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ABSTRACT
A study of transport in a quantum hot electron transistor made of an InAs/AlSb heterostructure is reported. It exhibited that the quantum hot electron transistors with a thick emitter efficiently prevented the parasitic base currents compared with transistors with a thin emitter. The static characteristics of the fabricated devices demonstrated an enhancement of the current gain of 9 and a collector breakdown voltage of 1.5 V with thick-emitter designed transistors. In optimized devices, the current is dominated by fast resonant tunneling that is promising for their future development of as high frequency transistors.

I. INTRODUCTION
Very high frequency electronic devices are essential for high speed telecommunications or sub-mm wave generation and detection systems. The performance of high frequency transistors is strongly related to the intrinsic transport properties of the used semiconductors. As a matter of fact, heterojunction bipolar transistors (HBTs) made of an InGaAs lattice matched on the InP substrate have significantly better frequency response as compared to GaAs-based devices. Furthermore, a significant increase in the cutoff frequency $f_c$ exceeding 700 GHz has been obtained thanks to the use of the indium-rich strained InGaAs material. Going further in this direction, smaller bandgap materials such as InAs and InSb are attractive candidates for even more rapid devices. In particular, electrons in InAs have a saturation velocity and mobility twice the values of InGaAs or InP. The work presented here aims to take benefit from these favorable properties of InAs-based structures to overcome the existing limits of electron transport in HBTs. InAs has already been proposed as a good material for bipolar transistors, but its use in HBTs was hampered by the lack of a large bandgap material, a lattice matched on the InAs substrate, that could provide the large valence band offset required for $pnp$ HBTs. A novel design of the transistor was proposed to overcome the current limitations above. This device that we call quantum hot electron transistor (QHET) is a unipolar transistor using the concept of a hot electron transistor (HET). Unlike the previously reported HET, it relies on a coherent vertical transport in a multi-quantum well InAs/AlSb heterostructure similar to that of quantum cascade lasers. It is not dependent on valence band offset alignments since only electrons in the conduction band are involved. For these reasons, its first advantage over $pnp$ HBTs is the high base sheet conductance accessible with moderate n-type doping levels (typically $n \sim 10^{18}$ cm$^{-3}$ as compared to $p \sim 10^{20}$ cm$^{-3}$ in HBTs), which is a key parameter for high speed operation. Second, electron transport in the short (typically 100 nm–200 nm) bulk InAs collector is mostly ballistic with calculated transit times much shorter than those in InP-based devices thanks to the larger saturation velocity. It also revealed that the InAs-based QHET with a submicronic footprint of the emitter exhibited low base resistance, high-current densities, and low operating voltage. Furthermore, the InAs-based QHET yielded the promising cut-off frequency of 77 GHz at room temperature for the first time. With the primary results, InAs-QHETs have been demonstrated to extend the frequency limit of vertical transport transistors toward the THz range. However, the physics of
transport of these devices are still unexploited, which is a necessary factor to continuously optimize the transistor structures as well as gain the dynamic performance.

In this Letter, we study the physics of transport in an original high speed vertical transport transistor based on InAs. To evaluate the physics of transport in InAs-based QHET, two QHET structures with different emitter designs have been fabricated and studied at cryogenic temperatures and room temperature. It demonstrated that the emitter with a thick barrier is sufficient to prevent the leakage currents from the emitters, which efficiently enhanced the gain statics.

II. EXPERIMENT
A. Design of QHET structures
The electron transport in the active part of the QHET is controlled by the quantum design of electron sub-bands. The complete layer structure, device geometry, and the conduction band diagram of the active part of the device are presented in Figs. 1 and 2. In Fig. 2, the band structure of the multi-quantum well part of the device is calculated using a home-made Schrödinger–Poisson solver based on the transfer matrix method. In order to account for the non-uniform electric field, each individual layer of the heterostructure was discretized in ten sections with a constant potential. We considered in the calculation of the energy levels and wavefunctions that the first and last AlSb barriers were infinitely thick. The boundary conditions for the electric field (F) were \( F = 0 \text{kV/cm} \) on the emitter side and \( F = 10 \text{kV/cm} \) on the collector side. Finally, we assumed that the electrons were distributed with independent thermal distributions in the emitter superlattice and in the base layer. The potential drop between the two associated Fermi levels corresponded to the junction bias \( V_{BE} = 0.45 \text{V} \). The material parameters used in the calculation are given in Ref. 15.

