We present a study of the use and limits of the Time-Delay Interferometry null channels for in flight estimation of the Laser Interferometer Space Antenna instrumental noise. The paper considers how the two main limiting noise sources, test-mass acceleration noise and interferometric phase measurement noise, propagate through different Time-Delay Interferometry channels: the Michelson combination $X$ that is the most sensitive to gravitational waves, then the less-sensitive combinations $\alpha$, and finally the null channel $\zeta$. We note that the null channel $\zeta$, which is known to be equivalent to any null channel, not only has a reduced sensitivity to the gravitational waves, but also feature a larger degree of cancellation of the test mass acceleration noise relative to the interferometry noise. This severely limits its use in quantifying the low frequency instrumental noise in the Michelson $X$ combination, which is expected to be dominated by acceleration noise. However, we show that one can still use in-flight noise estimations from $\zeta$ to put an upper bound on the considered noises entering in the $X$ channel, which allows to distinguish them from a strong stochastic gravitational wave background.

I. INTRODUCTION

The Laser Interferometer Space Antenna (LISA) gravitational wave observatory \cite{1} is expected to be continuously dominated by gravitational wave (GW) signals in its mHz frequency band. This implies a technical difficulty in quantifying and understanding the instrumental noise of LISA in the constant presence of GW signals, which is essential for maximizing the observatory scientific return and to identify possible Stochastic Gravitational Wave Backgrounds (SGWBs) \cite{1}. The LISA scientific observables are constructed from so-called Time-Delay Interferometry (TDI) combinations, which synthesize equal arm interferometers to cancel an otherwise overwhelming contribution from laser frequency noise \cite{2}. The primary GW observables will be obtained from TDI channels such as the Michelson $X$ channel. This channel represents a virtual interferometer with the same principle of measurements of a standard Michelson interferometer, as they are used in ground based GW observatories, such as LIGO, Virgo and Kagra\cite{2}.

In addition to these sensitive channels, it has been shown that so-called null channels can be constructed, which strongly suppress the GW signals at low frequencies and might therefore be used for characterizing the instrumental noise \cite{1,4}. An example is the null channel $\zeta$, which represents a symmetric measurement across all three arms of the constellation, strongly suppressing its response to GWs at low frequencies \cite{5}. It is shown in \cite{5} that all possible null channels can be derived from the $\zeta$ channel. Thus, it is sufficient to exclusively focus on $\zeta$ to study the relationship between TDI channels sensitive to GWs and null channels. The results obtained are then generally applicable to any null channel.

We notice that, compared to ground-based GW observatories, a null channel is particularly valuable in LISA. While ground-based observatories can exploit correlations between multiple detectors to discriminate the GWs from instrumental noise sources, LISA will be a single detector, such that the same kind of analysis might not be possible\cite{7}. Furthermore, as mentioned above, LISA is expected to be signal dominated, such that instrumental noise cannot be measured and characterized in flight in isolation from the GW signals.

Current approaches to the data analysis for LISA foresee a "global fit", in which an initial noise model\cite{6} will be defined to start subtracting resolvable sources in an iterative procedure \cite{7}. While the background residual noise model is refined in this iterative process, improving the identification of the known sources, the residual noise model does not necessarily distinguish between instrumental and gravitational noise. As our a priori knowledge of the instrumental noise background is likely to have limited accuracy, this poses a fundamental problem in the identification of a stochastic GW background. The goal of this paper is to analyze the extent to which the null

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\footnote{Technically, LIGO, Virgo and Kagra are Michelson interferometers with Fabry-Perot cavities.}

\footnote{In principle, LISA is not the only planned space-based GW detector targeting mHz frequencies, such that there is a possibility that multiple detectors will be operational at the same time \cite{5}. However, LISA is more advanced in terms of its development compared to other proposed projects, such that this possibility cannot be relied upon. Indeed, LISA has been selected in 2017 to be ESA’s third large-class mission \cite{1}.}

\footnote{A detailed noise model is also essential for the development of the mission, and is already in preparation and anchors the mission hardware requirements.}
channel can be used to characterize the dominant noise sources expected to affect the sensitive channels. We then further explore what this implies for the detectability of an isotropic SGWB of unknown spectral shape.

Following the LISA proposal [11], we consider two general groups of instrumental noise sources: the test mass (TM) acceleration noise and the effective total displacement noise in a one-way single link TM to TM measurement, which we abbreviate as Optical Metrology System (OMS) noise. As we will discuss in section II, each of these two groups of noise in reality represents a multitude of individual noise contributions driven by different physical effects, both known and possible unknown, such that the exact level and frequency dependence of these LISA instrumental noises cannot be reliably calculated a priori. As an example of the difficulty in accurately modeling noise a priori, the acceleration noise at 0.1 mHz measured by LISA Pathfinder (LPF), though compatible with the LISA noise requirements, exceeds by approximately a factor 4 the noise accounted for by the noise model [8].

Various key parameters of the LISA noise, including DC residual forces [9], magnetic field gradients [10], residual stray electrostatic fields [11], optical alignments [12, 13], among others, are all designed to be ideally zero, but with uncertainties that make their residual contribution to the observatory noise both difficult to predict and likely different among the different LISA TM or optical readouts. Other well known noise sources like Brownian noise from gas damping can have a non-trivial time dependence and thus an instantaneous noise Power Spectral Density (PSD) that is hard to predict [14]. As such, if noise knowledge is a key factor in extracting LISA science, either for a stochastic background or for noise priors on individual source parameters, then developing in situ techniques to quantify the instrument noise is an important task.

TM acceleration noise is expected to limit the GW sensitive channels, such as the Michelson X, at low frequencies (below a few mHz), while the OMS noise is most relevant for the sensitivity of these channels at high frequencies. On the contrary, we will show that TM acceleration noise is effectively suppressed in the $\zeta$ channel, such that it is dominated by OMS noise at all frequencies. This behaviour of the null channels has already been pointed out in [16, 17] for the null channel T that is built out of X and the two channels Y and Z (obtained from X by cyclic satellite permutations), considering the case of LISA with equal arm-lengths. However, as already shown in [18] and [4], T as a null channel is strongly compromised when the arm lengths are not exactly equal, especially at low frequencies. The null channel $\zeta$ remains

less sensitive to GW also in the more general unequal arm-lengths scenario. Therefore in the rest of the paper we will discuss only the properties of $\zeta$, and we will simplify the formulas presented to the equal arm-length approximation, which has negligible impact on the general conclusions. We will re-introduce the inequality of the arm-lengths when necessary to not bias the computations, and also perform time-domain simulations using realistic orbits to show our equal-arm-length models are accurate enough. We discuss in sections [II] and [IV] the impact of our findings, showing that the dominant OMS noise in the null channel strongly limits its effectiveness for noise characterization in the low frequency regime. For instance, for a null channel measurement to detect the TM acceleration noise relevant to the Michelson X combination at 0.1 mHz, the OMS noise would have to be at the 0.1 nm/$\sqrt{\text{Hz}}$ level at 0.1 mHz compared to the requirement of 6nm/$\sqrt{\text{Hz}}$ defined in the proposal [11], as shown in fig. 3. Such a low noise level is neither foreseen nor required for LISA GW observation in the TDI Michelson X, Y and Z channels (or the equivalent orthogonal combinations of these, A and E [19]).

