On the LQ Based Stabilization for a Class of Switched Dynamic Systems

M.A. Ortiz Castillo∗ M. Bonilla∗ J.J. Loiseau∗∗ M. Malabre∗∗∗ V. Azhmyakov∗∗∗

∗ CINVESTAV-IPN, Control Automático, UMI 3175 CINVESTAV-CNRS, A.P. 14-740.México 07000. mortiz@cinvestav.mx, mbonilla@cinvestav.mx

∗∗ CNRS, LS2N, UMR 6004, Nantes, France. Jean-Jacques.Loiseau@ls2n.fr, Michel.Malabre@ls2n.fr

∗∗∗ Department of Mathematical Sciences, Universidad EAFIT, Medellín, Republic of Colombia. vazhmyako@eafit.edu.co

Abstract: This paper deals with the stabilization of a class of time-dependent linear autonomous systems with a switched structure. For this aim, the switched dynamic system is modeled by means of an implicit representation combined with a Linear-Quadratic (LQ) type control design. The proposed control design stabilizes the resulting system for all of the possible realizations of its locations. In order to solve the Algebraic Riccati Equation (ARE) associated with the LQ control strategy one only needs the knowledge of the algebraic structure related to the switched system. We finally prove that the proposed optimal LQ type state feedback stabilizes the closed-loop switched system no matter which location is active. The proposed theoretical approaches are illustrated by a numerical example.

Keywords: switched dynamic systems, implicit control systems, linear quadratic regulator (LQR), Riccati equation, Lyapunov stability.

1. INTRODUCTION

In (Bonilla et al., 2015a) is shown that a wide class of time-dependent autonomous systems with a switched structure (Liberzon, 2003) can be adequately modeled by the state space representation:

\[ \dot{x} = A_q x + B u, \quad y = C_q x. \]  

(1.1)

where \( B \in \mathbb{R}^{n \times m} \) is an injective matrix and the \( A_q \) and \( C_q \) have the following structure (see e.g., (Narendra et al., 1994)):

\[ A_q = \overline{A}_0 + \overline{A}_1 \overline{D}(q) \quad \text{and} \quad C_q = \overline{C}_0 + \overline{C}_1 \overline{D}(q), \]  

(1.2)

The system remains in a specific location,

\[ q \in Q = \{ q_1, ..., q_\eta \} \quad \text{s.t.} \quad q_i \in \mathbb{R}^+, \quad i \in \{1, ..., \eta\}, \]  

(1.3)

for all time instants \( t \in [T_{i-1}, T_i) \), where \( T_i \in \mathbb{R}^+, T_0 = 0, T_{i-1} < T_i \), for all \( i \in \mathbb{N} \), \( \lim_{i \to \infty} T_i = \infty \), and \( s : \{ T_{i-1}, T_i \} \subset \mathbb{N} \rightarrow \mathbb{R}^+, \quad i \in \mathbb{N} \} \rightarrow Q, \quad s(T_{i-1}, T_i) = q \). Moreover, \( \overline{A}_0 \in \mathbb{R}^{\pi \times \pi} \) and \( \overline{A}_1 \in \mathbb{R}^{\pi \times \overline{n}} \) is an injective matrix and \( \overline{C}_0 \in \mathbb{R}^{\overline{n} \times \pi} \) and \( \overline{C}_1 \in \mathbb{R}^{\overline{n} \times \overline{n}} \) are surjective matrices with the property \( \overline{D}(0) = 0 \). In (Bonilla et al., 2015b) authors additionally propose a specific variable structure decoupling control strategy based on the ideal proportional and derivative (PD) feedback control law. Finally a proper practical approximation of the above ideal PD feedback is developed. Such feedback control strategies reject the initially given “variable structure” and make it possible to establish the required stability property of both control strategies.

In this paper we consider the stabilizing problem for a class of time-dependent switched dynamic systems equipped with a relative simple static state feedback. The paper is organized as follows: in Section 2 we formally introduce a class of switched systems and represent the given dynamics using the global implicit representation technique from (Bonilla et al., 2019). Section 3 contains a proper self-closed solution procedure for the main LQR design problem under specific assumption of unknown location. In Section 4 we show that the developed LQ type optimal feedback control additionally stabilizes the switched dynamic system. In Section 5 we discuss an illustrative numerical example and Section 6 summarizes our paper.

