Ultra-Low-Complexity Algorithms with Structurally Optimal Multi-Group Multicast Beamforming in Large-Scale Systems

Chong Zhang, Student Member, IEEE, Min Dong, Senior Member, IEEE, and Ben Liang, Fellow, IEEE

Abstract—In this work, we propose ultra-low-complexity design solutions for multi-group multicast beamforming in large-scale systems. For the quality-of-service (QoS) problem, by utilizing the optimal multicast beamforming structure obtained recently in [2], we convert the original problem into a non-convex weight optimization problem of a lower dimension and propose two fast first-order algorithms to solve it. Both algorithms are based on successive convex approximation (SCA) and provide fast iterative updates to solve each SCA subproblem. The first algorithm uses a saddle point reformulation in the dual domain and applies the extragradient method with an adaptive step-size procedure to find the saddle point with simple closed-form updates. The second algorithm adopts the alternating direction method of multipliers (ADMM) method by converting each SCA subproblem into a favorable ADMM structure. The structure leads to simple closed-form ADMM updates, where the problem in each update block can be further decomposed into parallel subproblems of small sizes, for which closed-form solutions are obtained. We also propose efficient initialization methods to obtain favorable initial points that facilitate fast convergence. Furthermore, taking advantage of the proposed fast algorithms, for the max-min fair (MMF) problem, we propose a simple closed-form scaling scheme that directly uses the solution obtained from the QoS problem, avoiding the conventional computationally expensive method that iteratively solves the inverse QoS problem. We further develop lower and upper bounds on the performance of this scaling scheme. Simulation results show that the proposed algorithms offer near-optimal performance with substantially lower computational complexity than the state-of-the-art algorithms for large-scale systems.

Index Terms—Multicast beamforming, optimal beamforming structure, large-scale optimization, extragradient algorithm, alternating direction method of multipliers, low complexity.

I. INTRODUCTION

Content distribution through wireless multicasting has been increasingly popular among new wireless services and applications and is expected to dominate future wireless traffic. Multi-antenna multicast beamforming is an efficient transmission technique to support high-speed content distribution to multiple user groups simultaneously. With massive multiple-input multiple-output (MIMO) being the essential technology for future networks [3], it is critical for multicast beamforming solutions to be scalable to meet the ultra-low complexity requirement for large-scale systems.

Downlink multicast beamforming has been studied for traditional multi-antenna systems in various scenarios, such as single-group or multi-group multicasting [4]–[6], multi-cell networks [7], [8], relay networks [9], [10], and cognitive radio networks [11], [12]. Multicast beamforming problems are challenging to solve, as they are generally non-convex and NP-hard [4]. Semi-definite relaxation (SDR) has been a prior state-of-the-art method considered in the existing works [13]. For traditional multi-antenna systems with relatively small problem sizes, SDR provides a good approximate solution [4]–[6], [8]. However, SDR is not a scalable method for large-scale problems, where the computational complexity becomes very high and the performance deteriorates significantly as the problem size grows.

With the increasing number of transmit antennas, successive convex approximation (SCA) [14] has become a more attractive approach for solving the multi-group multicast beamforming problems due to its computational and performance advantages over SDR [15]–[17]. SCA-based methods convexify the non-convex problem into a sequence of convex approximation subproblems, where the convex subproblems are typically solved by the interior-point method (IPM) [18]. However, IPM is a second-order algorithm. For computing a multicast beamforming solution in large-scale massive MIMO systems, using IPM still results in a relatively high computational complexity.

Following this, several methods have been proposed to further reduce the computational complexity at each SCA iteration. For the single-group case, first-order methods have been proposed to solve the convex subproblems [19], [20]. However, these methods are not directly applicable to multiple groups with inter-group interference. For the multi-group scenario with per-antenna power constraints, the alternating direction method of multipliers (ADMM) [21] has been considered for solving the subproblem in each SCA iteration. In [22], zero-forcing pre-processing for interference elimination has been proposed to reduce the multi-group case to a single-group equivalent problem, for which SCA is applied. These methods provide much lower complexity than the original SCA-based method. However, since the dimension of beamforming vectors is dictated by the number of antennas, the computational complexity of these methods still grows with the number of antennas in polynomial time. This renders these methods still computationally costly for massive MIMO systems. Alternatively, for multi-cell systems, low-complexity beamforming schemes using weighted maximum ratio transmission (MRT) in combination with SDR have been developed, where only the weights, one for each user, need to be optimized, and thus the size of the optimization problem is reduced [23], [24]. Despite all the above advancements in computational algorithms, they do not optimally utilize the multicast beamforming structure.

The optimal multi-group multicast beamforming structure has been recently obtained in [2]. It is shown that the optimal solution is a weighted minimum mean-square error (MMSE) filter with an inherent low-dimensional structure for the unknown weights to be computed. With this structure, the mul-
Multicast beamforming problems considered in all the above-mentioned works typically are cast into two problem formulations: a quality-of-service (QoS) problem for transmit power minimization with signal-to-interference-and-noise (SINR) guarantees for all users, or a max-min fair (MMF) problem for maximizing the minimum SINR among all users subject to a transmit power budget. Although both types of problems are non-convex and NP-hard, the MMF problem is more complicated to solve than the QoS problem. Typically, the solution to the MMF problem is obtained through iteratively solving its inverse QoS problem along with a bi-section search over the value of minimum SINR [2], [5], [6], [17], [21]. This additional layer of bi-section iterations results in high computational complexity for the MMF problem in large-scale systems. Therefore, developing a low-complexity method to obtain a good solution for the MMF problem is also critically important.

Besides the above-mentioned works on algorithm development for multicast beamforming design, asymptotic multicast beamforming in massive MIMO systems has been analyzed in [25], [26] without inter-group interference consideration, and in [2] with inter-group interference consideration. Multicast beamforming has also been investigated for energy efficiency maximization [27] and for joint unicast and multicast transmission in massive MIMO systems [28], [29]. For overloaded systems with fewer antennas than the users, the rate-splitting-based multicast beamforming strategies have been proposed [30]–[32]. These studies address different problems from the one considered in our work.

A. Contributions

In this paper, for downlink multi-group multicast beamforming in large-scale massive MIMO systems, we propose two fast first-order algorithms for the QoS problem, which are scalable in both antenna and user dimensions. Utilizing the optimal multicast beamforming structure, we convert the original QoS problem into a non-convex weight optimization problem of a much lower dimension. We then develop two fast algorithms to solve this weight optimization problem to obtain our multicast beamforming solution. The two algorithms are SCA-based methods, referred to as the extragradient-based SCA (ESCA) and ADMM-based SCA (ASCA). They are developed using two different first-order optimization techniques. Using these algorithms, we also propose a simple closed-form scaling scheme for solving the MMF problem. The main novelty and contribution of this work are summarized below:

- We propose ESCA to solve each SCA subproblem in the dual domain. In particular, we construct a saddle point reformulation of the dual problem, which can be further cast as a variational inequality problem for us to apply the extragradient method [33] to find the saddle point. Instead of directly considering the primal problem, our approach explores the dual problem where the extragradient method can be used efficiently, and we obtain simple closed-form gradient updates in each updating step, which is the key for our algorithm to compute the solution with low complexity. Furthermore, to avoid finding the Lipschitz constant of the gradient function to set the step size, which is generally difficult to obtain, we adopt a prediction-correction procedure for an adaptive step size in each update to ensure convergence to the saddle point for the optimal solution. We also consider two initialization methods, a fast extragradient-based initialization method and an SDR-based method, for generating favorable initial feasible points for ESCA to facilitate fast convergence.

- We also propose a fast ADMM-based algorithm for the SCA subproblems, referred to as ASCA. We transform each SCA subproblem into a favorable form to construct the ADMM procedure with two ADMM updating blocks. In particular, the structure of the transformed problem leads to simple updates in each ADMM update block, where the problem in each update block can be further decomposed into parallel subproblems of small sizes, each of which yields a closed-form solution. Thus, ASCA takes advantage of the closed-form updates in the ADMM procedure for fast computation and is guaranteed to converge to the optimal solution. The similar procedure is used to provide a ADMM-based fast initialization method to facilitate fast convergence of the algorithm.

- Taking advantage of the proposed fast algorithms for the QoS problems, we propose a simple closed-form scaling scheme to obtain a multicast beamforming solution for the MMF problem. The scheme directly scales the beamforming solution obtained from the QoS problem to meet the transmit power budget. It thus has substantially lower computational complexity than the conventional method of iteratively solving the inverse QoS problem with bi-section search. We also provide lower and upper bounds on the performance of the scaling scheme.

- Simulation results demonstrate that both ESCA and ASCA with their initialization methods provide near-optimal performance for the QoS problem. Their computational complexity is substantially lower than the state-of-the-art algorithms for large-scale systems as the number of antennas and users increases, demonstrating the algorithm scalability in large-scale systems. Between the two algorithms, ASCA is preferable with faster computation for small to moderately large systems where the number of antennas is less than 300 and the number of users per group is less than 20, while ESCA provides faster computation for larger systems. In addition, the proposed scaling scheme for the MMF problem is shown to result in only a mild loss to the optimal performance as the number of antennas becomes large but substantially faster to compute the solution.

We note that a multicast beamforming problem under per-antenna power constraints has been considered in [21], where an ADMM-based algorithm has been proposed. Besides the problem being different from ours, the optimal beamforming
structure is not considered there, and the iterative algorithm targets directly computing the beamforming vector. The resulting algorithm in [21] has three ADMM updating blocks, while our ASCA contains two ADMM blocks. In general, different from the existing algorithms [19]–[24], our proposed two algorithms, ESCA and ASCA, exploit both optimal multicast beamforming structure and the numerical optimization techniques, which lead to the simple closed-form updating steps to compute the multicast beamforming solution with ultra-low computational complexity for large-scale systems.

B. Organization and Notations

The rest of this paper is organized as follows. Section II presents the system model and problem formulation for multi-group multicast beamforming. Section III reviews the optimal multicast beamforming structure and the SCA method. In Sections IV and V, we present two fast algorithms, ESCA and ASCA, for the QoS problem. In Section VI, we propose a simple closed-form scaling scheme to find the solution for the MMF ASCA, for the QoS problem. In Section VII, we present two fast algorithms, ESCA and ASCA, for the problem. In Section VIII, we propose a simple closed-form scaling scheme to find the solution for the MMF problem. Simulation results are provided in Section VII, and the conclusion is presented in Section VIII.

Notations: Hermitian, transpose, and conjugate are denoted as $(\cdot)^H$, $(\cdot)^T$, and $(\cdot)^*$, respectively. The real and imaginary parts of a complex number are denoted as $\Re\{\cdot\}$ and $\Im\{\cdot\}$, respectively. The Euclidean norm of a vector is denoted as $\|\cdot\|$. The notation $x \succ 0$ indicates element-wise non-negative. The identity matrix is denoted as $I$. The notation $z \sim \mathcal{CN}(0, C)$ means $z$ is a complex Gaussian random vector with zero mean and covariance $C$. The non-negative real Euclidean space is denoted as $\mathbb{R}_+$. 

II. SYSTEM MODEL AND PROBLEM FORMULATION

We consider a downlink multi-group multicast beamforming scenario, where the base station (BS) equipped with $N$ antennas provides service for $G$ user groups. Each group receives a common message that is independent of the messages to other groups. Denote the set of group indices by $G = \{1, \ldots, G\}$. Assume that there are $K_i$ single-antenna users in group $i$, and the set of user indices in the group is denoted by $K_i = \{1, \ldots, K_i\}, i \in G$. Define the total number of users in all groups as $K_{\text{tot}} = \sum_{i=1}^{G} K_i$.

