Computational Intelligence Methods for Identifying Voltage Sag in Smart Grid

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ABSTRACT
In recent years pattern recognition of power quality (PQ) disturbances in smart grids has developed into crucial topic for system equipments and end-users. Undoubtedly analyzing the PQ disturbances develop and maintain smart grids effectiveness. Voltage sags are the most common events that affect power quality. These faults are also the most costly. This paper represents performance comparisons of different computer intelligence methods for voltage sag identification. PQube Analyzer which is installed in Ondokuz Mayis University Computer Laboratory for collecting real time disturbances data for each three phases in order to test for proposed algorithms. Firstly, we used Hilbert Huang Transform to generate Instantaneous Amplitude (IA) feature signal. Then Characteristic features are attained from IA. The 4 features, mean, standard deviation, skewness, kurtosis of IA are calculated. Support Vector Machines (SVMs) and C4.5 Decision Tree methods are conducted for classification of the disturbance. Secondly we used Fishers Discriminant Ratio for selecting statistical features such as mean, standard deviation, skewness and kurtosis of the normal and voltage sag signals for this part K Means Clustering Method were performed for classification of the disturbance. Consequently, SVMs, C4.5 Decision Tree and K Means Clustering Methods were performed also their achievements were matched for error rates and CPU timing.

1. Introduction

PQ problems occur due to frequency, voltage swells, voltage sags, light flicker, voltage transients, harmonics, unbalanced 3-phase signal. All equipments influence the network by the reproduction harmonics, voltage variations. Also the performance of transformers and generators are affected by the power quality on the distribution network. Unanticipated alteration of the voltage and current signal from healthy conditions can devastate or stoppage the critical electrical machineries created for various aim [1-5].

1.1. Voltage sag

Voltage sags are short-duration (less than 1 second) reductions in voltage magnitude. The sag is mostly related with the short circuit faults single-line to ground, line to line, and for three-phase to ground faults. This type of disturbance is nowadays one of the main PQ disturbances. Momentary increase of current has many origins in power systems such as energizing of transformers, short circuits, earth faults and starting of induction motors. Voltage sags are the most typical disturbances that affect power quality in smart grids. Moreover, they are the most costly. When voltage sag happens, the power supply inside electronic devices uses some of its stored energy to make up for the failure of input voltage. If enough energy is lost owing to the sag, then the power supply may fail its task to maintain adequate DC voltage to components. Even for nanoseconds such integrated circuits, inside the device are affected by. This is very short moment to destroy data in microprocessor based electronics and to bring about malfunctions of digital equipment especially PCs. Malfunction of PCs integrated, in a real-time system is linked with conceivably bigger results, because there are losses correlate with the controlled process. The sensitivity of PCs to voltage sags is mostly expressed only in terms of the magnitude and duration of the voltage sag [25, 26, 30, 31, 40]. Considering previous studies about PQ disturbances analysis, Short-Time Fourier Transform (STFT) have been applied for stationary and periodic signals in frequency domains. But when handling non-stationary signals, it is problematic to analyze with STFT [47, 48, 49]. It was experimentally verified and demonstrated the WT is better than STFT [47]. Wavelets are employed in PQ when the time information is crucial however, it is not crucial to perceive the specific frequency of a fault in waveforms. Using WT, high time resolution is arranged for high- frequency components and low time resolution is attained for low–frequency components of the disturbance. Another method, ST is acquired by multiplying WT...
with a phase correction factor. This method’s disadvantage is having large computation time. Since ST is a Fourier-based technique, it also deteriorates from the hand of Fast Fourier Transform (FFT) [18, 20, 47, 48, 49]. On the other hand, the Hilbert Huang Transform (HHT), signals are extracted for different oscillation modes called intrinsic mode functions (IMFs). These functions are related to different time scales. Application of Hilbert transform to these IMFs can indicate the instantaneous amplitudes (IA) and the instantaneous frequencies (IF). Therefore, HHT can give a reasonably better time-frequency pattern representations for non-stationary signals [6, 11, 12, 13, 18, 19, 21]. In the light of this information, we used the HHT for generating IA signals from the voltage sag signal Firstly, for 3 phases (L1-N, L2-N, L3-N) real time processing the first intrinsic mode function is removed with the addition (superposition) of remain components to reconstruct the analyzed signal. Secondly, we generate features with the help of statistical functions such as mean, standard deviation, skewness and kurtosis. For Diagnosis, we generate features with the help of statistical functions such as mean, standard deviation, skewness and kurtosis. For Diagnosis, we generate features with the help of statistical functions such as mean, standard deviation, skewness and kurtosis.