2. NON-STATIONARY SWITCHED SYSTEM WITH TIME-DRIVEN SWITCHING STRUCTURE

Let us consider the global implicit representation:

\[ \dot{x} = Ax + Bu, \quad y = Cx. \]  

(2.1)

Here

\[ E = \begin{bmatrix} E_1 & 0 \end{bmatrix}, \quad A = \begin{bmatrix} \overline{A}_0 & -\overline{A}_1 \end{bmatrix}, \quad B = \begin{bmatrix} B_1 \end{bmatrix}, \]  

(2.2)

and

\[ E = \begin{bmatrix} I \end{bmatrix}, \quad A = \begin{bmatrix} \overline{A}_0 & -\overline{A}_1 \end{bmatrix}, \quad C = \begin{bmatrix} \overline{C}_0 & -\overline{C}_1 \end{bmatrix}, \]  

(2.3)

We now assume that the locations set has a specific structure described by the following Hypothesis.

\[ H_1. \quad \text{(Bonilla et al., 2015b)} \] Given \( q_0, q_1, ..., q_\ell \in Q \), \[ g = [g_1, ..., g_\ell]^T, \quad g_1, ..., g_\ell \in \mathbb{R}^+, \] the locations \( q_i \in Q \) belong to the convex set described as follows...
\[ \mathcal{Q}_{0}(g) = \left\{ q_i \in Q \mid q_i = q_0 + \sum_{j=1}^{\ell} \gamma(i,j)g_j \xi_j, \ i \in \{1, \ldots, \ell\} \right\}, \]  
and moreover, for each \([T_{i-1}, T_i) \in \sigma^{-1}(q), \gamma(i,j)\) takes constant values in the closed subset of \(\mathbb{R} : [0,1]\).  

H2. (Bonilla et al., 2015b) There exist \(\Delta_0, \Delta_1, \ldots, \Delta_{\ell} \) such that

\[ \mathcal{D}(q_i) = \Delta_0 - \sum_{j=1}^{\ell} \gamma(i,j)g_j \Delta_j, \]

where \(\gamma(i,j)\) and \(g_j\) are determined by H1.

3. THE MAIN LQ-TYPE OPTIMAL CONTROL PROBLEM

Let us now formulate the following natural problem associated with a dynamics.

**Problem 1.** Given a switched system represented by (1.1) and (1.2) (viz. (2.1)-(2.3)) with a nonzero initial condition \(\pi_0\), and one unknown \(q \in Q\), determine a control design such that the descriptor variable \(\pi\) tends to zero and moreover, the following objective

\[ J = \int_{0}^{\infty} (\pi^TQ(\pi)\pi + u^Tr_0u)dt \]

attains its minimal value, where \(Q(q) = Q(q)^T \geq 0\) and \(R = R^T > 0\).

For the concrete treatment of the above Problem 1 we next follow Kailath (1980). Let us also refer to Lewis et al (2012) for more technical details. We now define the conventionally augmented objective associated with the originally given costs functional:

\[ J_{\lambda}(x, u, \lambda) = \int_{0}^{\infty} (H(x, u, \lambda, t) - \lambda^T \mathcal{E} \dot{x}) dt, \]

where the system’s Hamiltonian \(H\) is defined as follows

\[ H(x, u, \lambda, t) := L(x, u, t) + \lambda^T (\mathcal{A}x + \mathcal{B}u), \]

\[ L(x, u, t) := \frac{1}{2} (x^TQ(x) + u^TRu). \]

The above formalism implies the associated Euler-Lagrange equation (see Azhmyakov (2019) for details)

\[ \frac{\partial J}{\partial \dot{v}} - \frac{d}{dt} \left( \frac{\partial J}{\partial v(\dot{v})} \right) = 0, \]

where: \(f = H - \lambda^T \mathcal{E} \dot{x}\), and \(v \in \{ x, u, \lambda \} \) have the following constructive definition (Luenberger, 1969):

\[ v = x : \quad Q(q)x + \mathcal{A}^T \lambda = -\mathcal{E}^T \lambda, \]

\[ v = \lambda : \quad \mathcal{E} \dot{x} = \mathcal{A}x + \mathcal{B}u, \]

\[ v = u : \quad u = -R^{-1}B^T \lambda. \]