Let $h_{ik} \in \mathbb{C}^{N \times 1}$ be the channel vector from the BS to user $k$ in group $i$, for $k \in K_i, i \in G$. Let $w_i \in \mathbb{C}^{N \times 1}$ be the multicast beamforming vector for group $i \in G$. The received signal at user $k$ in group $i$ is given by

$$y_{ik} = w_i^H h_{ik} s_i + \sum_{j \neq i} w_j^H h_{ik} s_j + n_{ik}$$

where $s_i$ is the data symbol transmitted to group $i$ with $E(|s_i|^2) = 1$, and $n_{ik}$ is the receiver additive white Gaussian noise with zero mean and variance $\sigma^2$. The received SINR at user $k$ in group $i$ is expressed as

$$\text{SINR}_{ik} = \frac{|w_i^H h_{ik}|^2}{\sum_{j \neq i} |w_j^H h_{ik}|^2 + \sigma^2}.$$  

The transmit power at the BS is given by $P_i = \sum_{i=1}^G \|w_i\|^2$.

Two types of problem formulations are typically considered for the multicast beamforming design. The QoS problem is to minimize the transmit power while meeting the received SINR target at each user, which is formulated as

$$P_o : \min_{w} \sum_{i=1}^{G} \|w_i\|^2 \quad \text{s.t.} \quad \text{SINR}_{ik} \geq \gamma_{ik}, \ k \in K_i, i \in G \quad (1)$$

where $w_i \in \{w_i^H, \ldots, w_i^H\}^H$, and $\gamma_{ik}$ is the SINR target for user $k$ in group $i$. The other problem formulation is the (weighted) MMF problem, which is to maximize the minimum (weighted) SINR among all users subject to the transmit power constraint at the BS:

$$S_o : \max_{w} \min_{i,k} \frac{\text{SINR}_{ik}}{\gamma_{ik}} \quad \text{s.t.} \quad \sum_{i=1}^{G} \|w_i\|^2 \leq P \quad (2)$$

where $P$ is the transmit power budget at the BS, and $\gamma_{ik} > 0$, $\forall i, k$, here serves as the weight to control the trade-off of service or fairness among users. It is well-known that both $P_o$ and $S_o$ are non-convex and NP-hard. The existing works have proposed various computational optimization methods to find good suboptimal solutions. We will first focus on the QoS problem $P_o$ and develop two fast first-order algorithms to obtain the solutions for $P_o$. Then, we discuss how to use the proposed algorithms to solve the MMF problem efficiently.

III. OPTIMAL MULTICAST BEAMFORMING STRUCTURE AND THE SCA METHOD

The optimal multicast beamforming structure has been obtained in [2], under this structure, problem $P_o$ is transformed into an equivalent weight optimization problem of a much lower dimension that is independent of the number of antennas. This leads to a significant computational saving and provides opportunities for efficient algorithm designs for massive MIMO systems. To facilitate our algorithm development later, we briefly describe the optimal multicast beamforming structure obtained in [2], the transformed weight optimization problem, and the SCA method for the problem.

A. Optimal Multicast Beamforming Structure

It is shown in [2] that the optimal solution to $P_o$ is a weighted MMSE filter given by

$$w_i^o = R^{-1}(\lambda^o)H_i \alpha_i^o, \ i \in G \quad (3)$$

where $H_i = [h_{i1}, \ldots, h_{iK_i}] \in \mathbb{C}^{N \times K_i}$ is the channel matrix for group $i$, $\alpha_i^o \in \mathbb{C}^{K_i \times 1}$ is the optimal weight vector for group $i$, and $R(\lambda^o) = I + \sum_{i=1}^{G} \sum_{k=1}^{K_i} \lambda_{ik}^o |h_{ik}|^2$ is the (normalized) noise plus weighted channel covariance matrix, with $\lambda^o \in \mathbb{R}^{K_{\text{tot}} \times 1}$ containing the optimal Lagrangian multipliers $\{\lambda_{ik}^o\}$ associated with the SINR constraints in (1). The $k$th weight element $\alpha_{ik}^o$ indicates the relative significance of user $k$’s channel $h_{ik}$ in the overall group-channel direction $H_i \alpha_i^o$.

To determine $w_i^o$ in (3), we need to numerically compute both $\lambda^o$ and $\{\alpha_i^o\}$. The parameter $\lambda^o$ can be approximately computed by the simple fixed-point iterative method proposed in [2]. Given $\lambda^o$, based on the optimal solution structure $w_i^o$ in (3), the original problem $P_o$ is transformed into the following equivalent weight optimization problem for $\{\alpha_i\}$

$$P_1 : \min_{\alpha} \sum_{i=1}^{G} ||C_i \alpha_i||^2$$

where $C_i$ is the channel covariance matrix for group $i$. The optimal solution to $P_1$ is given by $\alpha_i^o = C_i^{-1/2} \alpha_i^o, \ i \in G$.
where $a^H_i, \ldots, a^H_G, C_i \triangleq \mathbf{R}^{-1}(\lambda)H_i \in \mathbb{C}^{N \times K_i}$, $f_{jik} \triangleq C_j^Hh_{ik} \in \mathbb{C}^{K_j \times 1}$, $k \in K_i, i, j \in G$. Different from the original problem $\mathcal{P}_o$ with $G$ variables, the converted problem $\mathcal{P}_1$ has $K_{tot}$ variables, which does not depend on the number of antennas $N$. The problem dimension of $\mathcal{P}_1$ is much smaller than that of $\mathcal{P}_o$ in massive MIMO systems with $K_i \ll N$.

The inherent structure of the optimal multicast beamforming solution in (3) paves the way for developing low-complexity fast algorithms for multicast beamforming design in large-scale massive MIMO systems.

### B. Obtaining Weights $\{a_i\}$ via SCA

The weight optimization problem $\mathcal{P}_1$ is still a non-convex NP-hard problem. The SCA method can be adopted to solve $\mathcal{P}_1$, which is guaranteed to converge to the local minimum [14]. Specifically, given any auxiliary vector $v_i \in \mathbb{C}^{K_i \times 1}, i \in G$, based on the inequality $(a_i - v_i)^Hf_{jik} a_i - v_i \geq 0$, we have $|a_i|^2 f_{jik}^2 \geq 2\Re\{a_i^Hf_{jik}\} - |v_i|^2 f_{jik}^2$. Replacing the numerator of the SINR expression in (4) by the right-hand side (RHS) of the above inequality, we convexify the SINR constraint and change $\mathcal{P}_1$ to the following convex optimization problem

$$\mathcal{P}_{SCA}(v) : \min_v \sum_{i=1}^{G} ||C_i v_i||^2$$

s.t. $\gamma_{ik} \sum_{j \neq i} |a_j^Hf_{jik}|^2 + |v_i^Hf_{jik}|^2 + \gamma_{ik}\sigma^2$

$$-2\Re\{a_i^Hf_{jik}\}v_i \leq 0, \quad k \in K_i, i \in G$$

where $v \triangleq [v_1^H, \ldots, v_G^H]^H$. Note that the solution to $\mathcal{P}_{SCA}(v)$ is always feasible to $\mathcal{P}_1$. By updating $v$ with the solution $v$ to $\mathcal{P}_{SCA}(v)$, we iteratively solve $\mathcal{P}_{SCA}(v)$ until convergence.

Since $\mathcal{P}_{SCA}(v)$ is convex, it can be solved by IPM available through standard convex solvers. However, IPM is a second-order algorithm (i.e., based on the Hessian matrix of the objective function), whose best computational complexity is $O(K_{tot}^4)$ and worst is $O(K_{tot}^4)$. Thus, iteratively solving $\mathcal{P}_{SCA}(v)$ via IPM still incurs relatively high computational complexity for large-scale systems when $K_i$ is large. To address this, for the rest of this paper, we develop two fast first-order algorithms to solve $\mathcal{P}_{SCA}(v)$ that maintain a low complexity in computing the solution for large-scale systems.

### IV. EXTRAGRADIENT-BASED SCA ALGORITHM

#### A. Dual Saddle Point Reformulation

For the purpose of computation, we rewrite problem $\mathcal{P}_{SCA}(v)$ using the real and imaginary parts of each complex quantity. Define $x_i \triangleq [\mathcal{R}e\{a_i\}^T, \mathcal{I}m\{a_i\}^T]^T, y_i \triangleq [\mathcal{R}e\{v_i\}^T, \mathcal{I}m\{v_i\}^T]^T$. Also, define

$$A_i \triangleq [\mathcal{R}e\{C_i\} - \mathcal{I}m\{C_i\}, \mathcal{I}m\{C_i\}], F_{jik} \triangleq [\mathcal{R}e\{f_{jik}\} - \mathcal{I}m\{f_{jik}\}, \mathcal{I}m\{f_{jik}\}].$$

for $k \in K_i, i, j \in G$. Then, we have $||C_i a_i||^2 = ||A_i x_i||^2$ and $|a_i|^2 f_{jik}^2 = ||x_j^T F_{jik} x_j||^2 = x_j^T F_{jik} x_j$, where $F_{jik} \triangleq F_{jik}^* F_{jik}$. Define $x \triangleq [x_1^T, \ldots, x_G^T]^T$ and $y \triangleq [y_1^T, \ldots, y_G^T]^T$. With these new variables, $\mathcal{P}_{SCA}(v)$ can be equivalently expressed as

$$\mathcal{P}_{SCA}(y) : \min_x \sum_{i=1}^{G} ||A_i x_i||^2$$

s.t. $y_i^T F_{jik} y_i + \gamma_{ik} \sum_{j \neq i} x_j^T F_{jik} x_j - 2y_i^T F_{jik} x_i + \gamma_{ik}\sigma^2 \leq 0, \quad k \in K_i, i \in G.$

We first describe the class of variational inequality problems [34] below.

**Definition 1 (Variational Inequality).** Given $\mathbb{Z} \subseteq \mathbb{R}^n$ and a mapping $\psi : \mathbb{Z} \rightarrow \mathbb{R}^n$, the variational inequality is to find $z \in \mathbb{Z}$ satisfying $\psi(z)^T(z' - z) \geq 0, \forall z' \in \mathbb{Z}$. Operator $\psi(\cdot)$ is said to be monotone on $\mathbb{Z}$ if $[\psi(z) - \psi(z')^T(z - z') \geq 0, \forall z, z' \in \mathbb{Z}$.

The problem is monotone if operator $\psi(\cdot)$ is monotone.

The projection methods belong to a class of iterative algorithms that solve the monotone variational inequality problems [34]. At each iteration, a projection method uses the updating step to compute the point (i.e., the value of the optimization variable) and then projects it onto the feasible set of the problem to ensure the updated point is feasible. Note that the projection method may not be an efficient method. In general, the projection methods are only computationally cheap when the projection is easy to compute.

Note that $\mathcal{P}_{SCA}(y)$ is convex, and the objective function is differentiable. Let operator $\psi(x)$ be the gradient of the objective function of $\mathcal{P}_{SCA}(y)$. Following the optimality criterion for a convex optimization problem, $\psi(x)$ is monotone, and thus $\mathcal{P}_{SCA}(y)$ is a monotone variational inequality problem. However, it is difficult to find a closed-form expression for the projection onto the feasible set of $\mathcal{P}_{SCA}(y)$. Thus, directly applying the projection method to solve $\mathcal{P}_{SCA}(y)$ is not computationally attractive. To overcome this difficulty, we resort to the Lagrange dual problem of $\mathcal{P}_{SCA}(y)$.