The base is made of a n-type InAs quantum well (QW) doped with silicon to \( 10^{18} \text{cm}^{-3} \). With a typical electron mobility greater than \( 10^4 \text{cm}^2/V\text{s} \), it corresponds to a base sheet resistance lower than \( 300 \Omega/\Box \). The emitter–base junction and collector barrier consist of graded InAs/AlSb superlattices. The collector transit region is made of 200 nm-thick undoped InAs. When compared to quantum cascade laser (QCL), the major difference in the band structure is the position of the extraction miniband that is resonant with the upper state of the active QW (the base of transistor) instead of the lower state in a QCL. Majority electrons provided by the doping in the base are confined in the QW ground state thanks to the collector barrier. When the QHET is supplied with suitable emitter–base bias \( (V_{BE}) \), electrons transit from the emitter to the collector (current \( I_C \)) through the coherent transport via the QW excited state that is resonantly coupled with the emitter and collector barrier states (Level 4, corresponding to the solid wavefunction in Fig. 2) and are transferred to the high field region of the base–collector junction made of bulk InAs. This resonant state is coupled to the continuum of states in the bulk collector; hence, the transit time of electrons through the base should be close to their dephasing time, typically a few tens of femtoseconds. The fraction of electrons scattered down to the lower states in the base produces the base current \( I_B \). An additional parasitic base current may arise from the direct injection from the emitter to the lower state of the base. Two samples were designed to observe the parasitic base currents. Herein, sample H5 contained thinner barriers in the emitter, which was reported in the previous

In sample H2, the layer sequence of the emitter consisted of
\[ 1.0/3.2/1.1/2.9/1.1/2.8/1.1/2.6/1.1/2.5/1.2/2.4/1.2/2.3/1.3/2.3/1.5/2.2/1.7/2.2/2.0; \]
and the collector barrier was \( 1.0/1.8/1.0/1.9/1.0/1.9/1.0/2.0/1.0. \)
In sample H5, the layer sequence of the emitter consisted of
\[ 1.2/3.2/1.2/3.0/1.2/2.8/1.2/2.7/1.2/2.5/1.2/2.3/1.2/2.1/2.1/2.0/2.1/2.0/2.1/1.0; \]
and the collector barrier was \( 1.0/1.5/0.6/1.2/0.6/1.2/0.6/1.2/0.6/1.2/0.6/1.4/0.6. \)
Here bold numbers are the AlSb thicknesses in nm and the roman numbers are the InAs thicknesses in nm; underlined layers were silicon-doped to \( 10^{18} \text{cm}^{-3} \).

![FIG. 1. (a) Layer sequence of the studied structures. (b) Schematic geometry of the studied double-mesa devices. (c) Optical micrograph of a fabricated QHET using double-mesa techniques, with 10×10 μm² emitter mesa and 20×30 μm² base mesa.](image-url)
A new structure called sample H2 was designed with the thicker barriers in the emitter compared with that of sample H5. The exact layer sequence is given in Fig. 1. The reduced barrier thickness resulted in a stronger coupling of the emitter states with the base quantized states. A quantitative estimation of this coupling is given by the level splitting at resonance ($2\hbar\Omega$) between the injector and the fourth base state. We calculated a value of $2\hbar\Omega = 11$ meV for sample H2 and $2\hbar\Omega = 45$ meV for sample H5. The strong coupling in H5, obtained with 1 nm-thick AlSb barriers was intended to provide a large emitter current density ($J_E$) at resonance. This is an important parameter for high frequency operation: using the common analysis wavefunctions are represented. Under normal operation, a resonant electron current flows from the emitter Fermi level (e) through the fourth quantized state (solid wavefunction) of the base and propagates ballistically in the collector.

