Body Shape and Complex Permittivity Determination Using the Method of Auxiliary Sources

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Abstract—In this article, the body shape and complex permittivity determination employing inverse electromagnetic scattering problem solution for two-dimensional cases is considered. The method of auxiliary sources (MAS) is used as a mathematical apparatus. Several body shape cases are considered, and the efficiency of the approach is shown. The program package is created based on this method, and the numerical experiment results are presented.

1. INTRODUCTION

For fifty years, inverse scattering problem has kept its importance and popularity because its ultimate purpose to determine physical and geometric properties of an inaccessible object without invasive methods which is always studied and appeals researchers through this branch of science. In such problems, the body can have mainly two essential properties. They are geometric properties which are the position and shape of the object, and physical properties which can be expressed with constitutive parameters. To find these main features, generally, the information contained in the scattered field distribution by this object is studied, when the object is illuminated by the wave source. Here, we only deal with electromagnetic waves in order to find main properties of the object. In this study, the body is located at an unbounded and homogeneous free space. Such a case is crucial for military, space, telecommunication, satellite applications.

There exist several methods for this problem solution. By the authors of this article, there are several published articles in this direction \cite{1–4}. In studies \cite{1, 2}, the authors only focused on the body shape and location determination. Under some specific conditions and estimations, there exist many highly accurate analytical and numerical methods for solving the inverse scattering problem. In article \cite{5}, the author used the double Fourier transform for the inverse problem solution. Furthermore, there are some other analytical and numerical works related to this topic \cite{6–14}. There are also other authors who used the Method of Auxiliary Sources to solve direct scattering problem \cite{15–17}.

A underlying assumption has been commonly made in many studies related to cylindrical bodies with infinite length in a specific direction. Studies starting with this assumption are done very frequently because of two reasons. First, the investigation is slightly simpler than other problems, and secondly, such objects have a similar geometry to real-world objects. In this article, we also deal with such an object. Here, a two-dimensional case is considered when the object is infinite in the third dimension.

The novelty of our article is that it solves this problem using MAS and expands the possibility of this method not only for body shape determination but also for complex permittivity determination. The body of interest is illuminated from many different locations, and the scattered field is measured at some circular curve contour. During the real experiment, it is needed to realize many measurements from different positions at several frequencies in order to get higher accuracy of the solution. In the
scope of this article, our investigation is based on the simulation. Note that in order to get the scattered field, the direct problem is also solved.

In Section 2, the theoretical background for the direct and inverse problem and the numerical approach is mainly presented. In the following section, the results are given. In Section 4, the comparison of the results obtained with our method and the other method is given. After that, the conclusion is drawn.

2. THEORETICAL BACKGROUND

In the theoretical part, the problem statement, field extremum on the boundary of two media, direct problem solution using the MAS, inverse problem solution again using the MAS and determination of the object complex permittivity are explained in detail.

2.1. Problem Statement

The problem statement, mathematical background, and application point of view are drawn in the section. Suppose that there is an object which in general has unknown complex Permittivity. In Figure 1, the problem statement is given. This non-magnetic object is radiated by a point source located near the object (Figure 1). The object reradiates scattered field. The scattered field \( E_{sc} \) satisfies the next well-known relation given in Eq. (1). Keep in mind that, here, the two-dimensional problem has only \( z \)-component.

\[
E_{sc}(\vec{r}_{out}) = \int_C H_0^{(1)}(k|\vec{r}_{out} - \vec{r}|) \nu(\vec{r}) [E_{inc}(\vec{r}) + E_{out}(\vec{r})]dx dy
\]  

(1)

Here, \( E_{sc}(\vec{r}_{out}) \) is the scattered field at some point outside the object of interest; \( \nu \) is called the object function given in Eq. (2); and \( C \) in the integral is the contour of the object.

\[
\nu(\vec{r}) = \frac{1}{\epsilon_0} \left( \epsilon(\vec{r}) + i \frac{\sigma(\vec{r})}{\omega} \right) - 1
\]  

