Toward Length-Versatile and Noise-Robust Radio Frequency Fingerprint Identification

Guanxiong Shen, Junqing Zhang, Alan Marshall, Mikko Valkama, and Joseph R. Cavallaro

Abstract—Radio frequency fingerprint identification (RFFI) can classify wireless devices by analyzing the signal distortions caused by intrinsic hardware impairments. Recently, state-of-the-art neural networks have been adopted for RFFI. However, many neural networks, e.g., multilayer perceptron (MLP) and convolutional neural network (CNN), require fixed-size input data. In addition, many IoT devices work in low-signal-to-noise ratio (SNR) scenarios but the RFFI performance in such scenarios is often unsatisfactory. In this paper, we analyze the reason why MLP- and CNN-based RFFI systems are constrained by the input size. To overcome this, we propose four neural networks that can process signals of variable lengths, namely flatten-free CNN, long short-term memory (LSTM) network, gated recurrent unit (GRU) network, and transformer. We adopt data augmentation during training which can significantly improve the model’s robustness to noise. We compare two augmentation schemes, namely offline and online augmentation. The results show the online one performs better. During the inference, a multi-packet inference approach is further leveraged to improve the classification accuracy in low SNR scenarios. We take LoRa as a case study and evaluate the system by classifying 10 commercial-off-the-shelf LoRa devices in various SNR conditions. The online augmentation can boost the low-SNR classification accuracy by up to 50% and the mult-packet inference approach can further increase the accuracy by over 20%.

Index Terms—Internet of Things, LoRa, LoRaWAN, device authentication, radio frequency fingerprint, deep learning.

I. INTRODUCTION

RADIO frequency fingerprint identification (RFFI) is a promising technique for authenticating Internet of Things (IoT) devices. The analog front-end of wireless devices consists of hardware components such as an oscillator, a mixer, a power amplifier, etc. These components may have slight variations in their specifications from their standard values, known as hardware impairments. These impairments are unique and can be used as identifiers, similar to how biometric fingerprints are used for identification [1].

The RFFI system identifies wireless devices by analyzing the characteristics of the received signals. Traditionally, this is achieved by manual feature extraction algorithms in combination with conventional machine-learning classification models. The manually extracted features can be the estimated hardware characteristics, such as the estimated CFO, IQ imbalance, phase error [2], [3], [4], power amplifier non-linearity [5], beam pattern [6], etc. They can also be the statistics of the received signal, such as the kurtosis, spectral flatness, and brightness [7], [8]. In addition to these, some domain transformation results and variants of constellation figures can serve as features as well, such as power density spectrum [9], [10] and differential constellation trace figure (DCTF) [11]. For non-stationary signals such as LoRa, the received IQ samples are often transformed into the time-frequency domain for analysis. The well-known algorithms include short-time Fourier transform (STFT) [12], [13], [14], [15], [20], [21], [22], [23], [24], [25], [26], [27], [28], [29], [30]. Compared to conventional RFFI protocols that manually design feature extraction algorithms, deep learning-based approaches can automatically extract discriminative features after sufficient training. Typically, a deep learning-based RFFI system uses a neural network to directly classify received signals and predict the device label. Previous studies demonstrated that deep learning-based RFFI systems often outperform traditional approaches and have attracted wide attention [30], [31]. Although deep learning-powered RFFI systems have shown excellent identification performance, they are constrained by the fixed input size problem. Previous studies use deep

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neural networks (DNNs) [13], convolutional neural networks (CNNs) [1], [27], [28], [32], or recurrent neural networks (RNNs) [26] as intelligent engines. However, DNN and CNN are designed for a specific input size. After the input size is determined, they are unable to process signals with different lengths, which conflicts with the fact that many wireless signals are variable in length. To overcome this, some studies use packet preambles/segments with a fixed length as model inputs [13], [15], [26], [29], [32]. However, this rules out the payload part, and the preamble length may not be fixed in some wireless protocols, e.g., LoRa. The fixed-size input problem of DNN and CNN has been indicated in previous works but the reason and analysis are not given [28], [33]. Some studies have proposed slicing/splitting techniques, in which the signal is split into equal-length segments and then processed separately [23], [24], [27], [28]. However, the use of slicing/splitting methods can lead to a decrease in performance, which will be experimentally demonstrated in the paper. In fact, sequential models such as RNN are able to process variable-length inputs by design, but this important property has never been exploited in previous RFFI studies.

Improving RFFI performance at low SNR is also challenging. The transmission power of IoT end nodes should be kept as minimum as possible to reduce power consumption, but this makes the received signals more vulnerable to noise. One improvement approach is to design noise-resilient signal representations as model inputs. For instance, Ozturk et al. found that time-frequency data, i.e., spectrogram, is more resilient to noise than time-series data [15]. Xing et al. proposed a stacking algorithm for direct sequence spread spectrum (DSSS) systems to improve signal quality [34]. Alternatively, we can also obtain noise robustness by enhancing the capability of the deep learning model. Data augmentation is a common approach to train a channel-agnostic RFFI model [14], [27], [33], which can also be leveraged to improve the model’s noise robustness. However, specific analysis on the effect of data augmentation in low SNR scenarios is still missing.

In this paper, we take LoRa/LoRaWAN, a well-known low power wide area network (LPWAN) technology, as a case study to investigate the above two challenges. The design methodology is applicable to any RFFI system that needs to tackle variable inputs and low SNR conditions and is not limited to LoRa. LoRa defines the physical layer modulation, while LoRaWAN specifies higher layer protocols such as the medium access control (MAC) layer and network architecture. LoRaWAN leverages the adaptive data rate (ADR) mechanism that enables end nodes to adjust transmission configurations on the fly. This makes the LoRa preamble length variable. LoRa transmission power is low, and the long-range communication results in serious attenuation.