**B. Fabrication of QHET**

The samples were grown by molecular beam epitaxy on the InAs substrate, using a Riber Compact 21 solid source machine equipped with As and Sb valved cracking cells. Thanks to the expertise that we developed for many years in the growth of InAs/AlSb QCLs, state of the art heterostructures were available for this work. From the grown wafers, QHETs were formed by a double-mesa processing that produced $10 \times 10 \mu m^2$ emitter and $20 \times 30 \mu m^2$ base mesas (Fig. 1). Ohmic contacts were made of nonalloyed Cr/Au. In this device fabrication, the critical step was the accurate contact of the thin base. A high precision layer-by-layer selective wet etching process has been developed for this purpose. The devices were passivated using a hard-backed resist that also supported the contact pads of the emitter and the base. The collector contact was directly made on the n-doped InAs substrate.

**III. RESULTS AND DISCUSSION**

Figure 3 shows the common base characteristics (Gummel plot) of QHETs from H5 structures at room temperature and cryogenic temperature (80 K). The Gummel plot displays the currents $I_B$ and $I_C$ as a function of $V_{BE}$, for a constant base–collector bias ($V_{BC} = 0$). The curves measured at the low temperatures give useful information to understand the origin of the measured currents in the different bias ranges. On sample H5, one can see three resonances in the base current similar to what is observed in a more conventional resonant tunneling diode (solid green circle), which are represented as (e $\rightarrow$ 2), (e $\rightarrow$ 3), and (e $\rightarrow$ 4) in Fig. 3(a). They correspond to the successive alignments of the emitter Fermi level with the base QW Levels 2, 3, and 4, while $V_{BE}$ is biased at 0.1 V, 0.3 V, and 0.4 V [Figs. 3(b)–3(d), respectively]. In the meantime, there are also successive alignment of base QW Levels 3 and 4 with collector Fermi levels, which are known as (3 $\rightarrow$ c), and (4 $\rightarrow$ c), respectively. When the fourth state of the base enters in resonance with the emitter and the collector, the expected current density is very high (larger than 500 kA/cm$^2$). In the present study, the large size of the emitter mesa and the device geometry did not allow reaching such high currents, notably because of the base access resistance, as shown in Ref. 6. As a consequence, the (e $\rightarrow$ 4) resonance is only observed as an increase in the collector current limited by the access resistance. In addition, the real potential drop in the emitter–base junction is significantly lower than the applied $V_{BE}$ bias. A significant collector current is only measured when electrons are injected directly into Level 4, i.e., for $V_{BE} > 0.6$ V since it is the only level that is coupled to the collector. Nevertheless, a small escape current can be seen from the electron injected into Level 3 at a bias $V_{BE} ∼ 0.4$ V (solid purple circle). Hence, the base current measured for $V_{BE} < 0.6$ V appears to be mostly a parasitic current injected directly from the emitter to the three lower base states. At room temperature, these resonances are not visible since they are broadened by thermal activation. One can, however, conclude that the strong base current measured below 0.6 V also originates from the same parasitic channels. In the same way, the large ideality factor of current $I_C$ is due to the current flowing from Level 3 to the collector at low bias $V_{BE}$. For sample H5, the current gain (i.e., $I_C / I_B$) is only observed when $V_{BE} > 0.8$ V. The static gain $\beta = I_C / I_B$ reaches a maximum value of 5 at large bias.

Sample H2 has been designed in order to reduce the parasitic current observed in sample H5. This has been done by reducing the emitter–base coupling thanks to the thicker AlSb barrier in the end of the emitter superlattice, as already explained, and also by adjusting the alignment of the levels originating from the last emitter QWs with the Level 4 of the base. At low temperatures, one can still see the resonance e $\rightarrow$ 3 for $V_{BE} ∼ 0.4$ V, but this parasitic current has been reduced by three orders of magnitudes as compared to sample H5 [Fig. 4(a)]. The currents in the QHET H2 are then dominated by the currents flowing through the resonant Level 4, as ideally wanted [Fig. 4(b)]. At room temperature, the Gummel plot of transistors H2
demonstrated a current gain for $V_{BE} > 0.3$ V with a maximal value of $\beta = 9$.