The remaining article is divided in four main sections. In section [II] we introduce the noise models defined in the LISA proposal and discuss in detail the TDI outputs. In section [III] we discuss the use of these null channels to calibrate and measure the instrumental noise during operations and the implication for distinguishing between instrumental noise and SGWB. In section [IV] we compare our analytical calculations with time domain simulations using the tools LISA Instrument, LISA GW-Response and PyTDI, which we configured to reflect the noise models given in the LISA proposal [20] and our semi-analytical GW response computation. We also study the response of the significantly less sensitive (compared to X) Sagnac channel $\alpha$, and we explore to which extent the noise entering in $\alpha$ could be combined with the null channel $\zeta$. Finally, we use $\zeta$ to compute an upper limit on the instrumental noise in X, which allows us to identify a strong SGWB in the X data stream.

In the last section we report our conclusion and future perspective.
II. INSTRUMENTAL NOISE MODELING AND TDI OUTPUTS

In this section, we briefly introduce the main limiting noise sources left after TDI suppression of laser frequency noise, and any possible further subtraction of any known calibrated and measured instrumental noise sources, such as, for example, the optical tilt to length cross-coupling to spacecraft motion \[^{[13]}\] and \[^{[12]}\].

The remaining noises, for which we have neither a measurement for coherent subtraction nor a high precision a priori model, falls into two broad categories \[^{[1]}\], the acceleration noise of each individual TM and an overall optical metrology noise term for each single link measurement. For the GW sensitive TDI channels, the former is expected to be the limiting noise source at low frequencies, while the latter is most relevant at high frequencies. We then compute how these noise sources propagate through different TDI channels, and discuss to which extent and at which frequencies the null channel can be a useful noise monitor for the GW sensitive channels.

We will express the phase outputs measured by LISA, used to build the TDI channels, as an effective displacement signal in units of meters.

A. TM acceleration noise

In the assumption of a perfect spacecraft jitter subtraction, we can ignore the complications of the split interferometry scheme \[^{[1]}\] and assume for the calculation of the TDI outputs that the two optical-benches, say OB\(_i\) and OB\(_j\), were two free-falling particles that accelerate along their relative line of sight towards each other with accelerations \(g_i\) and \(g_j\) respectively. \(g_i\) and \(g_j\) describe the TMs acceleration noise with respect to the local inertial frame, that for LISA, can be associated with the one defined by the incoming laser beam (See appendix \[^{[3]}\]).

We will denote the overall TM acceleration noise PSD of a single TM as \(S_{g_{ij}}\). We assume for simplicity that all TM acceleration noises are fully uncorrelated to each other, although this might not be the case in reality\(^{[1]}\). Furthermore, in our assumptions of free-falling optical benches, we can directly convert the acceleration noise of a single TM to an equivalent displacement of the correspondent optical bench, whose PSD is given as

\[
S_{g_{ij}}^{disp} = S_{g_{ij}} / (2\pi f)^4.
\]  

We will denote the time series associated with this displacement as \(x_{g_{ij}}(t)\).

Note that the exact noise shape and amplitude of each individual \(S_{g_{ij}}\) will result from the superposition of a multitude of physical effects. While many of these effects have been characterized during the very successful LPF mission, the total measured noise is considerably larger than the sum of these known sources, indicating the difficulty in achieving a complete, accurate model \[^{[8]}\] \[^{[21]}\]. We therefore cannot assume to have accurate prior knowledge of the overall acceleration noise, and would need to rely on in-flight measurements to constrain its value for each TM.

B. Optical Metrology System noise

We summarize as OMS noise any imperfection in the ability of the OMS to determine the separation between two TMs in a single link.

Similar to the overall acceleration noise acting on each TM, the overall OMS noise affecting a single link will be a superposition of many physical effects. In addition, the overall OMS noise summarizes noise entering due to different instrumental subsystems, such as the telescope, optical bench, phase measurement system, laser, clock and TDI processing \[^{[1]}\]. Note that some of these noise sources can again be correlated (similar to the TM acceleration noise), while here we assume they are not.

For simplicity, we consider only a single uncorrelated noise term in each single link measurement, whose PSD we denote by \(S_{oms}(f)\). We will denote the time series of these single link OMS as \(x_{oms}^{ij}(t)\).

We remark that while the TM acceleration was measured by LPF in realistic flight conditions, we expect new challenges and additional uncertainty in the pre-flight characterization of the OMS. While many terms in the OMS budget are well calculated from models and ground testing (such as shot noise and phasemeter noise), the end-to-end inter-spacecraft LISA optical measurement has never been performed and we can expect unknowns. This is especially true for the low frequency regime, as there will likely be no possible on-ground long term testing before flight (> 1 month). We also cannot assume this noise to be stationary over the mission duration, as we must expect that some of the physical parameters governing its level will change during the mission duration, such as tilt-to-length (TTL) effects \[^{[13]}\].

C. Analytical calculation of noise couplings into TDI \(X, \alpha\) and \(\zeta\)

In this section we illustrate the transfer of TM acceleration and OMS noises into the relevant TDI variables with simplified expressions valid in the low frequency limit (for angular frequencies \(\omega \ll 1/\tau\)). The full time dependence will be used in the calculations that follow in section \[^{[IV]}\].

We derive these noise couplings in the assumption that the light propagation time across all three LISA arms is equal to the same value \(\tau \approx 8.3\) s. As discussed in appendix \[^{[3]}\], the two noise sources we consider enter into a

\[^{[5]}\] For example, both TM inside one spacecraft might be affected by common-mode effects such as temperature fluctuations or tilt-to-length couplings due to rotation of the spacecraft.
single link as

$$\eta_{ij}(t) = x_{ij}^g(t - \tau) + x_{ij}^q(t) + x_{ij}^m(t).$$  \hspace{1cm} (2)$$

Here, $\eta_{ij}(t)$ represents the so-called intermediary TDI variables representing the single link TM to TM measurement. The first index $i$ represents the spacecraft the measurement is performed on at time $t$, while the second index $j$ denotes the distant spacecraft light was emitted from at time $t - \tau$.

From these measurements it is possible to build the TDI channels $X, \alpha$ and $\zeta$ as shown in table I. The table makes use of the the time shift operators which act on time dependent functions by evaluating them at another time, see appendix A.

As a preliminary analysis of the usefulness of $\zeta$ for noise characterization, it is instructive to consider the expression for $\zeta$ in table I in the equal-arm-length limit, where it simplifies to

$$\zeta = (1 - D) (\eta_{12} - \eta_{13} + \eta_{23} - \eta_{21} + \eta_{31} - \eta_{32}),$$  \hspace{1cm} (3)$$

with $D$ as the equal-arm delay operator.

We observe that $\zeta$ is insensitive to noise which is correlated such that it enters both of the two single-link measurements recorded on-board a single spacecraft in exactly the same way. More generally, noise entering correlated (but not exactly equal) in the two measurements, such as noise in the measurements $\eta_{12}$ and $\eta_{13}$ in Eq. [4] will be suppressed, while measurement noise entering anti-correlated will be amplified with respect to the uncorrelated case.

Considering the expressions for $\alpha$ and $X$ in table I in the equal-arm-length limit, we see that this is not the case for these channels, where the spacecraft links enter asymmetrically, and equal noise terms do not cancel in the same way. This means $\zeta$ cannot be used to characterize noise with these correlation properties. Furthermore, as we will discuss in the following, noise entering correlated in the two directions of a link (such as the TM acceleration noise) will also be suppressed in $\zeta$ with respect to noise which is fully uncorrelated in each link.