This paper designs RFFI protocols that can tackle input data with variable lengths and proposes solutions for low SNR scenarios. We investigate four neural network architectures that are not constrained by the input size. We leverage data augmentation to train noise-robust models and compare the performance of different augmentation strategies. Finally, we propose a multi-packet inference approach, which can significantly improve classification accuracy in low SNR scenarios. Experimental evaluation is carried out using 10 commercial-off-the-shelf LoRa devices and a USRP N210 software-defined radio (SDR) platform. Our contributions are highlighted as follows:

- We summarize and discuss the input size constraint of previously used neural networks, such as DNN and CNN. Four length-versatile neural networks are proposed to overcome the constraint, namely flatten-free CNN, long short-term memory (LSTM) network, gated recurrent unit (GRU) network and transformer. They are all capable of classifying LoRa devices from variable-length preambles. Transformer can achieve satisfactory performance with minimum complexity. To the best knowledge of the authors, this is the first time to explore how to apply the state-of-the-art transformer model to RFFI. We also compare their performance with the previously used slicing technique, which is the state-of-the-art solution for inputs with variable lengths. Our solution is shown to perform better at low SNRs than the slicing technique.

- We investigate the effect of data augmentation on enhancing RFFI noise robustness. We compare the performance of online, offline and no augmentation strategies. The models trained with online augmentation outperform the rest. Taking flatten-free CNN as an example, the model trained with online augmentation, offline augmentation, and no augmentation achieves accuracy of near 90%, 80%, and 10%, respectively, at 15 dB.

- We propose a lightweight multiple-packet inference method that can significantly improve classification accuracy in low SNR scenarios. Specifically, the accuracy can be boosted from 20% to over 50% at 0 dB, and from 60% to about 90% at 10 dB.

The codes1 and datasets2 for this research are accessible online.

In our previous work [35], we employed a transformer model to process variable-length signals. We also investigated the effect of data augmentation and multi-packet inference on improving the low-SNR RFFI performance. In this paper, we have significantly extended our contribution by studying three additional length-versatile neural networks, namely flatten-free CNN, LSTM and GRU. We also compare their performance with the slicing/splitting technique.

The rest of this paper is organized as follows. Section II presents the background and motivations. Section III introduces the RFFI system in detail and Section IV shows the architectures of the proposed length-versatile neural networks. Section V provides experimental settings, results and discussion. Section VI is the comparison with the slicing technique. Section VII introduces related work and Section VIII finally concludes the paper.

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1https://github.com/gxhen/lengthVersatileRFFI
2https://ieeexplore.ieee.org/document/lorarfiddatasetdifferentspreadingfactors
Fig. 1. Waveform (in-phase branch) of a LoRa preamble under different SF configurations. The preamble length increases with SF. (a) SF = 7. (b) SF = 8. (c) SF = 9.

Device 1
Device 2
Device K
Fig. 2. An RFFI system that can infer device labels from variable-length signals.

II. BACKGROUND AND MOTIVATIONS

A. LoRa/LoRaWAN Primer

LoRa uses chirp spread spectrum (CSS) as the physical layer modulation technique. There are several basic up-chirps at the beginning of a LoRa packet, named preambles. A baseband LoRa preamble is given as

\[ x'(t) = Ae^{j(-\pi Bt + \pi B^2 t^2)}, \quad (0 \leq t \leq T), \]

where \( A, B \) and \( SF \) denote signal amplitude, bandwidth and spreading factor, respectively. \( T \) is the duration of a LoRa symbol, given as

\[ T = \frac{2SF}{B}. \tag{2} \]

The bandwidth increases from 125 kHz to 500 kHz whenever the SF ranges from 7 to 12. In our configuration, the entire preamble part \( x(t) \) consists of eight repeating \( x'(t) \).

The ADR mechanism is adopted in LoRaWAN, which allows LoRa end-nodes to optimize the data rate/SF adaptively according to the estimated SNR at the gateway. The SF will increase whenever a LoRa end-node moves further away from the gateway, in order to maintain a higher link budget [36]. As formulated in (2), a higher SF leads to a longer LoRa symbol duration. Therefore, the LoRaWAN ADR mechanism makes the length of the preamble variable. Fig. 1 shows the waveform of one preamble under different SF configurations.

B. RFFI

As shown in Fig. 2, an RFFI system involves \( K \) IoT devices to be classified and a gateway as the authenticator. When the gateway receives a packet from a device, the RFFI protocol analyzes the received signal to infer from which end node the packet is sent.

For a specific IoT device \( i \), its signal \( x(t) \) is distorted by hardware impairments and propagated in the wireless channel. At the receiver side, the received signal \( r_i(t) \) is mathematically expressed as

\[ r_i(t) = h(t) * F_i(x(t)) + v(t), \tag{3} \]

where \( F_i(\cdot) \) denotes the specific hardware distortion of device \( i \), \( h(t) \) is the channel impulse response, \( v(t) \) is the additive white Gaussian noise (AWGN) and \( * \) denotes the convolution operation. When the receiver captures a packet from device \( i \), it extracts the preamble part \( r_i(t) \) and digitizes it, denoted as \( r_i[n] \). Then an RFFI system is leveraged to accurately map this \( r_i[n] \) to the device label \( i \), acting as a classifier. This is formulated as

\[ \mathcal{R}(r_i[n]) \rightarrow i, \tag{4} \]

where \( \mathcal{R}(\cdot) \) denotes operation of the RFFI system.