From (3.5)-(3.7) we next deduce the Hamiltonian equation

\[ \begin{bmatrix} \mathcal{E} \dot{x} \\ \mathcal{E}^T \lambda \end{bmatrix} = \begin{bmatrix} \mathcal{A} & -\mathcal{B}R^{-1}B^T \\ -Q(q) & -\mathcal{A}^T \end{bmatrix} \begin{bmatrix} x \\ \lambda \end{bmatrix}. \]

and from (2.2)-(3.8) we additionally obtain the following useful expressions

\[ \begin{bmatrix} \mathcal{A}^T P + PA_q - PBR^{-1}B^T P + \mathcal{Q}_q \end{bmatrix} = 0, \]

Taking into consideration the above relation (2.3), we have the final expressions:

\[ \begin{bmatrix} \dot{x} \\ \dot{\lambda} \end{bmatrix} = \begin{bmatrix} \mathcal{A}_0 - \mathcal{A}_1 - \mathcal{B}^T \mathcal{R}^{-1} \mathcal{B} \\ -\mathcal{Q} \end{bmatrix} \begin{bmatrix} x \\ \lambda \end{bmatrix}. \]

Let us now define \( q = \begin{bmatrix} x_c \\ x_r \end{bmatrix} \) and \( \lambda = \begin{bmatrix} \lambda_c \\ \lambda_r \end{bmatrix} \). Additionally let us denote

\[ Q(q) = \begin{bmatrix} \mathcal{Q}_0 & 0 \\ 0 & 0 \end{bmatrix}, \quad \mathcal{Q}_q = \mathcal{Q}_q^T \geq 0. \]

Taking into account the above formalism, we obtain the formal consequences:

\[ \begin{bmatrix} I & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & I & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \mathcal{A}_0 & -\mathcal{A}_1 & -\mathcal{B}^T \mathcal{R}^{-1} \mathcal{B} & 0 \\ -\mathcal{Q}_0 & 0 & 0 & 0 \\ 0 & -\mathcal{A}_1^T & -\mathcal{D}^T(q) & 0 \\ 0 & 0 & -\mathcal{A}_1^T & -I \end{bmatrix} = \begin{bmatrix} x_c \\ x_r \\ \lambda_c \\ \lambda_r \end{bmatrix}. \]

We next formally describe the constructive “separation idea” and split (3.10) into two subsystems. The obtained “dynamic part” has the following form

\[ \begin{bmatrix} \dot{x}_c \\ \dot{\lambda}_c \end{bmatrix} = \begin{bmatrix} \mathcal{A}_0 & -\mathcal{A}_1 & -\mathcal{B}^T \mathcal{R}^{-1} \mathcal{B} & 0 \\ -\mathcal{Q}_0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_c \\ \lambda_c \end{bmatrix} + \begin{bmatrix} \mathcal{A}_1^T \mathcal{Q}_q & 0 \\ 0 & \mathcal{D}^T(q) \end{bmatrix} \begin{bmatrix} x_r \\ \lambda_r \end{bmatrix}. \]

The resulting algebraic part of the proposed “separation” has the corresponding representation:

\[ \begin{bmatrix} x_r \\ \lambda_r \end{bmatrix} = \begin{bmatrix} -\mathcal{Q}_q \end{bmatrix} \begin{bmatrix} x_c \\ \lambda_c \end{bmatrix}. \]

Now from (3.12) and (3.11) we can easily deduce the next relations

\[ \begin{bmatrix} \dot{x}_c \\ \dot{\lambda}_c \end{bmatrix} = \begin{bmatrix} \mathcal{A}_0 & -\mathcal{B}^T \mathcal{R}^{-1} \mathcal{B} \\ -\mathcal{Q}_0 & -\mathcal{A}_1^T \mathcal{Q}_q \end{bmatrix} \begin{bmatrix} x_c \\ \lambda_c \end{bmatrix} + \begin{bmatrix} \mathcal{A}_1^T \mathcal{Q}_q & 0 \\ 0 & \mathcal{D}^T(q) \end{bmatrix} \begin{bmatrix} x_r \\ \lambda_r \end{bmatrix}, \]

which are equivalent to the condition

\[ \begin{bmatrix} \dot{x}_c \\ \dot{\lambda}_c \end{bmatrix} = \begin{bmatrix} \mathcal{A}_0 & \mathcal{A}_1^T \mathcal{D}(q) \]
where \( q \) is one unknown element of the given locations set \( \{q_1, \ldots, q_\ell \} \).