Define $\phi_i(k) \triangleq y_i^T F_{iik} y_i + \gamma_{ik} \sum_{j \neq i} x_j^T F_{jik} x_j - 2y_i^T F_{iik} x_i + \gamma_{ik}\sigma^2$, for $k \in K_i, i \in G$. The Lagrangian of $\mathcal{P}_{SCA}(y)$ is given by

$$L(x, \eta) = \sum_{i=1}^{G} (||A_i x_i||^2 + \eta_{i}^T \phi_i(x))$$

where $\eta_i \geq 0$ is the dual variable associated with constraint (5), and $\eta \triangleq [\eta_1^T, \ldots, \eta_G^T]^T$, for $\eta_i \triangleq [\eta_{1i}^T, \ldots, \eta_{K_i i}^T]^T$. The Lagrange dual problem of $\mathcal{P}_{SCA}(y)$ is given by

$$D_{L}(\eta) : \max_{\eta \geq 0} \min_{x} \sum_{i=1}^{G} (||A_i x_i||^2 + \eta_{i}^T \phi_i(x)).$$

Let $x^o$ and $\eta^o$ be the primal and dual optimal solutions for $\mathcal{P}_{SCA}(y)$. Since $\mathcal{P}_{SCA}(y)$ is convex and Slater’s condition holds, the strong duality holds. It follows that $u^o \triangleq (x^o, \eta^o)$ is a saddle point of the Lagrangian $L(x, \eta)$ [18]. Define operator $g(u)$ as

$$g(u) = g(x, \eta) \triangleq \begin{bmatrix} \nabla_u L(x, \eta) \\ -\nabla_\eta L(x, \eta) \end{bmatrix}, \quad u \in U$$

where $U \triangleq \mathbb{R}^{2K_{tot}\times K_{tot}}$ is a closed convex set. Then, $D_{L}(\eta)$ can be interpreted as finding the saddle point $u^o$. It is shown in [34] that the problem of finding the saddle point $u^o$ can be cast...
as the variational inequality problem: Find \( u^o \in \mathcal{U} \) that satisfies the following variational inequality

\[
g(u^n)^T(u - u^n) \geq 0, \quad \forall u \in \mathcal{U}. \tag{8}
\]

B. Extragradient-Based SCA

To solve problem (8), one may consider the basic projection algorithm (BPA) [34], which iteratively updates \( u \) using \( g(u) \) and then projects it onto \( \mathcal{U} \). However, the convergence of BPA requires operator \( g(u) \) to be strongly monotone [34]. Following Definition 1, operator \( \psi(\cdot) \) is said to be strongly monotone on \( \mathcal{Z} \), if there exists a constant \( c > 0 \) such that

\[
\psi(z) - \psi(z')^T(z - z') \geq c\|z - z'\|^2, \quad \forall z, z' \in \mathcal{Z}.
\]

Since \( \mathcal{L}(x, \eta) \) is linear with respect to (w.r.t.) \( \eta \), operator \( g(u) \) is not strongly monotone on \( \mathcal{U} \). Thus, we cannot apply BPA to our problem due to no convergence guarantee.

Instead of BPA, in this work, we adopt the extragradient method, a variant of BPA, proposed by Korpelevich in [33]. Compared with BPA, the extragradient method can guarantee convergence for a monotone operator, at the cost of an extra update-and-projection step at each iteration. For operator \( g(u) \) in (7), since \( \mathcal{L}(x, \eta) = \mathcal{L}(\eta) = \mathcal{L}(x) \) is convex in \( x \in \mathbb{R}^{2Kn} \) and \( -\mathcal{L}(x, \eta) \) is convex in \( \eta \in \mathbb{R}^n \), \( g(u) \) is monotone on \( \mathcal{U} \) by Definition 1. Thus, we develop a fast iterative algorithm to solve \( \mathcal{P}_{SI}(y) \) by applying the extragradient algorithm in the problem dual domain.

The updating procedure of the extragradient algorithm for finding the saddle point is summarized as follows [33], [34]: At iteration \( n + 1 \),

\[
\hat{x}^{(n)} = x^{(n)} - \alpha \nabla \mathcal{L}(x^{(n)}, \eta^{(n)}), \tag{9}
\]

\[
\bar{\eta}^{(n)} = [\eta^{(n)} + \alpha \eta^{(n)}]^{+} = \left[ \eta^{(n)} + \alpha \eta^{(n)} \right]_{+}, \tag{10}
\]

\[
x^{(n+1)} = x^{(n)} - \alpha \nabla \mathcal{L}(x^{(n)}, \eta^{(n)}), \tag{11}
\]

\[
\eta^{(n+1)} = [\eta^{(n)} + \alpha \eta^{(n)}]^{+} \tag{12}
\]

where \( x^{(n)} \) and \( \eta^{(n)} \) are the intermediate updates in the extra update-and-projection step in (9)-(10), \( \alpha \) is the step size, and notation \( [z]_{+} = \max\{z, 0\} \) for any \( z \in \mathbb{R}^n \).

The gradient \( \nabla \mathcal{L}(x, \eta) \) can be denoted as \( \nabla \mathcal{L}(x, \eta) = (\nabla \mathcal{L}(x, \eta))^T, \cdots, \nabla \mathcal{L}(x, \eta)^T \) from (6), we obtain

\[
\nabla \mathcal{L}(x, \eta) = 2A_{\tau}^TA_{\tau}x_i + 2\left( \sum_{j \in \mathcal{K}_i} \gamma_{jk} \eta_{jk} F_{ijk} \right) x_i
\]

\[
-2\left[ \sum_{k=1}^{K_i} \eta_{ik} F_{ik} \right] y_i, \quad i \in \mathcal{G}. \tag{13}
\]

Also from (6), we obtain the gradient \( \nabla \eta \mathcal{L}(x, \eta) \) as

\[
\nabla \eta \mathcal{L}(x, \eta) = [\phi^T_{\tau}(x), \cdots, \phi^T_{\tau}(x)]^T. \tag{14}
\]

Substituting the expressions in (13) and (14) into (9)-(12), we obtain the closed-form updates for \( x^{(n+1)} \) and \( \eta^{(n+1)} \).

For a monotone variational inequality problem with operator being \( L \)-Lipschitz continuous, the extragradient algorithm is guaranteed to converge to the optimal solution, provided that the step size satisfies \( \alpha < 1/L \) [34]. Unfortunately, it is difficult to determine Lipschitz constant \( L \) for \( g(u) \) in our problem. To overcome this difficulty, instead of a constant step size \( \alpha \), we adopt an adaptive strategy based on the prediction-correction procedure [35], [36] to adaptively set the step size \( \alpha^{(n)} \) for each iteration.

Given a fixed step size \( \alpha \), the prediction-correction procedure contains two steps at iteration \( n + 1 \):

1) Prediction: Obtain \( \tilde{u}^{(n)} = (\tilde{x}^{(n)}, \tilde{\eta}^{(n)}) \) from (9) and (10) with fixed \( \alpha \). Compute \( \tilde{d}_u^{(n)} = \|\tilde{u}^{(n)} - u^{(n)}\| \) and \( \tilde{d}_g^{(n)} = \|g(\tilde{u}^{(n)}) - g(u^{(n)})\| \). Compute step size \( \alpha^{(n)} = \frac{\tilde{d}_g^{(n)}}{\tilde{d}_u^{(n)}} \), where \( c \in (0, 1) \) is a constant.

2) Correction: Set \( \alpha^{(n)} = \min\{\alpha, \alpha^{(n)}\} \) for iteration \( n + 1 \) to update \( x^{(n+1)} \) and \( \eta^{(n+1)} \) in (9)-(12).

We now show that the prediction-correction procedure guarantees the extragradient algorithm converges to the saddle point \( u^o \) of problem (8). If we replace \( \alpha \) by \( \alpha^{(n)} \) in (9)-(12) of the extragradient algorithm, then for any step size sequence \( \{\alpha^{(n)}\} \), the following holds [35], [36]

\[
\|u^{(n+1)} - u^o\|^2 \leq \|u^{(n)} - u^o\|^2 - \|\tilde{u}^{(n)} - u^{(n)}\|^2 \left( 1 - \left( \frac{\alpha^{(n)} d_g^{(n)}}{d_u^{(n)}} \right)^2 \right). \tag{15}
\]

If we set \( \alpha^{(n)} \) as in the correction step of the above prediction-correction procedure, then \( \|u^{(n+1)} - u^o\| < \|u^{(n)} - u^o\| \) and \( \{u^{(n)}\} \) move towards the saddle point \( u^o \) of problem (8). It follows that the algorithm converges to the optimal solution to \( \mathcal{P}_{SICA}(y) \) in each SCA iteration.

Overall, the ESCA algorithm to solve \( \mathcal{P}_1 \) is summarized in Algorithm 1. The main computational complexity of ESCA lies in computing the gradients \( \nabla_{x\ell}(x, \eta) \) using (13) and \( \nabla_{\eta\ell}(x, \eta) \) in (14). At each extragradient iteration, the related matrix-vector computation for \( \nabla_{x\ell}(x, \eta) \) requires \( \sum_{i=1}^{G} \left( \sum_{j=1}^{K_i} 4K_j K_i^2 + 32K_i^2 \right) \) flops, and for \( \nabla_{\eta\ell}(x, \eta) \) requires \( \sum_{i=1}^{G} \left( \sum_{j=1}^{K_i} (16K_j^2 + 8K_i) K_i - 8K_i^2 \right) \) flops. Note that ESCA consists of two layers of iterations: the outer-layer SCA and the inner-layer extragradient-based algorithm to solve each \( \mathcal{P}_{SICA}(\nu) \). The number of iterations at the two layers for convergence depend on the system parameters \( N \) and \( K_{\text{tot}} \). From our experiments, for \( N = 100 \sim 500 \) antennas and \( K_{\text{tot}} = 15 \sim 60 \) users, it typically takes 20 ~ 200 iterations for the inner-layer extragradient algorithm to converge at each SCA iteration, and the outer layer typically takes 2 ~ 60 iterations to converge.

**Remark 1.** As mentioned earlier, BPA consists of only one update-and-projection step at each iteration (similar to (9)(10)), but requires the operator \( g(u) \) to be strongly monotone for convergence. If \( g(u) \) is only monotone, the updated point (the same as \( u^{(n)} \) in (9)(10) in iteration \( n + 1 \)) may move away from, instead of closer to, the optimal point \( u^o \) at each iteration. Thus, the updating procedure may not converge over iterations. In contrast, the extragradient algorithm adds an extra update-and-projection step as in (11)(12) at each iteration. This extra step can ensure convergence for a monotone operator \( g(u) \). Specifically, it is shown in [34] that in iteration \( n + 1 \), after obtaining \( u^{(n)} \) at the first update-and-projection step in (9)(10), a hyperplane \( \mathcal{H}(n) \) can be constructed at \( u^{(n)} \) with normal vector \( g(u^{(n)}) \). For the monotone operator \( g(u) \), it is proven that this hyperplane \( \mathcal{H}(n) \) separates the point \( u^{(n)} \) and the optimal point \( u^o \), where \( u^o \) is in the halfspace in the direction \( -g(u^{(n)}) \). The second update-and-projection step in (11)(12) then utilizes \( -g(u^{(n)}) \) to update
Algorithm 1 ESCA Algorithm to Solve $\mathcal{P}_1$

1: **Initialization:** Set feasible initial point $x(0)$. Set $\alpha$ and $c$.
2: $l = 0$.

3: **repeat**
4: $n = 0$.

5: **repeat**
6: Compute $\bar{y}(n)$ by (9)–(10) using $\alpha$.
7: Compute $g(\bar{y}(n))$ by (13)–(14).
8: Compute $d_l(n)$ and $d_l^{(n)}$.