1. Analytical computation of the acceleration noise for the TDI $X, \alpha$ and $\zeta$

Assuming equal arm lengths, we find that the TM acceleration noise for the combinations $X, \alpha$ and $\zeta$ can be approximated as:

$$X_g(t) \approx 16\tau^2 \left( x_{12}^g(t) - x_{13}^g(t) + x_{21}^g(t) - x_{31}^g(t) \right),$$ \hspace{1cm} (4a)$$

$$\alpha_g(t) \approx 3\tau \left( 3x_{12}^g(t) - 3x_{13}^g(t) + x_{21}^g(t) + x_{23}^g(t) - x_{31}^g(t) - x_{32}^g(t) \right),$$ \hspace{1cm} (4b)$$

$$\zeta_g(t) \approx \tau \left( x_{12}^g(t) - x_{13}^g(t) - x_{21}^g(t) + x_{23}^g(t) + x_{31}^g(t) - x_{32}^g(t) \right).$$ \hspace{1cm} (4c)$$

where we have expanded to leading order in the average light travel time $\tau$. This expansion is only valid at timescales much greater than $\tau \approx 8.3$ s.

They allow us to see immediately which TMs dominate the noise for each of the TDI combinations. The full expressions without expansion can be found in appendix C.

Under these assumptions, we can see that for the TM acceleration noise, the TDI combination $\zeta$ measures a signal that is a combination of all the 6 TMs. Similarly, $\alpha$ also measures a combination of all the six TMs but with different coefficients. The Michelson $X$ measures instead a combination of only four TMs.

2. Analytical computation of the metrology noise for the TDI $X, \alpha$ and $\zeta$

Following the same steps as in the previous section we find that the propagation of the OMS noise through the different TDI variables is:

$$X_{oms}(t) \approx 8\tau^2 \left( x_{12}^{m,\prime}(t) - x_{13}^{m,\prime}(t) + x_{21}^{m,\prime}(t) - x_{31}^{m,\prime}(t) \right),$$ \hspace{1cm} (5a)$$

$$\alpha_{oms}(t) \approx 3\tau \left( x_{12}^{m,\prime}(t) - x_{13}^{m,\prime}(t) - x_{21}^{m,\prime}(t) + x_{23}^{m,\prime}(t) + x_{31}^{m,\prime}(t) - x_{32}^{m,\prime}(t) \right),$$ \hspace{1cm} (5b)$$

$$\zeta_{oms}(t) \approx \tau \left( x_{12}^{m,\prime}(t) - x_{13}^{m,\prime}(t) - x_{21}^{m,\prime}(t) + x_{23}^{m,\prime}(t) + x_{31}^{m,\prime}(t) - x_{32}^{m,\prime}(t) \right).$$ \hspace{1cm} (5c)$$

Here, we expanded again to leading order in the average light travel time $\tau$ to see what the contributions of the OMS noise for each of the TDI combinations are at low frequency. This expansion is therefore only valid at timescales much greater than $\tau \approx 8.3$ s. The full expressions for the OMS noise of the aforementioned combinations without expansion can again be found in appendix C.

We see that for the OMS noise, at low frequency,
\( \alpha_{\text{oms}}(\tau) \approx 3 \zeta_{\text{oms}}(\tau) \). Furthermore, we observe that the OMS noise enters \( \alpha \) and \( \zeta \) only as a first derivative, while the TM acceleration noise entered as a second derivative. This reflects a low frequency suppression of TM acceleration noise relative to OMS noise terms, due to the difference between the two single link measurements (see eq. [1] and table I) along the same arm, which contain the same TM acceleration noise terms but with different delays. For the TDI X, on the other hand, both OMS noise and TM acceleration noise enter as a second derivative (compare eq. (4a) vs. eq. (5a)).

\[ \begin{align*}
\alpha & = \frac{1}{1 - D\alpha_{D2}D_{21} + (D_{12} - D\alpha_{D2}D_{21})\eta_{12} + (D_{12} - D\alpha_{D2}D_{21})\eta_{21} + (D_{12} - D\alpha_{D2}D_{21})\eta_{11} - (D_{12} - D\alpha_{D2}D_{21})\eta_{11}} \\
\zeta & = \frac{1}{1 - D\alpha_{D2}A_{21} - (D_{12} - D\alpha_{D2}A_{21})\eta_{12} + (D_{12} - D\alpha_{D2}A_{21})\eta_{21} + (D_{12} - D\alpha_{D2}A_{21})\eta_{11} - (D_{12} - D\alpha_{D2}A_{21})\eta_{11}} \\
X & = \frac{1}{1 - D\alpha_{D2}A_{21} - (D_{12} - D\alpha_{D2}A_{21})\eta_{12} + (D_{12} - D\alpha_{D2}A_{21})\eta_{21} + (D_{12} - D\alpha_{D2}A_{21})\eta_{11} - (D_{12} - D\alpha_{D2}A_{21})\eta_{11}}
\end{align*} \]

Table I. List of the TDI \( \alpha, \zeta \) and X as given in [1], expressed in terms of time shifts applied to the intermediary TDI variables \( \eta_{ij} \). The table has been adapted from [22].

\[ \begin{align*}
S_{\text{oms}}^{\text{noise}} & \approx 64\pi^4 \omega^4 \left( 4 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right), \\
S_{\zeta}^{\text{noise}} & \approx \tau^2 \omega^2 \left( \tau^2 \omega^2 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right).
\end{align*} \]

III. PRELIMINARY DISCUSSION

As known from the literature [2], and also shown in fig. 1 the Michelson X channel is sensitive to GWs. One of the expected GW sources for LISA is the SGWB, which in principle could be observed across the whole frequency band [24]. Such a SGWB will be superimposed with the instrumental noise entering in the X channel, such that we should measure an excess in noise power with respect to the real instrumental noise in order to detect a SGWB. However, as discussed in section I we cannot rely on noise modelling and on-ground testing to fully characterize the instrumental noise, such that we would need to measure it in-flight.

One option for measuring the instrumental noise would be to consider the output of a null channel like \( \zeta \) which, at least at low frequencies, is insensitive to GWs [24]. Figure 1 shows the sensitivities for the TDI X, \( \alpha \) and \( \zeta \), computed as described in appendix D. For X and \( \alpha \), we find the sensitivity to be unaffected by an armlength mismatch, while \( \zeta \) becomes slightly more sensitive to GWs when considering three unequal constant arms instead of three equal constant arms. Nevertheless, in both cases \( \zeta \) remains less sensitive than X by many orders of magnitude, such that we will consider the simpler equal arm-length case for computing the noises and GW response of the TDI variables in the following.

We can do a preliminary calculation by computing the total noise PSDs for TDI X and \( \zeta \), which we denote as \( S_{\alpha}^{\text{noise}} \) and \( S_{\zeta}^{\text{noise}} \). We compute them as the linear sum of the OMS and TM acceleration noises, respectively, using the low-frequency expansions given in eqs. (4) and (5).

\[ \begin{align*}
S_{\alpha}^{\text{noise}} & \approx 64\pi^4 \omega^4 \left( 4 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right), \\
S_{\zeta}^{\text{noise}} & \approx \tau^2 \omega^2 \left( \tau^2 \omega^2 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right).
\end{align*} \]

\[ \begin{align*}
S_{\alpha}^{\text{noise}} & \approx 64\pi^4 \omega^4 \left( 4 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right), \\
S_{\zeta}^{\text{noise}} & \approx \tau^2 \omega^2 \left( \tau^2 \omega^2 \sum_{ij \in T\!\!\!\!X} S_{gij}^{\text{disp}} + \sum_{ij \in T\!\!\!\!X} S_{\text{oms},ij} \right).
\end{align*} \]
as the overall noise entering in the two channels, valid for \( \omega \tau \ll 1 \). Here, we introduced the index sets \( I_x = \{12, 21, 13, 31\} \) and \( I_\zeta = \{12, 23, 31, 13, 32, 21\} \) for the four and six optical links (received at spacecraft \( i \) from spacecraft \( j \)) and TM acceleration noise terms (TM in spacecraft \( i \) accelerated towards spacecraft \( j \)) appearing in \( X \) and \( \zeta \), respectively.