2. Material and Methods

2.1. Signal Processing Methods for Feature Generation

Unlike the traditional methods such as STFT (Short time Fourier Transform), Wavelet Transform (WT) and Stockwell Transform (ST), Hilbert Huang Transform (HHT), especially for noise assisted PQ disturbances signals, provides better and useful representation in order to detect the disturbance. Disadvantages and advantages of signal processing methods for feature generation procedure summarized in Table I.

Table I. Comparison of feature extraction methods for PQ disturbances

| Method | Disadvantages                                      | Advantages                                                   |
|--------|---------------------------------------------------|--------------------------------------------------------------|
| STFT   | Basic: Non-adaptive                              | Profitably handle for stationary                              |
|        | Completely Theoretical                           | signals where properties of signals do not derive in time.   |
|        | Not acceptable for non-stationary                | Quite simple in application.                                  |
|        | signal as it does not follow signal               |                                                              |
|        | oscillation accurately because of limitation of   |                                                              |
|        | fixed window width                               |                                                              |
| WT     | Basic: Non-adaptive                              | Gives local illustration in both time and frequency.         |
|        | Completely Theoretical                           | Thus, suitable good time–frequency resolution is             |
|        | Fully affected by noise present in the signal,   | appropriate.                                                 |
|        | adversity from spectral                          |                                                              |
|        | leakage effects                                  |                                                              |
| ST     | Basis: Non-adaptive                              | Maintain time and frequency representation. Good time-frequency |
|        | Completely Theoretical                           | resolution.                                                  |
|        | Does not accomplish real-time requirement based   |                                                              |
|        | on block processing, false harmonics             |                                                              |
|        | measurement owing to dependency of frequency     |                                                              |
|        | window width                                     |                                                              |
| HHT    | Limited for narrow band conditions                | Basic: Adaptive                                               |
|        | Mode mixing and end effect problems               | Empirical                                                    |
|        |                                                   | Conveniently applicable to non-linear and non-stationary      |
|        |                                                   | processes.                                                   |
|        |                                                   | Generates perpendicular IMFs                                 |
|        |                                                   | whereby instantaneous amplitude and phase can be easily        |
|        |                                                   | assessed be easily evaluated                                  |

2.2. Stockwell transform (S-transform)

The S-transform is a time–frequency plane gadget produced by the mixture of wavelet and short time fourier transform It generates a time–frequency plane of a time series. It individually merges a frequency dependent resolution that contains the real and imaginary part of spectrum. Gaussian modulated co-sinusoids are the primary function for the S-transform [6, 7]. The Short Time Fourier Transform (STFT) of signal \( x(t) \) is explained in equation 1.

\[
STFT(\tau, f) = \int_{-\infty}^{\infty} x(t) g(\tau-t) e^{-j2\pi ft} dt
\]

\( \tau \) and \( f \) stand for the time of spectral localization and Fourier frequency, \( g(t) \) shows a window function. The S transform could stem from (1) by putting into \( g(t) \) with the Gaussian function, explained in equation 2.

\[
g(t) = \frac{|f|}{\sqrt{2\pi}} e^{-\frac{r^2 f^2}{2}}
\]

Ultimately the S transform is achieved in equation 3.

\[
S(\tau, f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi \frac{r^2 f^2}{2}} e^{-j2\pi ft} dt
\]

If the window of S transform is wider in time, it gives better frequency resolution for lower frequency. Whereas the window is narrower, it gives superior time resolution for higher frequency, namely trade-off between time and frequency resolution [18, 20].

![S Transform Contours of Normal Condition (Phase A)](image)

Figure 1. STransform Contours of Normal – Voltage Sag Cases

S-Transform is not capable of detecting noise as much as HHT in real time measurements [8, 32] (Figure 1). However, it is clearly shown that ST can produce valid features for detecting voltage sag.

2.3. Proposed Method: Hilbert Huang transform

HHT, a self-adaptive signal-processing method, which has been applied in non-stationary signal-processing successfully, was developed by Huang. EMD method is a sifting iteration, to decompose one Intrinsic Mode Function (IMF) component needs many times sifting till obtaining last IMF, residual monotonic function [9, 10]. Main steps of HHT shown in figure 2.
The EMD [11, 12, 13] algorithm the steps:

**Beginning of Sifting Process**

1. Find all maximas and minimas of the signal \( x(t) \).
2. Attain the upper and lower envelope by the cubic spline interpolation.
3. Find the mean function of the upper and lower envelope \( \text{mean}(t) \).
4. Calculate the difference signal \( f(t) = x(t) - \text{mean}(t) \).
5. When \( f(t) \) becomes a zero-mean process, then the iteration stops and \( f(t) \) is an Intrinsic Mode Function \( \text{IMF}_1, c_1(t) \); otherwise, go first step and replace \( x(t) \) with \( f(t) \).
6. Calculate the residue signal \( \text{res}(t) = x(t) - c_1(t) \).
   Iterate the procedure from steps (i) to (vi) to obtain \( \text{IMF}_2, c_2(t) \). To attain \( c_n(t) \), continue steps (1) till step (6) after \( n \) iterations.