C. Motivation 1: Variable-Length Signal vs. Fixed Input Length of Neural Networks

Many previous deep learning-based RFFI systems can only process fixed-length inputs [13], [28], [31], [32]. However, the length of \( r_i(t) \) might be variable. For instance, the lengths of LoRa preambles, i.e., \( T_i \) in Fig. 2, are different due to the LoRaWAN ADR mechanism.

1) Input Constraints of DNN: DNN, also known as fully connected neural network and multilayer perceptron (MLP), is used to build RFFI protocols in some previous works [12], [26], [31], which only accept fixed-length inputs.

The DNN is entirely composed of dense layers. A dense layer defines a linear transformation between its input vector \( x \) and output vector \( y \), formulated as

\[ y = xA^T + b, \tag{5} \]

where \( A \) and \( b \) denote the weights and biases learned during training, respectively. The length of \( x \) must equal the number of rows of \( A^T \) to perform the multiplication. When an input longer than \( x \) is fed into the DNN, the matrix and vector dimension do not agree thus the first dense layer does not work. Therefore, DNN is constrained by the input size.

2) Input Constraints of CNN: CNN is the most widely used model in RFFI. Previous studies have indicated that CNN can only handle signals of fixed size [23], [28], [33], but a proper explanation is missing.

We show that CNN requires fixed-size input due to the flatten and dense layers. CNN consists of two modules, convolutional/pooling layers for feature extraction and dense layers for classification. They are connected by a flatten layer which converts several two-dimension (2D) feature maps to a one-dimension (1D) vector. In fact, the feature extraction module is not sensitive to input length. The convolutional and pooling layers can process any size of the input but will output feature maps of different sizes. This further makes the flatten layer output a vector of different lengths. However, as explained in Section II-C.1, dense layers are designed for a fixed input length, which constrains the input size of the entire CNN.

As in the CNN exemplified in Fig. 3(a), the input is first processed by several convolutional/pooling layers for feature extraction, and a \((H_1, W_1, C)\) tensor is returned. This denotes \( C \) feature maps of size \((H_1, W_1)\), where \( C \) equals the number...
of filters in the last convolutional layer. Since dense layers can only process 1D vectors, a flatten layer is employed to convert the \((H_1, W_1, C)\) tensor to a vector containing \((H_1 \times W_1 \times C)\) elements. The dense layers then make predictions from it.

However, the designed CNN cannot process signals of different lengths. As shown in Fig. 3(b), when a long signal is an input into the CNN, the output of convolutional/pooling layers is a \((H_2, W_2, C)\) tensor. Note that \(H_2 \geq H_1\) and \(W_2 > W_1\) since the input is longer. The flatten layer then converts the tensor to a vector with \((H_2 \times W_2 \times C)\) elements, which cannot be processed by the dense layers. Therefore, the reason that CNN can only process fixed-size inputs is that dense layers cannot process variable-length vectors output from the flatten layer.

D. Motivation 2: Low SNR of Received Signals

RFFI systems should be able to identify wireless devices from low SNR signals. The SNR of \(r_i[n]\) in the decibel scale is formulated as

\[
y = 10 \log_{10} \left( \frac{RMS(h[n] \ast F_i(x[n]))}{RMS(v[n])} \right)
\]

where \(RMS(\cdot)\) returns the root mean square value. The IoT devices are often configured with a low transmission power to save energy, resulting in a low amplitude emitted signal \(F_i(x[n])\). Moreover, a long distance between the transmitter and receiver results in significant attenuation, which affects the magnitude of channel taps in \(h[n]\). These two factors jointly result in the low SNR of the received signal. It is necessary and challenging to improve RFFI performance in such low SNR scenarios.

III. RFFI Protocol Design

A. Overview

As shown in Fig. 4, the proposed RFFI protocol involves two stages, namely training and inference. First, we collect extensive labelled LoRa packets from \(K\) training devices, preprocess and store them as the training dataset. We also build a neural network that is capable of processing variable-size data. Then its parameters are initialized and trained with the collected dataset. Data augmentation is adopted during training to improve robustness to noise. Once the model training is completed, it can classify the newly received LoRa packet, i.e., the inference stage. The LoRa packet is preprocessed and converted to a channel-independent spectrogram to mitigate channel effects. Then it is fed into the trained neural network and an inference is made. In low SNR scenarios, we can leverage multiple packets for joint inference to obtain a more accurate prediction.

B. Preprocessing

As shown in Fig. 4, the RFFI protocol needs to preprocess the received LoRa packet in both the training and inference stages. We use the same preprocessing algorithms as in [37].

We employ packet detection and synchronization algorithms to capture LoRa transmissions. After that, we extract the packet preamble part \(r_i[n]\) for classification to prevent the model from learning protocol-specific information. Then, carrier frequency offset (CFO) compensation is carried out as CFO is sensitive to temperature variation and reduces the system stability [12], [13]. Finally, the signal is normalized by dividing its root mean square, to prevent the model from classifying devices based on the received signal strength, which is location dependent.

C. Channel-Independent Spectrogram

During both training and inference stages, the received IQ samples \(r_i[n]\) are converted to channel-independent spectrograms to mitigate channel effects [37]. Channel-independent spectrogram is adopted because it has been demonstrated to be resilient to channel variations [37].

