Exploiting User Mobility for WiFi RTT Positioning: A Geometric Approach

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Abstract—Due to the massive deployment of WiFi APs and its accessibility to various positioning elements, WiFi positioning is a key enabler to provide seamless and ubiquitous location services to users. There are various kinds of WiFi positioning technologies, depending on the concerned positioning element. Among them, round-trip time (RTT) measured by a fine-timing measurement protocol has received great attention recently. It provides an acceptable ranging accuracy near the service requirements in favorable environments when a line-of-sight (LOS) path exists. Otherwise, a signal is detoured along with non-LOS paths, making the resultant ranging results different from the ground-truth. The difference between the two is called an RTT bias, which is the main reason for poor positioning performance. To address it, we aim at leveraging the history of user mobility detected by a smartphone’s inertial measurement units, called pedestrian dead reckoning (PDR). Specifically, PDR provides the geographic relation among adjacent locations, guiding the resultant positioning estimates’ sequence not to deviate from the user trajectory. To this end, we describe their relations as multiple geometric equations, enabling us to render a novel positioning algorithm with acceptable accuracy. The algorithm is designed into two phases. First, an RTT bias of each AP can be compensated by leveraging the geometric relation mentioned above. It provides a user’s relative trajectory defined on the local coordinate system of the concerned AP. Second, the user’s absolute trajectory can be found by rotating every relative trajectory to be aligned, called trajectory alignment. The proposed algorithm gives a unique position when the number of detected steps and APs is at least 4 and 3 for linear mobility and 5 and 2 for arbitrary mobility. Various field experiments extensively verify the proposed algorithm’s effectiveness that the average positioning error is approximately 0.369 (m) and 1.705 (m) in LOS and NLOS environments, respectively.

I. INTRODUCTION

Estimating a user’s location, called positioning, has become vital as a clue to assimilate his behaviors and predict the demands to realize the vision of smart cities [1], [2]. Along with the appearance of smartphones equipping various built-in sensors and communication modules, positioning has become an attractive research area due to its promising potential capable of combining various techniques. Among them, this work focuses on round-trip times (RTTs) from multiple WiFi access points (APs) and a user’s mobility pattern measured by the built-in sensors, each of which has a different view from the positioning perspective. A WiFi RTT-based approach gives a macroscopic view of the region where the user is likely to be located during the movement. On the other hand, the user movement pattern provides a microscopic view of how the user is moving at a specific instant. As a result, combining the two renders the relation between point estimates, forming a geometric representation concerning the given measurements. It enables the design of a novel algorithm to achieve accurate positioning.

A. WiFi Positioning

As the massive number of WiFi APs have been deployed in our surroundings, WiFi positioning has received significant attention to providing users seamless positioning services anytime and anywhere [3]. Depending on the means used to capture specific physical properties of radio signals, two kinds of approaches exist in the literature: received signal strength (RSS)-based and RTT-based approaches. The former uses a mathematical path-loss model to translate the measured RSS to the distance to a WiFi AP. Given the ranging results to at least 3 APs, the user’s position is uniquely estimated using a multilateration method. RSS has been widely used as a fundamental positioning element due to its simple accessibility that RSS is measurable in a commercial off-the-shelf Android smartphone. However, the randomness of a radio signal, e.g., shadowing and short-term fading, makes it challenging to obtain reliable ranging results, especially in indoor environments where numerous reflectors and blockages exist. On the other hand, RTT is a relatively reliable measurement since it exploits one basic theory of classical physics that the speed of signal propagation is constant at light speed $c = 3 \cdot 10^8$ (m/sec). The measurement of RTT is enabled by fine timing measurement (FTM) protocol, which was firstly introduced in IEEE 802.11 mc [4]. After several handshaking signals between a smartphone and the paired AP, the timing instants of the signal receptions are shared, enabling the computation of RTT. With several super-resolution methods, FTM returns a precise RTT estimate at picosecond granularity in favorable environments, i.e., a line-of-sight (LOS) scenario [5]. Nevertheless, complicated surroundings with numerous reflectors and blockages make the concerned radio signal detour differently from a LOS path, called a non-LOS (NLOS) path. The resultant ranging results is thus biased, as mentioned in [6], which is the main reason behind the performance limit. Several methods have been suggested in the literature to overcome the limitation. A primary way is to identify whether the observed signal path is LOS or NLOS and calibrate the bias. In [7], the likelihood of a LOS path is derived, following the assumption that RSS is a Gaussian random variable. The recent trend of machine learning enables the
LOS/NLOS identification without any statistical hypothesis. For example, in [8] and [9], LOS and NLOS paths are identified by supervised learning techniques, e.g., support vector machine and artificial neural network respectively, of which the performances depend on the number of labeled data. In [10], the technique of unsupervised learning is used to infer the bias by designing a novel cost function underlying the fact that the spatial information, e.g., location, distance, and velocity, is temporally correlated. However, all machine learning-based techniques mentioned above need an offline phase such that support vectors or neural networks should be trained in advance concerning all possible sites, making their usability and scalability limited.

B. Estimating User Mobility

Understanding user mobility gives significant benefits for seamless positioning services by enabling a user to estimate his location in global positioning system (GPS)-restricted areas, i.e., inside buildings or tunnels. A user’s current location can be updated from the latest GPS signal by integrating the trajectory of user movement, called dead reckoning (DR). The performance of DR relies on the accuracy of estimating a user’s mobility, and DR works efficiently for vehicular positioning since its on-board sensors, called inertial measurement units (IMUs), can accurately measure velocity, orientation, etc [11].

Noting that modern smartphones also possess IMUs, the concept of DR can be applied to the positioning using a smartphone, defined as pedestrian DR (PDR) [12]. Basically, the IMUs of a typical smartphone comprise accelerometer, gyroscope, and magnetometer, each of which observes a user’s mobility from a different aspect. An accelerometer can count a user’s number of walking steps from the repetitive patterns of up-and-down accelerations, translated into the corresponding total moving distance by multiplying his step length. Next, a gyroscope can recognize turning direction to right or left sides when its measurement is suddenly changed. Last, a magnetometer can detect a heading direction in favorable conditions such as outdoor-like spaces [13]. Combining them leads to constructing the user’s full trajectory.

Despite its advantages mentioned above, the effectiveness of PDR as a standalone positioning technique is questionable due to the following reasons. First, a user’s step length should be known in advance as a prerequisite to translate the number of steps into the distance. Several formulas have been proposed in the literature to estimate it without the user’s direct input (see, e.g., [14] and [15]), most of which are designed based on the rule of a thumb. On the other hand, it should reflect the user’s characteristics such as height, weight, and speed and acceleration of walking, hindering the generalization into a simple formula. Second, a heading direction estimated by a magnetometer has a significant offset, affected by a magnetic distortion due to several indoor materials and a misalignment between the smartphone’s heading direction and the real moving direction [16]. Third, the concerned scenarios of PDR are mostly indoor, where a GPS signal is hardly detected. In other words, its accuracy cannot be guaranteed and deteriorates as time passes due to the accumulation of estimation errors. To cope with the above issues, it is recommended to incorporate PDR into other positioning systems by providing additional location information to calibrate these measurements. The recent advancement of this area can be found in numerous surveys, such as [17].

C. Integrating WiFi Positioning and PDR

The limitations of WiFi RTT positioning and PDR mentioned above are complementary and can be overcome by their integration, which is the main theme of this work. On the one hand, the IMUs of a smartphone excluding a magnetometer work independently of surrounding environments, playing a pivotal role in compensating an environment-dependent RTT bias. On the other hand, the positions estimated by WiFi positioning can fill the missing information to complete the user’s trajectory. Motivated by the synergy effect, a few recent works have been studied in the literature, most of which rely on a technique of an extended Kalman filter (EKF). EKF is a well-known nonlinear state estimator utilizing a series of sequential observations with measurement noises. In [18], for example, WiFi positioning’s RTT bias, PDR’s step length, and heading direction are jointly calibrated using EKF by inputting the raw measurements of PDR and WiFi RTTs. A similar approach is made in [19], where the positions estimated by PDR and the distances converted from WiFi RTTs are used as inputs of EKF. In [20], the PDR’s measurements are pre-calibrated by EKF. The result is then fused with the estimated distances from WiFi RTTs to enable 3D positioning by an unscented particle filter, another state estimation filter. It is shown that all approaches mentioned above can provide more accurate positioning results than standalone techniques. On the other hand, the convergence of EKF is not guaranteed and sometimes diverges due to the lack of statistical knowledge of measurement noises and the linearization of nonlinear functions required to compute the inputs’ covariance matrices. The resultant location estimates can be inconsistent depending on the selection of initial settings, calling for diversifying positioning approaches other than EKF.

