces are taken simultaneously with a dc bias voltage of $V_{SD} = 0.2 \, \text{mV}$ and an rf excitation of $v_{exc} \sim 0.1 \, \text{mV}$ (rms for all ac amplitudes). The best charge sensitivity is obtained at the largest slope, $|dV_i/dV_c|$, denoted by the arrow labelled by $\alpha$ in the figure. We estimate the charge sensitivity, i.e., the minimal detectable charge on the island, by applying a sinusoidal modulation to the $V_L$ as follows. The
inset of Fig. 2 shows the frequency spectrum of transmitted signal $V_t$ at the modulation frequency, $f_{mod} = 10$ kHz. The signal intensity is linearly dependent on the modulation amplitude, $V_{mod}$, as shown in Fig. 2. We obtain a charge sensitivity of $\delta q = 3.6 \times 10^{-5} e/\sqrt{Hz}$ at an excitation of $v_{ex} \sim 0.4$ mV. This is larger than the thermal noise and the shot noise, but is dominated by the noise of the amplifier.

Figure 3(a) shows the spectra of the transmitted signal. The thick trace measured at the best charge sensitivity ($\alpha$ in Fig. 1(a)) shows the SET noise, while the thin trace measured at zero conductance of the SET shows the noise of the measurement system. The SET shows $1/f$ noise below 1 kHz, similar to that reported in metal SETs [6,13]. For frequencies above 1 kHz, the white noise is dominated by the HEMT amplifier. Better amplifiers and a higher resonator $Q$ should reduce the system noise.

We also find a RTS in our dot. The specific trap we study has a relatively fast switching time and is a good example for demonstrating the fast response of the rf-SET. Figure 1(c) shows another CB oscillation with a jump seen at about -710 mV, labelled $\beta$ in the figure, where the jump appears at the largest $|dV_t/dV_g|$ of the CB peak for the best sensitivity. The jump is caused by the emission and capture of an electron from a trap in the vicinity of the quantum dot. The trap may be a defect in the crystal or a potential hollow (maybe a quantum dot) unintentionally created during fabrication. This specific trap is located close to the left gate rather than the right gate (schematically shown in the left inset of Fig. 4(a)).

The RTS noise is observed on a relatively short time scale for this particular trap, as shown in Fig. 4(a). The emission and capture statistics are derived from the RTS noise in the time domain. The duration time for the trap being empty (lower signal in the figure) corresponds to the capture rate, and the duration time for the trap being occupied (higher signal) corresponds to the emission rate. These durations are widely distributed, as shown in Fig. 4(b), and well characterized by an exponential dependence, $\exp(-t/\tau)$, with time constant $\tau$. The mean ($\mu$) and the standard deviation ($\sigma$) of the durations are practically the same as $\tau$. These observations indicate a Poisson process, i.e., each emission and capture process is an independent event. The time constant for capture, $\tau_c$, and emission, $\tau_e$, are strongly dependent on the energy level of the trap, $E_t$, or on the gate voltage $V_g$ (see Fig. 4(c)). Simply assuming a constant attempt rate for both transitions, $\tau_0^{-1}$, and a Fermi distribution, $f_{FD}(E) \equiv (1 + \exp(E/k_BT_{eff}))^{-1}$, in the source electrode, the capture and emission rates should be given by $\tau_c^{-1} = \tau_0^{-1} f_{FD}(E_t - E_F)$ and $\tau_e^{-1} = \tau_0^{-1} (1 - f_{FD}(E_t - E_F))$, respectively. The measured time-constants are reproduced with parameters $\tau_0 = 12 \mu s$ and $k_BT_{eff} = 0.46 \text{ meV}$, as shown by the solid lines in Fig. 4(c). The effective temperature, $T_{eff}$, is comparable to the rf excitation voltage across the SET, $v_{ex} \sim 0.4$ mV. We find that $T_{eff}$ increases with $v_{ex}$.