LoRa uses CSS modulation whose frequency changes over time. Hence STFT is used to reveal how the frequency varies over time. A channel-independent spectrogram, \(S^i\), can be further generated to mitigate channel effects, with each matrix element \(S^i_{k,m}\) mathematically given as

\[
S^i_{k,m} = 10 \log_{10} \left( \frac{\sum_{n=0}^{N-1} r_i[n]w[n-(m-1)R]e^{-j2\pi \frac{k}{N}}}{\sum_{n=0}^{N-1} r_i[n]w[n-mR]e^{-j2\pi \frac{k}{N}}} \right)^2
\]

\times for \(k = 1, 2, \ldots, N\) and \(m = 1, \ldots, M - 1\),

(7)

where \(w[n]\) is a rectangular window of length \(N\), and \(R\) is the hop size. In this paper, \(N\) and \(R\) are always set to 64.
and 32, respectively. Please refer to [37] for the detailed derivation of the channel-independent spectrogram.

The LoRa ADR mechanism introduced in Section II affects the size of the channel-independent spectrogram. As formulated in (7), the generated channel independent spectrogram $S'$ is a $N \times (M - 1)$ matrix. The height of the channel-independent spectrogram, $N$, is equal to the length of the window function $w[n]$. It can be configured to be the same for all SFs. However, the width of $S'$, $(M - 1)$, is related to the length of the received signal $r[n]$, which is formulated as

$$M - 1 = \frac{8 \cdot s_f}{B} \cdot \frac{1}{T_s} - N,$$

where $T_s$ is the receiver sampling interval. As shown in the formula, the width of $S'$ depends on the SF configuration $s_f$, receiver sampling interval $T_s$, and transmission bandwidth $B$. Therefore, the LoRaWAN ADR mechanism results in channel-independent spectrograms $S'$ of different widths. Fig. 5 shows channel-independent spectrograms when the SF is configured to 7, 8, and 9, respectively. They have the same height but different widths. When SF is set to 9, the channel-independent spectrogram is about four times as wide as when SF is set to 7. This demonstrates the input sizes of the neural network should not be fixed in LoRa-RFFI. Moreover, it validates the need for a neural network that can process channel-independent spectrograms of variable size.

### D. Collecting Training Dataset and Building Neural Networks

In the training stage, we collect a large number of LoRa packets, preprocess them and generate a training dataset. Note that we need to collect LoRa packets of various SF settings for training so that the system can handle different SFs in the inference stage. In this paper, we use SF 7, 8, and 9 as an example.

We also build a neural network by defining its architecture and initializing its parameters. The neural network architectures are described in detail in Section IV. Note that all the neural networks can accept variable-length inputs therefore they can be trained with all the collected SF 7, 8 and 9 LoRa packets. After defining the architecture, the network parameters are initialized.

### E. Training With Augmentation

The parameters of the initial neural network can be trained and updated with the collected training packets. We investigate two training pipelines with different augmentation strategies, namely offline and online augmentation, as shown in Fig. 6.

Data augmentation is usually employed in the training stage of an RFFI protocol to increase the model robustness against channel variations [14], [27], [33], [37]. In this paper, we focus on different SNR scenarios, therefore an AWGN channel model is employed for augmentation.

1) **Offline Augmentation:** The offline augmentation is performed on the original training dataset. We pass the training signals through AWGN channels with various SNR levels and store the augmented ones as the augmented training dataset.

At the beginning of each epoch during training, a number of samples are randomly selected from the augmented training dataset to form a mini-batch. The IQ samples in the mini-batch are then converted into channel-independent spectrograms and fed into the initial neural network. The gradients are calculated to update the neural network parameters. This update process continues until training stop conditions are met. In offline augmentation, the number of noisy signals learned during training equals the number of training samples.

This can be increased by replicating the training dataset but will be bounded by the storage space limit.

2) **Online Augmentation:** Online augmentation is also known as augmentation on-the-fly. Different from the offline one, online augmentation is performed on the mini-batches. As shown in Fig. 6, the mini-batch is selected from the training dataset and fed into AWGN channels, obtaining an augmented mini-batch. The other training steps remain the same as the offline augmentation.

In online augmentation, the model is trained with (steps × mini-batch size) noisy signals.

The SNR used for both online and offline augmentation should match the practical test scenarios. In the current study, the SNR uniformly ranges from 0 dB to 40 dB when generating the AWGN channels.

### F. Multi-Packet Inference

Once the training is completed, the neural network can act as a classifier to infer device identity. The received packet
is preprocessed and converted to the channel-independent spectrogram. Then it is fed into the trained neural network and the probability vector, $\hat{p}$, will be returned by the softmax layer. $\hat{p}$ can be regarded as confidence levels over $K$ devices.

Multi-packet inference refers to making decisions with multiple LoRa transmissions. More specifically, we average the inference $\hat{p}$ with $(N_{\text{pkt}} - 1)$ historical inferences to derive a merged prediction $\hat{p}'$, mathematically expressed as

$$\hat{p}' = \frac{1}{N_{\text{pkt}}} \sum_{n=1}^{N_{\text{pkt}}} \hat{p}^n.$$  

(9)

where $\hat{p}^n$ is the prediction from the $n^{th}$ packet. Then the predicted label can be derived by selecting the index with the highest probability, which is formulated as

$$\text{label} = \arg\max_k (\hat{p}').$$  

(10)

According to the experimental results, the multi-packet inference protocol is particularly effective in low SNR scenarios. However, in high SNR scenarios, this protocol can be disabled to save computing resources. A more detailed discussion can be found in Section V.

IV. LENGTH-VERSATILE NEURAL NETWORKS

The neural network input is the channel-independent spectrogram introduced in Section III-C. As shown in Fig. 5, it has different widths at different SF settings. In this section, we will propose four models that can process channel-independent spectrograms of any width, namely flatten-free CNN, LSTM, GRU, and transformer.