D. Main Contributions

This work aims at designing a new hybrid positioning design combining WiFi RTTs and PDR measurements without EKF. To this end, we attempt to adopt a geometric approach describing the relationship between multiple measurements in mathematical form. This approach provides twofold benefits from the positioning perspective. First, the relations mentioned above are summarized as a list of equations, forming a system of equations (SOE) whose unknowns are related to RTT biases, a user’s step length, heading direction, etc. The SOE can be iteratively solved using well-known optimization techniques, guaranteeing the convergence to a local optimal, and possible to solve it by a single matrix inversion if the SOE is linear. Second, we can rigorously provide requirements to guarantee the uniqueness and existence of the positioning result, e.g., the minimum numbers of detected walking steps and connected APs, which are equivalent to the conditions.
for unique positioning. It helps design not only a practical positioning algorithm, always returning an accurate position but also efficient WiFi deployments in a concerned area.

There have been several works adopting a geometric approach in different systems such as cellular positioning [21], RADAR [22], and vehicular positioning [23]. On the other hand, to the best of our knowledge, this work represents the first attempt to design a geometry-based positioning algorithm concerning the integration between FTM and PDR. The main contributions are summarized below.

- **Joint estimations of an RTT bias and a step length:**
  We aim at enhancing an RTT-based distance estimation by offsetting the RTT bias, coupled with PDR information like the number of walking steps, direction changes, and step length. Noting that all information excluding step length can be measurable by PDR, an SOE is formed to jointly estimate two unknowns of RTT bias and step length. In the linear mobility case, the SOE can be transformed into a linear structure solvable by a matrix inversion when at least 4 steps are detected. In the arbitrary mobility case, a one-dimensional (1D) search enables us to solve the SOE, guaranteeing its unique solution if at least 5 steps are detected. Besides, the positioning error due to measurement noises can be significantly reduced by leveraging diversified positioning results through multiple SOEs formed by different step combinations.

- **New positioning method using a trajectory alignment:**
  The user’s sequential positions during his movement can be estimated by aligning multiple trajectories derived from the estimated RTT bias of each WiFi AP and step length. It is equivalent to find the user’s initial heading direction, another information challenging to be estimated using a smartphone, as aforementioned. The minimum number of WiFi APs required to finding the user’s position uniquely is 3 or 2 depending on the linear or arbitrary mobility pattern, respectively. Given the requirement of the minimum number of APs and with measurement noises, we develop a 1D search-based algorithm to make all trajectories aligned as closely as possible.

- **Verification by field experiments:**
  The proposed positioning algorithms are evaluated based on field experiments and found to be effective. Specifically, the resultant average positioning error is reduced to 1.71 (m) in unfavorable environments like an underground parking lot, where numerous obstacles exist, such as vehicles, walls, and pillars.

The remainder of the paper is organized as follows. Section II introduces the system model, measurement procedures, and the overview of the entire positioning algorithm. Section III presents a technique jointly estimating RTT bias and step length for a single WiFi AP. Given the enhanced ranging results of multiple WiFi APs, the positioning algorithm finding the user’s location is developed in Section IV. Experiment results are presented in Section V, followed by concluding remarks in Section VI.

II. SYSTEM MODEL

Consider the scenario comprising one user with a smartphone and \( M \) WiFi APs, denoted by a set \( \mathbb{M} = \{1, \cdots, M\} \), as shown in Fig. 1. The smartphone has built-in IMU sensors and a WiFi module required to measure the user’s mobility pattern and RTTs, respectively. The detailed measurement procedures are firstly explained. Next, the geometric relations between the measurements and the user’s locations are derived. Last, the overview of the proposed approach is described.

A. Measurements

This subsection explains the measurement procedures and the corresponding outputs required to form geometric relations in the next subsection.

1) Mobility Pattern: Among IMU sensors in the smartphone, we use an accelerometer and a gyroscope to detect walking steps and turning directions, respectively. To be specific, the accelerometer’s up-and-down acceleration is referred to as the user’s one walking step. Consider that \( N \) walking steps are detected, each of which the index is \( n \in \mathbb{N} \), where \( \mathbb{N} = \{1, \cdots, N\} \). The instant of detecting the \( n \)-th step is denoted by \( t_n \). When detecting a walking step at \( t_n \), the smartphone checks its gyroscope between \( t_n \) and \( t_{n+1} \), whether the user’s movement direction is changed or not. Denote \( \mu_n \) the change of his moving direction when the \( n \)-th step is detected. We set \( \mu_0 = 0 \) unless the direction is changed. Besides, denote \( \theta_n \) the accumulate direction change by the \( n \)-th step, namely, \( \theta_n = \sum_{k=1}^{n} \mu_k \). The initial values of \( \mu_1 \) and \( \theta_1 \) are zero without loss of generality. All measurements mentioned above are illustrated in Fig. 2(a).

2) RTT: The smartphone’s WiFi modules and all WiFi APs support IEEE 802.11 mc or later specifications, enabling the measurement of RTTs between them via the FTM protocol. The FTM protocol is initiated whenever a walking step is detected at time \( t_n \). Denote \( \tau_n = [\tau_{n}^{(1)}, \cdots, \tau_{n}^{(M)}] \) the vector of the measured RTTs corresponding to the \( n \)-th walking step.

\(^1\)This work does not use a magnetometer due to its severe distortions as stated in Sec. I-B. Incorporating the measurement of a magnetometer with advanced calibration techniques such as [24] and [25] is interesting, deserving further investigation in the future.
walking pattern (see, e.g., [3]). Both $\omega$ and $d$ are unknowns to be estimated.

Second, the relation of the RTT $\tau_n^{(m)}$ to the user’s location $p_n$ is described as follows. The RTT $\tau_n^{(m)}$ can be translated into the propagation distance by multiplying $\frac{c}{2}$ where $c$ is light speed. It is in general biased and larger than the direct distance $\| p_{AP}^{(m)} - p_n \|$, where $p_{AP}^{(m)} = [x_{AP}^{(m)}, y_{AP}^{(m)}]^T$ is AP $m$’s coordinates and $\| \cdot \|$ represents the Euclidean distance. The relation between the two is thus given as

$$\frac{c \cdot \tau_n^{(m)}}{2} = \| p_{AP}^{(m)} - p_n \| + b, \ \forall n \in \mathbb{N}, \ \forall m \in \mathbb{M},$$

where $b$ represents the AP $m$’s RTT bias. The bias $b$ is a random variable drawn from an unknown stochastic distribution, since it is affected by various factors such as carrier frequency, bandwidth, and surrounding materials [26], and it is too complex to derive the distribution in a tractable form. The randomness of $b$ is one dominant reason hampering accurate positioning. On the other hand, we regard $b$ as a deterministic value as stated in the following assumption, which helps design a tractable algorithm in the sequel.

**Assumption 1** (Deterministic Bias). The bias experienced from the same WiFi AP is assumed to be constant but unknown, denoted by $b^{(m)}$ for all $m \in \mathbb{M}$.

**Remark 1** (Validity of Deterministic Bias). The underlying intuition behind Assumption 1 is that the bias dominantly depends on the concerned AP’s surroundings, which are unlikely to be changed within a short duration. The algorithm proposed in the sequel relies on the period when RTT bias is relatively constant, playing a pivotal role in determining positioning accuracy. The algorithm thus works well if Assumption 1 is valid within a finite period.