In resonant conditions, the current gain can be seen as the ratio of the inter-sub-band relaxation time $\tau_{rel}$ in the base to the transit time $\tau_{tr}$ for electrons in the resonant state,

$$\beta = \frac{\tau_{rel}}{\tau_{tr}}. \quad (1)$$

The time $\tau_{rel}$ can be estimated by the calculated LO phonon scattering rates from Level 4 to the three lower levels. We obtained a value of $\tau_{rel} = 0.7$ ps. However, this is a probably overestimated value since additional scattering mechanisms, such as interface roughness or electron–electron scattering may play a significant role. We estimate that electron–electron scattering is still limited thanks to the relatively low doping level of the base. We can derive base transit

FIG. 3. (a) Common-base static characteristics (Gummel plots) of the studied samples HS at 80 K and 300 K. The emitter mesa is $10 \times 10 \mu m^2$ and the base mesa is $20 \times 30 \mu m^2$; (b)–(d) calculated conduction band diagram of the active part of the device, when biased at, $V_{BE} = 0.1$ V, 0.3 V and 0.4 V, respectively.
FIG. 4. (a) Common-base static characteristics (Gummel plots) of the studied samples H2 at 80 K and 300 K. The emitter mesa is $10 \times 10 \, \mu m^2$ and the base mesa is $20 \times 30 \, \mu m^2$; (b) calculated conduction band diagram of the active part of the device, with thick barrier layers for reducing the parasitic currents.

time $t_B$ by converting the equation above,

$$t_B \approx \frac{\tau_{rel}}{\beta}.$$  \hspace{1cm} (2)

From the measured gain $\beta = 9$, we can conclude that the transit time for resonant tunneling is $t_B = \tau_{rel}/\beta < 78$ fs. This analysis indicates the future directions to explore in order to further increase the current gain in a QHET structure: reduce the extraction time with a more transparent collector barrier or increase the inter-sub-band relaxation time with a different base QW design. The base resistance is estimated to be about 350 $\Omega$–500 $\Omega$, extracted from the Gummel plot for both structures of H2 and H5. Evidently, this value, due to the large dimension of the devices, is not suitable for high frequency operation that would require submicron dimensions and a large aspect ratio of the emitter mesa.

The static common-emitter output characteristics $I_c(V_{ce})$ of QHET H2, measured at both cryogenic and room temperatures, are shown in Fig. 5. A clear saturation of current $I_c$ is obtained with a differential current gain of about 9, consistent with the value of $\beta$ measured in common-base configuration. The threshold voltage is $0.5 \, V$ and collector breakdown voltage $B_{V_{CE}}$ is larger than $1.5 \, V$. However, it must be noted that these values are probably overestimated because of the parasitic series resistances due to the geometry of the device. The relatively high access resistances also limited the maximum current density to $J_E \approx 30 \, kA/cm^2$. Low values of the breakdown voltage were often presented as the major drawback of low bandgap materials, such as InAs. The data of Fig. 5 demonstrate that good operating conditions can nevertheless be obtained in the QHETs. Under operating conditions, the voltage drop across the bulk InAs collector transit region is approximately equal to $V_{CE} = 0.6 \, V$ (see the band diagram on Fig. 2). Hence, the breakdown starts when the voltage applied to the collector junction is about $0.9 \, V$. It corresponds to an electric field of $45 \, kV/cm$ in the $200 \, nm$-thick collector, which is a sensible value. The two characteristics at low and room temperature are quite similar, as expected for a device in which transport is mainly controlled by resonant tunneling. A noteworthy difference is, however, the stronger collector leakage current at room temperature as seen on the lower curve ($I_B = 0$). This is explained by the thermally activated transport of electrons from the base ground state to the collector barrier miniband.

IV. CONCLUSIONS

In summary, we demonstrated the feasibility of the novel vertical transport QHETs based on InAs. A suitable current gain and operating voltages are obtained under static operation. Electron transport is shown to be controlled by potentially very fast coherent resonant tunneling through the base excited state. Together with
previously reported very short theoretical collector transit times\textsuperscript{14} and excellent experimental base sheet resistance,\textsuperscript{6,13} this study provides an additional evidence that the QHET technology has all required properties for producing a transistor with cut-off frequencies in the THz range. The future steps toward the demonstration of high speed operation are, first, to optimize the transistor structures to enhance the static gain and, second, use a metamorphic growth on semi-insulating GaAs\textsuperscript{4} in order to allow high frequency characterizations. The improved design proposed here should result in a significantly better emitter differential resistance and, hence, improve the cutoff frequencies as compared to Ref. 6.

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DATA AVAILABILITY

The data that support the findings of this study are available from the corresponding author upon reasonable request.

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