Scalable All-Printed Microwave Microfluidic Sensor for Multi-Liquid Characterization based on a Stub-Loaded Microstrip Line

Mahmoud Wagih, Member, IEEE, and Junjie Shi

Abstract—Microwave microfluidic sensors are typically designed with a channel in vicinity of a resonator’s fringing electric (E)-fields, to characterize the material properties of a single fluid. This paper leverages hybrid 3D and dispenser printing to realize a scalable microfluidic sensor utilizing the parallel-plate capacitance of an open-ended microstrip stub, enabling, for the first time, a tunable sensitivity. A stub-loaded microstrip line is then proposed for characterizing multiple microfluidic samples simultaneously using a simple two-port multi-band resonator. The physical constrains which limit the scalability of the proposed sensors have been analyzed analytically and numerically, prior to implementing a three-channel triple-band sensor. The microfluidic channels have been fabricated using stereolithography 3D printing with the microstrip line directly dispenser printed on a conformable polyimide substrate. To accommodate varying channel thicknesses, a tapered microstrip line is proposed to maintain the impedance matching. The fabricated sensor is characterized using binary water-IPA mixtures to evaluate its sensitivity, comparing favorably with reported 3D-printed sensors. The proposed sensor achieves over 90% accuracy in determining the real permittivity following a simple water-based calibration across the different channels, for samples with $16 < \varepsilon_r < 78$. In addition to achieving a sensitivity in-line with state-of-the-art single-channel printed sensors, the sensor’s applicability to temperature sensing is demonstrated experimentally with over 2.4–5 MHz/°C temperature sensitivity across all channels.

Index Terms—Antennas, Filter, Microfluidic, Microstrip, Resonant Sensors, Sensor,

I. INTRODUCTION

M ICROWAVE resonators have recently attracted significant interest for a plethora of sensing applications [1]–[3]. From chipless wireless RFID “smart” sensing [1], to on-chip microwave, mmWave [4], and THz material characterization [5], microwave sensing covers a range of industrial to healthcare applications [6], [7]. Compared to chemical methods, microwave sensors are non-destructive, non-invasive, and can sense various stimulants such as gas concentrations [8], glucose levels [7], and bacterial growth in real-time [6].

Resonant dielectric sensors are preferred over broadband phase-delay sensors for applications where a high sensitivity is required for small changes in the relative permittivity of a microfluidic Material under Test (MUT) [9]. Planar resonators such as split ring resonators (SRRs) [7], [10], [11], series metamaterial-like LC resonator [12], interdigitated capacitors within an SRR [9], spoof surface plasmons [13] and more recently substrate integrated waveguide (SIW) cavity [14] resonators have been widely investigated for microfluidic and material sensing. The quality factor and the capacitance or inductance of the sensor are often discussed when improving the sensitivity of such sensors [13]. However, most planar sensors are based on a microfluidic channel placed on top of or going through the planar resonator, [7], [9], [10], [13], [15], utilizing the fringing electric (E)-fields going through the MUT for sensing. Moreover, while certain resonator designs may improve the sensitivity based on a high Q-factor, designing a sensor based on fringing E-fields will only yield minimal improvements to the sensitivity due to the existence of multiple capacitive elements, not contributing to sensing [16]. Therefore, the use of the parallel plate capacitance as opposed to the fringing E-fields was proposed for improving the sensitivity compared to conventional planar resonators [16].

Additive manufacturing, using direct-write 3D and conductor printing, has been widely used to realize microwave and mmWave components such as filters, antennas, and transmission lines [2], [14], [15], [17]–[23]. Recently, additive manufacturing using fused deposition modelling (FDM) has been proposed a method of realizing microfluidic channels for integration with resonators implemented on a conventional printed circuit board (PCB) [15]. To this end, the resonator and the channel are designed as separate entities, which are then integrated after fabrication. This restricts the MUT to being deployed over a planar microstrip resonator, typically existing in its fringing field region [15]. However, utilizing the fringing capacitance of a complex structure such as a split ring resonator reduces the overall sensitivity of the sensor due to the existence of several capacitive elements in the equivalent circuit of the sensing resonator [16]. Therefore, the use of the parallel-plate capacitance of a microstrip was proposed for high-sensitivity sensing [24]. However, the frugal realization of the sensor limits its scalability [16]. 3D printing was used to realize a microfluidic channel burried beneath a microstrip line [22]. The use of stereolithography (SLA) printing to realize a buried meander-line microfluidic channel, buried within a substrate-integrated waveguide (SIW) resonant cavity, was also proposed [14]. While these implementations shows the flexibility of 3D printed channels and their improved integration.
with microstrip structures, the ability to 3D print the channel was not applied to improving the sensor’s sensitivity, or to enable scalable multi-parameter sensors.