We can observe that a TM displacement due to TM acceleration noise and the OMS noise enter with almost the same transfer function into the \( X \) channel, up to an additional factor 4 in the TM displacement. Conversely, in \( \zeta \) the TM acceleration noise is suppressed towards low frequencies by a factor \( \tau^2 \omega^2 \) relative to the OMS noise. This implies that while TM acceleration noise becomes dominant in \( X \) for frequencies in which (on average) \( S_{\text{oms}} \ll 4 S_{\text{disp}} \), for \( \zeta \) the same holds only if \( S_{\text{oms}} \ll \tau^2 \omega^2 S_{\text{disp}} \). Considering frequencies in the range \( 10^{-3} \text{ Hz} \) to \( 10^{-3} \text{ Hz} \), the TM acceleration noise pre-factor \( \tau^2 \omega^2 \) (cf. eq. (7)) is between \( 2.7 \times 10^{-5} \) and \( 2.7 \times 10^{-3} \).

This means that the OMS noise would have to be from ten parts in a million to one part in a thousand smaller in power than the TM acceleration noise in order for the latter to have the same order of magnitude as the OMS noise in the \( \zeta \) channel in the sub-mHz band.

As such, the null channels ability to monitor noise in the GW sensitive channels at low frequencies is limited. \( \zeta \) could only be used to reliably detect the relevant sub-mHz noise in a worst case scenario where the TM acceleration noise is orders of magnitude larger than the OMS noise in these frequency ranges, such that it overcomes the scaling factor \( \tau^2 \omega^2 \) and becomes dominant in both \( \zeta \) and \( X \).

As we will discuss in the next section, the currently assumed requirements for TM acceleration and OMS noises are very far away from these values. Nevertheless, we can still formulate upper and lower bounds on a SGWB signal based on \( X \) and \( \zeta \) for the full LISA frequency band.

## IV. UPPER LIMITS, EXPECTED NOISE LEVELS AND SIMULATIONS

After the preliminary analysis in section III, let us now drop the low-frequency approximation and discuss the accuracy to which we can use \( X \), \( \alpha \) and \( \zeta \) to identify a potential SGWB with LISA.

To this end, we briefly introduce the currently assumed noise levels given in the literature [24]. Note that these should be thought of as the performance requirements we aim to reach with as much margin as possible, not as accurate predictions of the actual in-flight performance. We also perform time domains simulations using LISA Instrument [25] and pyTDI [26] to test our expressions for how these noises couple into the different TDI variables. Similarly, we also perform time domain simulations to test our semi-analytical computation of the GW response of different TDI variables presented in appendix C using the tool GW-response [27]. Using simulations allows us to compare our (semi)-analytical expressions, computed assuming equal arm-lengths, with data generated using realistic LISA orbits provided by ESA.

### A. Analytical model and simulations

#### 1. Instrumental Noise

Considering the analytical computation in time domain of the TM acceleration and OMS noises for the TDI \( X \), \( \alpha \), \( \zeta \) in appendix C, we can estimate the PSD of the aforementioned TDI combinations assuming all TM acceleration and OMS noises to be uncorrelated, which yields

\[
S_{X_\varphi} = 256 \sin^4(\tau \omega) \cos^2(\tau \omega) \left( (S_{g12}^{\text{disp}} + S_{g13}^{\text{disp}}) \cos^2(\tau \omega) + S_{g21}^{\text{disp}} + S_{g31}^{\text{disp}} \right) ,
\]

(8a)

\[
S_{\alpha_\varphi} = 16 \sin^2 \left( \frac{\tau \omega}{2} \right) \sin^2 \left( \frac{3 \tau \omega}{2} \right) \left( (1 + 2 \cos(\tau \omega))^2 (S_{g12}^{\text{disp}} + S_{g13}^{\text{disp}}) + S_{g21}^{\text{disp}} + S_{g23}^{\text{disp}} + S_{g31}^{\text{disp}} + S_{g32}^{\text{disp}} \right) ,
\]

(8b)

\[
S_{\zeta_\varphi} = 16 \sin^4 \left( \frac{\tau \omega}{2} \right) (S_{g12}^{\text{disp}} + S_{g13}^{\text{disp}} + S_{g21}^{\text{disp}} + S_{g23}^{\text{disp}} + S_{g31}^{\text{disp}} + S_{g32}^{\text{disp}} ) ,
\]

(8c)
FIG. 2. Upper plots: the left one shows the PSD of the TDI X and the right one the PSD of $\zeta$ for the TM acceleration noise, the optical metrology noise and the total noise as simulated with LISA Instrument compared with the respective analytical models. Lower plot: response to GW for TDI X and $\zeta$ as simulated with LISA Instrument and LISA GW-Response compared with the semi-analytical models computed considering equal arm length for X and both equal and three fixed unequal arm length for $\zeta$ as described in appendix [17].

and

\begin{align}
S_{X_{oms}} &= 64 \sin^4(\tau \omega) \cos^2(\tau \omega) \left( S_{oms_{12}} + S_{oms_{13}} + S_{oms_{21}} + S_{oms_{31}} \right), \\
S_{\alpha_{oms}} &= 4 \sin^2 \left( \frac{3 \tau \omega}{2} \right) \left( S_{oms_{12}} + S_{oms_{13}} + S_{oms_{21}} + S_{oms_{23}} + S_{oms_{31}} + S_{oms_{32}} \right), \\
S_{\zeta_{oms}} &= 4 \sin^2 \left( \frac{\tau \omega}{2} \right) \left( S_{oms_{12}} + S_{oms_{13}} + S_{oms_{21}} + S_{oms_{23}} + S_{oms_{31}} + S_{oms_{32}} \right).
\end{align}

We verify the validity of these equations (derived in the equal-arm limit) using time domain simulations with realistic orbits. We disabled all noise sources available in LISA Instrument except TM acceleration noise and OMS noise in the inter-spacecraft interferometer, and set all noises of the same type to the same level, as given in [24].

For the TM acceleration noises, this means a value of

\begin{equation}
S_{g_{ij}}(f) = \left( 3 \times 10^{-15} \frac{m}{s^2 \sqrt{Hz}} \right)^2 \times \left( 1 + \left( \frac{0.4 \text{ mHz}}{f} \right)^2 \right) \left( 1 + \left( \frac{f}{8 \text{ mHz}} \right)^4 \right),
\end{equation}
which translates to
\[ S_{\text{disp}}(f) = S_{g_{ij}}(f)/(2\pi f)^4 \] (11)
in terms of displacement.

The noise level of the OMS is instead given as
\[ S_{\text{oms}_{ij}}(f) = \left(15 \text{ pm}/\sqrt{\text{Hz}}\right)^2 \times \left(1 + \left(\frac{2 \text{ mHz}}{f}\right)^4\right), \] (12)
where the factor \(1 + \left(\frac{2 \text{ mHz}}{f}\right)^4\) is a low frequency relaxation term introduced to take into account our difficulties in measuring that noise below a few mHz from on-ground laboratory experiments. This relaxation is further justified by the fact that it has no impact on the low frequency GW sensitivity in X, as OMS noise remains very subdominant in X compared to TM acceleration noise even when including this factor, as visible in the left plot in fig. 2.