(2)

Here, \( \epsilon(\vec{r}) \) is the complex permittivity of the object which is, in general, the function of the coordinate, conductivity \( \sigma(\vec{r}) \), angular frequency \( \omega \), wavenumber \( k \), and incident electric field \( E_{inc}(\vec{r}) \). If the Born approximation is used, inside the integral, there will be only the incident electric field as in Eq. (3).

\[
E_{sc}(\vec{r}_{out}) = \int_C H_0^{(1)}(k|\vec{r}_{out} - \vec{r}|) \nu(\vec{r}) E_{inc}(\vec{r}) dx dy
\]  

(3)

Note that the approximation works for small values of the object function and wave number. The term which is neglected has asymptotical behavior \( O(k^4 ||\nu||^2) \). Initially, the object function is unknown. If this function is determined, the complex permittivity given in Eq. (4) can be found.

\[
\epsilon(\vec{r}) + i \frac{\sigma(\vec{r})}{\omega} = \epsilon_0(\nu(\vec{r}) + 1)
\]  

(4)

Figure 1. The problem statement.
Consider the case when there is a smooth object with constant permittivity $\epsilon(\vec{r}) = c$. Here, $c$ is a constant. In this case, the object function can be taken outside the integral.

$$E_{sc}(\vec{r}_{out}) = \nu \int_C H_0^{(1)}(k |\vec{r}_{out} - \vec{r}|) E_{inc}(\vec{r}) \, dx \, dy$$

(5)

In order to find $\nu$, it is needed to take the integral (the incident field distribution is known) and to measure the scattered field in a point. Therefore, this problem is very easily solved. However, the difficulty arises when the contour of integration is not known. In other words, the shape of the object and its exact location are not known. This means that one cannot measure the object, directly. For example, it is the case when a satellite orbits around some asteroids and tries to determine the electric properties of the asteroids. Also, there can be the case when the object is not reachable for human, and the only way is to send a drone to inspect the object by using electromagnetic waves. There can be other situations on earth when one cannot reach the object of interest to measure it, but one can go around the object, illuminate with a source, and measure the scattered field. Therefore, the first goal is to determine the shape of the object and its location in space. This problem could be generalized when the object of interest is embedded in some media. In that case, the hidden object is not seen directly, and the only way is to investigate its properties using electromagnetic radiation. However, this article only considers the case when the object is located in free space.

2.2. Determination of the Field Extremum on the Boundary of Two Media

Before going to the main problem, it is better to highlight the fundamental property of the electromagnetic fields at the border of two dielectrics. For simplicity, it is an acceptable idea to consider the case when we have the infinite planar boundary of two media with permittivity’s $\epsilon_1$ and $\epsilon_2$ given in Figure 2. Note that the incidence wave is a plane wave and time dependency $e^{i\omega t}$.

![Figure 2](image)

**Figure 2.** The diffraction by the infinite planar boundary of the two media.

Respectively, the incident, reflected and transmitted field can be given as in Eq. (6).

$$E_{inc}(\vec{r}) \hat{a}_z = E^0(\vec{r}) = E_0 e^{(-ik(x \sin \phi - (y-y_0) \cos \phi))} \hat{a}_z$$

$$E^- \hat{a}_z = E_0 \rho_\perp e^{(-ik(x \sin \phi + (y-y_0) \cos \phi))} \hat{a}_z$$

$$E^+ \hat{a}_z = E_0 \tau_\perp e^{(-ik(x \sin \phi + (y-y_0) \cos \psi))} \hat{a}_z$$

(6)

Here, $\rho_\perp$ and $\tau_\perp$ are the reflection and transmission coefficients, respectively, and $\hat{a}_z$ is the unit vector directed through the $z$-axis. Let us introduce the next integral given in Eq. (7) and solve it.

$$P(y) = \int_{k_1}^{k_2} \left| E_0 + E^- \right|^2 \, dk = E_0^2 (k_2 - k_1) \left( 1 + \rho_\perp^2 + 2 \rho_\perp \frac{\sin \alpha}{\alpha} \cos \beta \right)$$

(7)
where, \( \beta = (y - y_0)(k_2 + k_1) \cos \phi \) and \( \alpha = (y - y_0)(k_2 - k_1) \cos \phi \).