This paper proposes a stub-loaded microstrip line for simultaneously characterizing multiple liquid dielectric samples, representing the first scalable microwave sensor capable of characterizing more than two samples. In Section II, microstrip stubs are investigated from a dielectric sensing perspective, followed by closed-form circuit analysis and full-wave simulations. Section III discusses the fabrication of the sensor using low-cost 3D substrate and dispenser microstrip printing along with the measured results. Finally, the dielectric extraction, sensitivity analysis and sensor accuracy evaluated are presented in Section IV, along with a qualitative and quantitative comparison with state-of-the-art sensors.

II. MICROSTRIP SENSOR DESIGN AND SIMULATION

A. Parallel-Plate Resonant Microstrip Sensor Design

A parallel plate microstrip capacitor exhibits a significantly higher E-field density over a larger area, compared to the fringing fields of a coupled resonator. While an open-ended microstrip stub was used in [25] for sensing applications, the MUT was placed above the parallel line, implemented on a standard PCB, resulting in a standard fringing capacitance sensor similar to [26]. 3D printing was previously used to implement a channel buried underneath a microstrip line on a standard PCB [22].

With the MUT buried in the microstrip substrate, as shown in Fig. 1(b), the microfluidic channel can now be regarded as a multi-layered substrate [27]. The equivalent circuit of the dielectric sensor can then be simplified to a single LC element. The capacitive element is dependent on the MUT’s permittivity ($\epsilon_{MUT}$). However, for the multi-material substrate, the equivalent real permittivity, $\epsilon_{Eq}$, is given by

$$\epsilon_{Eq} = \sum_{n=1}^{N} \left( \frac{t_n}{\epsilon_n} \right)^{-1} \times H$$

where $H$ is the total height of the substrate, $t_n$ is the thickness of the individual layers from Fig. 1, and $\epsilon_n$ is the permittivity of layer $n$ [27].

As varying the MUT varies $\epsilon_{MUT}$ and hence $\epsilon_{Eq}$, the maximum change in the resonant frequency $\Delta f_r$ for a sensing microstrip stub can be given by the

$$\Delta f_r = \frac{1}{2\pi\sqrt{LAC}}$$

where $L$ represents the inductance of the resonant microstrip stub, and $\Delta C_n$ represents the change in the capacitance induced by the MUT [16]. The sensitivity $S$ of a resonant dielectric sensor is in-turn given by

$$S = \frac{f_{air} - f_{\epsilon_r}}{f_{air}(\epsilon_r - 1)}$$

where $f_{air}$ and $f_{\epsilon_r}$ are the resonant frequencies in air and for an MUT-loaded sensor [14].

Typically, maximizing $S$ is regarded as a figure of merit in dielectric sensor design [16], [28]. Observing (1) and (2), it can be seen that varying $\epsilon_{Eq}$, through controlling the height of the substrate and the channel, can increase the variation in capacitance and hence, enhancing $S$.

To demonstrate the effect of varying thicknesses on $S$, $H=3$ mm is considered with varying $t_c$ to $t_1$ and $t_2$ ratios, in the case of a single resonant microstrip stub. MUT $\epsilon_r$ of 78 and 1.0, representative of water and air, are assumed for the loaded and unloaded cases. $\epsilon_{Eq}$ has been calculated analytically using (1), for varying $t_c/H$ ratios. A substrate $\epsilon_s=2.7$ is assumed in this calculation. Fig. 2(a) shows the analytically-calculated variation in $\epsilon_{Eq}$ as a function of the channel-to-substrate thickness, where it can be observed that orders of magnitude change in $\epsilon_{Eq}$ could be introduced to the sensor by controlling the channel’s height.