Note that the estimated OMS noise model for LPF also includes a low-frequency relaxation to account for possible thermally driven effects [28]. However, these were likely buried in the LPF noise at lower frequencies where TM acceleration noise is believed to dominate. The upper part of fig. 2 shows the results of three simulation runs with LISA Instrument where we enable either one, the other or both of these noise sources. We use PyTDI to compute the Michelson X and \(\zeta\) variables.

First, we note that in all cases, the simulated data, with realistic, unequal arm orbits, agrees well with the simplified equal-arm analytic expressions derived for the noise. We see that in the \(\zeta\) channels the OMS noise is dominant over the TM acceleration noise at all frequency, while TM acceleration noise becomes the dominant noise source for X below a few mHz.

Moreover, if we assume all noises of the same type to have the same noise level, we can use eqs. (10c) and (11c) to compute that for \(\zeta\) we would need an OMS noise level of \(S_{\text{oms}} = 24 \sin^2\left(\frac{\tau}{2}\right) \times 4 \sin^2\left(\frac{\tau}{2}\right) S_{\text{disp}}^g\) such that OMS and TM acceleration noises appear at the same magnitude. This can be translated in the single link OMS noise contribution with the value of \(4 \sin^2\left(\frac{\tau}{2}\right) S_{\text{disp}}^g\), which we compare in fig. 3 to the requirement for the OMS noise given in eq. (12).

We observe that this noise level is likely impossible to achieve as the new required level of OMS noise is 160 pm/\(\sqrt{\text{Hz}}\) at 0.1 mHz, orders of magnitude below the currently assumed value. It must be also kept in mind that this conclusion is true keeping fixed the TM acceleration noise level to the nominal value, while drastically lowering the OMS noise level. Any improvement of the TM acceleration noise in LISA would make the upper limit achieved by the null channel even less relevant. However, fig. 2 shows that, at least assuming nominal noise levels, both X and \(\zeta\) are dominated by OMS noise above 4 mHz, which might suggest that \(\zeta\) can put a stronger constraint on the instrumental noise in this frequency range.

2. Gravitational wave response

We denote the PSD of the X and \(\zeta\) channels due to GWs as
\[ S_{X_h} = T_{X_h} S_h, \quad S_{\zeta_h} = T_{\zeta_h} S_h. \] (13)

\(S_h\) is expressed as a dimensionless stochastic GW strain in Hz\(^{-1}\), while \(S_{X_h}\) and \(S_{\zeta_h}\) are expressed in m\(^2\)/Hz. Therefore, the response functions \(T_{X_h}\) and \(T_{\zeta_h}\) each include a conversion factor that has units m\(^2\)/Hz.

The lower plot in fig. 2 shows the GW responses to a SGWB of TDI X and \(\zeta\) for 51 stochastic GW sources isotropically distributed over the sky, computed in the frequency domain as described in appendix D. To verify the validity of these equal arm-length models, we compare them to time domain simulations using the tools LISA Instrument, GW-response and PyTDI which use realistic ESA orbits. We inject an isotropic SGWB computed from \(N = 48\) sources into the time-domain simulation and disable all instrumental noises. The two strain time series \(h_{X,}\) \(h_{\zeta}\) for each source are computed as a white noise of amplitude \(1/N\), where \(N\) is the number of GW sources, to overall simulate a sky-averaged response to a unit amplitude SGWB.

We see that for X our model for equal arms agrees with the simulations, while the equal arm-length model for \(\zeta\) diverges from the simulations for frequencies smaller than 60 mHz. Considering three un-equal but constant arms for our semi-analytical response calculation for \(\zeta\) extends the validity of the model to almost the entire LISA required frequency range, while we still see a divergence.

9 The GW-response tool used to compute the time domain response only allows certain fixed numbers of stochastic sources. \(N = 48\) is the closest valid value to what we used in the Fourier domain computation.
between simulations and the model at very low frequencies below 0.3 mHz. A preliminary study indicates that the mismatch is probably linked to the fact that we neglect the Sagnac effect in our model, i.e., that we assume the delays across the two directions of each arm to be equal.

However, as the mismatch mostly occurs outside the required LISA frequency band and the response of $\zeta$ remains sufficiently small compared to that of $X$ inside the LISA band down to 0.1 mHz this does not significantly impact our conclusions.

### B. Combining Sagnac channels

First, let us consider the apparent possibility to use $\zeta$ to characterize and subtract the noise in $\alpha$. As discussed in section II.C.2 the OMS noise contributions in $\alpha$ and $\zeta$ fulfill $\alpha_{oms}(\tau) \approx 3\zeta_{oms}(\tau)$ at low frequencies. This implies that subtracting $3\zeta$ from $\alpha$ allows you to remove the common OMS noise. In fig. 4 we report the simulations of the TM acceleration noise plus OMS noise for $\alpha - 3\zeta$ compared to the respective analytical models, which confirms what was predicted by the analytical calculation. What is then left as dominant noise source at low frequencies is a combination of the following four TMs:

$$\left[\alpha_{oms,g} - 3\zeta_{oms,g}\right](t) \approx 6T^2\left(x_{12}''(t) - x_{13}''(t) + x_{21}''(t) - x_{31}''(t)\right).$$  

(14)

We notice that eq. (14) is equal, up to a constant factor, to the low frequency TM acceleration noise of the TDI $X$ channel (see eq. (4a)). This means that using the null channel $\zeta$ to reduce the noise, with the purpose of retrieving the GW signal in $\alpha$, gives you back the channel $X$, which is sensitive to GWs as well as to the TM acceleration noise. We therefore focus on just the $X$ and $\zeta$ channels in the following.

### C. Upper limit on instrumental noise in $X$

Let us now consider the expression for the OMS noise and TM acceleration noise for TDI $\zeta$ and $X$ without considering the presence of GW signals in our data. To put an upper limit on the instrumental noise in $X$ we are looking for a frequency dependent factor $F$ such that $FS_\zeta \geq S_X$, which implies

$$F(S_{\zeta_{oms}} + S_{\zeta_{disp}}) \geq S_{X_{oms}} + S_{X_{disp}}.$$  

(15)

Referring to eqs. (8a), (8c), (9a) and (9c), this means

$$FT_{\zeta_{oms}} \sum_{ij \in I_\zeta} S_{oms_{ij}} + FT_{\zeta_{disp}} \sum_{ij \in I_\zeta} S_{g_{ij}} \geq T_{X_{oms}} \sum_{ij \in I_X} S_{oms_{ij}} + T_{X_{disp}} \left(\left(S_{g_{12}}^{disp} + S_{g_{13}}^{disp}\right)\cos^2(\tau\omega) + S_{g_{21}}^{disp} + S_{g_{31}}^{disp}\right).$$  

(16)

Since $I_X \subset I_\zeta$ and $S_{oms_{ij}}$ and $S_{g_{ij}}$ are strictly positive, we have

$$\sum_{ij \in I_X} S_{oms_{ij}} \geq \sum_{ij \in I_\zeta} S_{oms_{ij}}.$$  

(17)

and further considering that $\cos^2(\tau\omega) \leq 1$ we get

$$\sum_{ij \in I_X} S_{g_{ij}}^{disp} \geq \left(\left(S_{g_{12}}^{disp} + S_{g_{13}}^{disp}\right)\cos^2(\tau\omega) + S_{g_{21}}^{disp} + S_{g_{31}}^{disp}\right).$$  

(18)