Now, it is time to find the extremums of Eq. (7). It can be easily shown that

\[
\begin{align*}
\min [P(y)] &= E_0^2 (k_2 - k_1)(1 + \rho_\perp)^2 = P(y_0) \quad \text{if } \rho_\perp < 0 \text{ and } \epsilon_2 > \epsilon_1 \\
\max [P(y)] &= E_0^2 (k_2 - k_1)(1 + \rho_\perp)^2 = P(y_0) \quad \text{if } \rho_\perp > 0 \text{ and } \epsilon_2 < \epsilon_1
\end{align*}
\]

Here, \( \min \) and \( \max \) stand for the minimum and maximum values of the function, respectively.

In both cases, we have the extremum at the boundary of two dielectric media, and this result can be generalized for other shapes of the boundary. Therefore, it is needed to know the scattered field distribution along the boundary of two media, and then, the integration of the electric field in frequency range from \( k_1 \) to \( k_2 \) is required as given in Eq. (8). Therefore, we have to find the analog of the \( P(y) \) function which will be, in our case, the function of coordinate variables \( P(x, y) \).

\[
P(x, y) = \int_{k_1}^{k_2} \left| E_0(x, y) + E^-(x, y) \right|^2 dk
\]

During the numerical experiment, the integral is substituted by the sum as in Eq. (9).

\[
P(x, y) = \sum_{k=k_1}^{k_2} \left| E_0(x, y) + E^-(x, y) \right|^2 \Delta k
\]

We should choose the range of \( x \in [x_1, x_2] \) and \( y \in [y_1, y_2] \) so that it should occupy the object of interest. The first aim is to find the distribution of \( P(x, y) \) function. The next step is to find the minimum (or maximum) area in \( P(x, y) \) which corresponds to the boundary of two media. This gives an ability to reconstruct the shape of the scattering object. However, in order to determine \( P(x, y) \) function, it is needed to know the scattered electric field value in the ranges \( x \in [x_1, x_2] \) and \( y \in [y_1, y_2] \). Keep in mind that the incident field is initially given and known. During the real experiment, it is necessary to conduct measurements of the scattered field. We have to take into account the fact that \( P(x, y) \) requires the scattered electric field distribution for different frequencies in the range of wavenumber \( k \in [k_1, k_2] \).

In our case, the real experiments is substituted by the direct problem solution of the diffraction problem. For this, MAS is used which is discussed in the following part [4].

### 2.3. Direct Problem Solution Using the MAS

The MAS gives an ability to solve the diffraction problem of electromagnetic waves of the cylinder source by the 2D complex shape object (with the contour \( S \)) with complex permittivity \( \epsilon \) (Figure 3). The permittivity of the area outside the object contour corresponds to the vacuum.

The electric field of the cylinder source can be expressed as \( E_{\text{inc}}(\vec{r}) = H_0^{(1)}(k |\vec{r} - \vec{r}_{\text{inc}}|) \). \( H_0^{(1)} \) is the Hankel’s function of the first kind and zero-order, and \( \vec{r}_{\text{inc}} \) is the location of the line source. The observation point is given as \( \vec{r} = (x, y) \). The electric field has only \( z \)-component which is normal to the object contour. Outside and inside the \( S \) contour, there are constructed inner and outer auxiliary contours. Inner contour describes the electric field distribution outside the scatterer. Outer contour describes the field inside the scatterer. Mathematically, the electric field outside the scatterer is the sum of the field by the incident source and the field radiated by the inner auxiliary surface as given in Eq. (10).

\[
E_{\text{out}}(x, y) = \sum_{n=1}^{N} X_n^{\text{in}} H_0^{1} \left( k_{\text{out}} \sqrt{(x - x_{\text{in}})^2 + (y - y_{\text{in}})^2} \right) + E_{\text{inc}}(x, y)
\]