The power spectrum of the RTS noise (Fig. 3(b)) is flat up to a few kHz and decreases above 5 kHz. The spectrum can be fitted with a Lorentzian curve (dashed line), $\propto (1 + (2\pi f/T)^2)^{-1}$. The fitting parameter, $T = 16 \mu s$, is comparable to $\tau_c\tau_e/(\tau_c + \tau_e)$ from the time domain analysis, which also indicates a Poisson process.

The RTS noise has been studied in sub-micrometer Si metal-oxide-semiconductor structures [5], AlGaAs/GaAs narrow channels [14], and quantum dots [17]. A higher sample resistance in general restricts the lower frequency range that can be measured. Thus, most of the measurements on SETs ($R > 26 \text{ k} \Omega$ in principle) are restricted to the $1/f$ noise region. Our observation of RTS in the quantum dot is consistent with those previously reported. The maximum bandwidth $B$, where the change in the charge, $\Delta q$, is distinguished from the noise, $\delta q$, is given by $B \sim \frac{1}{\tau}(\Delta q/\delta q)^2$. This yields $B \sim 1 \text{ MHz}$ for the trap we investigated. We observe a duration as short as few $\mu s$ and rise/fall time shorter than $1 \mu s$ (the right inset of Fig. 4(a)), when the bandwidth is set to 2.5 MHz. If the trap is replaced with another quantum dot connecting another set of leads [5], single electrons will tunnel through the dot with a frequency of $I/e$ (600 kHz at a driving current, $I$, of 0.1 pA). The rf-SET should resolve each tunneling event of such a current.

In summary, we have demonstrated a transmission-type rf-SET using a quantum dot for detecting charge kinetics of a trap in an AlGaAs/GaAs heterostructure. Our results are encouraging for the direct measurement of ac response from a single-electron state.

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FIG. 1. (a) Schematic setup for rf-SET measurements. The quantum dot, d, fabricated in an AlGaAs/GaAs heterostructure is controlled with two gate voltages, $V_L$ and $V_R$. The charging energy of the dot is about 3 meV. The SET is placed in an LC resonator ($L = 100$ nH, $C \sim 0.5$ pF, resonant frequency $\omega/2\pi = 700$ MHz, and quality factor about 4). Signal $V_{t}e^{i\omega t}$ is amplified and detected with a homodyne configuration. The sample is cooled to 30 mK and measured at zero magnetic field. (b) Typical Coulomb blockade (CB) oscillation measured by dc current (upper traces) and rf-transmission (lower traces). (c) CB oscillation in another case where the SET probed a trap.

FIG. 2. Transmitted signal $V_t$ measured by applying a small modulation signal, $V_{mod}\sin(2\pi f_{mod})$, to the left gate ($f_{mod} = 10$ kHz). Static gate voltages are $V_L = -700$ mV and $V_R = -540$ mV (labelled $\alpha$ in Fig. 1(b)). Transmitted power at $f_{mod}$ is plotted against $V_{mod}$. The horizontal bar indicates the noise level measured with a resolution bandwidth of 1 Hz. The inset shows the spectrum of the transmitted signal at $V_{mod} = 1.7$ mV.
FIG. 3. Noise spectrum of the transmitted signal. The equivalent charge with respect to the quantum dot is on the right. (a) Noise of the SET electrometer. Thick trace shows the SET noise measured at $\alpha$ in Fig. 1(b), while the thin trace indicates noise from the measurement system (taken where SET is highly resistive). The dashed line shows a $1/f$ dependence. (b) Noise when the random telegraph signal (RTS) is dominant ($\beta$ in Fig. 1(c)). The thin line is the system noise. The dashed curve is a Lorentzian fitted to the RTS spectrum.

FIG. 4. (a) Random telegraph signal (RTS) measured at different gate voltages. Signal $V_t$ is high if the trap is occupied by an electron and low if the trap is empty. The right inset shows the shortest duration of the occupied state. The left inset shows the location of the trap T relative to the dot d. (b) Typical histogram of the duration of the empty state. The solid line is exponentially fitted to the histogram. (c) Time constants $\tau_c$ for capture and $\tau_e$ for emission. The horizontal bar is the energy scale of the trap state, $E_t$. The solid lines are fitted from the Fermi distribution of electrons in the leads.