Fig. 2(b) shows the closed-form simulated forward transmission response of a microstrip stub with length $L=30$ mm, simulated using Keysight ADS, for varying MUT-to-substrate height ratios. It can be clearly observed that the change in the resonance frequency $\Delta f_r$ is highly dependent on the channel’s height, where the sensitivity can be improved by up to four-folds, by changing the MUT-to-substrate ratio from 20% to 80%. Therefore, by combining the simple parallel-plate capacitance of a microstrip resonator with high-resolution additive manufacturing, microwave microfluidic sensors with high tunable sensitivities can be realized.
B. Multi-Fluid Sensor Design

So far, only a single sensing element has been discussed. As shown in Fig. 1, a microstrip line could be loaded with multiple sensing elements, i.e., resonant stubs, for simultaneously characterizing multiple fluids using the same read-out circuit, such as an oscillator [29], significantly reducing the testing time, connections, and cost. The limit of such sensor scaling mechanism is the bandwidth. To explain, the open-ended line resonates at at \( L = \lambda / 4 \) and \( 3\lambda / 4 \). Therefore, should additional sensing elements be added to the line, their resonant frequency has to fall between the first and second-order resonances of the lowest-frequency stub.

A numerical example is considered to illustrate the bandwidth limits of a multi-stub sensor. Stub \( L = 44 \) mm, substrate \( H = 1.44 \) mm, and \( W_S = 1.4 \) mm are assumed. Unloaded \( \epsilon_{r,0} = 1.6 \) and 1.55 are investigated for loaded \( \Delta \epsilon_{r,0} = 2.9 \) and 6.5, respectively. The microstrip stubs’ broadband bandwidth which can be used to add more stubs to the line, their resonant frequency has to fall between the first and second-order resonances of the lowest-frequency stub.

Consequently, to enable multi-stub/multi-fluid sensors to be realized, the sensitivity could be dynamically tuned by varying the MUT-to-substrate thickness ratio. The maximum number \( N \) of stubs which can be realized in the scalable bandwidth of a resonant microstrip sensor is hence given by

\[
N = \frac{3f_0,MUT - f_{0,r}}{\Delta f_n},
\]

where \( \Delta f_n \) is the sensing bandwidth of the \( n \)th stub, \( f_{0,MUT} \) is the resonant frequency with the maximum MUT \( \epsilon_r \), and \( f_{0,r} \) is the unloaded resonance. Thus, as long as the sensing bandwidth of each resonant stub does not overlap the other elements on the same two-port network, a scalable multi-fluid sensor can be realized based on stub-loaded microstrip line. However, this assumes that closely-spaced sensing elements do not interact or mutually-couple, this is investigated using 3D full-wave simulations, detailed in the next sub-section.

C. Sensor Design and Full-Wave Simulation

For the sensor to be fabricated using low-cost standard SLA 3D printing, the total substrate height should be over 2 mm [14]. Due to the high thickness, standard edge PCB RF connectors cannot be directly mounted on the substrate. To overcome the thickness issue, a ramped and tapered non-uniform microstrip line is proposed. Tapered non-uniform microstrip lines first emerged for MMIC applications [30], [31], and were more recently applied to 3D mmWave multi-layered packaging [32], enabled by the advances in 3D printing. The width and taper rate are optimized to maintain \( Z_0 = 50 \, \Omega \) throughout the structure. The layout and dimensions of the fabricated prototype are shown in Fig. 4.

So far, the closed-form microstrip circuit model has been used to simulate the sensor. However, such a model assumes the stubs are fully isolated and have no influence on each other’s \( f_n \). 3D Full-wave simulation in CST Microwave Studio was used to observe the sensor’s response for MUT \( \epsilon_r = 1.0 \) and 78, representing the two ends of the possible MUTs based on recently reported sensors [28]. The first step in designing the multi-stub sensor is to identify the minimum distance required between the adjacent stubs to mitigate coupling.

The sensor has been simulated in CST Microwave Studio for varying separation, \( W_G \), between the stubs. It was found that, based on a substrate’s \( \epsilon_r = 2.7 \), the minimum \( W_G \) required to minimize the coupling between the three separate stubs and \( f_n \) is 10.5 mm, \( \approx 10 \times W_S \), as shown in Fig. 4. The simulated response of the three resonators/channels for varying MUT \( \epsilon_r \)
is shown in Fig. 5. For each sweep, the other channels were simulated with a $\varepsilon_r = 78$ filling, representative of active operation with a water-based MUT, which emulates various real applications such as glucose monitoring [7]. Observing Fig. 5, changing $\varepsilon_r$ in each channel introduces minimal variations in the other channels, with the most obvious change in the case of the MUT in channel 2, in Fig. 5(b), influencing $f_1$ and $f_3$ when filled with water. Such additional interference could be calibrated using a sample of known $\varepsilon_r$, e.g. water, which is widely used to calibrate microwave permittivity sensors [33].