9: **Prediction:** Compute $\delta_n = c d_l^{(n)} / d_l^{(n)}$.
10: **Correction:** Update step size $\alpha(n) = \min\{\alpha, \delta_n\}$.
11: If $\alpha = \alpha(n)$ then
12: Update $x(n+1)$ and $y(n+1)$ by (11)–(12).
13: Else
14: Update $x(n+1)$ and $y(n+1)$ by (9)–(12) using $\alpha(n)$.
15: **end if**
16: $n \leftarrow n + 1$.
17: **until convergence**
18: Set $y(l+1) = x(n)$.
19: **until convergence**

This ensures that $u^{(n+1)}$ move towards the optimal point $u^*$ and thus is closer to $u^*$ than $u(n)$ is.

### C. Initialization for ESCA

A challenge in using SCA to solve $\mathcal{P}_1$ is that it requires a feasible initial point satisfying the SINR constraint (4) for $\mathcal{P}_{\text{SCA}}(v)$ (equivalently $y(0)$ for $\mathcal{P}_{\text{SCA}}(y)$). It is necessary to generate a feasible initial point with a low computational complexity. Furthermore, a good initial point closer to the (locally) optimal point of $\mathcal{P}_1$ could accelerate the convergence. Below, we consider two initialization methods for ESCA.

1) **Extragradient-based initialization method (EIM):** Based on the extragradient-based algorithm in Section IV-B, we propose EIM as follows. EIM generates a feasible point by solving the following feasibility problem

$$\mathcal{P}_{\text{feas}} : \text{Find } \{x\}$$

$$\text{s.t.} \quad \sum_{j \neq i} x_j^T F_{i j} x_i + \sigma^2 \geq \gamma_{i k}, \quad k \in \mathcal{K}_i, i \in \mathcal{G}$$  \hspace{1cm} (16)

where $x$ and $F_{i j}$ are defined at the beginning of Section IV-A, and constraint (16) is an equivalent real representation of constraint (4) in $\mathcal{P}_1$ based on $|a_i F_{i j}|^2 = x_j^T F_{i j} x_i$.

We solve $\mathcal{P}_{\text{feas}}$ by applying the extragradient method with the adaptive step size proposed in Section IV-B. Specifically, the Lagrangian of $\mathcal{P}_{\text{feas}}$ is given by $\mathcal{L}(x, \eta) = \sum_{i=1}^{G} \phi_i^T (x), \eta_i$, where $\eta_{i k}$ is the dual variable for constraint (16), $\eta \triangleq [\eta_1, \ldots, \eta_{G}]^T$ is the Lagrange multiplier, and $\phi_i(x) \triangleq [\phi_{i 1}(x), \ldots, \phi_{i K_i}(x)]^T$.

The gradient $\nabla_x \mathcal{L}(x, \eta)$, for $i \in \mathcal{G}$, is given by

$$\nabla_x \mathcal{L}(x, \eta) = 2 \sum_{j \neq i} \sum_{k=1}^{K_i} \gamma_{i j} \eta_{i k} F_{i j} x_i - 2 \sum_{k=1}^{K_i} \eta_{i k} F_{i k} x_i.$$  \hspace{1cm} (17)

Similar to (14), the gradient $\nabla_{\eta} \mathcal{L}(x, \eta)$ is given by $\nabla_{\eta} \mathcal{L}(x, \eta) = [\bar{\phi}_i^T (x), \ldots, \bar{\phi}_G^T (x)]^T$. The updating procedure of the extragradient method in (9)–(12) is then applied for solving $\mathcal{P}_{\text{feas}}$, with $\eta$, $\nabla_x \mathcal{L}(x, \eta)$, and $\nabla_{\eta} \mathcal{L}(x, \eta)$ being replaced by $\bar{\eta}$, $\nabla_{\eta} \mathcal{L}(x, \eta)$, and $\nabla_{\eta} \mathcal{L}(x, \eta)$, respectively. Furthermore, the prediction-correction procedure in Section IV-B is used to set the adaptive step size at each iteration.

For simplicity, we randomly choose the initial point for EIM. Note that this point may not be feasible for SINR constraint (4) in $\mathcal{P}_1$. Also, since $\mathcal{P}_{\text{feas}}$ is a non-convex optimization problem, EIM may not be guaranteed to converge or the terminating point may not be feasible as required for $y(0)$ for $\mathcal{P}_{\text{SCA}}(v)$. Thus, EIM is served as a heuristic algorithm. If the point produced by EIM is infeasible, we may consider different initial points for EIM until a feasible is obtained. Our extensive simulation experiments show that EIM converges and provides a feasible initial point with a very high probability.

2) **SDR-based initialization method:** Similar to [2], we can apply SDR along with Gaussian randomization to $\mathcal{P}_1$ to obtain an approximate solution as a feasible initial point $y(0)$ to $\mathcal{P}_{\text{SCA}}(y)$ for ESCA. Recall that $\mathcal{P}_1$ is converted from the original problem $\mathcal{P}_o$ and is of relatively smaller size ($K_o$ variables and constraints). SDR can provide a good initial point when the problem is of small to moderate size, leading to fast convergence for ESCA. However, as $K_o$ becomes large, the computational complexity of SDR will increase significantly, and the quality of the initial point it computes deteriorates. Thus, SDR for initialization is only suitable for small to moderate problems.

### V. ADMM-BASED SCA ALGORITHM

In this section, we develop another computationally efficient algorithm based on ADMM [37], which is a different optimization technique from ESCA. ADMM has drawn growing popularity in recent years as a robust and fast numerical method for solving large-scale problems. We propose an ADMM-based algorithm to solve each SCA subproblem $\mathcal{P}_{\text{SCA}}(v)$. Define the auxiliary variables $d_{i k} \triangleq a_i^H F_{i k}$ for $k \in \mathcal{K}_i, i, j \in \mathcal{G}$. Define $d \triangleq [d_{i k}^H, \ldots, d_{G K_G}^H]^T \in \mathbb{C}^{G K_G}$, with $d_{i k} \triangleq [d_{i k}, \ldots, d_{i K_i}]^T \in \mathbb{C}^G$. Then, the problem $\mathcal{P}_{\text{SCA}}(v)$ can be equivalently expressed as

$$\mathcal{P}_{\text{ADMM}}(v) : \text{min}_{a, d} \sum_{j=1}^{G} \|C_j a_j\|^2$$

$$\text{s.t.} \quad d_{i k} = a_i^H F_{i k}, \quad k \in \mathcal{K}_i, i, j \in \mathcal{G}$$

$$\gamma_{i k} \sum_{j \neq i} |d_{i k}|^2 + |v_{i k}|^2 \|C_j a_j\|^2 - 2 Re\{d_{i k}^H v_{i k}\} \leq 0, \quad k \in \mathcal{K}_i, i \in \mathcal{G}.$$  \hspace{1cm} (18)

Denote the feasible set for the inequality constraint (18) by $\mathcal{F}$. Define the indicator function for the set $\mathcal{F}$ as

$$I_{\mathcal{F}}(d) = \begin{cases} 0 & \text{if } d \in \mathcal{F}, \\ \infty & \text{otherwise.} \end{cases}$$

Then, $\mathcal{P}_{\text{ADMM}}(v)$ is equivalent to the following problem

$$\mathcal{P}_{\text{ADMM}}'(v) : \text{min}_{a, d} \sum_{j=1}^{G} \|C_j a_j\|^2 + I_{\mathcal{F}}(d)$$

$$\text{s.t.} \quad d_{i k} = a_i^H F_{i k}, \quad k \in \mathcal{K}_i, i, j \in \mathcal{G}.$$  \hspace{1cm} (19)
By introducing the auxiliary vector $d$ in constraint (18), we construct the equality-constrained problem $P_{\text{ADMM}}(v)$, where the objective function contains two separate terms for $d$ and $a$ only. This allows us to apply ADMM to decompose the problem into separate subproblems [37]. The augmented Lagrangian of $P_{\text{ADMM}}(v)$ is given by

$$
L_p(d, a, q) = \sum_{j=1}^G \|C_j a_j\|^2 + I_P(d) + \frac{\rho}{2} \sum_{j=1}^G \sum_{i=1}^G \sum_{k=1}^{K_i} |d_j - a_j^H f_{ijk} + q_{ijk}|^2
$$

where $\rho > 0$ is the penalty parameter, and $q_{ijk} \in \mathbb{C}$ is the dual variable associated with constraint (19), and $q \triangleq [q_1^H, \ldots, q_{GK_i}^H]^H$, with $q_{ik} \triangleq [q_{1ik}, \ldots, q_{GK_i k}]^T$. We now decompose $L_p(d, a, q)$ into two subproblems for $d$ and $a$ separately, and update $\{d, a, q\}$ alternatively. Our proposed ADMM-based updating procedure for solving $P_{\text{ADMM}}(v)$ is given as follows:

$$
d^{(n+1)} = \arg \min_d L_p(d, a^{(n)}, q^{(n)}),
$$

$$
a^{(n+1)} = \arg \min_a L_p(d^{(n+1)}, a, q^{(n)}),
$$

$$
q_{ijk}^{(n+1)} = q_{ijk}^{(n)} + (d_j^{(n+1)} - a_j^{(n+1)} H f_{ijk})
$$

where $n$ is the iteration index. Since $P_{\text{ADMM}}(v)$ is convex, the above ADMM procedure is guaranteed to converge to the optimal solution of $P_{\text{ADMM}}(v)$ [37].

The two main updating steps in (21) and (22) involve solving two optimization problems. In the following, we derive the closed-form solutions for the two optimization problems in (21) and (22), which leads to fast computation at each iteration.

A. Closed-Form $d$-Update

Given $a^{(n)}$ and $q^{(n)}$, from (20), the update of $d$ in (21) is equivalent to solving the following problem

$$
P_d(v) : \min_d \sum_{j=1}^G \sum_{i=1}^G \sum_{k=1}^{K_i} |d_j - a_j^H f_{ijk} + q_{ijk}|^2 \quad \text{s.t.} \quad \sum_{j=1}^G \sum_{i=1}^G \sum_{k=1}^{K_i} |d_j - a_j^H f_{ijk} + q_{ijk}|^2 \leq \lambda (18).$$

Problem $P_d(v)$ can be decomposed into $K_{\text{tot}}$ convex subproblems, one for each user $k$ in group $i$ given by

$$
P_{d,ik}(v) : \min_d \sum_{j=1}^G |d_j - \epsilon_{1,jik}^{(n)}|^2 \quad \text{s.t.} \quad \sum_{j=1}^G \sum_{k=1}^{K_i} |d_j - \epsilon_{1,jik}^{(n)}|^2 \leq \lambda_{24}(24).$$

where

$$
e_{1,jik}^{(n)} \triangleq a_j^{(n)} f_{ijk} - q_{ijk}^{(n)}, \quad e_{2,ik}^{(n)} \triangleq |v_i^H f_{ik}|^2 + \gamma_{ik}^2, \quad e_{3,ik}^{(n)} \triangleq \epsilon_{1,jik}^{(n)}$$

Problem $P_{d,ik}(v)$ is a convex QCQP-1 problem, whose solution may be derived in closed-form. In particular, a problem of similar structure has been considered in [21], where the closed-form solution is derived in [21, Appendix A]. We directly use this result and state the closed-form solution below. The optimal solution $d_{ijk}^*$ for $P_{d,ik}(v)$ is given by

$$
d_{ijk}^* = \begin{cases} 
e_{1,jik}^{(n)} + \nu_{ik}^* e_{3,ik} & \text{if } j = i, \\
v_{ik}^* - \nu_{ik}^* e_{3,ik} & \text{otherwise} \end{cases} \quad \text{where} \quad \nu_{ik}^* = 0 \text{ if } e_{2,ik} + \gamma_{ik} \sum_{j \neq i} \epsilon_{1,jik}^{(n)} \leq 2 \text{Re} \{\epsilon_{3,ik} e_{1,jik}^{(n)}\} \leq 0; \nu_{ik}^* \text{ is the unique real positive root of the following cubic equation of } \nu_{ik}^*:$$

$$
e_{2,ik} + \gamma_{ik} \sum_{j \neq i} \epsilon_{1,jik}^{(n)} \leq 2 \text{Re} \{\epsilon_{3,ik} e_{1,jik}^{(n)}\} - 2 \nu_{ik} \epsilon_{3,ik}^2 = 0. \quad (27)$$

Since the roots of (27) are given by the cubic formula, $v_{ik}^*$ is obtained in closed-form. For the sake of completeness, the key steps leading to the above solution is provided in Appendix A.