As a result, the above is rewritten as

$$\| p_{AP}^{(m)} - p_n \| + b^{(m)} = \frac{c \cdot \tau_n^{(m)}}{2} = r_n^{(m)}, \ \forall n \in \mathbb{N}, \ \forall m \in \mathbb{M},$$

where $r_n^{(m)}$ represents the propagation distance directly obtained from the RTT $\tau_n^{(m)}$.

**C. Procedure Overview**

This subsection previews the entire procedure to estimate the user’s locations $\{p_n\}$ as follows.

1) **Joint Bias-and-Step Length Estimation**: First, the RTT bias of each AP and the user’s step length are jointly estimated. Let us explain the case of AP $m$ as an example. Collect all RTTs measured from AP $m$ and the user’s direction changes, denoted by $\tau^{(m)} = [\tau_1^{(m)}, \cdots, \tau_N^{(m)}]$ and $\theta = [\theta_1, \cdots, \theta_N]$, respectively. Using $\tau^{(m)}$ and $\theta$, the AP’s deterministic bias $b^{(m)}$ and the user’s step length, say $d^{(m)}$, can be found by solving SOE derived from the geometric relations of (1) and (2). The detailed procedure is elaborated in Section III.
2) User Location Estimation: Second, the sequence of the user’s locations \( \{p_n\} \) are estimated using the above estimations of \( d(m) \) and \( b(m) \) that provide the user’s relative locations from AP \( m \). All relative locations can be aligned into single ones using a new positioning method, corresponding to the user’s real locations \( \{p_n\} \). The detailed procedure is elaborated in Section IV.

III. ENHANCING RTT-BASED RANGING VIA JOINT BIAS & STEP LENGTH ESTIMATION

In this section, we aim at improving RTT-based ranging performance by jointly estimating the RTT bias and step length. First, a coordinate system is transformed to facilitate algorithm designs. Next, joint bias-and-step length estimation algorithms are developed for cases of linear and arbitrary mobilities. Last, a new algorithm based on multiple combinations of steps is proposed to reduce performance degradation due to measurement noises.

A. Transformation to a Local Coordinate System

This section focuses on a pair of the user and a single WiFi AP. For brevity, the location and the bias of the concerned AP are denoted by \( p_{AP} \) and \( b \), respectively, by omitting the AP’s index \( m \). Then, we transform a global coordinate system into a local coordinate one by shifting the origin into \( p_{AP} \) and rotating the X-axis aligned with the heading direction \( \omega \), as shown in Fig. 3. Denote \( z_n = [q_n, u_n]^T \) the user’s location redefined in the local coordinates. Then, (1) and (2) are rewritten as

\[
\begin{align*}
    z_{n+1} &= z_n + d \begin{bmatrix} \cos(\theta_n), \sin(\theta_n) \end{bmatrix}^T \\
    &= z_1 + d \begin{bmatrix} \sum_{j=1}^{n} \cos(\theta_j), \sum_{j=1}^{n} \sin(\theta_j) \end{bmatrix}^T, \\
    r_n &= \|z_n\| + b.
\end{align*}
\]

Plugging (3) into (4) gives

\[
    r_n = \left\| z_1 + d \begin{bmatrix} \sum_{j=1}^{n-1} \cos(\theta_j), \sum_{j=1}^{n-1} \sin(\theta_j) \end{bmatrix} \right\| + b, 
\]

which is rewritten in terms of \( q_1 \) and \( u_1 \) as

\[
    (q_1 + dc_n)^2 + (u_1 + ds_n)^2 = (r_n - b)^2, \quad n \in \mathbb{N},
\]

where \( c_n = \sum_{j=1}^{n-1} \cos(\theta_j) \) and \( s_n = \sum_{j=1}^{n-1} \sin(\theta_j) \). Depending on the user’s mobility patterns, different algorithms are designed, introduced in the following subsections.

B. Case 1: Linear Mobility

Consider the case when the user is moving in a straight line without any direction change, which can be detectable by observing \( \{\theta_n\} \), namely, \( \theta_n = 0, \forall n \in \mathbb{N} \) [see Fig. 3(a)]. It makes \( c_n = 1 \) and \( s_n = 0 \) for all \( n \), and (6) can be reduced as

\[
    q_1^2 + 2aq_1d + d^2n^2d^2 + u_1^2 = r_n^2 - 2r_nb + b^2, \quad n \in \mathbb{N}. \tag{7}
\]

Next, the nonlinear terms in (7) related to \( q_1 \) and \( b \) are eliminated by choosing two reference steps (RSs), denoted by the subset \( \mathbb{S} = \{a_1, a_2\} \subset \mathbb{N} \). The algorithm covering the selection of RSs is explained in the sequel. Given \( \mathbb{S} \), subtracting (7) of the reference steps \( a_1 \) or \( a_2 \) from that of the \( n \)-step gives

\[
    2(n-a)q_1d + (n^2-a^2)d^2 = r_n^2 - r_a^2 - 2b(r_n - r_a), \quad a \in \mathbb{S}. \tag{8}
\]

The first term can be called out by manipulating the above two as

\[
    d^2(a_1 - a_2) + b \cdot 2 \left( \frac{r_n - r_{a_1}}{n - a_1} - \frac{r_n - r_{a_2}}{n - a_2} \right)
    = \left( \frac{r_n^2 - r_{a_1}^2}{n - a_1} - \frac{r_n^2 - r_{a_2}^2}{n - a_2} \right), \quad n \in \mathbb{N} \cap \mathbb{S}^c, \tag{9}
\]

which is linear of \( d^2 \) and \( b \). As a result, we formulate a system of linear equations with two unknowns \( d^2 \) and \( b \) as

\[
    A(\mathbb{S})x = b(\mathbb{S}), \tag{E1}
\]
where \( x = [d^2, b]^T \). For matrix \( A(S) \) and vector \( b(S) \), we have

\[
A(S) = \begin{bmatrix}
a_1 - a_2 & 2 \left( \frac{r_1 - r_{a_1}}{1 - a_2} - \frac{r_1 - r_{a_2}}{1 - a_2} \right) \\
\vdots & \vdots \\
a_1 - a_2 & 2 \left( \frac{r_N - r_{a_1}}{N - a_1} - \frac{r_N - r_{a_2}}{N - a_2} \right)
\end{bmatrix} \in \mathbb{R}^{(N-2) \times 2},
\]

\[
b(S) = \begin{bmatrix}
r_2^2 - r_{a_1}^2 - \frac{r_1^2 - r_{a_2}^2}{1 - a_2} \\
\vdots \\
n_2^2 - r_{a_1}^2 - \frac{r_N^2 - r_{a_2}^2}{N - a_2}
\end{bmatrix} \in \mathbb{R}^{(N-2) \times 1}.
\]

(10)

Problem E1 comprises \((N-2)\) equations with two unknowns of \( d \) and \( b \). It is thus straightforward to provide the feasibility condition of E1 as follows.

**Proposition 1** (Feasible Condition: Linear Mobility). Problem E1 has a unique solution if the number of detected steps are at least 4, namely, \( N \geq 4 \).

With \( N \geq 4 \), E1 can be solved by

\[
x^*(S) = \frac{[d^*(S)]^T, b^*(S)]^T}{[A(S)^T A(S)]^{-1} A(S)^T b(S)}.
\]

(11)

In the presence of significant measurement noise, the matrix \( A(S) \) and the vector \( b(S) \) are corrupted, denoted by \( \hat{A}(S) \) and \( \hat{b}(S) \). Then, E1 is replaced as the following minimization problem:

\[
x^*(S) = \arg \min \| \hat{A}(S)x - \hat{b}(S) \|
\]

\[
= [\hat{A}(S)^T \hat{A}(S)]^{-1} \hat{A}(S)^T \hat{b}(S),
\]

(12)

which has the same structure as (11).

Given the pair of \( d^*(S) \) and \( b^*(S) \), we derive the coordinates of \( z_1 \), denoted by \( z_1^*(S) = [q_1^*(S), u_1^*(S)] \).