As a remark, instead of utilizing $\zeta$ to remove the excess OMS noise in $\alpha$, one could also construct the optimal channels $A$ and $E$ out of the Sagnac variables, in which the dominant OMS noise terms also cancel. This follows readily from the result stated in footnote 7 that the Sagnac channels fulfill $\alpha \approx \beta \approx \gamma$ for low-frequency OMS noise. Thus, OMS noise is cancelled to first order in both $A \approx \alpha - \gamma$ and $E \approx \alpha - 2\beta + \gamma$, giving these channels the same sensitivity as their Michelson-equivalents.
We can therefore define our noise estimate factor as

\[ F \geq T_{X_{\text{oms}}}/T_{\zeta_{\text{oms}}} \quad \text{and} \quad F \geq T_{X_{g}}/T_{\zeta_{g}}. \]

We can therefore define our noise estimate factor as

\[ F = \max(T_{X_{\text{oms}}}/T_{\zeta_{\text{oms}}}, T_{X_{g}}/T_{\zeta_{g}}) = 256 \cos^4 \left( \frac{\omega T}{2} \right) \cos^2(\omega T). \]

Note that by inspection of eqs. (8a), (8c), (9a) and (9c), we find the ratio \( T_{X_{\text{oms}}}/T_{\zeta_{\text{oms}}} \) to be dominant at all frequencies, which allows us evaluate the maximum in the previous equation.

We show in fig. 5 in the left plot, the overall noise upper limit \( F_{\text{noise}} \) obtained in this way next to the actual noise \( S_{X_{\text{noise}}} \) in \( X \). In addition, we show the two individual upper limits we would obtain for just the OMS noise and just the TM acceleration noise by considering only the contribution of \( T_{X_{\text{oms}}}/T_{\zeta_{\text{oms}}} S_{\zeta_{\text{oms}}} \) and \( T_{X_{g}}/T_{\zeta_{g}} S_{\zeta_{g}} \), respectively.

Inspecting the right plot of fig. 5, we observe that (assuming noise at the required levels) the upper limit on the instrumental noise in \( X \) posed by \( \zeta \) is up to a factor 50 in amplitude above the actual noise level, in particular at low frequencies. This results in a rather weak upper limit, reflecting OMS noise in a frequency band where only TM acceleration noise is relevant. At high frequencies, on the other hand, where both \( X \) and \( \zeta \) are dominated by OMS noise, the estimate is significantly more stringent, and stays below a factor 2 in amplitude from 25 to 100 mHz.

We want to underline that the derivation of the expression \( F_{\text{noise}} \) for the noise upper limit does not rely on any assumptions on the actual noise levels of the individual TM and OMS noise terms, as only sums over all TM and OMS channels affects eq. (16). Additionally, it could be evaluated at any time, and is therefore robust against non-stationarity of the noise. However, this upper limit does rely on our assumptions on noise correlations made in section II and the particular outcome we show in fig. 5 reflects the nominal values assumed for the level of TM acceleration and OMS noise.

D. Upper and lower limits on a SGWB

We now additionally consider the presence of possible SGWBs in our data, on which we can put lower and upper bounds as follows. As before for the instrumental noise, we will remain agnostic to the spectral shape and amplitude of the SGWB. We do however assume to know the response function of the different TDI channels, which we compute as described in appendix I for the case of an isotropic SGWB.

In the presence of such a SGWB we can introduce

\[ S_{X_{\text{meas}}} = S_{X_{\text{noise}}} + S_{X_{h}}, \]

as the combination of instrumental noise and GW signal that we can actually measure in the TDI X channel.

We remind that eq. (13) together with eq. (21) immediately allows us to put an upper bound on a possible SGWBs,

\[ S_{h} \leq S_{X_{\text{meas}}}/T_{X_{h}}, \]

as any model predicting a higher value of \( S_{h} \) would be incompatible with our measurements \( S_{X_{\text{meas}}} \).

To put a lower bound on \( S_{h} \) based on our data (i.e., claim a detection), we can make use of our previously derived upper bound \( F_{\text{noise}} \) on the instrumental noise in \( X \).

To this end, as \( \zeta \) is not perfectly insensitive to GWs (cf. fig. 1), we need to define

\[ S_{\zeta_{\text{meas}}} = S_{\zeta_{\text{noise}}} + S_{\zeta_{h}}. \]

The upper bound (cf. eq. (15)) on the instrumental noise now becomes \( F(S_{\zeta_{\text{meas}}} - S_{\zeta_{h}}) \geq S_{X_{\text{noise}}} \), which allows us to write

\[ S_{X_{\text{meas}}} \leq F(S_{\zeta_{\text{meas}}} - S_{\zeta_{h}}) + S_{X_{h}}, \]

where we simply add \( S_{X_{h}} \) on both sides of the previous inequality and consider the definition of \( S_{X_{\text{meas}}} \). Then, considering eq. (13) this implies the lower bound

\[ S_{h} \geq \frac{S_{X_{\text{meas}}} - F S_{\zeta_{\text{meas}}}}{T_{X_{h}} - FT_{\zeta_{h}}}, \quad \text{valid if } T_{X_{h}} > FT_{\zeta_{h}}. \]

Note that the right-hand side of this equation can be negative even if there is a SGWB, in which case it is compatible with \( S_{h} = 0 \) and we cannot claim a detection of a SGWB. On the other hand, if it is positive, this would indicate presence of a GW background, at least in the assumptions used here.

Assuming that we only consider the frequency range in which eq. (25) is valid, i.e., if \( T_{X_{h}} > FT_{\zeta_{h}} \), the right-hand side of eq. (25) will be positive if \( S_{X_{\text{meas}}} > F S_{\zeta_{\text{meas}}} \), which means we are able to identify a SGWB if

\[ S_{X_{\text{noise}}} + T_{X_{h}} S_{h} > F(S_{\zeta_{\text{meas}}} + T_{\zeta_{h}} S_{h}) \]

\[ \iff S_{h} > \frac{F S_{\zeta_{\text{meas}}} - S_{X_{\text{noise}}}}{T_{X_{h}} - FT_{\zeta_{h}}}. \]

Equation (26b) allows us to define a detection threshold assuming known noise levels, as depicted in the left
FIG. 5. Left: Upper limit on the instrumental noise in X derived from the noise observed in ζ, for the case where all TM acceleration and optical metrology noises are at the levels specified by eqs. (11) and (12), respectively. Moreover, the dotted yellow and blue curves show the estimate one could put on just the OMS or TM noise, respectively. Right: Ratio between actual noise in X and the overall upper limit, in amplitude assuming required levels of TM acceleration and OMS noise.

plot of fig. 6, where it is shown alongside the upper limit defined in eq. (22) in case we don’t have a GW background. Note that both these quantities now apply to the fundamentally unknowable instrumental noise levels, and cannot be evaluated from the raw data. We remind that the scaling factor $F$ used in this derivation was computed in the equal-arm assumption. While we showed in section IV A that the equal-arm noise models are generally valid across most of the LISA frequency band, we expect them to diverge in small frequency bands around the zeros of the TDI transfer function.  

This issue might be circumvented by using a different set of second generation TDI variables which lack zeros at such low frequencies, as for example described in [22].

V. CONCLUSION

The LISA data analysis, particularly in the search for a SGWB, should be as robust as possible to ignorance of the noise model and to variations of the noise from the different components of the instrumental setup.

It will likely be impossible to accurately predict and faithfully model the instrumental noise performance pre-flight, such that efforts to characterize the noise based on in-flight observables should be exploited as much as possible. We present here how one can use the ζ channel to estimate the level at which two of the main noise sources, the uncorrelated TM acceleration and OMS noise, will affect the GW sensitive X channel. This is a rather conservative estimate, in the sense that it assumes nothing about actual instrumental noise levels, homogeneity between different TM acceleration or OMS noise terms and noise stationarity. However, there are potential limits due to our assumptions on noise correlations, as ζ is highly insensitive to correlated noise entering both single-link measurements on-board a single spacecraft, while X is not.