**End of the Sifting Process**

7. The algorithm stops when the final residual signal \( \text{res}(t) \) becomes a monotonic function.

2.3.1. Ensemble EMD (EEMD) Algorithm

The major challenge of the EMD algorithm is the appearance of mode mixing. Therefore the EEMD algorithm was performed for overcoming the problem in this study [12, 13].

Analyzed signal is decomposed using the ensemble empirical mode decomposition (EEMD) into intrinsic mode function (IMF) that includes instantaneous frequencies (IF) and instantaneous amplitudes (IA). When the signal is turn into IMFs, the Hilbert transform can formerly be carried out to each IMF giving the IA and IF versus time plot. This merging of EEMD process and Hilbert transform is known as the HHT [6, 11, 12, 13, 18, 19, 21].

Figure 3a. Feature vectors (IMF1 (noise) to IMF3) for a voltage sag signal with EEMD

Figure 3b. Feature vectors (IMF4 to Residual (IMF12)) for a voltage sag signal with EEMD

For 3 phases (L1-N, L2-N, L3-N) real time processing the first intrinsic mode function is removed with the addition (superposition) of remain components to reconstruct the analyzed signal. Fig 3 illustrates first component IMF1 the noise (lowest
magnitude highest frequency signal) on the line (L1-N) to the lowest frequency component residual, last IMF12.

$$
\sum_{i}^{N} \frac{1}{d_i} \rightarrow \text{Entropy}
$$

3. PQ Disturbances Computational Intelligence Methods

3.1. Support vector machine (SVM)

SVM methods, is developed by Vapnik, whereby statistical learning technique being the basis contributes a novel machine learning method. SVMs are linked supervised learning methods used for classification and regression. SVMs target obtaining optimal hyper-planes different classes of input data in a high dimensional feature space, thus new test data can be classified by acquired the hyper-planes Support vectors lie closest to the optimal hyper-plane [7, 14, 18, 21, 28, 34].

Let \( \{x_i, y_i\}, i = 1, 2, ..., M \) be \( M \) training data vectors \( x_i \) with class label \( y_i \). Given an input vector \( x \), an SVM constructs a classifier of the form:

$$
f(x) = \text{sign}(\sum_{i=1}^{M} a_i y_i K(x_i, x) + b)
$$

(4)

\( a_i \) denotes non-negative Lagrange multipliers each of which corresponds to \( a \) training data, \( b \) denotes bias, \( K(, ,) \) is a kernel fulfilling the conditions of Mercer's theorem [11, 34]. In this study we performed two kernel functions are the polynomial kernel \( K(x_i, x_j) = (x_i, x_j + 1)^d \) we tested for \( d=2, 3 \). Gaussian Radial Basis Function (RBF) \( K(x_i, x_j) = e^{-\frac{|x_i-x_j|^2}{2\sigma^2}} \) we tested sigma 0.01 and 1 in this study.

3.2. Decision Trees

Decision trees are methods that utilize divide-and-conquer approaches as structure learning by induction [15, 16]. The C4.5 algorithm was developed by Qinlan, contains the generation of a tree whereby a training set, finding the information gain criterion with the help of Shannon Entropy in order to find the finest attribute/feature to be used at each node.

3.2.1. Shannon Entropy

It is given a probability distribution \( D = (d_1, d_2, ..., d_m) \) the Information carried by this distribution entropy of \( D \) is explained in equation 5.

$$
\text{Entropy}(D) = -\sum_{i=1}^{m} d_i \log(d_i)
$$

(4)

3.2.2. The Information Gain

The Information defines the gain for a test \( t \) and a position \( d \)

$$
\text{Gain}(D, t) = \text{Entropy}(D) - \sum_{j=1}^{m} d_j \text{Entropy}(d_j)
$$

(5)
that values \( dj \) is the set of all possible values for attribute \( t \). Equation 5 can calculate highest ranking attributes and build the decision tree where at each node is located the attribute with the highest information gain. Furthermore, the algorithm applies the post pruning approach to diminish the size of the tree and prohibit over fitting. C4.5 is a technique for approximating discrete-valued functions that is powerful tool to noisy data and suitable for learning distinctive statements [16, 17, 18, 19, 20, 21, 28, 29].