First, plugging \( \{d^*(S), b^*(S)\} \) of (11) or (12) into (8) leads to deriving \( q_1^*(S) \) as

\[
q_1^*(S) = \frac{\sum_{a \in S} \sum_{n \in N, n \neq a} r_n^2 - r_n^2 - 2b^*(S)(r_n - r_a) - (n^2 - a^2)(d^*(S))^2}{2(N-1)}. \]

(13)

Next, there exist two possible solutions for \( u_1^*(S) \) satisfying (7) as \( u_1^*(S) \) and \( -u_1^*(S) \), where

\[
u_1^*(S) = \frac{1}{N} \sum_{n \in N} r_n^2 - 2r_nb^*(S) + (b^*(S))^2
\]

\[- (q_1(S))^2 + 2
\]

\[- (q_1(S))^2 + 2
\]

\[- (q_1(S))^2 + 2
\]

It is challenging to distinguish which one is correct under the current setting of a single WiFi AP. On the other hand, it is possible to discriminate the correct one if multiple APs are given, explained in the next section.

\[2\text{For ease of exposition, the matrix } A \text{ and the vector } b \text{ in (10) and (18) are expressed assuming } S \cap \{1,N\} = \emptyset.\]

\[C. \text{ Case 2: Arbitrary Mobility}\]

Consider the case when the user is randomly moving, namely, \( \theta_n \neq 0 \) for some \( n \in \mathbb{N} \) [see Fig. 3(b)]. Letting \( R = \sqrt{q_1^2 + u_1^2} \) and \( \gamma = \tan^{-1} \left( \frac{u_1}{q_1} \right) \) makes (6) as

\[
q_1^2 + u_1^2 + 2d(q_1c_n + u_1s_n) + d^2(c_n^2 + s_n^2)
\]

\[
= R^2 + 2dR \sum_{i=1}^{n-1} \cos(\theta_i - \gamma) + d^2(c_n^2 + s_n^2)
\]

\[
r_n^2 - 2r_nb + b^2,
\]

(15)

where (a) follows from \( q_1 \cos(\theta_n) + u_1 \sin(\theta_n) = R \cos(\theta_n - \gamma) \). Using RSSs \( S = \{a_1, a_2\} \) as in the previous case of linear mobility, we have

\[
2dRf_{n,a}(\gamma) + d^2\eta_{n,a}
\]

\[
r_n^2 - r_n^2 - 2b(r_n - r_a), \quad a \in S, \quad n \in \mathbb{N} \cap \mathbb{S}^c,
\]

(16)

where \( f_{n,a}(\gamma) = \sum_{i=1}^{n-1} \cos(\theta_i - \gamma) - \sum_{i=1}^{a-1} \cos(\theta_a - \gamma) \) and \( \eta_{n,a} = (c_n^2 + s_n^2) - (c_a^2 + s_a^2) \).

The nonlinear term \( 2dRf_{n,a}(\gamma) \) is canceled out by manipulating the above two as

\[
d^2 \left[ f_{n,a}(\gamma) \eta_{n,a} - f_{n,a}(\gamma) \eta_{n,a} \right]
\]

\[
= \alpha_n(S, \gamma)
\]

\[
+ b \cdot 2 \left[ f_{n,a}(\gamma) (r_n - r_a) - f_{n,a}(\gamma) (r_n - r_a) \right]
\]

\[
= \beta_n(S, \gamma)
\]

\[
f_{n,a}(\gamma) (r_n^2 - r_a^2) - f_{n,a}(\gamma) (r_n^2 - r_a^2), \quad n \in \mathbb{N} \cap \mathbb{S}^c.
\]

\[
= \gamma_n(S, \gamma)
\]

(17)

We formulate another system of linear equations with two unknowns \( x = [d^2, b]^T \) as

\[
A(S, \gamma)x = b(S, \gamma),
\]

(12)

where

\[
A(S, \gamma) = \begin{bmatrix}
\alpha_1(S, \gamma) & \beta_1(S, \gamma) \\
\vdots & \vdots \\
\alpha_N(S, \gamma) & \beta_n(S, \gamma)
\end{bmatrix} \in \mathbb{R}^{(N-2) \times 2},
\]

\[
b(S, \gamma) = \begin{bmatrix}
\gamma_1(S, \gamma) \\
\vdots \\
\gamma_n(S, \gamma)
\end{bmatrix} \in \mathbb{R}^{(N-2) \times 1}.
\]

(18)

Given \( S \), E2 has \((N-2)\) equations with three unknowns of \( d, b, \) and \( \gamma \), providing the following feasible condition.

**Proposition 2** (Feasible Condition: Arbitrary Mobility). Unless \( f_{a_1, a_2}(\gamma) = 0 \), Problem E2 has a unique solution if the number of detected steps are at least 5, namely, \( N \geq 5 \).

**Proof:** See Appendix A. \( \square \)

**Remark 2** (Underdetermined System). Define \( \hat{\gamma} \) the angle satisfying \( f_{a_1, a_2}(\gamma) = 0 \). The rank of \( A(S, \hat{\gamma}) \) is 1, making E2 an underdetermined system that has infinite number of solutions. Using the condition \( f_{a_1, a_2}(\gamma) = 0 \), it is straightforward to identify whether the concerned \( \gamma \) is \( \hat{\gamma} \).
With $N \geq 5$ and $f_{a1,a2}(\gamma) \neq 0$, the solution for E2 has a similar form to the linear mobility counterpart as

$$
x(S, \gamma) = [(d(S, \gamma))^2, b(S, \gamma)]^T
= \left[ A(S, \gamma)^T A(S, \gamma) \right]^{-1} A(S, \gamma)^T b(S, \gamma),
$$

which is valid only when $\gamma$ is correctly picked. Otherwise, $x(S, \gamma)$ of (19) does not satisfy E2, i.e., $A(S, \gamma)x(S, \gamma) \neq b(S, \gamma)$. Prompted by the fact, a correct $\gamma^*(S)$ can be easily found by a simple 1D search over $[0, \pi]$ satisfying the following criteria:

$$
\gamma^*(S) = \{ \gamma | e_1(S, \gamma) = 0, \ \gamma \in [0, \pi) \},
$$

where

$$
e_1(S, \gamma) = \| A(S, \gamma)x(S, \gamma) - b(S, \gamma) \|.
$$

Remark 3 (Ambiguity of $\gamma$). Noting the period of $\tan^{-1}(x)$ being $\pi$, two possible solutions of $\gamma^* S \in [0, \pi)$ and $[\pi, 2\pi)$, say $\gamma_1^*(S)$ and $\gamma_2^*(S)$, where $\gamma_1^*(S) + \pi = \gamma_2^*(S)$. Despite the ambiguity, the resultant solutions of $d$ and $b$ are not changed regardless of $\gamma_1^*(S)$ or $\gamma_2^*(S)$, namely, $d(S, \gamma_1^*(S)) = d(S, \gamma_2^*(S))$ and $b(S, \gamma_1^*(S)) = b(S, \gamma_2^*(S))$. As a result, we focus on finding $\gamma_1^*(S)$ in the range of $[0, \pi)$, and it is considered as $\gamma^* S$.

When the measurement are corrupted, the noisy versions of the matrix $A(S, \gamma)$ and the vector $b(S, \gamma)$ are given, denoted by $\tilde{A}(S, \gamma)$ and $\tilde{b}(S, \gamma)$, making it difficult to use the discriminant of (20) directly. Instead, we develop the following two-stage approach.

1) Finding $x$: Given $\gamma$, we formulate the following minimization problem as

$$
x(S, \gamma) = \arg \min_x \| \tilde{A}(S, \gamma)x - \tilde{b}(S, \gamma) \|
= \left[ \tilde{A}(S, \gamma)^T \tilde{A}(S, \gamma) \right]^{-1} \tilde{A}(S, \gamma)^T \tilde{b}(S, \gamma),
$$

which follows an equivalent structure of (19).

