Multi-gigahertz operation of photon counting InGaAs avalanche photodiodes

Z. L. Yuan, A. W. Sharpe, J. F. Dynes, A. R. Dixon and A. J. Shields

Toshiba Research Europe Ltd, Cambridge Research Laboratory, 208 Cambridge Science Park, Milton Road, Cambridge, CB4 0GZ, UK

(Dated: 6 January 2010)

We report a 2 GHz operation of InGaAs avalanche photodiodes for efficient single photon detection at telecom wavelengths. Employing a self-differencing circuit that incorporates tuneability in both frequency and arm balancing, extremely weak avalanches can now be sensed so as to suppress afterpulsing. The afterpulse probability is characterized as 4.84% and 1.42% for a photon detection efficiency of 23.5% and 11.8%, respectively. The device will further increase the secure bit rate for fiber wavelength quantum key distribution.

Semiconductor InGaAs avalanche photodiodes (APDs) offer a practical solution for single photon detection in the near infrared, because they are rugged and compact, cryogen-free and of low power consumption. Due to their practicality, InGaAs APDs are widely used in a number of important applications, such as quantum key distribution as well as eye-safe laser ranging and semiconductor device characterization. Despite their afterpulse problem, the photon counting performance has been steadily improved over the past decade. Under gated mode, the gating frequency has been increased from initially 100 kHz to presently over 1 GHz due mainly to advances in detection circuitry. A count rate of 497 MHz has recently been demonstrated.

Detecting weak avalanches is the key to high speed single photon detection. In gated mode, weak avalanches are usually not immediately accessible because of capacitive response by the APD to the gate signal. They can be recovered through removal of the capacitive response, for which both sine-wave gating and self-differencing (SD) have been demonstrated as highly effective. Sine-wave gating relies on effectively filtering the capacitive response that is purely sinusoidal, while SD removes the capacitive signal by subtracting an identical copy that is delayed by an integer number of clock cycles. Contrary to the sine wave gating, any form of gating, including sine waves and low-duty cycle pulses, is acceptable by an SD as long as it is periodical. Through SD or sine-wave gating, gigahertz operation has been achieved with their respective performances summarized in Table I. InGaAs APDs have been applied to Mbits/s secure key rate quantum key distribution generation of high quality random numbers and long distance entanglement distribution.

As pointed out in Ref. 8, continuous frequency tuning was not possible in the original SD circuit, preventing an ideal performance due to the difficulty in perfect frequency match to a given photon source. This inflexibility can be totally relieved by adding adjustability into the circuit, as shown in Fig. 1(a). In comparison to the original SD, the tuneable circuit replaces the 50/50 splitter with an adjustable splitter and also a co-axial line stretcher. Adjustability in the splitting ratio is realized by adding a potentiometer in a triangular resistive splitter. The line stretcher has a tuneable delay range of 45 ps, thus allowing the SD to match a frequency within the range from 0.987 to 1.033 GHz with an arbitrarily high precision. Thanks to the adjustability in splitting ratio and frequency, the SD circuit allows greater suppression of the capacitive response. The suppression is measured to be 62 dB for the 1 GHz frequency compo-

![FIG. 1. (a) Schematic for a tunable SD circuit with 1-ns delay; LS: line-stretcher. (2) Avalanche waveforms for dark events recorded by an oscilloscope after the 1-ns SD circuit is turned on.](https://example.com)
FIG. 2. Photon current and count rate as a function of the laser pulse delay measured by an InGaAs APD under a square wave gating of (a) 1 GHz, (b) 2 GHz, and (c) 3 GHz.

ments. The overall cancelation\(^{15}\) for the capacitive response is approximately 38 dB, which is 17 dB better than the non-tuneable SD circuit\(^{8}\).

In what follows, we report a significantly improved performance of InGaAs APDs due to the tuneable SD. A gating frequency of 2 GHz operation is demonstrated for single photon counting with a high detection efficiency of 23.5% and a low afterpulse probability of 4.85%. As compared with our previous results (see Table I), the present detector operates at almost twice the frequency and twice the efficiency.

The InGaAs APD under test is cooled electrically to \(-30^\circ\text{C}\). A square wave in combination with a DC voltage source is used as the gating signal to bias the device, as illustrated in Fig. 1(a). The APD signal, after passing through the SD, is amplified before analysis by an oscilloscope or a time-correlated photon counter. The electrical current through the DC path is monitored.