B. Closed-Form $a$-Update

Given $d^{(n+1)}$ and $q^{(n)}$, the update of $a$ in (22) is equivalent to solving the following problem

$$
P_a(v) : \min_a \sum_{j=1}^G \|C_j a_j\|^2 + \frac{\rho}{2} \sum_{i=1}^G \sum_{k=1}^{K_i} |q_{ijk}^{(n+1)} - a_j^H f_{ijk} + q_{ijk}^{(n)}|^2.$$
Algorithm 2 ASCA Algorithm to Solve $P_1$

1: **Initialization:** Set feasible initial point $v^{(0)}$. Set $\rho_i, l = 0$.

2: **repeat**

3: // solve $P_{\text{ADMM}}(v^{(l)})$

4: **Initialization:** $a^{(0)} = v^{(l)}$, $d^{(0)} = 0$, $q^{(0)} = 0$, $n = 0$.

5: **repeat**

6: Compute $d^{(n+1)}$ by (21) with $a^{(n)}$ and $q^{(n)}$.

7: Compute $a^{(n+1)}$ by (22) with $d^{(n+1)}$ and $q^{(n)}$.

8: Compute $q^{(n+1)}$ by (23) with $d^{(n+1)}$ and $a^{(n+1)}$.

9: $n \leftarrow n + 1$.

10: **until** convergence

11: Set $v^{(l+1)} = a^{(n)}$. Set $l \leftarrow l + 1$.

12: **until** convergence

In Section VII, we will provide the simulation study to compare the convergence, performance, and the computational time of ESCA and ASCA.

C. Initialization for ASCA

As discussed earlier in Section IV-C, a feasible initial point is required by SCA to solve $P_1$. Below we propose an ADMM-based initialization method (AIM). Using the ADMM-based algorithm above, we propose AIM for ASCA. The feasible point is computed by solving the following feasibility problem

$$P_{\text{feasADMM}}: \text{Find } \{a\} \quad \text{s.t.} \quad (4).$$

Using the auxiliary variables $d_{ijk} = a_{ijH}^T f_{ijk}$, for $k \in K_i$, $i, j \in \mathcal{G}$, defined at the beginning of Section V, $P_{\text{feas}}$ is equivalently expressed as

$$P_{\text{feasADMM}}: \text{Find } \{a, d\} \quad \text{s.t.} \quad d_{ijk} = a_{ijH}^T f_{ijk}, \ k \in K_i, \ i, j \in \mathcal{G},$$

$$\sum_{j \neq i} |d_{ijk}|^2 + |\sigma|^2 \geq \gamma_{ik}, \ k \in K_i, \ i \in \mathcal{G}. \quad (29)$$

Denote the feasible set for constraint (29) by $\tilde{F}$. The augmented Lagrangian of $P_{\text{feasADMM}}$ is given by

$$\tilde{L}_\rho(d, a, \tilde{q}) = I_{\tilde{F}}(d) + \frac{\rho}{2} \sum_{j=1}^G \sum_{i=1}^K \sum_{k=1}^i |d_{ijk} - a_{ijH}^T f_{ijk} + \tilde{q}_{ijk}|^2 \quad (30)$$

where indicator function $I$, penalty parameter $\rho$, dual variable $\tilde{q}_{ijk}$ are defined similarly as those in (20). Similar to the ADMM updating procedures in (21)-(23), based on (30), the AIM updating procedure for solving $P_{\text{feasADMM}}$ is given as follows: At iteration $n + 1$,

$$a^{(n+1)} = \arg \min_a \sum_{j=1}^G \sum_{i=1}^K \sum_{k=1}^i |d_{ijk} - a_{ijH}^T f_{ijk} + \tilde{q}_{ijk}(n)|^2, \quad (31)$$

$$a^{(n+1)} = \arg \min_a \sum_{j=1}^G \sum_{i=1}^K \sum_{k=1}^i |d_{ijk} - a_{ijH}^T f_{ijk} + \tilde{q}_{ijk}(n+1)|^2, \quad (32)$$

$$\tilde{q}_{ijk} = \tilde{q}_{ijk} + (d_{ijk} - a_{ijH}^T f_{ijk}(n+1)). \quad (33)$$

The updating steps in (31) and (32) can be derived in closed-form, which are provided in Appendix B.

The initial point for AIM is randomly chosen, which may not be feasible for SINR constraint (29) in $P_{\text{feasADMM}}$. However, if AIM converges, the terminating point will satisfy SINR constraint (29), and the produced point $a$ for $v^{(l)}$ is feasible to $P_{\text{SCA}}(v)$. Note that since $P_{\text{feas}}$ is a non-convex optimization problem, the ADMM procedure for AIM described above may not be guaranteed to converge. Thus, AIM is served as a heuristic algorithm. Our extensive simulation studies show that AIM converges to a feasible point with a very high probability.

Besides AIM, we can also use the SDR-based initialization method discussed in Section IV-C for initialization of ASCA, where a feasible initial point $v^{(0)}$ for ASCA is obtained by applying SDR along with Gaussian randomization to solve $P_1$.

VI. SCALING SCHEME FOR THE MMF PROBLEM

In this section, with the proposed ESCA and ASCA for the QoS problem, we propose an efficient scheme to obtain a solution to the MMF problem $S_o$. We transform the original MMF problem $S_o$ into the following equivalent problem

$$S_o^*(\gamma, P): \max_{w, t} \ t$$

$$\text{s.t.} \quad \text{SINR}_{ik} \geq t_{\gamma_{ik}}, \ k \in K_i, \ i \in G$$

$$\sum_{i=1}^G \|w_i\|^2 \leq P$$

where vector $\gamma$ contains all SINR targets $\{\gamma_{ik}\}$ of all users in all groups. Furthermore, we parameterize the QoS problem $P_o$ as $P_o(\gamma)$. It has been shown that $S_o^*(\gamma, P)$ and $P_o(\gamma)$ are the inverse problems to each other [5]. Specifically, denote the maximum objective value of $S_o^*(\gamma, P)$ by $t_o = S_o^*(\gamma, P)$. Let the minimum power objective value of $P_o(\gamma')$ be $P = P_o(\gamma')$ for some $\gamma'$. Then, we have the following inverse relation:

$$t_o = S_o^*(\gamma, P_o(t_o)), \quad P = P_o(S_o^*(\gamma, P_o(t_o))). \quad (34)$$

Based on this relation, in the literature, the MMF problem $S_o^*(\gamma, P)$ is typically solved iteratively solving the QoS problem $P_o(\gamma')$ along with a bi-section search over $t$ until the transmit power objective in $P_o(\gamma')$ is equal to $P$ [2], [5], [6], [17], [21]. However, this approach is computationally expensive. As mentioned at the end of Section III-B, existing works use either SDR or SCA to compute a solution to $P_o(\gamma)$, where the second-order IPM is used to solve either relaxed or convexified approximate problem. As a result, iteratively solving $P_o(\gamma')$ incurs high computational complexity not suitable for large-scale problems. Using the optimal beamforming structure $w_i^g$ in (3) can significantly reduce the required computation in the above approach, where $P_o(\gamma)$ can be converted to a much smaller weight optimization problem as in $P_1$. Nonetheless, resorting to the additional layer of iterations to solve the QoS problem is still computationally costly for the MMF problem as compared with the QoS problem itself, especially for large-scale problems [2].

To avoid the additional iterative procedure, we develop a closed-form scaling scheme for finding a solution to $S_o$ directly. Specifically, we first obtain the solution to $P_o(\gamma)$ by solving the smaller weight optimization problem $P_1$ using either ESCA or ASCA proposed in Sections IV and V. Then, we scale this solution to $P_o(\gamma)$ to obtain a solution to $S_o^*(\gamma, P)$, such that the transmit power budget $P$ is met. Parameterize $S_o$ as $S_o(\gamma, P)$. 

Algorithm 3 The Closed-Form Scaling Scheme for $S_o(\gamma, P)$

1. Solve $P_o(\gamma)$ and attain solution $w^P(\gamma)$.
2. Compute $P^o(\gamma) = \sum_{i,k}^{G} |w_i^0(\gamma)|^2$.
3. Compute $w^o(\gamma, P) = \sqrt{\frac{P}{P^o(\gamma)}} w^P(\gamma)$ as the solution to $S_o(\gamma, P)$.

This proposed scaling scheme is summarized in Algorithm 3. We show below that this scaling scheme provides a feasible solution to $S_o(\gamma, P)$, and we also bound its performance.

**Proposition 1.** Let $w^o(\gamma)$ be a feasible beamforming solution to $P_o(\gamma)$ produced by a given algorithm, with the achieved objective value denoted by $P^o(\gamma)$. Define $P^i_{ik}(\gamma) \triangleq \sum_{j \neq i} |b_{ik}^H w_j^o(\gamma)|^2$. Then, the scaled beamforming vector $w^o(\gamma, P)$ is a feasible solution to $S_o(\gamma, P)$: the corresponding achieved objective value, denoted by $t^i(\gamma, P)$, satisfies

$$\frac{P}{P^o(\gamma)} \text{min}_{i,k} \frac{P^i_{ik}(\gamma) + \sigma^2}{P^o(\gamma) P^i_{ik}(\gamma) + \sigma^2} \leq t^i(\gamma, P) \leq \frac{P}{P^o(\gamma)} \text{max}_{i,k} \frac{P^i_{ik}(\gamma) + \sigma^2}{P^o(\gamma) P^i_{ik}(\gamma) + \sigma^2} \quad (35)$$

where $P^o(\gamma)$ denotes the optimal objective value of $P_o(\gamma)$.

**Proof:** See Appendix C.

Note that the tightness of the lower and upper bounds for $t^i(\gamma, P)$ in (35) depends on transmit power $P^o(\gamma)$, which is obtained by a given algorithm for $P_o(\gamma)$ with solution $w^o(\gamma)$. If the power budget $P$ for the MMF problem $S_o(\gamma, P)$ is more than the optimal power value for $P_o(\gamma)$, i.e., $P \geq P^o(\gamma)$, and the algorithm provides $P^o(\gamma) = P$, then $w^o(\gamma, P) = w^o(\gamma)$, and the bounds in (35) in this case are simplified to $1 \leq t^i(\gamma, P) \leq \frac{P}{P^o(\gamma)}$. Note that, often for the given algorithm, the solution $w^o(\gamma)$ results in at least one SINR constraint being attained with equality, i.e., $\text{SINR}_{ik} = \gamma_{ik}$, for some $i, k$, then we have $t^i(\gamma, P) = 1$. In a special case when $P^o(\gamma) = P = P^o(\gamma)$, we have $t^i(\gamma, P) = 1$, and also both the upper and lower bounds become 1. In this case, since $P = P^o(\gamma)$, from the inverse relation in (34), we have $t^o = 1$. Thus, $t^i(\gamma, P) = t^o = 1$, and the bounds in (35) are tight.