A. Flatten-Free CNN

As discussed in Section II-C, CNN can only process fixed-size inputs because dense layers cannot process the flattened variable-length vectors. To overcome this, we replace the flatten layer with a global average pooling 2D layer so that the dense layers can always receive a fixed-length input. The architecture of the proposed flatten-free CNN is shown in Fig. 7(a), which refers to the design of ResNet [21] but is more lightweight and optimized for the RFFI task. It consists of ten convolutional layers, one max pooling layer, one global average pooling 2D layer and one dense layer of $K$ neurons activated by the softmax function. Skip connections are employed. $1 \times 1$ convolution is performed on the output of the fifth convolutional layer so that it can be added to the output of the seventh convolutional layer. The output of all the convolutional layers is zero-padded to be of the same dimension as the input and then activated by the ReLU function.

The input channel independent spectrogram is fed into the convolutional and max pooling layers for feature extraction, and 128 feature maps of size $(32, (M - 1)/2)$ are returned, forming a $(32, (M - 1)/2, 128)$ tensor. The first two dimensions $(32, (M - 1)/2)$ depend on the size of the input channel-independent spectrogram and the third dimension equals the number of filters in the last convolutional layer.

A global average pooling 2D layer is leveraged to make the input size to the dense layers independent of the channel-independent spectrogram size. It is highlighted in orange in Fig. 7(a). The global average pooling 2D layer calculates the average value of each individual feature map and outputs a fixed-length (128) vector that acts as the input to the dense layer. In other words, the size of the dense layer input only depends on the number of filters in the previous convolutional layer.

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B. LSTM Network

RNN is another popular class of deep learning models, which has been employed in some previous RFFI studies [13], [14], [26], [30]. RNN is designed for sequential data, such
as speech and human language, whose length is naturally variable. Therefore, RNN models are not constrained by input size by design. However, this valuable property has never been discussed and investigated in previous RFFI studies.

A popular variant of RNN is the LSTM network [38]. RNN-structured neural networks can process variable-width channel-independent spectrograms. The LSTM network is designed as shown in Fig. 7(b), which consists of two 256-unit LSTM layers, one global average pooling 1D layer and one softmax-activated dense layer. The LSTM layers are configured to return the full sequence instead of the last output. Similar to convolutional layers, LSTM layers are insensitive to the input size as well. Therefore, how to make the input of the dense layer independent of the channel-independent spectrogram width is the critical aspect of this design.

As shown in Fig. 7(b), the channel independent spectrogram is processed by two LSTM layers and a tensor of shape (256, M − 1) is obtained. The first dimension (features) equals the unit number 256 of the second LSTM layer, and the second dimension (timesteps) of the output tensor equals the width of the channel-independent spectrogram, M − 1. Therefore, the shape of the LSTM layers output (256, M − 1) varies with the width of the input channel independent spectrogram.

A global average pooling 1D layer is leveraged to make the dense layer input size constant. It is highlighted in green in Fig. 7(b). It averages the tensor along the timesteps and outputs a (256) vector. By this operation, the dense layers always receive a (256) vector regardless of the width of the input channel-independent spectrogram.

C. GRU Network

GRU network is another variant of RNN, which can also handle variable-length inputs [39]. GRU units have fewer parameters compared to LSTM units. The designed GRU network is shown in Fig. 7(c), whose architecture is nearly the same as the LSTM network, except that the LSTM layers are replaced by the GRU layers. Similar to LSTM layers, the GRU layers also output a (256, M − 1) tensor and the global average pooling 1D layer can convert it to a fixed-length (256) vector regardless of the input size. Therefore, GRU networks can handle variable-size inputs as well.

D. Transformer

In addition to RNN models, the transformer is another state-of-the-art model for processing variable-length sequential data [40]. As shown in Fig. 7(d), the proposed transformer for RFFI consists of two multi-head attention (MHA) layers, two point-wise feed-forward (point-wise FFN) layers, one global average pooling 1D layer and a dense layer with softmax activation. Similar to convolutional and LSTM layers, the MHA layer and the point-wise FFN layer can also process the input of any length. Hence the goal of transformer design is to make the input of the dense layer a constant length.

The model design refers to the encoder part proposed in [40]. First, the position embedding is added to the input channel independent spectrogram to enable the transformer to learn the temporal correlations. The result of addition is then fed into MHA and point-wise FFN layers for feature extraction. Skip connections and layer normalization are leveraged to improve the feature learning ability. The output of the second point-wise FFN layer is a tensor of shape (M − 1, 64), which is exactly the same size as the input channel independent spectrogram. As shown in Fig. 5, channel independent spectrograms under different SF configurations have the same height N = 64. Therefore, the tensor shape is simplified to (64, M − 1), which is affected by the input width.

Similar to the LSTM model, a global average pooling 1D layer is used to average the tensor along timesteps. It converts the (64, M − 1) tensor to a (64) vector that acts as the input of the dense layer. The vector is fixed-length regardless of the model input size, so the transformer can handle channel-independent spectrograms with any SF configuration.

E. Summary

The critical design of a length-versatile classification neural network is to make the input to the dense layer constant, which is achieved by using global average pooling 2D/1D layers in our schemes. The global average pooling 2D layer is used in CNN models. It is placed after the convolutional layers and converts the 3D tensor to a 1D vector by averaging each individual feature map. The output size of the global average pooling 2D layer only depends on the number of filters in the last convolutional layer. The global average pooling 1D layer is used in RNN and transformer models. It averages along the time steps to convert the LSTM/GRU/FFN layer outputs to a constant-length vector. With the help of global average pooling layers, the dense layer always receives a fixed-length vector regardless of the model input size, thus being able to process variable-length received signals r_i(t).