2) Finding $\gamma$: We use a 1D search based on the following criteria:

$$
\gamma^*(S) = \arg \min_{\gamma \in \mathbb{P}(S)} \left[ w_1 \cdot e_1(S, \gamma) + w_2 \cdot e_2(S, \gamma) \right],
$$

where $\mathbb{P}(S)$ is a range of feasible $\gamma$ defined as all calibrated distances $\{r_n - b(S, \gamma)\}$ and estimated step length $d(S, \gamma)$ being positive, namely,

$$
\mathbb{P} = \left\{ \gamma \left| \min_{n \in \mathbb{N}} [r_n - b(S, \gamma)] > 0, \ \ d(S, \gamma) > 0, \ \ \gamma \in [0, \pi) \right\}.
$$

Two error functions are considered whose weighed factors $\{w_1, w_2\}$ satisfy $w_1 + w_2 = 1$. The first function $e_1(S, \gamma)$ is specified in (21). For the second one $e_2(S, \gamma)$, denote the matrix $R(S, \gamma) = \left[ R_{1,1}(S, \gamma), \cdots , R_{N,1}(S, \gamma); R_{1,2}(S, \gamma), \cdots , R_{N,2}(S, \gamma) \right] \in \mathbb{R}^{2 \times (N-1)}$, where $R_{n,a}(S, \gamma)$ is the estimated $R$ obtained by plugging $d(S, \gamma)$ and $b(S, \gamma)$ into the $n$-th equation of (16) as

$$
R_{n,a}(S, \gamma) = \frac{r^2_n - r^2_a - 2(r_n - r_a) \cdot b(S, \gamma) - (d(S, \gamma))^2 \eta_{n,a}}{2f_{n,a}(\gamma) \cdot d(S, \gamma)}, \ \ a \in S.
$$

Given $R(S, \gamma)$, the error function $e_2(S, \gamma)$ is defined as

$$
e_2(S, \gamma) = \text{std}(R(S, \gamma)),
$$

where $\text{std}(\cdot)$ represents the standard deviation of all elements therein.

Remark 4 (Error Functions). Given $S$, the first error function $e_1(S, \gamma)$ represents the error of E2 itself by focusing its explicit solution $x = [d^2, b]^T$. On the other hand, the second error function $e_2(S, \gamma)$ captures the error on a latent variable $R$ that is eliminated in E2 due to the linearization process of (16) and (17). Selecting the weight factors $\{w_1, w_2\}$ is discussed in Section V.

In both cases with and without measurement noises, a pair of the optimal solution $d^*(S)$ and $b^*(S)$ can be obtained from the solution $x^*(S) = x(S, \gamma^*(S))$. Given $\{d^*(S), b^*(S)\}$, we derive the coordinates of $z_1$, say $z_1^*(S) = [q_1^*(S), u_1^*(S)]^T$ as follows. First, linear equations with $q_1$ and $u_1$ are derived from (16) as

$$
q_1(c_n - c_a) + u_1(a_n - a_a) = \frac{r^2_n - r^2_a - 2(b^*(S) \cdot (r_n - r_a) - (d^*(S))^2 \eta_{n,a})}{2d^*(S)} \eta_{n,a}, \ a \in S, n \neq a,
$$

leading to formulating a system of linear equations as

$$
H(S)z_1 = g(S), \quad (E3)
$$

where $H(S) = \left[ H_{a1}; H_{a2} \right] \in \mathbb{R}^{2(N-1) \times 2}$ and $g(S) = \left[ g_{a1}(S); g_{a2}(S) \right] \in \mathbb{R}^{2(N-1) \times 1}$ with

$$
H_a = \begin{bmatrix}
c_1 - c_a & s_1 - s_a \\
\vdots & \vdots \\
c_{N-1} - c_a & s_{N-1} - s_a
\end{bmatrix} \in \mathbb{R}^{(N-1) \times 2},
$$

$$
g_a(S) = \begin{bmatrix}
g_{a1}(S) \\
\vdots \\
g_{aN}(S)
\end{bmatrix} \in \mathbb{R}^{(N-1) \times 1}, \ a \in S. \quad (28)
$$

Given the feasible condition of E2 stated in Proposition 2, E3 has a unique solution obtained by a single matrix inversion as

$$
z_1^*(S) = [q_1^*(S), u_1^*(S)]^T = \arg \min_{z_1} \| H(S)z_1 - g(S) \|
= [H(S)^T H(S)]^{-1} H(S)^T g(S). \quad (29)
$$

D. Using Multiple Combination of Reference Steps

This subsection deals with the remaining issue of selecting RSs $S$, helping mitigate the positioning error due to significant measurement noises. To this end, multiple combinations of RSs are utilized to achieve more accurate positioning than a single RS-based scheme, based on a common statistical belief that more observations make the estimate less deviated from a ground-truth. The detailed procedure is explained as follows.
1) Selecting Candidate RSs: First, several steps are picked as RS’s candidates, denote by $C$, based on the initial propagation distance estimates $\{r_n\}$ defined in (2). In general, smaller $r_n$ means that AP $n$ is located in proximity whose RSS is likely to be high. It is thus reasonable to consider the resultant ranging result is relatively accurate. Motivated by this intuition, the set $C$ contains a step’s index, say $n$, if $r_n$ is in the top $C$ smallest, namely,

$$C = \{n \in \mathbb{N} | r_n \leq r_k, n \in C, k \in C', |C| = C\},$$ (30)

where $C$ is the cardinality constraint of $C$, whose effect is verified by field experiments in Section V.

2) Estimating the Bias and Step Length: Two indices of $C$ are picked as $S$. It is possible to make up to $L = \binom{C}{2}$ combinations of $S$. Denote $S_\ell$ the $\ell$-th set of RSs, $\ell \in \{1, \cdots, L\}$. Given $S_\ell$, compute $d^*(S_\ell)$ and $b^*(S_\ell)$ by following the procedures in Sec. III-B and Sec. III-C, depending on the cases of linear and arbitrary mobilities, respectively. Given all individual estimates $\{d^*(S_\ell)\}$ and $\{b^*(S_\ell)\}$ representative estimates, denoted by $d^*$ and $b^*$ respectively, are computed using their medians, namely,

$$d^* = \text{median}(d^*(S_\ell)), \quad b^* = \text{median}(b^*(S_\ell)).$$ (31)

3) Estimating the relative coordinates of $z_1^*$: Given $d^*$ and $b^*$, compute $\{z_1^*(S_\ell)\}$ for all possible sets of RSs using (13) and (14) for the case of linear mobility, or (29) for the case of arbitrary mobility. Given all individual estimates $\{z_1^*(S_\ell)\}$, representative estimate, denoted by $z_1^*$ is computed using their medians, namely,

$$z_1^* = \text{median}(z_1^*(S_\ell)).$$ (32)

Remark 5 (Mean vs. Median). While a mean-based estimation has been widely used as a de facto standard approach, it is prone to a few outliers severely deviated from a ground-truth value. On the other hand, a median-based estimation can ignore these outliers. Thus, it is more suitable to design a positioning algorithm based on the median, which is simple yet robust from measurement noises, such as [27] and [28].

IV. POSITIONING VIA TRAJECTORY ALIGNMENT

In this section, we aim at positioning the user’s locations by aligning multiple trajectories based on the measurements of different APs, called trajectory alignment (TA). First, the user’s relative trajectory defined on the local coordinate system of each AP is derived based on the estimations in the preceding section. Next, a basic principle of TA is mathematically explained assuming the case without measurement noise. Last, a practical algorithm is designed able to work in the case with measurement noise.

A. Relative Trajectory Derivation

This section derives the sequence of the user’s locations, denoted by $Z = \{z_1^*\} = \{[q_n^*, u_n^*]\}$, corresponding to the user’s relative trajectory defined on the local coordinate system. From the preceding estimations of the initial location $z_1^* = [q_1^*, u_1^*]$ in (32), it is possible to derive the following locations using (3) and the step length estimation $d^*$. Depending on the case of linear or arbitrary mobilities, we have different results explained below.