Here, \( X_n^{\text{in}} \) are unknown complex amplitudes of the auxiliary sources distributed on the inner auxiliary contour. Then, the electric field inside the scatterer is expressed as in Eq. (11)

\[
E_{\text{in}}(x, y) = \sum_{n=1}^{N} X_n^{\text{out}} H_0^{1} \left( k_{\text{in}} \sqrt{(x - x_{\text{out}})^2 + (y - y_{\text{out}})^2} \right)
\]
Here, $X_{\text{out}}^n$ are unknown complex amplitudes of the auxiliary sources distributed on the outer auxiliary contour. In order to find these unknown coefficients, the boundary condition satisfaction on the scatterer’s surface is required. For the dielectric media, boundary condition means that the tangential components of the electric and magnetic fields should be continuous along the scatterer surface.

$$E_{\text{in}}(x_m, y) = E_{\text{out}}(x_m, y), \quad H_{\text{in}}(x_m, y_m) = H_{\text{out}}(x_m, y_m).$$

$(x_m, y_m)$ are the points on the $S$ surface.

Magnetic field strength ($\vec{H}$) is found by Maxwell’s equation given as follows.

$$\text{rot} \vec{E} = -i \omega \mu \vec{H}$$

After requiring these boundary conditions on the scatterer’s surface, the linear algebraic equation system with respect to the unknowns $X_{\text{in}}^n$ and $X_{\text{out}}^n$ is achieved. After solving this linear algebraic equation system, unknown coefficients are determined. Then, the field distributions due to the scatterer are found inside and outside. The values of the scattered field will be used for the next step when the inverse problem is solved.

### 2.4. Inverse Problem Solution Using the MAS

We determine the scattered field with the method described in Section 2.3 on a semicircle (Figure 4) which contains the scatterer object.

During the real experiment, instead of solving the diffraction problem by the scatterer, we should measure the field values on this semicircle (Figure 4) which we call measured surface. The MAS gives an ability to reconstruct the scattered field in the rectangular area shown in Figure 4.

Let’s formulate the MAS for the inverse problem solution. We construct an auxiliary semicircle contour near the measured surface. The reconstructed scattered field $E_{\text{rec}}(x, y)$ can be expressed as Eq. (12).

$$E_{\text{rec}}(x, y) = \sum_{n=1}^{N} X_n H^{(2)}_0 \sqrt{(x-x_n)^2 + (y-y_n)^2}$$

Here, $X_n$ are unknown coefficients of the auxiliary sources to be determined. $H^{(2)}_0$ is the Hankel’s function of the second kind and zero-order, which describes the absorbing wave. In order to find $X_n$ coefficients, the boundary condition satisfaction on the measured surface is required. In Eq. (12), $x_n$ and $y_n$ are the auxiliary source locations, and $x$ and $y$ are any points in the reconstructed area. The reconstructed field on this surface must be the same as the measured field as given in Eq. (13).

$$E_{\text{sc}}(x_n, y_n) = E_{\text{rec}}(x_n, y_n) \quad n = 1, 2, 3...N$$
Here, $N$ is the number of collocation points where the boundary condition satisfaction is required. As a result, we get the linear algebraic equation system after the solution of which we find the unknown coefficients. After that, it is possible to reconstruct the scattered field in the rectangular area shown in Figure 4. Then, it is possible to find function $P(x, y)$ with Eq. (9).

The next step is to find its minimum value of $P(x, y)$ and then, to find the body shape and location. However, if we radiate the object from one side, it can reconstruct only a part of the object surface, because the auxiliary surface is not closed. Therefore, in order to find the whole surface of the scatterer, we have to conduct many measurements. We should change the location of the incident field source and at the same time, should rotate the measured surface. Finally, we have to add the obtained $P(x, y)$ function distribution for different source locations, and we will see the extremum of $P(x, y)$ on this distribution. After that, it is necessary to extract the body contour from this minimum area. Because we have the numerical data, it is easy to determine in which points $P(x, y)$ is minimum. As a result, we get the object contour with some thickness. Thickness depends on the choice of the wavelength range. Finally, we should filter the data to extract the object contour.