Comparing Algorithm 3 with the iterative bi-section search method using (34), we see that our proposed closed-form scaling scheme avoids iteratively solving $P_2(t^o)$ along with a bi-section search over $t$, and thereby, enjoys a significant reduction of computational complexity. Moreover, we can directly apply the fast algorithm ESCA proposed in Section IV or ASCA proposed in Section V along with the optimal structure in (3) to provide a solution to $P_o(\gamma)$ with this scaling scheme. This approach leads to two fast first-order algorithms for solving $S_o(\gamma, P)$ in large-scale systems.

**Remark 2.** We point out that our scaling scheme for the multi-group multistatic beamforming MMF problem is different from a similar scaling scheme proposed for the MMF problem in a single-group scenario in [22]. Specifically, the scheme in [22] first applies ZF beamforming for each group to eliminate inter-group interference. Once the interference is removed, for the equivalent single-group multistatic beamforming problem, [22] scales the beamforming solution of the single-group QoS problem to obtain a feasible solution to the MMF problem. In our scheme, we directly handle the original multi-group MMF problem containing inter-group interference. Our scheme scales the solution of the multi-group QoS problem $P_o(\gamma)$ to solve the original MMF problem $S_o(\gamma, P)$. Note that the scheme in [22] has the lower and upper bounds for the objective value $t$ (similar to $t$ in $S'_o(\gamma, P)$) as $\left[\frac{P}{P^o(\gamma)}, \frac{P}{P^o(\gamma)}\right]$ with no interference present. In contrast, for our scheme, the lower and upper bounds in (35) contain additional terms w.r.t. $P^i_{ik}(\gamma)$. Note that $I^i_{ik}(\gamma)$ represents the inter-group interference to user $k$ in group $i$ by solution $w^o(\gamma)$ of $P^o(\gamma)$. Following this, both terms $I^i_{ik}(\gamma) + \sigma^2$ and $\frac{P^i_{ik}(\gamma)}{P^o(\gamma)}$ in (35) are the interference plus noise term for user $k$ in group $i$, where the latter is based on the scaled beamforming solution $w^o(\gamma, P)$. These additional terms in the lower and upper bounds in (35) represent the minimum and maximum inter-group interference ratios, respectively. In the special case of single group $G = 1$, the bounds in (35) reduces to $\left[\frac{P}{P^o(\gamma)}, \frac{P}{P^o(\gamma)}\right]$ in [22]. Thus, the bounds in (35) can be viewed as a generalization of the bounds in [22] from single-group to multi-group settings with inter-group interference.

**VII. SIMULATION RESULTS**

We consider a default setup for downlink multi-group multicast beamforming, where $G = 3$ groups, $K_i = K$ users/group, $\forall i \in G$, and the same SINR target for all users as $\gamma_{ik} = \gamma = 10$ dB, $\forall k, i$. The user channels are generated independently with an identical distribution as $h_{ik} \sim \mathcal{CN}(0, 1)$. The performance plots are obtained by averaging the results over 100 channel realizations per user.

For QoS problem $P_o$, we consider the proposed two fast algorithms: ESCA in Algorithm 1 and ASCA in Algorithm 2. For ESCA, we set the step size $\alpha = 0.1$ and the constant $c = 0.8$. For ASCA, we set the penalty parameter $\rho = 0.2$.\footnote{We have studied different values of $\rho$ and found $\rho = 0.2$ generally provides overall good performance and convergence speed for ASCA.} We consider different initialization methods discussed in Sections IV-C and V-C for ESCA and ASCA, respectively. Therefore, we refer to our algorithms as follows: 1) EIM-ESCA: ESCA with EIM initialization; 2) SDR-ESCA: ESCA with SDR initialization; 3) AIM-ASCA: ASCA with AIM initialization; 4) SDR-ASCA: ASCA with SDR initialization. Note that each algorithm solves the weight optimization problem $P_1$ using the optimal beamforming structure in (3).\footnote{Note that we only consider SDR-SCSA as the benchmark for comparison against our proposed algorithms. This is because SDR-SCSA is the state-of-the-art method with a near-optimal performance and substantially lower computational complexity than other existing algorithms in the literature. The comparison of SDR-SCSA and other existing algorithms in both performance and computational complexity has already been provided in [2], and adding other algorithms here will not provide additional insight or observation than what has been shown in [2].} Besides the above four algorithms for solving $P_1$, we also consider the following methods for comparison:

- **Lower Bound for $P_o$:** Obtained by solving the relaxed version of $P_o$ via SDR.
- **SDR-CSGA [2]$: Solve $P_1$ via SCA using $P_{1SCGA}(\nu)$.

*Note that in (3), the exact expression of $R(\lambda^o)$ is used and its computation is discussed below (3). One may utilize the asymptotic expression of $R(\lambda^o)$ obtained in [2] to simplify the computation of $R(\lambda^o)$. Since the fixed-point iterative method [2] for computing $R(\lambda^o)$ is computationally expensive, using the asymptotic expression of $R(\lambda^o)$ only brings a minor reduction of computation cost and thus is not considered in the simulation.*
where each $P_{\text{ISCA}}(v)$ is solved by the standard convex solver CVX, which uses IPM. The SDR method is used to generate an initial point.

For MMF problem $S_o$, we consider ESCA and ASCA with the proposed scaling scheme in Algorithm 3. They are denoted as ESCA-Scaling and ASCA-Scaling, respectively. For comparison, we also consider the following methods:

- Upper Bound for $S_o$: Obtained by solving the relaxed version of $P_o$ using SDR along with the bi-section search over $t$.
- ESCA-Bisection: Solve $S_o$ via iteratively solving $P_o$ along with bi-section search over $t$. For solving $P_o$, SDR-ESCA is used, where the optimal beamforming structure in (3) with the asymptotic expression of $R(\lambda^v)$ is applied.
- ASCA-Bisection: Similar to ESCA-Bisection, except that SDR-ASCA is used to solve $P_o$.
- CSCA-Bisection [2]: Similar to ESCA-Bisection, except that SDR-CSCA is used to solve $P_o$.
- SDR-Bisection [2]: Solve $S_o$ via iteratively solving $P_o$ along with bi-section search over $t$. For solving $P_o$, SDR along with the Gaussian randomization method is used, where the optimal beamforming structure in (3) with the asymptotic expression of $R(\lambda^v)$ is applied.

A. Convergence Analysis

In this subsection, we study the convergence behavior of the two proposed fast algorithms (ESCA and ASCA) for the QoS problem $P_o$. Note that each algorithm consists of two layers of iterations: Outer-layer SCA and the inner-layer iterative algorithm to solve each $P_{\text{ISCA}}(v)$. Let $P_t = \sum_{i=1}^{G} \|w_i\|^2$ denote the total transmit power. Fig. 1 shows the trajectory of normalized transmit power $P_t/\sigma^2$ over the inner-layer iterations at the first outer-layer SCA iteration. We set $N = 500$ and $K = 10$. We observe that SDR-SCA and SDR-ASCA converge faster and result in lower $P_t/\sigma^2$ than EIM-ESCA and AIM-ASCA. This shows that the SDR initialization method provides a better initial point than the other initialization methods. Between SDR-ESCA and SDR-ASCA, we observe a similar convergence rate. Comparing EIM-ESCA and AIM-ASCA, we see that AIM-ASCA converges faster than EIM-ESCA. Next, we consider the convergence behavior of the relative difference of the optimization variable for each algorithm. Specifically, we consider the normalized difference $\frac{\|x^{(n+1)} - x^{(n)}\|}{\|x^{(n+1)}\|}$ in two consecutive inner-layer iterations of ESCA in Algorithm 1, and similarly, $\frac{\|a^{(n+1)} - a^{(n)}\|}{\|a^{(n+1)}\|}$ for ASCA in Algorithm 2. Fig. 2 shows these relative differences for different algorithms at the first outer-layer SCA iteration for $N = 500$ and $K = 10$. Again, we observe that the SDR initialization method results in a faster convergence than the other initialization methods. Both SDR-ESCA and SDR-ASCA reach a relative difference of $10^{-3}$ within 50 iterations, with SDR-ASCA converging slightly faster than SDR-ESCA. Comparing EIM-ESCA and AIM-ASCA, we see that AIM-ASCA provides a faster convergence than EIM-ESCA, reaching a relative difference of $10^{-3}$ within 50 iterations.

We now study the convergence behavior of different algorithms over the outer-layer SCA iterations. Fig. 3 shows the convergence behavior of our proposed algorithms at the outer-layer SCA iterations, for $N = 500$ and $K = 10$. We set the inner-layer convergence threshold for the proposed algorithms such that they converge to nearly the same value. For EIM-ESCA and SDR-ESCA, we set the inner-layer convergence threshold to be $\frac{\|x^{(n+1)} - x^{(n)}\|}{\|x^{(n+1)}\|} \leq 10^{-3}$. Note that although Fig. 1
shows that at the first outer-layer SCA iteration, EIM-ESCA converges to a higher value of $P_t/\sigma^2$ than that of SDR-ESCA over the inner-layer iterations. For ESCA, we set the inner-layer convergence threshold as the rest of algorithms at the outer-layer SCA iterations. As we see in Fig. 3, all algorithms converge to the same value of $P_t/\sigma^2$ as that of SDR-ESCA and SDR-ASCA. The reason is due to the high computational complexity involved in generating the lower bound, which increases fast with $N$ and becomes impractical for $N$ beyond 300. Similarly, the upper bound shown in Fig. 6 is provided until $N = 200$.

| $N$   | 1   | 100 | 200 | 300 | 400 | 500 |
|-------|-----|-----|-----|-----|-----|-----|
| EIM-ESCA | 2.694 | 3.032 | 2.708 | 2.589 | 2.711 |
| AIM-ESCA | 1.150 | 2.415 | 3.195 | 4.092 | 4.339 |
| SDR-ESCA | 0.327 | 0.237 | 0.233 | 0.224 | 0.243 |
| SDR-ASCA | 0.076 | 0.051 | 0.061 | 0.071 | 0.088 |
| SDR-CSCA [2] | 7.179 | 6.126 | 6.400 | 5.963 | 6.500 |
| EIM (Init. method) | 0.037 | 0.044 | 0.045 | 0.050 | 0.057 |
| AIM (Init. method) | 0.0068 | 0.0058 | 0.0059 | 0.0063 | 0.0073 |
| SDR (Init. method) | 1.050 | 1.038 | 1.104 | 1.103 | 1.137 |

**Table II**

QoS: Average Computation Time over $K$ (sec.) ($G = 3, N = 100$).

| $K$   | 1   | 5    | 7    | 10   | 15   | 20   | 35   |
|-------|-----|------|------|------|------|------|------|
| EIM-ESCA | 1.098 | 1.689 | 2.737 | 5.498 | 8.685 | 14.33 |
| AIM-ESCA | 0.519 | 0.725 | 1.234 | 2.043 | 3.146 | 5.426 |
| SDR-ESCA | 0.056 | 0.112 | 0.373 | 0.856 | 1.666 | 4.883 |
| SDR-ASCA | 0.012 | 0.024 | 0.087 | 0.216 | 0.433 | 2.594 |
| SDR-CSCA [2] | 1.747 | 3.502 | 7.592 | 13.79 | 21.52 | 57.85 |
| EIM (Init. method) | 0.019 | 0.025 | 0.038 | 0.063 | 0.090 | 0.450 |
| AIM (Init. method) | 0.0048 | 0.0055 | 0.0067 | 0.0072 | 0.012 | 0.018 | 0.077 |
| SDR (Init. method) | 0.545 | 0.727 | 1.094 | 1.852 | 3.489 | 22.79 |