1) Linear mobility: Recalling that there exist two candidates of $u_1^*$ [see (14)], two possible local trajectories are thus made, say $Z_+$ and $Z_-$, given as

$$Z_+ = \{[q_n^*, u_n^*] | q_n^* = q_1^* + (n-1)d, \quad u_n^* = +u_1^*, \quad \forall n \in \mathbb{N} \},$$
$$Z_- = \{[q_n^*, u_n^*] | q_n^* = q_1^* + (n-1)d, \quad u_n^* = -u_1^*, \quad \forall n \in \mathbb{N} \},$$ (33)

where $q_1^*$ and $u_1^*$ are specified in (13) and (14), respectively. Either $Z_+$ or $Z_-$ is the real trajectory $Z$, differentiated by the positioning algorithm introduced in the sequel.

2) Arbitrary mobility: Contrary to the linear mobility counterpart, no ambiguity of the initial location exists. The resultant local trajectory $Z$ is given as

$$Z = \{[q_n^*, u_n^*] | q_n^* = q_1^* + dc_n, \quad u_n^* = u_1^* + ds_n, \quad \forall n \in \mathbb{N} \},$$ (34)

where the coefficient $c_n$ and $s_n$ are specified in (6).

B. Trajectory Alignment

This subsection introduces the concept of TA and explains its feasible conditions. Consider the user’s relative trajectory estimated by the RTT measurements from AP $m$, say $Z^{(m)} = \{z_n^{(m)*}\}$. It is converted into a global coordinate system when the initial direction $\omega$ is given, namely,

$$p_n^{(m)}(\omega) = p_{\omega n}^{(m)} + \begin{bmatrix} \cos(\omega) & -\sin(\omega) \\ \sin(\omega) & \cos(\omega) \end{bmatrix} z_n^{(m)*},$$ (35)
where \( p^{(m)}_{AP} \) is AP \( m \)'s location assumed to be known in advance. Noting that the user’s location is unique regardless of which AP is used for positioning, the following condition should be met if \( \omega \) is correctly selected, denoted by \( \omega^* \):

\[
p_n = p_n^{(1)}(\omega^*) = p_n^{(2)}(\omega^*) = \cdots = p_n^{(M)}(\omega^*), \quad \forall n \in \mathbb{N},
\]

which is said that all trajectories are perfectly aligned. The aligned trajectory after TA is equivalent to the trajectory defined on the global coordinate system, denoted by \( \mathcal{P} = \{p_n\} \), if it exists uniquely. The following proposition gives different feasible conditions of TA for linear and arbitrary mobilities.

**Proposition 3** (Feasible Condition of Trajectory Alignment). There exists a unique \( \omega^* \) satisfying the condition of (36) if the number of APs \( M \) not on a straight line is at least 3 for linear mobility or the number of APs \( M \) is at least 2 for arbitrary mobility.

**Proof**: See Appendix B. \( \square \)

Fig. 4 and 5 respectively represent the graphical examples of TA for linear and arbitrary mobilities, showing that one more AP is required to identify whether \( Z^{(m)} \) is \( Z^{+} \) or \( Z^{-} \). The comparison between the two from the perspective of entire procedure is discussed in the following remarks.

**Remark 6** (Linear vs. Arbitrary Mobilities). Recall the feasible conditions of the number of detected steps, say \( N \), is 4 or 5 depending on linear or arbitrary mobilities, respectively (See Propositions 1 and 2). Considering the number \( M \)'s requirement in Proposition 3, the sum of \( N \) and \( M \) should be at least 7, regarded as the overall requirement for both cases’ unique positioning.

### C. Algorithm Design

In an ideal case without a measurement noise, it is possible to find \( \omega^* \) making all trajectories aligned perfectly, equivalent to satisfying the condition (36). In contrast, it may be challenging to do in practical cases with measurement noises, since several APs rather deteriorates the positioning accuracy if their estimation errors of bias and step length are severe. It is overcome by excluding these APs in advance and minimizing a new error function defined for TA. The detailed algorithm is explained below.

1) **Feasible AP Selection**: First, we aim at excluding APs unlikely to contribute to accurate positioning. To this end, we define the set of feasible APs \( \mathcal{F} \), whose element’s bias and step length estimations, say \( b^{(m)}\ast \) and \( d^{(m)}\ast \), satisfy the following condition:

\[
\mathcal{F} = \left\{ m \biggr| \min_{n \in \mathbb{N}} \left[ \frac{b^{(m)} - b^{(m)}\ast}{d^{(m)}\ast} \right] > 0, \quad d^{(m)}\ast > 0, \quad z^{(m)}_{1}\ast \in \mathbb{R}, \quad m \in \mathbb{M} \right\},
\]

where the first and second conditions mean that the distance estimation after deducting the bias and the step length estimation should be positive, and the third condition means that the coordinates of estimated position are real numbers. The APs not in \( \mathcal{F} \) are excluded and the relative trajectories \( \{Z^{(m)}\ast\}_{m \in \mathcal{F}} \) are used for the next step.

2) **The Optimal Heading Direction Derivation**: Depending on the mobility pattern being arbitrary or linear, different methods are used to find the optimal heading direction as follows.

a) **Arbitrary mobility**: Given \( \mathcal{F} \), we aim at aligning all trajectories as closely as possible. Specifically, we define an error function \( e_3(\omega) \) as the sum of the Euclidean distance between two relative trajectories in \( \mathcal{F} \), given as

\[
e_3(\omega) = \sum_{i,j \in \mathcal{F}} \sum_{n=1}^{N} \| p^{(i)}_n(\omega) - p^{(j)}_n(\omega) \|.
\]

By a 1D search, it is possible to find \( \omega^* \) to minimize the error function \( e_3(\omega) \), namely,

\[
\omega^* = \arg \min_{\omega \in [0, 2\pi]} e_3(\omega).
\]

b) **Linear mobility**: Recall that there exists the ambiguity of relative trajectories in the case of linear mobility, say \( \{Z^{(m)}_{\ast}, Z^{(m)}_{\ast}\}_{m \in \mathcal{F}} \) specified in (33). To remove this ambiguity, we utilize the relation between relative trajectories of different APs, as stated in the following proposition.

**Proposition 4** (The Relation Between Relative Trajectories). The distance between relative positions of two APs is always equal to the distance between the two APs, namely,

\[
\| z^{(m_1)}_{\ast} - z^{(m_2)}_{\ast} \| = \| p^{(m_1)}_{AP} - p^{(m_2)}_{AP} \|, \quad \forall n \in \mathbb{N}, \quad \forall m_1, m_2 \in \mathbb{M}.
\]

**Proof**: See Appendix C. \( \square \)

Select one reference AP whose index is denoted by \( r \). Depending on \( Z_{\ast}^{(r)} \) or \( Z_{\ast}^{(r)} \), there exist two possible sequences.
of relative trajectories, denoted by $\mathcal{Y}_{+}^{(r)}$ and $\mathcal{Y}_{-}^{(r)}$, initialized as $\{Z_{+}^{(r)}\}$ and $\{Z_{-}^{(r)}\}$, respectively. For example, given $Z_{+}^{(r)}$, either C1 or C2 holds, given as

$$
\psi_1 = ||p_{n}(r) - p_{AP}^{(m)}|| - ||z_{n}^{(r)} - z_{n}^{(m)}|| = 0,
$$
$$
\forall n \in N, \quad z_{n}^{(r)} \in Z_{+}^{(r)}, \quad z_{n}^{(m)} \in Z_{+}^{(m)}, \quad (C1)
$$

$$
\psi_2 = ||p_{n}(r) - p_{AP}^{(m)}|| - ||z_{n}^{(r)} - z_{n}^{(m)}|| = 0,
$$
$$
\forall n \in N, \quad z_{n}^{(r)} \in Z_{+}^{(r)}, \quad z_{n}^{(m)} \in Z_{+}^{(m)}, \quad (C2)
$$