2.5. Determination of the Object Complex Permittivity

As mentioned above, the Born approximation given in Eq. (5) is used in order to evaluate complex permittivity. This time, it needs one more extra measurement to get the scattered field in the point outside the scatterer. In our case, we find this value by the direct problem solution described in Section 2.3. The wavenumber is chosen so that it is compatible with the Born approximation. In this study, the wavenumber equals $k = 0.25$. Because in the previous stage, we have determined the contour shape, we can take integral given in the right part of Eq. (5). Here, the incident wave is initially given, and the Hankel argument contains all the known variables. Therefore, we divide the scattered field given in the left part by the integral value, and we get object function $\nu$. Finally, from Eq. (4), the value of the complex permittivity can be found.

3. RESULTS OF THE NUMERICAL SIMULATION

In this section, results of the numerical simulations are presented for the stages of reconstruction procedure and also for different shapes.

The program package is created for this problem solution. First, the circle shape dielectric with complex permittivity is considered. For simplicity in the considered cases, the permittivity has a real component and equals $\varepsilon = 4$. The radius of the cylinder is taken as 1 meter, so the problem in the frequency range with corresponding wavenumbers $k$ is from $1 \text{m}^{-1}$ to $10 \text{m}^{-1}$ with the step $\Delta k = 0.25 \text{m}^{-1}$. As mentioned above, the scattered field is found by first solving the direct problem. Figure 5 shows several partial reconstructions of the circle surface for different incident angles and also
Figure 5. Step by step reconstruction of the circle surface ($P(x, y)$ distribution function for different incident angles).

Figure 6. The reconstructed contour of the scatterer: (a) In first approximation, (b) the final contour.

As seen, we get a circle shape spot object as a result if we extract corresponding points of space, and we get the next image in Figure 6(a). We should extract the contour with possibly minimal width from this aggregate of points. We do it using a computer program. We filter these points and get the contour as in Figure 6(b).

As the contour is found, we can find the object function $v$ from Eq. (5) and determine the complex values for the permittivity. The reconstructed value of the permittivity is $\epsilon = 3.94$. The accuracy of the solution is 98.5%. The error that we get is the result of reconstructed contour shape deviation from the real contour.

We also consider the case of an ellipse (with parameters $1 \, \text{m} \times 0.5 \, \text{m}$), and the corresponding results are shown in Figure 7. In Figure 7, (a) corresponds to the reconstructed function $P(x, y)$; (b) is the contour reconstruction in the first approximation; and (c) is the final reconstructed contour.

In the case of an ellipse, the reconstructed value is $\epsilon = 3.92$. The error of the permittivity value determination is a little higher and is about 2%. This is also due to the deviation of the reconstructed contour from the real one.

An object with a more complicated shape is also considered which is given on Figure 8. The geometry of the object is composed by line segments and circle sectors. Dimensions of the object are listed in Table 1.

The curvature radii $r_1$ and $r_2$ are comparatively small in order to simulate the effect of the rectangular corner. We do not put these radii as zero because we cannot solve the direct problem with the MAS if the object shape is not smooth [4]. Radius $r_3$ is comparatively bigger in order to
Figure 7. The reconstructed contour of the scatterer: (a) $P(x, y)$ function distribution, (b) the contour in the first approximation, (c) the final contour.

Figure 8. The geometry of the object with complicated shape.