**B. Performance Comparison for the QoS Problem**

We now compare the performance of different algorithms for the QoS problem. We set SINR target $\gamma = 10$ dB. Also, for all algorithms in comparison, we set the convergence threshold for the outer-layer SCA iterations as $\frac{\|\hat{a}^{(n+1)} - a^{(n)}\|}{\|a^{(n+1)}\|} \leq 10^{-3}$. Fig. 4 shows the normalized transmit power $P_t/\sigma^2$ vs. $N$ by different algorithms, for $G = 3$ and $K = 10$. We see that our proposed algorithms have a similar performance to that of SDR-CSCA in [2], and all algorithms nearly attain the lower bound for $P_o$. This demonstrates that our proposed fast algorithms achieve a near-optimal performance. The computational advantages of ESCA and ASCA are shown in Table I, where we list the average computation time by each algorithm used for the plots in Fig. 4. The first five rows show the computation times of different algorithms, excluding the initialization. We observe that, by using the optimal structure in (3), the computation times of all algorithms remain roughly unchanged as $N$ increases. Under the same SDR initialization method, the computation times of SDR-ESCA and SDR-ASCA are only about 4% and 1% of that of SDR-CSCA, respectively, and SDR-ASCA has a smaller computation time than SDR-ESCA. The computation times of EIM-ESCA and AIM-ASCA are both about 40% of that of SDR-CSCA. The computation time of AIM-ASCA increases with $N$ more noticeably than other algorithms. As a result, AIM-ASCA is initially faster than EIM-ESCA for $N \leq 200$, but its computation time increases and becomes slower than EIM-ESCA for $N \geq 300$. The reason is due to the quality of initialization as will be explained below. The last three rows in Table I show the computation times of different initialization methods. We see that the EIM is a fast initialization method, with its computation time about 4% of that of SDR. The AIM is the fastest one among all three initialization methods, with its computation time about 15% and 0.5% of that of EIM and SDR, respectively. However, the quality of AIM initialization deteriorates as $N$ increases.
Unlike other initialization methods, leading to more iterations for convergence and a longer computation time. This is evidenced by the computation time of AIM-ASCA, which increases more noticeably as $N$ increases.

Fig. 5 shows $P_i/\sigma^2$ vs. $K$ by different algorithms, for $G = 3$ and $N = 100$. Again, the performance of SDR-ESCA and SDR-ASCA are nearly identical to that of SDR-CSCA and nearly attain the lower bound. The performance gaps of EIM-ESCA and AIM-ASCA to the lower bound are more noticeable as $K$ becomes large, although they are still within $0.6 \text{ dB}$. Similar to Table I, the corresponding average computation times of these algorithms are shown in Table II. We see that both ESCA and ASCA are considerably faster than CSCA to compute the solution. In particular, unlike CSCA, their computation times only mildly increase with $K$. For the initialization methods, EIM and AIM are faster and much more scalable over $K$ than SDR. The complexity of SDR increases noticeably over $K$, and thus, it is not a suitable initialization method for very large systems. Overall, in terms of the total computation time (including initialization and the algorithm itself), for $N = 100$ and $K \leq 20$, SDR-ASCA is the fastest among all algorithms. AIM-ASCA offers comparable computation time as SDR-ASCA. For $N = 100$ and $K \geq 35$, AIM-ASCA and EIM-ESCA offers faster computation time than the rest using SDR for initialization.

Comparison Summary: Based on the above simulation analysis, we have the following comparison summary of the two proposed algorithms:

- Among the initialization methods (EIM, AIM, and SDR), SDR provides the best initialization point, which leads to faster computation for both ESCA and ASCA (i.e., SDR-ESCA and SDR-ASCA). However, the computational complexity of SDR is high. As $K$ becomes large (e.g., $K > 20$), SDR-based initialization becomes computationally expensive and is not recommended. EIM and AIM are both very low-complexity methods over $N$ and $K$. AIM provides faster initialization than EIM. However, the quality of the initial point that AIM provides deteriorates over $N$, leading to a noticeable increase of computation time by AIM-ASCA over $N$. In contrast, the computation time of EIM-ESCA remains roughly unchanged over $N$.
- Both ESCA and ASCA provide closed-form updates in each iteration. The computational complexity of ESCA and ASCA grow over $K$. From our study, we conclude that:
  - For the system with users per group $K \leq 15$, SDR-ASCA is generally the fastest among all algorithms. When $N \leq 100$, AIM-ASCA is similar to SDR-ASCA. However, the complexity of SDR-ASCA increases over $K$, and that of AIM-ASCA increases over $N$. As both $K$ and $N$ grow, SDR-ASCA and AIM-ASCA have higher computation time and are no longer preferred.
  - For a large-scale system where both $N$ and $K$ are large (e.g., $N \geq 500$, $K \geq 35$), EIM-ESCA is expected to provide the fastest computation time among these algorithms and thus is preferred.

### C. Performance Comparison for the MMF Problem

We now present the performance of our proposed Algorithm 3 for the MMF problem $S_o$. We set the maximum transmit power budget against noise variance as $P/\sigma^2 = 10 \text{ dB}$. Fig. 6 plots the average minimum SINR vs. $N$. Table III shows the corresponding computation time by these algorithms, for $G = 3$ and $K = 10$. Both ESCA-Bisection and ASCA-Bisection provide near-identical performance to that of CSCA-Bisection, and they are nearly optimal as compared with the upper bound, but with much lower computation times.

#### Table III

**MMF: Average Computation Time over $N$ (sec.) ($G = 3$, $K = 10$).**

| $N$    | 1  | 50 | 100 | 200 | 300 | 400 |
|--------|----|----|-----|-----|-----|-----|
| ESCA-Scaling | 0.482 | 0.327 | 0.237 | 0.233 | 0.224 |
| ASCA-Scaling | 0.099 | 0.076 | 0.051 | 0.061 | 0.071 |
| ESCA-Bisection | 16.43 | 21.11 | 33.02 | 46.18 | 44.63 |
| ASCA-Bisection | 12.35 | 12.11 | 13.06 | 14.09 | 17.94 |
| CSCA-Bisection [2] | 99.04 | 87.33 | 81.62 | 84.10 | 99.86 |
| SDR-Bisection [2] | 11.19 | 15.46 | 19.85 | 15.82 | 22.12 |

Fig. 6: MMF: Minimum SINR vs. $N$ ($G = 3$, $K = 10$).

VIII. CONCLUSION

In this work, exploiting the optimal multicast beamforming structure, we proposed two fast computational algorithms,
Appendix A

Derivation of $d_{ik}^o$ in (26)

The Lagrangian of $P_{	ext{dub}}(\nu)$ is given by

$$
\mathcal{L}(d_{ik}, \nu_{ik}) = \sum_{j=1}^{G} |d_{jik} - e_{1,j,ik}^{(n)}|^2 + \nu_{ik} e_{2,ik} + \nu_{ik} \gamma_{ik} \sum_{j \neq i} |d_{jik}|^2 - 2 \nu_{ik} \Re\{d_{ik} e_{3,ik}^o\}
$$

where $\nu_{ik} \geq 0$ is the Lagrange multiplier associated with constraint (24). Since $P_{	ext{dub}}(\nu)$ is convex, the optimal $d_{ik}^o$ and $\nu_{ik}^o$ satisfy the KKT conditions [18]. Setting the gradient $\nabla d_{ik} \mathcal{L}(d_{ik}, \nu_{ik}^o) = 0$, we obtain $d_{ik}^o$ in (26). Substituting the expression of $d_{ik}^o$ in (26) into the constraint of $P_{\text{dub}}(\nu)$ yields

$$
\nu_{ik}^o \leq e_{2,ik} + \gamma_{ik} \sum_{j \neq i} \frac{|e_{1,j,ik}^{(n)}|^2}{1 + \nu_{ik}^o \gamma_{ik}^2} - 2 \nu_{ik} \Re\{e_{3,ik} e_{1,ik}^{(n)}\}
$$

$$
- 2 \nu_{ik}^o |e_{3,ik}^o|^2 \leq 0.
$$

It is shown in [21, Appendix A] that the function $f(\nu_{ik}^o)$ is strictly decreasing for $\nu_{ik}^o \geq 0$. By the complementary slackness condition, we have $\nu_{ik}^o f(\nu_{ik}^o) = 0$. If $f(0) \leq 0$, then $f(\nu_{ik}^o) < 0$ for any $\nu_{ik}^o \geq 0$, and we have $\nu_{ik}^o = 0$. Otherwise, $\nu_{ik}^o = 0$ is not an feasible dual solution for $P_{\text{dub}}(\nu)$; thus, $f(\nu_{ik}^o) = 0$, and $\nu_{ik}^o$ is the unique real positive root of the cubic equation (27), whose roots can be obtained by the cubic formula [21].

Appendix B

Closed-Form Updating Steps for AIM

1) Closed-Form $d$-update in (31): Similar to that in Section V-A, given $a^{(n)}$ and $q^{(n)}$, the optimization problem in (31) can be decomposed into $K_{\text{aim}}$ subproblems, one for each user $k$ in group $i$ given by

$$
P'_{\text{dub}} : \min_{d_{ik}} \sum_{j=1}^{G} \left|d_{jik} - e_{1,j,ik}^{(n)} \right|^2$$

s.t. $\gamma_{ik} \sum_{j \neq i} |d_{jik}|^2 + \gamma_{ik} \sigma^2 - |d_{ik}|^2 \leq 0. \quad (36)$

where $e_{1,j,ik}^{(n)} \triangleq a_{ik}^{(n)} H_{jik} - d_{jik}$, for $k \in K_i, i,j \in G$. Problem $P'_{\text{dub}}$ is a non-convex QCQP-1 problem similar to $P_{\text{dub}}(\nu)$. Since it satisfies Slater’s condition, the strong duality holds [18, Appendix B], and the optimal $d_{ik}^o$ satisfies the KKT conditions. Let $\mu_{ik}^o \geq 0$ be the optimal Lagrange multiplier associated with constraint (36). For $P'_{\text{dub}}$ being feasible, we have $0 \leq \mu_{ik}^o \leq 1$ [18, Appendix B]. Thus, using the KKT conditions and with a procedure similar to that in Appendix A, we have the closed-form optimal solution $d_{ik}^o$ to $P'_{\text{dub}}$ given by

$$
d_{ik}^o = \begin{cases} 
\frac{e_{1,i,ik}^{(n)}}{1 + \mu_{ik}^o \gamma_{ik}} & \text{if } j = i, \\
\frac{e_{1,j,ik}^{(n)}}{1 + \mu_{ik}^o \gamma_{ik}} & \text{otherwise}
\end{cases}
$$

where $\mu_{ik}^o = 0$ if $\gamma_{ik} \sum_{j \neq i} |e_{1,j,ik}^{(n)}|^2 + \gamma_{ik} \sigma^2 - |e_{1,i,ik}^{(n)}|^2 \leq 0$; otherwise the following quartic equation is guaranteed to have a unique root in $(0,1)$, and $\mu_{ik}^o$ is this unique root:

$$
\gamma_{ik} \sum_{j \neq i} |e_{1,j,ik}^{(n)}|^2 + \gamma_{ik} \sigma^2 - \frac{|e_{1,i,ik}^{(n)}|^2}{(1 + \mu_{ik}^o \gamma_{ik})^2} = 0.
$$
2) Closed-Form $a$-update in (32): Given $\mathbf{d}^{(n+1)}$ and $\hat{\mathbf{q}}^{(n)}$, the optimization problem in (32) can be decomposed into $G$ subproblems, one for each group $j$ given by

$$
P_{\text{sub}}: \min_\mathbf{a} \sum_{i=1}^{G} \sum_{k=1}^{K_i} \left| a_j^{(n+1)} - a_j^H f_{ijk} + \hat{q}_{ijk}^{(n)} \right|^2,
$$

which is an unconstrained quadratic convex optimization problem. The closed-form solution to $P_{\text{sub}}$ is expressed as

$$
a_j^{(n+1)} = \left( \sum_{i=1}^{G} \sum_{k=1}^{K_i} f_{ijk}^H f_{ijk} \right)^{-1} \sum_{i=1}^{G} \sum_{k=1}^{K_i} \left( d_{ijk}^{(n+1)} + \hat{q}_{ijk}^{(n)} \right)^* f_{ijk}.
$$