We show that using ζ we estimate the noises under consideration in X within a factor 2 in amplitude in the band from 25 mHz to 100 mHz, while this estimate worsens to within a factor 50 in amplitude at the lowest frequencies (assuming the instrumental noise levels from the requirement). We can use this upper bound on the instrumental noise to compute a lower bound on the GW background needed to explain the overall observed PSDs of both ζ and X. To this end, both the response of X to gravitational waves, $T_{X,h}$, and that of our instrumental noise estimate to gravitational waves, $FT_{ζ,h}$, have to be considered. While $FT_{ζ,h}$ is strongly subdominant to $T_{X,h}$ at low frequencies, this relationship is inverted at high frequencies, such that the lower bound becomes less stringent than one would expect from the performance of the noise estimate alone, and eventually becomes invalid. As visible in the right plot of fig. 6, we have $T_{X,h} > FT_{ζ,h}$ only up to around 50 mHz.

Note that the fact that the noise estimate ζ provides at low frequencies is a factor 50 above the actual instrumental noise in X implies that, within the assumption of this study, we could identify a SGWB only if it were significantly larger than the TM acceleration noise expected to dominate X at these frequencies.

Still, even assuming the nominal instrumental noise levels, this lower bound would allow to detect big stochastic backgrounds in a large part of the frequency band.

---

11 For example, the first zero of the second generation Michelson variable lies at roughly 30 mHz. Assuming the arms of the constellation to be mismatched by 1 percent, the equal arm model is accurate to within 90 percent in a bandwidth of roughly 1 mHz around this zero.

12 Such a strong SGWB could potentially bury the coherent signals associated with most of the LISA science case in gravitational noise.
FIG. 6. Left: Comparison of the detection threshold derived from $\zeta$ (red curve) and the upper limit given by $X$ (blue curve), both expressed as relative arm length change (strain) assuming OMS and TM acceleration noise at the requirements level. Right: GW response functions of TDI $X$, $FS_\zeta$ and $S_\zeta$, computed assuming 51 stochastic GW sources isotropically distributed over the sky.

Given the large uncertainties in the range of possible stochastic background levels [30], including spectral shape and amplitude, as well as the demonstrated difficulty in predicting instrument noise, the results shown here might proof useful. As such, our paper addresses the idea of simultaneous signal plus noise measurement, and shows the limit of achieving this with the TDI null channel.

Note that while our approach is agnostic to the noise levels, the predicted performance is computed assuming OMS noise to be exactly at the required noise levels (which includes a strong low-frequency relaxation), but, especially at low frequency, this noise has high uncertainty. If the actual hardware turns out to perform better in-flight than what can be demonstrated on-ground, the estimate would consequently improve. For example, earlier studies which performed similar estimates (e.g., [31, 32]) assumed the OMS noise to be white across the whole frequency band, and came to the conclusion that we can make a better use of the null channels at low frequencies to estimate the SGWB. We remark that for the OMS noise in $\zeta$ to reach the same level as the TM acceleration noise (limiting the $X$ channels at low frequencies) would require order of magnitude improvements in the performance of the OMS.

We want to reinforce that the upper and lower bounds we compute here are agnostic to the actual instrument performance and don’t rely on any model of the individual noise spectral shapes or stationarities. This is in contrast to some other results in the literature (see for instance [16, 18, 33, 34]), which showed it is possible to put significantly more stringent bounds on the noise assuming stationarity over the whole mission duration and a fixed (and known) noise shape which only depends on a single amplitude parameter. If indeed such a priori knowledge of the noise level and shape were possible, it would be possible to resolve SGWB even below the threshold of the instrument noise.

The results presented here demonstrate the necessity of using realistic assumptions on the prior knowledge of the instrumental noise, noise correlations and stationarity. It is important to consider that the data analysis pipelines in LISA operations will likely rely on some model for the noise (even if Bayesian techniques for parameter’s estimation with unknown noise has been introduced in literature, see for instance [35]). Although procedures like those described in this manuscript do not translate naturally into a Bayesian data analysis framework we believe they might still proof useful to cross-check and interpret the results from a full Bayesian analysis, given the large number of parameters such a procedure has to determine. Additionally, the lower and upper bounds provided from our method could be used as priors in a Bayesian framework.

Further studies should be performed to quantify the real impact this has on achieving the LISA science objectives to detect SGWBs.

To conclude, we reiterate that this study is limited in that we only considered the two main classes of noise, TM acceleration and OMS noise, and that we further assume that these are fully uncorrelated for the six TMs and six one-way optical metrology links. Follow-up studies could investigate other known noise sources with different correlation properties, such as sideband modulation noise [36] or TTL couplings [37], to verify to which extend the results presented here hold for such noises. Furthermore, we only considered here the case of an isotropic SGWB for simplicity. Any anisotropic SGWB, such as the expected foreground from galactic binaries, will have an annual modulation in the response function, which might help to distinguish it better from the instrumental noise.

We want to remark that some instrumental noises might also show annual modulations due to the position of the LISA satellites along the orbit, which one should account for when studying this scenario.
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Appendix A: Time shift operators

We define the following notations related to time-shift operators and TDI combinations [22].

**Delay operator:**
\[ D_{ij} \eta(\tau) = \eta(\tau - d_{ij}(\tau)). \] (A1)
Given a time of reception \( \tau \) of a beam on spacecraft \( i \), evaluates the measurements \( \eta \) (we dropped the double index for simplicity) of that beam at the time of emission at spacecraft \( j \), which we write as \( \tau - d_{ij}(\tau) \). Note that depending on what frame \( \eta(\tau) \) is defined in, the computation of \( d_{ij} \) can include a change in reference frames and clock offsets, as discussed in [23].

**Advancement operator:**
\[ A_{ij} \eta(\tau) = \eta(\tau + a_{ij}(\tau)). \] (A2)
Given a time of emission \( \tau \) of a beam from spacecraft \( j \), evaluates the phase \( \eta \) of that beam at the time of reception on spacecraft \( i \), which we write as \( \tau + a_{ij}(\tau) \). This is the inverse operation to that of the delay operator, such that we have the identity \( A_{ij}D_{ji}\eta(t) = D_{ij}A_{ji}\eta(t) = \eta(t) \).

**Multiple Delay operators:**
\[ D_{ij}D_{jk} \eta(\tau) = \eta(\tau - d_{ij}(\tau) - d_{jk}(\tau - d_{ij}(\tau))). \] (A3)

**Multiple Delay and Advancement operators:**
\[ A_{ni}D_{ij}D_{jk} \eta(\tau) = \eta(\tau + a_{ni}(\tau) - d_{ij}(\tau + a_{ni}(\tau)) - d_{jk}(\tau + a_{ni}(\tau) - d_{ij}(\tau + a_{ni}(\tau))). \] (A4)

Only the delays \( d_{ij}(\tau) \) are directly accessible from the LISA measurements. The advancements \( a_{ij}(\tau) \) can be computed from them by iteratively solving
\[ a_{ij}(\tau) = d_{ij}(\tau + a_{ij}(\tau)), \] (A5)
which directly follows from \( A_{ij}D_{ji}\eta(t) = \eta(t) \).