The superiority of the tuneable SD is visible in the traces of avalanches that are due to dark events. The use of dark events helps to avoid complications due to the avalanche amplitude dependence on photon number\(^{16}\) which may give rise to a false confidence in the circuit performance. Figure 1(b) shows three dark events obtained with a 1-ns delay SD when the APD is gated at a frequency of 1, 2 and 3 GHz, respectively. For all frequencies, the SD shows an excellent level of cancelation. The uncanceled background is much weaker than the avalanche signal, which can be straightforwardly discriminated.

For high speed gating to be useful, it is important for an APD not to react strongly to photons arriving between gates, but only be sensitive to those arriving within an electrical gate. To verify this, we illuminate the APD with a 1 GHz pulsed laser whose intensity is attenuated to 3 pW, corresponding to 0.023 photons/pulse. The APD gating is synchronized with the laser at its harmonic frequencies of 1, 2, and 3 GHz. Photocurrent, as well as photon count response, is recorded as the laser pulse delay is varied relative to the gating. As shown in Fig. 2, both responses show very similar behavior: the measured values vary with the laser pulse delay, forming a series of peaks whose frequency corresponds well with the respective APD gating period. These peaks are attributed to the delays with which the incident photons coincide with the APD gates. Under such conditions, each absorbed photon have a high probability of inducing a macroscopic avalanche. When the laser pulse is tuned between the gates, the photo responses, including photocurrents, fall close to their dark values, suggesting that the off-gate photons have a negligible avalanche gain or detection probability. The full width half maximum (FWHM) for each response peak is approximately 100 ps, showing little dependence on the gating frequency.

To characterize the photon counting performance, the APD is illuminated with 50-ps optical pulses from a 1550 nm semiconductor laser diode synchronized at 1/64 of the APD gating frequency. The laser power is attenuated to 0.128 pW, corresponding a photon flux of \(\mu = 1.0 \times 10^6 \text{s}^{-1}\), before launching into the fiber pigtail of the APD. At 1 and 2 GHz gating, each illuminating pulse contains 0.064 and 0.032 photons, respectively, thus

FIG. 3. (a) Detector dark count probability \(P_d\), and (b) afterpulse probability \(P_a\) as a function of the net detection efficiency \(\eta_{\text{net}}\).
ensuring operation in the single photon counting regime. By fixing the gating amplitude at 7.1 V, we characterize the raw count rate \( (C) \), dark count \( (P_d) \), and afterpulse \( (P_a) \) probabilities as function of the APD DC voltage bias, using the method described previously. We point out that the dark count probability is measured per gate, while the afterpulse probability is measured per photon count. Excluding the dark and afterpulse counts, the net detection efficiency \( (\eta_{net}) \) is obtained using

\[
\eta_{net} = \frac{(C - P_d f)}{1 + P_a} \tag{1}
\]

where \( f \) is the clock frequency.

The characterization results, plotted in Fig. 2, shows a better performance for a gating frequency of 2 GHz than 1 GHz. For similar detection efficiencies, the APD shows much lower values for both \( P_d \) and \( P_a \) for 2 GHz operation. The detection efficiency is considerably better than previous results. For \( P_a < 10\% \), the net efficiencies are 27.8\% (\( P_a = 8.8\% \)) and 25.7\% (\( P_a = 7.2\% \)) for 1 and 2 GHz respectively. The corresponding dark count probabilities are \( 2.90 \times 10^{-5} \) and \( 1.71 \times 10^{-5} \). The afterpulse probability is also significantly smaller. At \( \eta_{net} = 11.8\% \), a typical efficiency for InGaAs APDs, the afterpulse probability is measured to be 1.43\% for 2 GHz. This value is four times lower than our previously reported value and is also 50\% lower than the recent value reported for a 1.5 GHz sine-wave gating APD. At an elevated efficiency of \( \eta_{net} = 23.5\% \), \( P_a \) only increases slightly to 4.84\%. The reduced afterpulse probability is advantageous for error-sensitive applications, such as quantum key distribution.

The reduced afterpulse probability agrees well with the reduction in the avalanche charge \( q \). The relation between \( P_a \) and \( q \) is plotted in Fig. 4. The value of \( q \) is obtained using the ratio of photo-current to count rate, representing the total charge per detection event that flows through the device. Two distinctive relations are observed: (i) \( P_a \) that decreases with \( q \) and (ii) a linear relation. The former is an artifact, because these data are obtained with low efficiency (see inset of Fig. 4), or low bias conditions, under which \( q \) is highly overestimated because a considerable fraction of avalanches are not sufficiently strong for detection. The latter is the actual relation, which explicitly reveals that the afterpulse probability increases linearly with the avalanche charge. The lowest \( q \) is measured as 0.068 pC for 1 GHz. The value is less than 50\% of a previous value measured at a similar frequency. For 2 GHz, \( q \) becomes even lower with a value of 0.035 pC recorded. This significant reduction in \( q \) helps the APD to achieve high efficiency with minimal afterpulse at high frequencies, and is the origin of the lower afterpulse rate at higher gating frequencies.