APPENDIX C

PROOF OF PROPOSITION 1

Proof: It is straightforward to check that the scaled beamforming vector $w^0(\gamma, P) = \sqrt{P/\|P\|} w^{(n+1)}(\gamma)$ satisfies constraint (2) and therefore is a feasible solution to $S_0(\gamma, P)$. The achieved objective value $t^0(\gamma, P)$ corresponding to $w^{(n+1)}(\gamma)$ is expressed in terms of $w^0(\gamma)$ as

$$
t^0(\gamma, P) = \min_{i,k} \frac{1}{P_{\text{sub}}(\gamma)} \left[ \sum_{i,k} \left( \frac{P_{\text{sub}}(\gamma)}{P} \right) \frac{1}{P_{\text{sub}}(\gamma)} \left| h_{ik}^H w_0(\gamma) \right|^2 + \sigma^2 \right].
$$

Define $t^0(\gamma, P) \triangleq \min_{i,k} \frac{1}{P_{\text{sub}}(\gamma)} \left[ \sum_{i,k} \left( \frac{P_{\text{sub}}(\gamma)}{P} \right) \frac{1}{P_{\text{sub}}(\gamma)} \left| h_{ik}^H w_0(\gamma) \right|^2 + \sigma^2 \right]$. Since $w_0(\gamma)$ satisfies constraint (1) in $P_0(\gamma)$ as a feasible solution, we have

$$
t^0(\gamma, P) \geq \frac{1}{P_{\text{sub}}(\gamma)} \sum_{i,k} \left( \frac{P_{\text{sub}}(\gamma)}{P} \right) \frac{1}{P_{\text{sub}}(\gamma)} \left| h_{ik}^H w_0(\gamma) \right|^2 + \sigma^2
$$

Based on this, from (37), we have

$$
t^0(\gamma, P) \geq \min_{i,k} \left( \frac{P_{\text{sub}}(\gamma)}{P} \right) \frac{1}{P_{\text{sub}}(\gamma)} \left| h_{ik}^H \frac{w_0(\gamma)}{\sqrt{P/\|P\|}} \right|^2 + \sigma^2.
$$

Combining (38) and (41), we have (35).

REFERENCES

[1] C. Zhang, M. Dong, and B. Liang, “First-order fast algorithm for structurally optimal multi-group multicast beamforming in large-scale systems,” in Proc. IEEE Int. Conf. Acoust., Speech, Signal Process., Jun. 2021, pp. 4790–4794.

[2] M. Dong and Q. Wang, “Multi-group multicast beamforming: Optimal and efficient algorithms,” IEEE Trans. Signal Process., vol. 68, pp. 3738–3753, May 2020.

[3] E. G. Larsson, O. Edfors, F. Tufvesson, and T. L. Marzetta, “Massive MIMO for next generation wireless systems,” IEEE Commun. Mag., vol. 52, no. 2, pp. 186–195, Feb. 2014.

[4] N. D. Sidiroopoulos, T. N. Davidson, and Z.-Q. Luo, “Transmit beamforming for physical-layer multicasting,” IEEE Trans. Signal Process., vol. 54, no. 5, pp. 2238–2251, May 2006.

[5] E. Karipidis, N. D. Sidiroopoulos, and Z.-Q. Luo, “Quality of service and max-min fair transmit beamforming to multiple cochannel multicast groups,” IEEE Trans. Signal Process., vol. 56, no. 3, pp. 1268–1279, Mar. 2008.

[6] D. Christopoulos, S. Chatzinotas, and B. Ottersten, “Weighted fair multicast multigroup beamforming under per-antenna power constraints,” IEEE Trans. Signal Process., vol. 62, no. 19, pp. 5132–5142, Oct. 2014.

[7] M. Jordan, X. Gong, and G. Ascheid, “Multicell multicast beamforming with delayed SNR feedback,” in 2009 IEEE Global Telecommun. Conf., Nov. 2009, pp. 1–6.

[8] Z. Xiang, M. Tao, and X. Wang, “Coordinated multicast beamforming in multicell networks,” IEEE Trans. Wireless Commun., vol. 12, no. 1, pp. 12–21, Jan. 2013.

[9] N. Bornhorst, M. Pesavento, and A. B. Gershman, “Distributed beamforming for multi-group multicasting relay networks,” IEEE Trans. Signal Process., vol. 60, no. 1, pp. 221–232, Jan. 2012.

[10] M. Dong and B. Liang, “Multicast relay beamforming through dual approach,” in Proc. IEEE Int. Workshop Comput. Advances Multi-Sensor Adaptive Process., Dec. 2013, pp. 492–495.

[11] K. T. Phan, S. A. Vorobyov, N. D. Sidiroopoulos, and C. Tellambura, “Spectrum sharing in wireless networks via QoS-aware secondary multicast beamforming,” IEEE Trans. Signal Process., vol. 57, no. 6, pp. 2323–2335, Jun. 2009.

[12] Y. Huang, Q. Li, W.-K. Ma, and S. Zhang, “Robust multicast beamforming for spectrum sharing-based cognitive radios,” IEEE Trans. Signal Process., vol. 60, no. 1, pp. 527–533, Jan. 2012.

[13] Z.-Q. Luo, W.-K. Ma, A. M.-C. So, Y. Ye, and S. Zhang, “Semidefinite relaxation of quadratic optimization problems,” IEEE Signal Process. Mag., vol. 27, no. 3, pp. 20–34, May 2010.

[14] R. Marks and G. P. Wright, “A general convex approximation algorithm for nonconvex mathematical programs,” Oper. Res., vol. 26, no. 4, pp. 681–683, Aug. 1978.

[15] L. Tran, M. F. Hanif, and M. Juntti, “A conic quadratic programming approach to physical layer multicasting for large-scale antenna arrays,” IEEE Signal Process. Lett., vol. 21, no. 1, pp. 114–117, Jan. 2014.

[16] O. Mehnata, K. Huang, B. Gopalakrishnan, A. Konar, and N. D. Sidiroopoulos, “Feasible point pursuit and successive approximation of non-convex QCQPs,” IEEE Signal Process. Lett., vol. 22, no. 7, pp. 804–808, Jul. 2015.

[17] D. Christopoulos, S. Chatzinotas, and B. Ottersten, “Multicast multigroup beamforming with per-antenna power constrained large-scale arrays,” in Proc. IEEE Int. Workshop Signal Process. Advances Wireless Commun., Jun. 2015, pp. 271–275.

[18] S. Boyd and L. Vandenberghe, Convex Optimization. Cambridge, U.K.: Cambridge Univ. Press, 2004.

[19] A. Konar and N. D. Sidiroopoulos, “Fast approximation algorithms for a class of non-convex QCQPs with combinatorial structures,” IEEE Trans. Signal Process., vol. 65, no. 13, pp. 3494–3509, Jul. 2017.

[20] M. S. Ibrahim, A. Konar, and N. D. Sidiroopoulos, “Fast algorithms for joint multicast beamforming and antenna selection in massive MIMO,” IEEE Trans. Signal Process., vol. 68, pp. 1897–1900, Mar. 2020.

[21] E. Chen and M. Tao, “ADMM-based fast algorithm for multi-group multicast beamforming in large-scale wireless systems,” IEEE Trans. Commun., vol. 65, no. 6, pp. 2685–2698, Jun. 2017.

[22] M. Sadeghi, L. Sangunietti, R. Cuiulet, and C. Yuen, “Reducing the computational complexity of multicasting in large-scale antenna systems,” IEEE Trans. Wireless Commun., vol. 16, no. 5, pp. 2963–2975, May 2017.
S. Boyd, N. Parikh, E. Chu, B. Peleato, and J. Eckstein, “Distributed optimization and statistical learning via the alternating direction method of multipliers,” Found. Trends Mach. Learn., vol. 3, no. 1, pp. 1–122, 2011.

E. N. Khobotov, “Modification of the extra-gradient method for solving variational inequalities and certain optimization problems,” USSR Comput. Math. Math. Phys., vol. 27, no. 5, pp. 120–127, 1987.

P. Marcotte, “Application of Khobotov’s algorithm to variational inequalities and network equilibrium problems,” Inf. Syst. Oper. Res., vol. 29, no. 4, pp. 258–270, Nov. 1991.

S. Boyd, N. Parikh, E. Chu, B. Peleato, and J. Eckstein, “Distributed optimization and statistical learning via the alternating direction method of multipliers,” Found. Trends Mach. Learn., vol. 3, no. 1, pp. 1–122, 2011.

Min Dong (Senior Member, IEEE) received the B.Eng. degree from Tsinghua University, Beijing, China, in 1998, and the Ph.D. degree in electrical and computer engineering with a minor in applied mathematics from Cornell University, Ithaca, NY, in 2004. From 2004 to 2008, she was with Qualcomm Research, Qualcomm Inc., San Diego, CA. Since 2008, she has been with Ontario Tech University, where she is currently a Professor in the Department of Electrical, Computer and Software Engineering. Her research interests include wireless communications, statistical signal processing, learning techniques, optimization and control applications in cyber-physical systems.

Dr. Dong was the recipient of the Early Researcher Award from the Ontario Ministry of Research and Innovation in 2012, the Best Paper Award at IEEE ICC in 2012, and the 2004 IEEE Signal Processing Society Best Paper Award. She is a co-author of the Best Student Paper at IEEE SPAWC 2021 and the Best Student Paper of Signal Processing for Communications and Networking at IEEE ICASSP 2016. She is an Editor for the IEEE TRANSACTIONS ON WIRELESS COMMUNICATIONS and IEEE OPEN JOURNAL OF SIGNAL PROCESSING. She was an Associate Editor for the IEEE TRANSACTIONS ON SIGNAL PROCESSING (2010-2014) and IEEE SIGNAL PROCESSING LETTERS (2009-2013). She was on the Steering Committee of IEEE TRANSACTIONS ON MOBILE COMPUTING (2019-2021). She was an Elected Member of the IEEE Signal Processing for Communications and Networking Technical Committee of IEEE Signal Processing Society (2013-2018). She was the lead Co-Chair of the Communications and Networks to Enable the Smart Grid Symposium at the IEEE International Conference on Smart Grid Communications in 2014.

Ben Liang (Fellow, IEEE) received honors-simultaneous B.Sc. (valedictorian) and M.Sc. degrees in Electrical Engineering from Polytechnic University (now the engineering school of New York University) in 1997 and the Ph.D. degree in Electrical Engineering with a minor in Computer Science from Cornell University in 2001. He was a visiting lecturer and post-doctoral research associate at Cornell University in the 2001–2002 academic year. He joined the Department of Electrical and Computer Engineering at the University of Toronto in 2002, where he is now Professor and L. Lau Chair in Electrical and Computer Engineering. His current research interests are in networked systems and mobile communications. He is an associate editor for the IEEE TRANSACTIONS ON MOBILE COMPUTING and has served on the editorial boards of the IEEE TRANSACTIONS ON COMMUNICATIONS, the IEEE TRANSACTIONS ON WIRELESS COMMUNICATIONS, and Security and Communication Networks (Wiley). He regularly serves on the organizational and technical committees of a number of conferences. He is a Fellow of IEEE and a member of ACM and Tau Beta Pi.