Appendix B: TM acceleration and displacement noise models

Following the convention that \( \hat{L}_{ji} \) is the link vector from the emitting satellite \( OB_j \) to the receiving one \( OB_i \), and \( \vec{g}_i \) the \( OB_i \) acceleration relative to its inertial reference frame, we can define the acceleration of \( OB_i \) that points towards \( OB_j \), at time \( t \), as:
\[ g_{ij}(t) \equiv \vec{g}_i(t) \cdot \hat{L}_{ji}. \] (B1)

Then, the relative acceleration \( \Delta g_{\text{single-link}}(t) \), between the two free-falling TMs along the line of sight of the unit vector \( \hat{L}_{ji} \), at time \( t \) on \( OB_i \), can be computed as:
\[ \Delta g_{\text{single-link}}(t) = (\vec{g}_i(t) - \vec{g}_j(t - \tau)) \cdot \hat{L}_{ji} \equiv (g_{ij}(t) + g_{ji}(t - \tau)), \] (B2)
where we have use the approximation that:
\[ \vec{g}_i(t) \cdot \hat{L}_{ji} \approx -\vec{g}_i(t) \cdot \hat{L}_{ij}. \] (B4)

We can estimate the PSD of \( \Delta g_{\text{single-link}} \) under the assumption of uncorrelated but statistically equivalent acceleration noises for the two TMs as:
\[ S_{\Delta g_{\text{single-link}}} (\omega) = 2S_{g_{ij}}, \] (B5)
where \( S_{g_{ij}} \) is the PSD of the single TM acceleration noise. To give an estimate of the OMS noise for the inter-spacecraft interferometer in a LISA link, we should consider that it enters just at the time \( t \) when we perform the measurement, as:
\[ \Delta x_{\text{single-link}}(t) = x_{ij}(t). \] (B6)
Here \( x_{ij} \) is the readout noise expressed in term of displacement at \( OB_j \) that faces the far \( OB_i \).

Appendix C: Analytical computation in time domain of the acceleration noise and displacement noise for the TDI \( X, \alpha, \zeta \)

We can compute how the TM acceleration noise propagates through the TDI \( X, \alpha \) and \( \zeta \), assuming equal and constant arm lengths as follows:
\[
X_g(t) = g_{12}(t - 8\tau) - 2g_{12}(t - 4\tau) - g_{13}(t - 8\tau) + 2g_{13}(t - 4\tau) + 2g_{21}(t - 7\tau) - 2g_{21}(t - 5\tau) - 2g_{21}(t - 3\tau) + 2g_{21}(t - \tau) - 2g_{31}(t - 7\tau) + 2g_{31}(t - 5\tau) + 2g_{31}(t - 3\tau) - 2g_{31}(t - \tau) + g_{21}(t) - g_{13}(t), \\
\alpha_g(\tau) = g_{12}(t - 6\tau) - 2g_{12}(t - 3\tau) - g_{13}(t - 6\tau) + 2g_{13}(t - 3\tau) + g_{21}(t - 5\tau) - g_{21}(t - 4\tau) - g_{21}(t - 2\tau) + g_{21}(t - \tau) + g_{23}(t - 5\tau) - g_{23}(t - 4\tau) - g_{31}(t - 2\tau) - g_{31}(t - \tau) - g_{32}(t - 5\tau) + g_{32}(t - 4\tau) + g_{32}(t - 2\tau) - g_{32}(t - \tau) + g_{12}(0) - g_{13}(0), \\
\zeta_g(\tau) = g_{12}(t - 2\tau) - 2g_{12}(t - \tau) - g_{13}(t - 2\tau) + 2g_{13}(t - \tau) - g_{21}(t - 2\tau) + 2g_{21}(t - \tau) + g_{23}(t - 2\tau) - 2g_{23}(t - \tau) + g_{31}(t - 2\tau) - 2g_{31}(t - \tau) - g_{32}(t - 2\tau) + 2g_{32}(t - \tau) + g_{12}(t) - g_{13}(t) - g_{21}(t) + g_{23}(t) + g_{31}(t) - g_{32}(t).
\] (C1a)

Following the same assumption we used for computing the TM acceleration noise, we can also compute how the OMS noise enters in the above mentioned TDI channels:

\[
X_{oms}(t) = x_{12}(t - 6\tau) - x_{12}(t - 4\tau) - x_{12}(t - 2\tau) - x_{12}(t) - 6\tau) + x_{13}(t - 4\tau) + x_{13}(t - 2\tau) + x_{13}(t - \tau) + x_{13}(t), \\
\alpha_{oms}(t) = - x_{12}(t - 3\tau) + x_{13}(t - 3\tau) + x_{21}(t - 5\tau) - x_{21}(t - 2\tau) - x_{23}(t - 5\tau) - x_{23}(t - 2\tau) + x_{31}(t - \tau) - x_{31}(t), \\
\zeta_{oms}(t) = - x_{12}(t - \tau) + x_{13}(t - \tau) + x_{21}(t - \tau) - x_{23}(t - \tau) - x_{31}(t - \tau) + x_{32}(t - \tau) + x_{32}(t - \tau) + x_{12}(t) - x_{13}(t) - x_{21}(t) + x_{23}(t) + x_{31}(t) - x_{32}(t).
\] (C2a)

**Appendix D: Computation of the Sensitivity**

Following [39] and [4], we consider stochastic sources with both plus and cross polarizations in their source frame. In the Solar System Barycenter (SSB), these will appear with \(h_+\) and \(h_x\) given by\( h_\pm^{SSB}(t, r) = h_\pm(t, r) \cos(2\psi) - h_x(t, r) \sin(2\psi)\) and \(h_\perp^{SSB}(t, r) = h_\pm(t, r) \sin(2\psi) + h_x(t, r) \cos(2\psi)\), where \(\psi\) is the polarization angle. The sensitivity to GW sources coming from different directions is computed for each source considering the relative frequency shift that an incoming GW causes on a LISA link as for example given in [24]. We then convert this frequency shift to an equivalent displacement. We computed both the case of three equal armlength and three unequal constant armlength. Assuming that our signal is made of superposition of many GW sources coming from different directions and with different polarizations, we can consider that the output of a TDI, given superpositions of \(n\) plane waves is:

\[
S_{j_k} = \sum_{i=1}^{n} T_{ji}^j(\omega) S_{h_i}(\omega),
\] (D1)

where \(S_{h_i}(\omega)\) is the PSD of the \(i\)th GW source expressed as dimensionless strain, and \(T_{ji}^j(\omega)\) is the absolute squared value transfer function for the \(j\)th TDI, including the conversion factor such that \(S_{j_h}\) is in units of \(m^2\text{Hz}^{-1}\). Labelling the PSD of the TM acceleration noise and OMS noise for each TDI \(j\) as \(S_{j_g}\) and \(S_{j_{oms}}\), respectively, the sensitivity of each TDI combination is computed by renormalising the total instrument noise Amplitude Spectral Density (ASD) by the GW transfer function as:

\[
S_{j_n/h} = S_{j_g}^{disp} + S_{j_{oms}} \frac{RMS\{T_{ji}^j(\omega)\}}{2},
\] (D2)

where the \(RMS\{\}\) denotes the root mean square over all sources \(i\) and as before \(S_{j_g}^{disp}\) is the TM acceleration noise expressed as an equivalent displacement. The response to a SGWB can also be written using a continuous integral over the whole sky as for example shown in [40] and [16]. The angular integral reported there is then evaluated numerically to get a result which is valid for the whole LISA frequency range. The computation reported in this paper is one possible method for numerically approximating the result of the continuous integral by replacing it with a sum over discrete stochastic sources from different directions. Indeed, in the limit of an infinite number of sources, this converges to the same integral.
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