As summarized in Table 1, the present APD operates almost double the frequency and also double the efficiency, compared to our previous result. Moreover, the higher clock frequency does not require a more stringent synchronization with the laser. Based on the detector response curves in Fig. 2, synchronization with the laser requires the same precision at 2 GHz as 1 GHz, which is already demonstrated in high bit rate quantum key distribution and long distance entanglement distribution. As a result, the improved performance of the detector is expected to allow a four-fold increase in the secure bit rate of quantum key distribution.

An outstanding issue remains to be addressed for multi-gigahertz gating. For this APD, the timing jitter is measured as 100, 120 and 380 ps for 1, 2 and 3 GHz gating, respectively, showing an increasing trend with gating frequency. The amount of jitter is acceptable for 1 and 2 GHz, but is prohibitively high for 3 GHz because it is considerably greater than the clock period of 333 ps at this frequency. The jitter, and thus the maximum gating frequency, of the detector is limited by the bandwidths of the electronics, the diode itself, or a combination of both. In the future it is intended to employ devices, as well as electronics, of higher bandwidths to circumvent this issue.

In conclusion, we have demonstrated high efficiency single photon detection at telecom wavelengths using self-differencing InGaAs APDs operated at multi-gigahertz gating frequencies. The device shows a detection efficiency of 23.5\% when operating at 2 GHz and with an afterpulse probability of 4.84\%. The enhanced efficiency and speed will significantly increase the bit rate of quantum key distribution.

\[1\] A. R. Dixon, Z. L. Yuan, J. F. Dynes, A. W. Sharpe, and A. J. Shields, Opt. Express 16, 18790 (2008).
\[2\] T. Maruyama, F. Narusawa, M. Kudo, and M. Tanaka, Opt. Eng. 41, 395 (2002).
\[3\] M. B. Ward, O. Z. Karimov, D. C. Unitt, Z. L. Yuan, P. See, D. G. Gevaux, A. J. Shields, P. Atkinson, and D. A. Ritchie, Appl. Phys. Lett. 86, 201111 (2005).
\[4\] D. S. Bethune and W. P. Risk, IEEE J. Quant. Electron. 36, 340 (2000).
\[5\] A. Tomita and K. Nakamura, Opt. Lett. 27, 1827 (2002).
6A. Yoshizawa, R. Kaji, and H. Tsuchida, Jpn. J. Appl. Phys. 43, L735 (2004).
7N. Namekata, S. Sasamori, and S. Inoue, Opt. Express 14, 10043 (2006).
8Z. L. Yuan, B. E. Kardynal, A. W. Sharpe, and A. J. Shields, Appl. Phys. Lett. 91, 041114 (2007).
9A. R. Dixon, J. F. Dynes, Z. L. Yuan, A. W. Sharpe, A. J. Bennett, and A. J. Shields, Appl. Phys. Lett. 94, 231113 (2009).
10J. Zhang, R. Thew, C. Barreiro, and H. Zbinden, Appl. Phys. Lett. 95, 091103 (2009).
11L. Xu, E. Wu, X. Gu, Y. Jian, G. Wu, and H. Zeng, Appl. Phys. Lett. 94, 161106 (2009).
12N. Namekata, S. Adachi, and S. Inoue, Opt. Express 17, 6275 (2009).
13J. F. Dynes, Z. L. Yuan, A. W. Sharpe, and A. J. Shields, Appl. Phys. Lett. 93, 031109 (2008).
14J. F. Dynes, H. Takesue, Z. L. Yuan, A. W. Sharpe, K. Harada, T. Honjo, H. Kamada, O. Tadanaga, Y. Nishida, M. Asobe, and A. J. Shields, Opt. Express 17, 11440 (2009).
15The cancelation is defined here as 20 log $v_2/v_1$, where $v_2$ and $v_1$ are peak-peak amplitudes for the remaining capacitive response after the SD with one arm broken or both arms attached respectively.
16B. E. Kardynal, Z. L. Yuan, and A. J. Shields, Nature Photon. 2, 425 (2008).
17Z. L. Yuan, A. R. Dixon, J. F. Dynes, A. W. Sharpe, and A. J. Shields, New J. Phys. 11, 045019 (2009).