The neural network architectures in Fig. 7 are designed with reference to the famous deep learning models such as ResNet [21] and transformer [40]. Their hyper-parameters, e.g., filter number in CNN, and unit number in LSTM/GRU, are optimized experimentally. A rule of thumb in building neural networks is that the more parameters, the stronger the learning ability. However, it is also worth noting this can result in overfitting especially when the training dataset is not sufficiently large. Therefore, the optimization strategy used in this work is to first build lightweight neural networks with the minimum number of parameters and then gradually increase them until the model performance stops improving.

V. EXPERIMENTAL EVALUATION

A. Experimental Settings

1) Data Collection Settings: 10 LoRa devices under test (DUTs) are employed. As shown in Fig. 8(a), five of them are LoPy4 and the others are Dragino LoRa shields, and they all use Semtech SX1276 chipsets. All LoRa DUTs are configured with a bandwidth of 125 kHz, a carrier frequency of 868.1 MHz, and a 0.5 second transmission interval. As shown in Fig. 8(b), a USRP N210 SDR platform is adopted as the receiver, and its sampling rate is set to 250 kHz. The LoRa DUTs and USRP N210 are placed half a meter apart with no obstacle between them (line-of-sight). Data collection was
TABLE I

| Model          | SF   | 0 dB  | 4 dB  | 8 dB  | 12 dB | 16 dB | 20 dB | 24 dB | 28 dB | 32 dB | 36 dB | 40 dB |
|----------------|------|-------|-------|-------|-------|-------|-------|-------|-------|-------|-------|-------|
| Flatten-Free CNN | SF 7 | 22.10% | 35.78% | 56.60% | 75.96% | 87.06% | 91.52% | 94.22% | 97.50% | 99.26% | 99.86% | 100.00% |
|                | SF 8 | 20.44% | 35.40% | 53.96% | 74.64% | 90.86% | 98.12% | 98.92% | 99.68% | 100.00% | 100.00% | 100.00% |
|                | SF 9 | 18.68% | 34.28% | 56.64% | 78.86% | 91.12% | 95.16% | 97.56% | 98.86% | 99.92% | 99.98% | 99.98% |
| LSTM           | SF 7 | 24.02% | 40.00% | 61.22% | 78.58% | 88.24% | 92.12% | 94.62% | 97.52% | 99.34% | 99.96% | 99.98% |
|                | SF 8 | 23.98% | 38.66% | 58.98% | 78.38% | 93.64% | 98.66% | 99.44% | 99.78% | 99.98% | 100.00% | 100.00% |
|                | SF 9 | 24.14% | 40.90% | 65.96% | 85.06% | 92.94% | 95.72% | 97.66% | 98.84% | 99.92% | 99.98% | 99.98% |
| GRU            | SF 7 | 23.82% | 35.86% | 58.42% | 77.04% | 87.38% | 91.54% | 94.28% | 97.28% | 99.38% | 99.90% | 99.96% |
|                | SF 8 | 22.88% | 34.56% | 55.08% | 75.36% | 91.28% | 97.88% | 99.26% | 99.78% | 100.00% | 100.00% | 100.00% |
|                | SF 9 | 22.08% | 37.48% | 62.10% | 81.52% | 91.52% | 95.28% | 96.64% | 98.44% | 99.44% | 99.90% | 99.94% |
| Transformer    | SF 7 | 21.98% | 35.86% | 56.22% | 76.46% | 87.78% | 92.56% | 94.64% | 97.88% | 99.26% | 99.84% | 99.92% |
|                | SF 8 | 21.52% | 34.14% | 51.60% | 73.96% | 90.98% | 97.74% | 99.38% | 99.78% | 100.00% | 100.00% | 100.00% |
|                | SF 9 | 21.08% | 37.06% | 61.90% | 81.22% | 91.58% | 95.02% | 96.92% | 98.72% | 99.54% | 99.96% | 99.98% |

 conductor device by device. Each LoRa DUT has the SF set to 7, 8, and 9 in turn and 3,000 packets are transmitted per SF.

2) Dataset Description: The experimental dataset contains 90,000 packets from 10 DUTs under three SF configurations. 75,000 of them (25,000 for each SF) are used for training and the rest 15,000 packets (5,000 packets for each SF) are for testing. The preamble part of packets with SF 7, 8, and 9 contains IQ samples with lengths of 2,048, 4,096, and 8,192, respectively. The SNR of all the received packets is estimated to be around 70 dB. Therefore, we treat them as noiseless signals and add AWGN to them to simulate different SNR conditions during the test.

3) Neural Network Training Details: The neural networks are trained according to the pipelines described in Section III-E. 10% training packets are randomly selected for validation. Adam is used as the optimizer and the mini-batch size is set to 32 [41]. A learning rate scheduler is adopted. The initial learning rate is 0.001 and is reduced by a factor of 0.2 every time the validation loss stops decreasing for five epochs. The training stops when the validation loss plateaus for ten epochs. The neural networks are implemented with the Tensorflow and Keras libraries. The training was carried out on a PC with an NVIDIA GeForce GTX 1660 GPU and an Intel Core i7-9700 CPU.

B. Performance of Different SF Configurations

The proposed flatten-free CNN, LSTM, GRU, and transformer models can process channel-independent spectrograms under different SF configurations (Fig. 5). Their classification results are shown in Table I.

The four sub-figures demonstrate that there is no evident performance gap among the four models. In addition to this, there is also no significant difference when classifying signals of different SFs.