Finally let us recall the common knowledge of the feedback-type optimal control design for the generic LQ problem (see for example Kailath (1980) and Lewis et al (2012))

\[
F_* = R^{-1} BT P, \quad u = -F_0 x. \tag{3.17}
\]

We next use the well-known facts from this section as a theoretical basis for the stabilization approach we propose.

3.1 On the ARE Involved Solution Approach

Which location \( q \) is currently active being unknown implies the conceptual difficulties in solving the ARE (3.16). Note that the solvability property of the “switched” ARE under consideration is assumed for every admissible location. For the concrete treatment of (3.16) we next assume that \( Q_q \) has the same structure as in (1.1).

That means

\[
Q_q = Q_0 + \sum_{j=1}^\ell g_j \gamma(i, j) Q_j, \tag{3.18}
\]

where \( Q_0 = Q_0^T > 0 \) and (see (2.5));

\[
Q_j = (\bar{A}_1 \Delta_j)^T P_0 + P_0 (\bar{A}_1 \Delta_j). \tag{3.19}
\]

\( P_0 \) is a positive definite matrix and a solution of the ARE:

\[
A_q^T P_0 + P_0 A_q - P_0 B R^{-1} B^T P_0 = - Q_0. \tag{3.20}
\]

From (1.2) and (3.16) we next conclude that

\[(\bar{A}_0 + \bar{A}_1 \Delta(q)) P + P (\bar{A}_0 + \bar{A}_1 \Delta(q)) - P B R^{-1} B^T P + Q_0 = 0 \tag{3.21}\]

and taking into account the basic relation (2.5) and (3.18) we finally get the useful relation

\[
\left( \bar{A}_0 + \bar{A}_1 \left( \Delta_0 - \sum_{j=1}^\ell g_j \gamma(i, j) \Delta_j \right) \right) P + P \left( \bar{A}_0 + \bar{A}_1 \left( \Delta_0 - \sum_{j=1}^\ell g_j \gamma(i, j) \Delta_j \right) \right) - P B R^{-1} B^T P + \sum_{j=1}^\ell g_j \gamma(i, j) Q_j = 0 \tag{3.22}
\]

viz:

\[(\bar{A}_0 + \bar{A}_1 \Delta_0)^T P + P (\bar{A}_0 + \bar{A}_1 \Delta_0) - P B R^{-1} B^T P + Q_0 = \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\bar{A}_1 \Delta_j)^T P + P (\bar{A}_1 \Delta_j) \right) - \sum_{j=1}^\ell g_j \gamma(i, j) Q_j. \tag{3.23}\]

Relations (3.19) and (3.23) imply the next formal consequence

\[
(\bar{A}_0 + \bar{A}_1 \Delta_0)^T P + P (\bar{A}_0 + \bar{A}_1 \Delta_0) - P B R^{-1} B^T P + Q_0 = \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\bar{A}_1 \Delta_j)^T P + P (\bar{A}_1 \Delta_j) - (A_1 \Delta_j)^T P_0 \right)
- P_0 (\bar{A}_1 \Delta_j). \tag{3.24}
\]

which finally involve (under assumption \( P = P_0 \)) the resulting ARE of the following type:

\[
(\bar{A}_0 + \bar{A}_1 \Delta_0)^T P_0 + P_0 (\bar{A}_0 + \bar{A}_1 \Delta_0) - P_0 B R^{-1} B^T P_0 + Q_0 = 0 \tag{3.25}
\]

Let us note that the obtained ARE (3.25) does not include any (unknown) active location. It depends only on the given structure of (2.1). The essential parameters, namely, \((\bar{A}_0, \bar{A}_1, \Delta_0)\) determine the above equation structure. Let us summarize the obtained result in the form of a theorem.