When C1 holds (i.e., $\psi_1 = 0$), $Z_{+}^{(m)}$ and $Z_{-}^{(m)}$ are added in $\mathcal{Y}_{+}^{(r)}$ and $\mathcal{Y}_{-}^{(r)}$, respectively. When C2 holds (i.e., $\psi_2 = 0$), reversely, $Z_{+}^{(m)}$ and $Z_{-}^{(m)}$ are added in $\mathcal{Y}_{+}^{(r)}$ and $\mathcal{Y}_{-}^{(r)}$, respectively. The addition process is continued until $|\mathcal{Y}_{+}^{(r)}| = |\mathcal{Y}_{-}^{(r)}| = F$. In the presence of measurement noises, neither C1 nor C2 can be satisfied. Instead, we relax C1 and C2 as $\psi_1 > \psi_2$ and $\psi_1 \leq \psi_2$, respectively. Given $\mathcal{Y}_{+}^{(r)}$ and $\mathcal{Y}_{-}^{(r)}$, relative trajectories are rotated using (35), denoted by $p_{n}^{(m)}(w; \mathcal{Y}_{+}^{(r)})$ and $p_{n}^{(m)}(w; \mathcal{Y}_{-}^{(r)})$, respectively. We define $e_{3}^{(r)}(\omega) = \min \left\{ e_{3}^{(r)}(\omega; \mathcal{Y}_{+}^{(r)}), e_{3}^{(r)}(\omega; \mathcal{Y}_{-}^{(r)}) \right\}$, where

$$
e_{3}^{(r)}(\omega; \mathcal{Y}_{+}^{(r)}) = \sum_{i,j \in F} \sum_{n=1}^{N} ||p_{n}^{(i)}(w; \mathcal{Y}_{+}^{(r)}) - p_{n}^{(j)}(w; \mathcal{Y}_{-}^{(r)})||,
$$

$\mathcal{Y}_{+}^{(r)} \in \{ \mathcal{Y}_{+}^{(r)}, \mathcal{Y}_{-}^{(r)} \}$. By a 1D search, it is possible to find the reference AP $r^*$ and the corresponding $\omega^*$ to minimize the error function as

$$
\{ r^*, \omega^* \} = \arg \min_{r \in F, \omega \in [0, 2\pi]} \left\{ e_{3}^{(r)}(\omega) \right\}.
$$

3) Determining the estimated trajectory: Last, the estimated trajectory, say $P = \{p_{n}\}$, is derived by averaging $\{p_{n}^{(m)}(\omega^*)\}_{m=1}^{F}$ as

$$
p_{n}^* = \frac{1}{|F|} \sum_{m \in F} p_{n}^{(m)}(\omega^*), \quad \forall n \in N.
$$

V. FIELD EXPERIMENTS

This section aims at verifying the proposed positioning algorithms using several field experiments at two different indoor sites, each of which has different environments from the positioning perspective, as shown in Fig. 6. The first experimental site, called site A, is a plaza located inside the Engineering building at Yonsei University, Seoul, Korea. Site A is an open space like outdoor environments where several LOS paths exist. On the other hand, the second experimental site, called site B, is a parking lot located under Building 11 in Korea Railroad Research Institute, Uiwang, Korea. Compared with site A, most signal propagations are followed by NLOS paths due to the presence of many obstacles such as parked vehicles, walls, and pillars. The detailed experiment setups are summarized in Table I, unless specified.

We use the algorithm in [15] to obtain the heading direction changes $\{\theta_{n}\}$ (see Fig. 7 as an example). As shown in Fig. 7, we consider $\theta_{n} \in \{0, \frac{\pi}{2}, \pi, \frac{3\pi}{2}\}$, assuming that the user’s moving direction only has finite choices depending on the surrounding arrangement (i.e., road, wall and et al.). Its effect is discussed in the sequel.

Two benchmarks are considered. The first one is based on raw RTT measurements without bias compensation for a conventional multilateration method, such as \textit{linear least square-reference selection} (LLS-RS) [29]. For the second one, compensated RTT measurements are utilized for a multilateration method but TA is not applied. We use the Euclidean distance of estimated positioning to ground-truth locations to represent a positioning error, i.e., $||p_{n}^* - p_{n}||$. 

![Floor plans of field experiments](image1)

![Sample trajectory](image2)

![Quantization of heading change](image3)
A. Positioning Accuracy

We verify the performance of the algorithm for the cases of linear and arbitrary mobilities. The key performance metrics are summarized in Table II.

1) Linear Mobility: First, we consider the cases of linear mobility. Figure 8 illustrates a graphical example of the estimated trajectories of the proposed one and two benchmarks, while Figure 9 shows cumulative distributional functions (CDFs) of the resultant positioning errors. Several key observations are made. First, the gain of the RTT bias compensation in Sec. III is verified by comparing two benchmarks, showing significant performance improvements for both Sites A and B. Second, TA explained in Sec. IV makes all estimated points tailored to the trajectories detected by PDR, leading to additional performance enhancement. As a result, the resultant positioning errors of approximately 90% are located within 2 (m) and 4 (m), and the average errors are 1.359 (m) and 1.915 (m) for Sites A and B respectively. The other performance metrics are summarized in Table II.

2) Arbitrary Mobility: Second, we consider the cases of arbitrary mobility. we illustrate a graphical example of the estimated trajectories in Figure 10 and cumulative distributional functions (CDFs) of the resultant positioning errors in Figure 11, showing similar tendencies to the linear mobility counterpart. Besides, it is shown that the algorithm for arbitrary mobility provides a more accurate positioning result than that for linear mobility such that the resultant positioning errors of 90% are approximately located within 0.5 (m) and 2.5 (m), and the average errors are 0.369 (m) and 1.705 (m) for Sites A and B, respectively. Compared with the linear mobility cases, the arbitrary mobility cases provide one more dimension of geometric information. It is fully utilized by the proposed algorithm including RTT bias compensation and TA, leading to significant performance enhancement.

B. Effect of Weighed Factors $w_1$ and $w_2$

Recall a pair of $w_1$ and $w_2$ weighting the error functions $e_1$ of (21) and $e_2$ of (26), respectively. As stated in Remark 4,
the error function $e_1$ focuses on minimizing the error on step length $d$ and the RTT bias $b$, while $e_2$ aims at minimizing the error on the initial position $z_1^*$. Fig. 12 represents the effect of weight factors on the average positioning error, showing that $\{w_1, w_2\} = \{0, 1\}$ provides the most accurate positioning result. It is because the errors on $d$ and $b$ can be compensated by the method of multiple combination of RSs introduced in Sec. III-D. Therefore, if a sufficient number of RSs is selected, it is optimal to focus on the minimization of $e_2$.

C. Effect of Multiple Combination of RSs

Fig. 13 shows the average positioning error as a function of the number of candidate RSs $C$, showing that the error is reduced from 4.702 (m) to 1.705 (m), when the number of RSs $C$ increases from 2 to 18. On the other hand, $C$ larger than 18 rather deteriorates a positioning result since a larger portion of RSs is likely to be severely corrupted by the measurement error. Through various simulation studies, the positioning error can be minimized in average sense by selecting the number of RSs $C$ as $0.25N$, where $N$ is the number of walking steps (e.g., $\frac{C}{N} = \frac{18}{70} \approx 0.257$ in the experiment setting of Fig. 13).

D. Effect of Heading Direction Error

Recall that heading direction changes $\{\theta_n\}$ are quantized into four levels as $\theta_n \in \{0, \frac{\pi}{2}, \pi, \frac{3\pi}{2}\}$, by exploiting the prior information of surrounding arrangement. It enables us to obtain the precise trajectory estimation, as shown in Fig. 8 and 10. To investigate its effect on positioning accuracy, Fig. 14 represents the estimated trajectory without the quantization, showing that the degradation of the positioning accuracy is marginal such that the average positioning error increases from 0.369 (m) to 0.494 (m) at Site A, and 1.705 (m) to 1.958 (m) at Site B.

VI. CONCLUDING REMARK

This work has presented a novel positioning algorithm to estimate a user’s location by integrating RTT and PDR mea-
measurements. Geometric relations between the two are formulated as mathematical form, enabling us to design a tractable and scalable positioning algorithm as well as provide the feasible conditions of the number of steps and WiFi APs for a unique positioning. The superiority of the proposed method has been well verified by field experiments that the positioning accuracy can be significantly improved than the conventional multilateration techniques.