### 3.3. K-Means Clustering

K-means clustering algorithm is a significant unsupervised classification technique to separation of set of dissimilar patterns. Same featured patterns are arranged into same clusters, different features are replaced in different sets [22, 23]. The steps of the algorithm:

1. Apply \( K \) points into the space illustrated by objects being clustered. These points symbolize first group centroids. Determine each object to the nearest centroid group.
2. As objects have been completed, recalculate the localizations of the \( K \) centroids.
3. Iterate Steps 1 and 2 until the centroids do not change anymore. This creates a separation of the objects into groups from which the metric with squared Euclidean distance objective function which is defined in equation 6, can be calculated.

The objective of this algorithm is to make an objective function, minimize the squared error function [35, 36, 37].

\[
Obj(x) = \sum_{i=1}^{K} \sum_{j \in S_i} \|x_j - \mu_i\|^2
\]

\( K \) denotes clusters \( S_i \), \( i = 1, 2, ..., k \), and \( \mu_i \) is the mean points of centroids.

#### 3.3.1. Fisher’s discriminant ratio (FDR)

FDR or SNR (in machine learning) is commonly performed to compute the inequitable way of specific features between two clusters. Linked with the values of a feature in two classes, \( \mu_1 \) and \( \mu_2 \) be the respective mean values and \( \sigma_1^2 \) and \( \sigma_2^2 \) the respective variances defined in equation 7 [38, 39].

\[
FDR = \frac{(\mu_1 - \mu_2)^2}{(\sigma_1^2 + \sigma_2^2)}
\]

### 4. Experimental Set up and Discussions

For experimental studies, we installed PQube Analyzer in Ondokuz Mayis University Computer Laboratory for collecting real time disturbances data for each three phases in order to test for proposed algorithms. 3 phases connected with totally 28 numbers of PC loads. PQube power meter installation properties explained in appendix of the paper. This paper includes computational intelligence methods in order to classify voltage sag disturbance. Figure 6 shows 3 phase voltage sag condition PQube sample measurements from Computer Laboratory in Ondokuz Mayis University including July 2016 to December 2016.

![Figure 6. Computer laboratory power quality monitor with PQube voltage sag](image)

In this study, we apply two methods for identification of voltage sag. Selecting appropriate features of voltage sag events are highly crucial for diagnosing of the fault properly. Classification process of real-time voltage sag disturbances are shown in Figure 7.

![Figure 7. Schematic model of Voltage Sag Classification](image)
First Method, firstly extracted features are produced from ensemble empirical mode decomposition (EEMD) method so as to classify the voltage sags in grid. After reconstruction signal without noisy part, removing first intrinsic mode function (IMF). Then instantaneous amplitude (IA) is generated by means of HHT. The following features were extracted: mean, standard deviation, skewness function results of IA.

Second Method, we generated the statistical features from PQube 3 phase line voltages. In order to select the best feature from the functions, we performed Fisher Discriminant Ratio criteria. Selected features which give high ratio for Phase A and B are standard deviation and kurtosis, for phase C are mean and skewness. Then we used generated features for K Means Clustering Methods. Figure 8 shows selected features space and cluster assignments and centroids of normal and voltage sag condition sets.

Table II. Performances of disturbance diagnose algorithms

|                | Phase A (L1-N) | Phase B (L2-N) | Phase C (L3-N) |
|----------------|----------------|----------------|----------------|
| Error rate     |                |                |                |
| (ER)(%)        |                |                |                |
| SVM-poly d=2   | 10             | 4.088          | 0              | 3.321          | 20             | 5.883          |
| SVM-poly d=3   | 10             | 3.929          | 0              | 3.636          | 20             | 5.489          |
| SVM-RBF sigma =0.01 | 50        | 0.738          | 0              | 3.064          | 20             | 5.805          |
| SVM-RBF sigma =1 | 0          | 5.254          | 0              | 5.368          | 50             | 0.674          |
| DecisionTree   |                |                |                |
| C4.5           | 20             | 6.20           | 20             | 6.835          | 20             | 3.514          |
| K-Means (Euclidean distance) | 6.66 | 0.693 | 16.6 | 0.510 | 13.3 | 0.480 |

For phase A, it is concluded that for radial basis kernel function sigma 1, gives more proper result precision of 100% and CPU time (5.254 sec) in non linear SVM. Kernel function for SVM is

\[ K(x_i, x_j) = e^{-\frac{|x_i - x_j|^2}{2(\sigma^2)}} \]

For phase B, it is deduced that for radial basis kernel function sigma 0.01, gives more accurate result precision of 100% and CPU time (3.064 sec) in non linear SVM.