**Theorem 1.** Assume that all the technical conditions of this Section are satisfied. Then the optimal feedback solution to the main Problem 1, where \( Q(q) \) is given by (3.9), (3.18) and (3.19), has the following form

\[
F_0 = R^{-1} B^T P_0, \quad u = -F_0 x. \tag{3.26}
\]

Here \( P_0 \) is a solution of the ARE (3.25).

The presented analytic result constitutes a theoretical basis for the stabilization problem studied in the next section.

### 4. The LQ Based Stabilization of Switched Systems

As a well-known stabilizability property of the classic optimal LQ control design (see for example Lewis et al (2012); Kailath (1980)), we next show that the generic optimal control from Theorem 1 also stabilizes system (1.1)–(1.3) even in the case of unknown locations \( q \in Q \).

Applying the optimal control feedback (3.26) to the switched system representation (1.1), we obtain the following closed loop state space form

\[
\dot{x} = A_q x - B F_q \dot{x} = (A_q - B F_q) \dot{x} \tag{4.1}
\]

Taking into consideration the previously derived formulae (1.2) and (2.5) in (4.1), we also get:

\[
\dot{x} = (\bar{A}_0 + \bar{A}_1 \Delta_0 - \sum_{j=1}^\ell g_j \gamma(i, j) \Delta_j) - B R^{-1} B^T P_0 = \dot{x} = (\bar{A}_0 + \bar{A}_1 \Delta_0 - \sum_{j=1}^\ell g_j \gamma(i, j) \Delta_j) - B R^{-1} B^T P_0 \tag{4.2}
\]

where: \( F(\Delta_j, P_0) = \bar{A}_0 + \bar{A}_1 \Delta_0 - \sum_{j=1}^\ell g_j \gamma(i, j) \bar{A}_1 \Delta_j - B R^{-1} B^T P_0 \). Let us define the Lyapunov function:

\[
V(x) = \pi(t) T P_0 \pi(t). \tag{4.3}
\]

The usual Lie derivative of (4.3) (along the trajectories of system (4.2)) next implies:

\[
\dot{V}(t) = \pi^T (t) P_0 \pi + \pi(t) T P_0 \pi \tag{4.4}
\]

Relations (3.19) and (3.23) imply the next formal consequence

\[
(\bar{A}_0 + \bar{A}_1 \Delta_0)^T P + P (\bar{A}_0 + \bar{A}_1 \Delta_0) - P B R^{-1} B^T P + Q_0 = \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\bar{A}_1 \Delta_j)^T P + P (\bar{A}_1 \Delta_j) - (A_1 \Delta_j)^T P_0 \right)
- P_0 (\bar{A}_1 \Delta_j). \tag{3.24}
\]
\[ \dot{V}(t) = \pi^T \left( (\overline{A}_0 + \overline{A}_1 \overline{A}_0) \dot{P}_0 + P_0 (\overline{A}_0 + \overline{A}_1 \overline{A}_0) \right) P_0 - P_0 B R^{-1} B^T P_0 \]
\[ + \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\overline{A}_1 \overline{A}_j) \right)^T P_0 \]
\[ + P_0 (\overline{A}_1 \overline{A}_j) \big) \right)\pi. \]

From (3.25) and (3.26) we deduce
\[ \dot{V}(\pi) = -\pi^T \left( \left[ \sqrt{Q_0} F \right] + \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\overline{A}_1 \overline{A}_j) \right)^T P_0 \right) \pi, \]
where
\[ Q_0 = \left[ \sqrt{Q_0} F \right] + \sum_{j=1}^\ell g_j \gamma(i, j) \left( (\overline{A}_1 \overline{A}_j) \right)^T P_0 \]

The analytic relations obtained above constitute in fact a formal proof of our next stability result. Let us also recall Theorem 5.10 of Chapter 6, Section 5 of Stewart (1973), and Corollary 2.6-2 of Kailath (1980).

The obtained result provides a stability criterion for the switched systems under consideration in absence of the exact a priori information about a concrete switching mechanism.

Starting from a model in the form (1.1), the procedure to design the feedback is summarized as follows.