In terms of accuracy, the RFFI performance is nearly perfect when SNR is over 30 dB. The accuracy is reduced to 60% at 10 dB, but the multi-packet inference can increase it to 90%, which will be shown in Fig. 10. Fig. 9 shows the confusion matrices of the flatten-free CNN model at various SNRs, which provides detailed classification results. Fig. 9(a) indicates that the RFFI protocol cannot successfully classify LoRa DUTs at 0 dB. The best case is DUT 4, where 218 out of the 500 packets are correctly classified. The worst case is DUT 9, where only 17 out of 500 packets are correctly classified. When the SNR is increased to 10 dB, as shown in Fig. 9(b), we can see that all the predictions are in the red boxes, which means the RFFI protocol is capable of distinguishing the types of DUT. More specifically, a LoPy4 DUT will rarely be classified as a Dragino LoRa shield since their characteristics are different. However, the system still cannot classify LoRa DUTs from the same manufacturer with high accuracy. In contrast, most DUTs can be correctly classified at the higher 20 dB, while only the accuracy of DUT 1 is unsatisfactory. This is probably because DUT 1 has very similar characteristics to DUT 2 and DUT 4. Therefore, misclassification can still occur even at 20 dB. At the nearly ideal 40 dB shown in Fig. 9(e), the RFFI protocol achieves perfect classification accuracy for all the DUTs.

C. Comparison of Augmentation Strategies

We trained each deep learning model (flatten-free CNN, LSTM, GRU and transformer) using online, offline and no augmentation strategies, respectively. For clarity, we only use SF 7 signals during inference to compare the performance of different augmentation strategies, respectively. For clarity, we only use SF 7 signals during inference to compare the performance of different augmentation strategies, since there is no significant performance gap among different SF settings.

Their performance at different SNRs is shown in Table II.
The models trained with online augmentation outperform the ones trained with offline augmentation since the models can learn many more noisy signals during the online augmentation compared to the offline one. We take the training of the flatten-free CNN as an example. The training with online augmentation stops after 140,000 steps. Therefore, the flatten-free CNN has learned 140,000 × 32 (mini-batch size) noisy signals. In contrast, offline augmentation can only provide 75,000 (training set size) noisy signals to the flatten-free CNN for learning. Although replicating the training dataset several times may increase the number of noisy signals to the same level as online augmentation, it will significantly increase the disk storage size and memory requirements. In our case, when using offline training, we need to replicate the dataset around 60 times to match the number of noisy signals provided during online training, which is unaffordable for a large-scale dataset. It can also be observed that the models trained without any augmentation perform well at high SNRs but poorly at low SNRs. This demonstrates data augmentation must be employed during training, otherwise, the RFFI protocol may not work even at 20 dB.

Online augmentation requires more training time than the offline one. Take the training of flatten-free CNN as an example, online augmentation costs 200 minutes, while offline costs only 70 minutes. The increasing training time can be tolerated as the neural network can be trained on a powerful workstation or cloud server and then deployed to the gateway.

D. Effect of Multi-Packet Inference

The proposed multi-packet inference approach can significantly improve RFFI performance, particularly in low SNR scenarios. The effect of packet number on classification accuracy is studied at four different SNRs, and the results are given in Fig. 10. As shown in the figure, the multi-packet inference can increase the classification accuracy of the four models by over 20% at 0 dB, 5 dB, and 10 dB, while obtaining marginal improvement at 20 dB. This demonstrates that multi-packet inference can be adopted in low SNR scenarios to effectively improve system performance. Furthermore, we can find that the classification accuracy gradually improves as the number of packets increases. However, the improvement is relatively limited after the number of packets exceeds 10. A trade-off should be considered as involving more packets also leads to higher complexity and requires more space to store historical inferences.

E. Complexity Analysis

We investigate the model complexity in terms of inference time and the number of parameters. Inference time determines...
the latency of the system, which is an important metric for RFFI. The number of parameters affects the size of the neural network. Larger models require more storage and memory for the embedded platform. As shown in Table III, their inference time on SF 7 signals is not much different. The difference is at most 2 ms. However, the gap becomes apparent when classifying the longer SF 9 signals. The transformer costs 33.8 ms, which is 9 ms faster than the slowest LSTM network.

Note that the inference time is limited by the deep learning library used and can be further accelerated. The transformer is also the most lightweight model, with only 348,938 learnable parameters.

Fig. 11. Sliced channel independent spectrograms of LoRa preambles. (a) SF 7 preambles are split into eight slices. (b) SF 8 preambles are split into 16 slices. (c) SF 9 preambles are split into 32 slices.

F. Results Summary

1) Comparison Among Neural Networks: The transformer can achieve competitive performance with the least parameters and the shortest inference time. Table I shows that the proposed four models are similar in classification accuracy when classifying signals of different SFs. Table III demonstrates that the transformer is the most lightweight model and its inference speed is also the fastest. Therefore, the transformer is the recommended neural network considering both performance and complexity.

2) Solutions to Low SNR RFFI: The RFFI performance in low SNR scenarios can be improved in two ways. One is applying online augmentation during training to increase model noise robustness, and another is merging the results of multiple packets/observations during inference. Online augmentation increases the training overhead which can be resolved by training on a cloud server. Using multi-packet inference is effective in low SNR scenarios, which can significantly improve the accuracy by over 20%.

VI. COMPARISON WITH Slicing Technique

An alternative solution to the variable-length input problem is the slicing/splitting technique, which is employed in [23], [24], [27], and [28]. In this section, we compare the performance of the slicing/splitting technique with our proposed length-versatile neural networks.