The current work can be extended in several directions. First, it can be applied to improve positioning accuracy by utilizing additional information such as a floor plan or to guess an unknown floor plan by fusion with the simultaneous localization and mapping (SLAM) algorithm. Second, our approach could be applied to correcting the positioning error in the autonomous driving of vehicles and drones, which is an area where many attempts are being made. We believe that the technique of carefully fusing multiple positioning methods will yield better results and will be further developed.

APPENDIX

A. Proof of Proposition 2

Noting that E2 has \((N - 2)\) equations, it is an overdetermined system when \(N \geq 5\) if the rank of the matrix \(A(S, \gamma)\) in (18) is 3. Then, there always exists \(\gamma^*\) satisfying \(e_1(S, \gamma^*) = 0\) since E2 is formulated from a geometric representation of multiple unknowns.

Consider a special case of \(\gamma = \hat{\gamma}\) where \(f_{a_1,a_2}(\hat{\gamma}) = 0\). At this case, \(f_{n,a_2}(\hat{\gamma}) = f_{n,a_1}(\hat{\gamma})\) because if \(a_1 < a_2\), \(f_{n,a_1}(\hat{\gamma}) = \sum_{i=1}^{2} \cos(\theta_i - \hat{\gamma}) - \left( \sum_{i=1}^{n} \cos(\theta_i - \hat{\gamma}) + \sum_{2 \leq a_2} \cos(\theta_i - \hat{\gamma}) \right)\) and \(f_{a_1,a_2}(\hat{\gamma}) = \sum_{2 \leq a_2} \cos(\theta_i - \hat{\gamma})\) by definition. From this, the matrix components \(\alpha_n(S, \hat{\gamma})\) and \(\beta_n(S, \hat{\gamma})\) can be changed as

\[
\alpha_n(S, \hat{\gamma}) = f_{n,a_2}(\hat{\gamma}) \eta_{n,a_1} - f_{n,a_1}(\hat{\gamma}) \eta_{n,a_2} = f_{n,a_1}(\hat{\gamma}) (\eta_{n,a_1} - \eta_{n,a_2}) = f_{n,a_1}(\hat{\gamma}) \eta_{a_2,a_1},
\]

\[
\beta_n(S, \hat{\gamma}) = 2 \left( f_{n,a_2}(\hat{\gamma})(r_n - r_{a_1}) - f_{n,a_1}(\hat{\gamma})(r_n - r_{a_2}) \right) = 2 \left( f_{n,a_1}(\hat{\gamma})(r_{a_2} - r_{a_1}) \right).
\]
Above (39) and (40) have common terms \( f_{n,a_1}(\gamma) \), the matrix \( A(S, \gamma) \) which defined as (18) becomes

\[
A(S, \gamma) = \left[ a_{x_2, a_1}, 2(a_{x_2} - a_{x_1}) \right]. \tag{41}
\]

Therefore, the rank of \( A(S, \gamma) \) is 1, it is underdetermined system. This finishes the proof.

**B. Proof of Proposition 3**

We provide separated proofs for arbitrary and linear mobilities.

1) **Arbitrary mobility**: Consider the case with APs 1 and 2. Using (35), a pair of relative locations at \( n = 1 \), say \( z^{(1)*} \) and \( z^{(2)*} \), are given as

\[
z^{(1)*} = \Theta(-\omega)(p^{(1)}(\omega) - p^{(2)}(\omega)), \quad z^{(2)*} = \Theta(-\omega)(p^{(2)}(\omega) - p^{(1)}(\omega)),
\]

where \( \Theta(\omega) = \begin{bmatrix} \cos(\omega) & -\sin(\omega) \\ \sin(\omega) & \cos(\omega) \end{bmatrix} \) is a rotation matrix. Their relation is derived by subtracting the above two as

\[
z^{(1)*} - z^{(2)*} = \Theta(-\omega) \left( \begin{bmatrix} p^{(1)}(\omega) - p^{(2)}(\omega) \\ p^{(2)}(\omega) - p^{(1)}(\omega) \end{bmatrix} \right).
\]

Noting that \( p^{(1)}(\omega) = p^{(2)}(\omega) \) if \( \omega = \omega^* \), the above is reduced as

\[
z^{(1)*} - z^{(2)*} = \Theta(-\omega^*)(p^{(2)} - p^{(1)}). \tag{44}
\]

where \( \omega^* \) is unique since the rotation matrix is not ambiguous between [0, 2\( \pi \)]. Next, the relation between two relative locations at \( n = 2 \), say \( z^{(1)*}_2 \) and \( z^{(2)*}_2 \), is given as

\[
z^{(1)*}_2 - z^{(2)*}_2 = \Theta(\omega) \left( \begin{bmatrix} z^{(1)*}_2 + d \cos(\theta_1), \sin(\theta_1) \end{bmatrix}^T \right) - \left( \begin{bmatrix} z^{(2)*}_2 + d \cos(\theta_1), \sin(\theta_1) \end{bmatrix}^T \right)
\]

\[
= \Theta(\omega^*) (z^{(1)*}_2 - z^{(2)*}_2). \tag{45}
\]

The above is straightforwardly extended into other relative locations at \( n \in \mathbb{N} \). In other words, two relative trajectories \( Z^{(1)} = \{ z^{(1)*}_n \} \) and \( Z^{(2)} = \{ z^{(2)*}_n \} \), are perfectly aligned when \( \omega = \omega^* \). We complete the proof.

2) **Linear Mobility**: Consider the case with APs 1, 2, and 3. Recall that there are two relative trajectories for each AP, say \( \{ Z^{(m)}_n \}_{m=1}^3 \). Following the same step in the arbitrary mobility counterpart, two possible global coordinates are made for APs 1 and 2, denoted by \( Z^{(1)}_n \) and \( Z^{(2)}_n \), which are symmetric concerning the line between the locations of APs 1 and 2 (See Fig. 15). Specifically, the corresponding heading directions, denoted by \( \omega^{(1)*}_2 \) and \( \omega^{(2)*}_2 \), has the following geometric relation as

\[
\omega^{(1)*}_2 + \omega^{(2)*}_2 = 2 \angle (p^{(1)}_{AP} - p^{(2)}_{AP}), \tag{46}
\]

where \( \angle(x) \) returns the angle of the vector \( x \). Between the two, one is real whereas the other is fake. Similarly, the resultant heading directions for APs 1 and 3, denoted by \( \omega^{(1)*}_3 \) and \( \omega^{(3)*}_3 \), gives

\[
\omega^{(1)*}_3 + \omega^{(3)*}_3 = 2 \angle (p^{(1)}_{AP} - p^{(3)}_{AP}). \tag{47}
\]

Using the above two equations, it is easy to identify which ones are real. For example, assume \( \omega^{(1)*}_2 \) and \( \omega^{(3)*}_3 \) are real, namely, \( \omega^* = \omega^{(1)*}_2 = \omega^{(3)*}_3 \). Unless \( \angle (p^{(1)}_{AP} - p^{(2)}_{AP}) = \angle (p^{(1)}_{AP} - p^{(3)}_{AP}) \), it is obvious to identify \( \omega^{(1)*}_2 \) and \( \omega^{(3)*}_3 \)
are fake since they are different. Note that if all APs exist on a straight line, it cannot be distinguished between real and fake ones because symmetry is maintained, completing the proof.

C. Proof of Proposition 4

It can easily be verify by the fact \( z^{(1)}_2 - z^{(2)}_2 = \Theta(-\omega^2)(p_{AP}^{(2)} - p_{AP}^{(1)}) \) as in (45). By putting norm on both sides,

\[
\|z^{(1)}_2 - z^{(2)}_2\|_2 = \|p_{AP}^{(2)} - p_{AP}^{(1)}\|_2.
\]

Also, \( z^{(1)}_2 - z^{(2)}_2 = z^{(1)}_1 - z^{(2)}_1 \) as proved in (46). The above manners are extended into other APs at \( m_1, m_2 \in \mathbb{N} \) and other relative positions at \( n \in \mathbb{N} \). This finishes the proof.

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