1. Identify the parameters of the implicit representation (2.1)–(2.5), in particular the matrices \( \overline{A}_0, \overline{A}_1, B, \) and \( \overline{A}_0 \).
2. Choose matrices \( R \) and \( Q(q) \) in the form defined by (3.9) and (3.18), satisfying one of the four conditions of Theorem 2, namely, (4.5), (4.6), (4.7) or (4.8).
3. Solve the Riccatti equation (3.25), and define the feedback by (3.26).

5. NUMERICAL ASPECTS

In this section we apply the theoretical results (stability results) developed in the previous parts of the manuscript and study an illustrative example taken from Azhmyakov (2019). Let us also refer to Bonilla et al. (2019) for some further examples and theoretical details. Consider now the state space representation (1.1) determined by the following matrices
\[ A_q = \begin{bmatrix} \alpha & \beta + 1 \\ \alpha + 1 & \beta \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_q = [-\alpha - \beta]. \] (5.1)

Comparing with (1.2), we can observe that
\[ \overline{A}_0 = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, \quad \overline{A}_1 = \begin{bmatrix} 1 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \end{bmatrix}, \quad \overline{D}_q = [\alpha \beta], \]
\[ \overline{Q}_0 = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}, \quad \overline{C}_1 = 1. \] (5.2)

The conventional systems transfer functions of (1.1) and (5.1) for each pair \((\alpha, \beta)\) are:
\[ F_q(s) = \frac{(\beta + 2)s - \alpha}{(s + 1)(s - (\alpha + \beta + 1))}. \] (5.3)

### 5.1 Global Implicit Representation

The global implicit representation associated with (1.1) and (5.1) can be formalized as
\[ \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{x} \\ \dot{\pi} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \beta & -1 \end{bmatrix} \begin{bmatrix} \pi \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \]
\[ y = [\alpha - \beta - |i - 1| \begin{bmatrix} \pi \\ \dot{x} \end{bmatrix}, \] (5.5)

viz (cf. \( \sum_{\alpha}^{gr} \mathbb{E}, \mathbb{A}, \mathbb{B}, \mathbb{C} \)). Moreover, we have
\[ \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} \dot{x} \\ \dot{\pi} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \alpha^2 & -1 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \] (5.6)
\[ y = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} x. \]
We now consider the systems controllability requirement and examine the characteristic determinant of the controllability matrix for the pair \((A_q, B)\) which is:
\[
\det [B \quad A_q] = \begin{vmatrix} 0 & \beta + 1 \\ \alpha & \beta \end{vmatrix} = -\beta - 1. \tag{5.7}
\]
The characteristic polynomial of \(A_q\) is:
\[
|sI_2 - A_q| = \begin{vmatrix} s & -\alpha \\ 0 & s \end{vmatrix} = (s - (\alpha + \beta)s - (\alpha + \beta + 1) = (s + 1)(s - (1 + \alpha + \beta)). \tag{5.8}
\]
Hence, we are ready to calculate the eigenvalues of \(A_q\)
\[
s_1 = -1
\]
\[
s_2 = 1 + \alpha + \beta \tag{5.9}
\]
From (5.7), we obtain the controllability region associated with the pair \((A_q, B)\):
\[
CR(A_q, B) = \{ \beta \in \mathbb{R} : \beta \neq -1 \}. \tag{5.10}
\]
From (5.8), we have the Hurwitz region of \(A_q\):
\[
HR_{A_q} = \{ (\alpha, \beta) \in \mathbb{R}^2 : 1 + \alpha + \beta < 0 \}. \tag{5.11}
\]
Let us note that the locations set (5.2) has the same structure as \(H1\), which implies the next constructive relation:
\[
q \in \mathbb{Q}_{(-\alpha, -\beta)} \Rightarrow \begin{cases} (\alpha, \beta) \in Q \\
(\alpha, \beta) = (\alpha - \beta) + \gamma(i, 1)(\alpha - \beta)q_1 + \gamma(i, 2)(\beta - \beta)q_2,
\end{cases} \quad \gamma(i, 1), \gamma(i, 2) \in [0, 1] \tag{5.12}
\]
where
\[
\bar{\alpha} = 1.5, \quad \bar{\beta} = 3, \quad \alpha = 1, \quad \beta = 0. \tag{5.13}
\]
Moreover, we also have \(q_0 = (\alpha - \beta) = (\alpha - 1, 0), \quad q_1 = (\alpha - 1, 0)\) and \(q_2 = (0, -1)\). In Table 1 we present the values of \(\gamma(i, j)\) and we additionally deduce:
\[
\Delta_0 = (\alpha - \beta) = (-1, 0), \quad \Delta_1 = -q_1 = (1, 0), \quad -\Delta_2 = -q_2 = (0, 1). \tag{5.14}
\]