Table 1. Parameters and their values for Figure 7.

| Parameter | Value ($m$) |
|-----------|-------------|
| $a$       | 1.00        |
| $b$       | 0.35        |
| $c$       | 0.80        |
| $r_1$     | 0.10        |
| $r_2$     | 0.05        |
| $r_3$     | 0.20        |

check how the method works with concave and convex geometries. The permittivity is again $\epsilon = 4$, but because the object dimension is smaller than previous cases, we need higher frequencies. That is why we change the wave number in the range $5 \text{ m}^{-1}$ to $20 \text{ m}^{-1}$ with steps $\Delta k = 0.25 \text{ m}^{-1}$. Figure 9(a) shows the reconstructed $P(x, y)$ distribution. Figure 9(b) shows the reconstructed object contour in the first approximation, and lastly, Figure 9(c) shows the final reconstructed contour.
Figure 9. The reconstructed contour of the scatterer: (a) \( P(x, y) \) function distribution, (b) the contour in the first approximation, (c) the final contour.

The reconstructed value of the permittivity is 3.86. The error of solution is 3.5%. The error is slightly higher than the previous cases. Again the reason is the contour reconstruction error. However, in all three cases, the error is less than 4%.

We also consider the case when the conductivity is not zero. For \( \sigma = 3000000 \, \text{Ω}^{-1} \), the reconstructed value of the permittivity \( (\epsilon) \) given in Eq. (4) is 3.86, and the reconstructed value of the conductivity is \( \sigma = 2991721 \, \text{Ω}^{-1} \). The error of the permittivity is again 3.5%, and error of conductivity is 0.3%. Note that we do not consider higher values of permittivity. After \( \epsilon > 5 \), the error becomes more than 5%. In order to increase accuracy, it is needed to consider higher frequencies, and more measurements need to be carried out.

4. COMPARISON WITH OTHER METHOD

In this section, we present the comparison of the results obtained by our method and the analytical method described in the work [18]. The object of interest is a circle with radius 1 and permittivity \( \epsilon = 1.1 \). The geometry for the inverse problem analytical solution is given in Figure 10. In order to
solve the inverse problem and find the permittivity distribution with analytical method [18], the object of interest is irradiated by the plane wave with the angle of incidence $\theta$. After that, the scattered field is measured on the circular surface (dotted circle Figure 10) for different $\varphi$ values. The next step is to find the Fourier transform of the object function which can be expressed as:

$$\hat{v} [k (\cos \varphi - \cos \theta), k (\sin \varphi - \sin \theta)] = \frac{\sqrt{8k\pi}}{k^2} e^{-\frac{i\pi}{4} \sqrt{|x|}} e^{-ik|x|} u^s(x)$$

where $k$ is the wave number; $x = (x_1, x_2)$ is the point in which the scattered field is measured (observation point); $u^s(x)$ is the scattered field measured in point $x$.

After that, we calculate the object function in point $x_0 = (x_{10}, x_{20})$ with the next formula:

$$v(x_{10}, x_{20}) = \frac{k^2}{(2\pi)^2} \int_{\pi}^{-\pi} d\theta \int_{\theta-\pi}^{\pi} \hat{v} [k (\cos \varphi - \cos \theta), k (\sin \varphi - \sin \theta)] e^{ik((\cos \varphi - \cos \theta)x_{10}+(\sin \varphi - \sin \theta)x_{20})} d\varphi$$

We calculate the object function with this formula along the $x_1$ axis in the range $(-1.5, 1.5)$ and after that evaluate the permittivity value using Eq. (4).

Figure 11 shows the comparison between the results obtained with the analytical method and that with the MAS. The value of wave number for the analytical method case is $k = 1.25$ m$^{-1}$, and the radius of the circle on which the scattered field is evaluated is $\rho = 50$ m.

![Figure 11](image)

**Figure 11.** Reconstruction of the permittivity distribution obtained by two different methods: with MAS and with the analytical method.

As shown in the figure, our method gives a stepwise solution with correct values of the permittivity, and the analytical method gives the results with smoothly changing values of permittivity with some deviation from real value. This comparison emphasizes the efficiency of our method.

5. CONCLUSION

In this article, body shape and permittivity determination are considered by solving the inverse problem. As the mathematical apparatus, the MAS is used. A program package is created for the numerical experiment realization, and some results are presented. The numerical experiments show the efficiency of the new method.
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