A. Slicing Technique

Slicing/splitting is to divide the received signal into shorter slices/segments with equal length. During the inference, the CNN infers from each slice individually and the softmax outputs of all the slices are averaged to predict the label. In this scheme, the CNN always accepts equal-length slices as inputs so that the variable-length problem can be avoided.

In our implementation, we divide the preamble part of each LoRa packet into several slices of 256 IQ samples and then convert them to $(64, 6)$ channel-independent spectrograms as CNN inputs. The SF 7, 8, and 9 signals contain 2,048, 4,096, and 8,192 IQ samples, respectively. Therefore, SF 7 preamble is split into eight slices, each corresponding to one LoRa SF 7 preamble. SF 8 and SF 9 preamble parts are divided into 16 and 32 slices, respectively. Each SF 8 slice contains half a LoRa SF 8 preamble and each SF 9 slice contains a quarter of a LoRa SF 9 preamble. The slices can be viewed in Fig 11.

We designed an ordinary CNN with a flatten layer to process the slices, this is termed the slicing CNN, whose structure is shown in Fig. 12. The slicing CNN has a flatten layer, but in the flatten-free CNN, it is replaced with a global average pooling 2D layer, which is the only difference between
these two architectures. Therefore, we can assume they have similar learning abilities. The slicing CNN is trained using exactly the same settings as the flatten-free CNN, including optimizer, learning rate, mini-batch size and stop conditions, etc. Therefore, a fair comparison can be conducted.

B. Performance Comparison

The performance comparison between the slicing CNN and the flatten-free CNN is shown in Fig. 13. The flatten-free CNN performs better than the slicing CNN in all three SF configurations. Although both of them can achieve near-perfect performance when SNRs are above 30 dB, the accuracy of flatten-free CNN is about 10% higher than the slicing technique at low SNRs. The reason for this performance gap will be discussed in Section VI-C.

We also analyzed the complexity of the slicing CNN, which is included in Table III. It has 797,194 parameters in total. The inference time (sum of all slices) of SF 7, 8 and 9 signals is 31.3, 33.9 and 46.4 ms, respectively. Compared to the complexity of the flatten-free CNN shown in Table III, the slicing CNN requires more storage space and longer inference time.

C. Characteristic Differences Among Preambles

In this section, we discuss the reason for the performance gap between the slicing and our proposed length-versatile techniques. We experimentally show that the characteristics of the eight LoRa preambles are different from each other, even though their payloads are the same. Therefore, when using length-versatile techniques, the input to the flatten-free CNN contains all eight preambles with different characteristics.

Then the flatten-free CNN can learn the correlation among the eight feature-different preambles. In contrast, the slicing technique only processes a short section of the signal each time.

We take SF 7 signals as an example to show that the eight preambles in a LoRa packet are different in characteristics. As shown in Fig. 11, the SF 7 channel independent spectrogram is splitting into eight slices, each slice corresponding to a LoRa preamble. We then trained eight CNNs with preambles in different positions. More specifically, CNN 1 is trained with the first preamble of all the training packets, while CNN 2 is trained with the second preamble, and so forth. Then we use the preamble in different positions to test the eight trained CNN models. The classification results are given in Table IV. The results demonstrate that the CNN trained with preambles in a specific position cannot accurately classify the preambles in other positions. For instance, the CNN 1 model is trained with the first preamble and it only achieves 49% accuracy in classifying the second preamble, which reveals that the first preamble has different characteristics from the second one.

Fig. 14 shows the waveform of the first and second LoRa preambles. They are almost identical except for the part in the purple box. The start of the first preamble shows a gradual increase in signal amplitude. We presume this is the signal transient part that is generated when the hardware components
are powered on, which serves as features in traditional RFFI techniques [10], [42]. Typically, these transient-based methods require equipment with a high sampling rate to collect sufficient IQ samples of the transient part. However, Fig. 14 demonstrates that the relatively low-cost USRP SDR can capture it as well. The first preamble contains the signal transient part, which differs from the other preambles. Therefore, as shown in Table IV, the CNN 1 that is trained with the first preamble does not work well on the others.

Even though the difference from the second to the eighth preamble cannot be observed directly from the waveform, it can be revealed by the classification results in Table IV. We highlight the accuracy above 90% and find that they are distributed along the diagonal. This demonstrates that the characteristics of the preamble vary in different positions. The above discussion shows that the characteristics of slices (preambles) are different even though each preamble is modulated with the same payload. Therefore, the flattened-free CNN can learn the difference/correlation among them since it processes eight preambles simultaneously. In contrast, the slicing CNN only receives a short slice each time, and all per-slice inferences are merged by simply averaging. This disables the CNN’s ability to learn the correlation among feature-different preambles, which results in the worse performance of the slicing CNN shown in Fig. 13.

VII. CONCLUSION

The research reported here aims to tackle the variable input size and low SNR problems in deep learning-based RFFI protocols. We use LoRa as a case study because it suffers from both of these challenges. LoRaWAN ADR mechanism results in the variable lengths of LoRa preambles, which requires the model to be able to process inputs of different lengths. Therefore, we present four length-versatile neural networks, namely flattened-free CNN, LSTM, GRU, and transformer. LoRa is an LPWAN technology whose received signals may have low SNR, making LoRa-RFFI challenging. Data augmentation can be an efficient approach to boost model noise robustness. We compare the performance of online, offline, and no augmentation strategies and found that online augmentation outperforms the others. Furthermore, we leverage a multi-packet inference approach that can considerably improve the system performance in low SNR scenarios. Experiments involving 10 LoRa devices and a USRP N210 SDR were carried out for evaluation. The results show that online augmentation and multi-packet inference are effective in improving the RFFI performance in low SNR conditions.

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