### 5.2 LQ Feedback Stabilization

We now solve the ARE (3.25) with
\[
\mathbb{Q}_0 = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, \quad \text{and} \quad R = 1. \tag{5.15}
\]
Note that a concrete numerical solution procedure for the resulting ARE can be easily found in MATLAB® and Python numerical packages. From (5.3), (5.14) and (5.15) we deduce that this solution of (3.25) has the following form
\[
P_0 = \begin{bmatrix} 0.4765 & 0.2168 \\ 0.2168 & 1.9714 \end{bmatrix}. \tag{5.16}
\]

Hence, the feedback (3.26) is:
\[
F_{e0} = R^{-1}B^TP_0 = [0.2168 \quad 1.9714], \quad u = -F_{e0}\bar{x}. \tag{5.17}
\]

As proposed in Section 4, the above optimal LQ - type control feedback is finally used for the system stability design. From (5.3), (5.14) and (5.16) we can deduce the concrete stability relations and the corresponding numerical parameters
\[
\mathbb{Q}_1 = (A_1A_1^T)P_0 + P_0(A_1A_1^T) = \begin{bmatrix} 1.3867 & 1.4142 \\ 1.4142 & 0 \end{bmatrix},
\]
\[
\mathbb{Q}_2 = (A_2A_2^T)P_0 + P_0(A_2A_2^T) = \begin{bmatrix} 0 & 0.6933 \\ 0.6933 & 2.8284 \end{bmatrix}. \tag{5.18}
\]
The spectra \(\sigma_j \in \{ -0.8817, 2.2684 \}, \quad \text{and} \quad \sigma_2 = \{-0.1608, 2.9892\}\).

Hence
\[
\lambda_{\min}(\mathbb{Q}_0) = 1, \quad \lambda_{\min}(\mathbb{Q}_1) = -0.8817, \quad \lambda_{\min}(\mathbb{Q}_2) = -0.1608. \tag{5.19}
\]
Finally, from (5.19) and (5.13), we get (cf. (5.4) with \(g_1 = (\alpha - \alpha)\) and \(g_2 = (\beta - \beta)\):
\[
\lambda_{\min}(\mathbb{Q}_0) + (\alpha - \alpha)\lambda_{\min}(\mathbb{Q}_1) + (\beta - \beta)\lambda_{\min}(\mathbb{Q}_2) = 1 + (0.5(-0.8817)) + (3)(-0.1608) = 0.0767 > 0. \tag{5.20}
\]
The above inequality implies the conventional Lyapunov stability conditions (4.5).

Figure 1 depicts all the feasible points that satisfy the sufficient condition. Analysing this Figure, we can conclude that all the locations are inside the sufficiently big stability region, where the formal analytic stability conditions are expressed by (4.5).
\[
A_{q1} = \begin{bmatrix} -1.5 & 0.2 \\ -0.5 & -0.8 \end{bmatrix}, 
A_{q2} = \begin{bmatrix} -1 & -1 \\ 0 & -2 \end{bmatrix}, 
A_{q3} = \begin{bmatrix} -1 & -1 \\ 0 & 0 \end{bmatrix}, 
A_{q4} = \begin{bmatrix} -1 & -2 \\ 0 & -3 \end{bmatrix}, 
A_{q5} = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}.
\]

The generated signal \(i(t)\) involves a random switching mechanism and hence the sub-systems and the realizations of the switching times are unknown. Figure 2 shows the time response under the initial condition \(\pi = [8 \ 3]^T\) and the switching signal \(i(t)\).

In fact, we have proven a kind of a “robustness” result with respect to a possible (admissible) switching mechanism: the LQ-type optimal control design from the main Theorem 1 stabilizes system (1.1)–(1.3) under the assumption of an unknown dynamic location \(q \in \mathcal{Q}\). Note that the formal proof of Theorem 2 involves some recent results from (Bonilla et al., 2015b).

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