For phase C, it is deduced that for K-Means clustering (Euclidean distance function \(d_{i,j}\)), gives better result precision of 86.7% and CPU time (0.480 sec).

\[ Euclidean(d_{i,j}) = \sqrt{\sum_{k=1}^{n} (x_{ik} - x_{jk})^2} \]

It is clearly shown in Table II when examining each phases K-means clustering algorithm is the fastest algorithm.

For phase A and B, SVM with RBF sigma=1 gives better results for detection voltage sag.

For three phases, C4.5 decision tree gives same 80% recognition rate.

(Note: the most proper and robust classifiers for each phase are showed by Red font in Table II).

5. Conclusion

In this study, we aimed to generate non-time consuming features for detection voltage sag in smart grid. We also process and classify real time 3 phase voltage sag signals. In contrast to literature studies we do not normalize the features. Thus we can relate load types and fault magnitude and duration. Also it is processed real-time noise-assisted data. For this reason this study gives effective ideas about real time disturbance detection. As a result, SVMs, C4.5 Decision Tree and K Means Clustering Methods were performed also their achievements were matched for error rates and CPU timing. Simulations indicate that K-means clustering algorithm is the fastest algorithm. For phase A and B, SVM with RBF sigma=1 gives better result for detection voltage sag. Simulation results will be using for relation with loads types and the magnitude and duration of the voltage sag.

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Table III provides discussions about related signal processing methods such as WT, ST HT and for decision part SVM, DT, k-nearest neighbor (k-nn), Probability Neural Network (PNN). It also gives percentages error rates literature studies and the proposed methods. It is clearly shown in Table III most of the studies reported the results on synthetic normalized an done phase data which are absolutely far from real PQ disturbances. Apart from the previous studies, detection capability validates the potential of the proposed algorithm. To analyze the real time power signals and classification techniques were employed with Matlab™ Toolboxes.

Table III. Performances of disturbance diagnose algorithms

| Reference        | Methods for PQ Detection | Data types with Noise or Not ? | Error rate (ER)(%) |
|------------------|--------------------------|--------------------------------|-------------------|
| Erstri and Demir [41] | WT + SVM                 | Noisy data+Practical data      | 4,19%             |
| Xiao et al. [42]  | ST + SVM                  | Without noise+Synthetic data    | 7,70%             |
| Samantaray [43]   | ST + DT                   | Noisy data+Practical data       | 2,44%             |
| Hu et al. [44]    | WT + SVM                  | Without noise+Synthetic data    | 1,50%             |
| Gargoom et al. [45]| HT + k-NN                 | Without noise+Synthetic data    | 19,40%            |
| Shukla et al. [46]| HT + PNN                  | Noisy data+Practical data       | 2,50%             |
| Proposed Methods  | *Table II                 | Noisy data+Practical data       |                   |
|                  |                          | (Not normalized data for each phase measurements) | *Table II |

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APPENDIX

Table IV. PQUBE MAINS VOLTAGE MEASURING CHANNELS [33]

| Connection            | L1, L2, L3, N PQube screw terminals |
|-----------------------|-------------------------------------|
| Frequency Range       | 40 Hz ~ 70 Hz and 320 Hz ~ 560 Hz, Nominal 50 Hz, 60 Hz, or 400 Hz auto, 320-560 Hz manually selected. |
| Specifications below apply at 50/60 Hz. | |
| Mains Configuration   | Single-phase, split-phase, delta, wye or star. User selected or auto-selected. |
| Range of Nominal Input Voltage | 100 VAC ~ 690 VAC L-L (69 VAC ~ 400 VAC L-N), User selected or auto-selected. |
| Measurement Channels  | Line-to-Neutral, Line-to-Line, Neutral-to-Earth. |
| Sampling Rate         | 256 samples per cycle, phase-locked to input frequency. |
| Measurement Range     | 0 VAC ~ 900 VAC L-L (320 VAC L-N) |
| Accuracy              | ±0.05% rdg ±0.05% FS typical (10%~150% of nominal). Factory tested at better than ±0.04% rdg ±0.04% FS. Note: FS = 345 VAC or 520 VAC, selected based on nominal line-to-earth voltage. |

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