### III. SENSOR FABRICATION AND MEASUREMENT

#### A. Hybrid Additive Manufacturing

To fabricate the sensor with the dimensions in Fig. 4, a hybrid additive manufacturing approach is utilized. First, the substrate with an embedded channel is 3D printed. The microstrip line is then dispenser printed (thick-film direct-write printing technique [23], [34]) on a conformable thin substrate before integration with the channel. Similar approaches have been proposed combining inkjet printing with 3D printing for antenna and system-on-package applications [32]. Dispenser printing on the other hand is more suited to rapid manufacturing and prototyping compared to inkjet printing, in addition to being able to realize layers of higher thickness (and hence conductivity) in a single print, significantly reducing fabrication time and cost [23]. Fig. 6 shows the fabrication process of the sensor.

Printing the substrate represents a more straightforward and lower cost fabrication method compared to molded PDMS microfluidic channels integrated with low-loss RF PCB substrates [35]. SLA 3D printing has been selected as opposed to the more conventional FDM, where an SLA-printed substrate will exhibit water-tight properties, preventing the substrate from absorbing the MUT. Moreover, the optical translucency of the resin (Formlabs Clear Resin) enables visualization of the MUT as it fills the channel [14]. The substrate was printed by 3D Hubs using a Formlabs SLA printer with ultraviolet (UV) post-curing. Owing to the increased thickness of the channel and the proposed ramped microstrip line, the designed sensor can be printed using low-cost low-resolution printers due to the large layer thickness of 0.7 mm used in this work.

A 1.6 mm-thick SLA substrate was printed for measuring the relative permittivity of the substrate, to guide the design of the microstrip sensor. Fig. 7 shows the closed-form (ADS) simulated and measured $S_{21}$ response of the T-resonator for characterizing the 3D-printed resin’s permittivity; $\varepsilon_r=2.7$ and $\tan\delta=0.02$.

Direct-write dispensor printing is used to fabricate the microstrip line and stub using conductive silver ink. The ink (Voltera flexible ink) was printed on a 75 μm flexible polyimide substrate ($\varepsilon_r=3.2$, $\tan\delta=0.002$) using a Voltera V-One printer. Dispenser printing has previously been used to design and prototype antennas and rectennas [23], [34], showing minimal variations compared to etched copper sheets of the same geometry [34]. Following the conductor curing at 170° on a standard hot-plate, the flexible microstrip line is adhered to the ramped substrate using a double-sided adhesive, of a <50 μm measured thickness. Photographs and micrographs
of the fabricated sensor are shown in Fig. 8.

Unlike previous works [14], [16], the proposed fabrication process is scalable and can be adjusted in real-time to control the microstrip line’s dimensions based on the substrate’s thickness. This overcomes using a manually-assembled copper tape ground plane as in [16], or an adhered copper film as in [14]. Therefore, the proposed fabrication method is a comparable lower cost alternative to standardized inkjet printing of microwave microfluidic sensors [?], [35].

B. Sensor Measurements and Characterization

The fabricated prototypes were measured using a two-port ZVB4 vector network analyzer (VNA), calibrated using a standard TOSM calibration. The 3D printed channels have been connected via long tubes, as shown in Fig. 9, to external syringes containing the MUT. The binary De-ionized (DI) water and IPA mixtures have been prepared by volume and kept in separate syringes. The measurements were carried out at a fixed room temperature to ensure the MUT’s permittivity temperature-dependence does not influence the accuracy of the results.

The simulated and measured forward transmission of the bare sensor, i.e. with empty channels, are shown in Fig. 10. As the 3D CST model is based on the measured thickness of the fabricated sensor, based on the micrograph in Fig. 8(c), and the measured dielectric properties of the substrate, the observed close agreement is anticipated. This further demonstrates the advantage of using direct-write printing to realize the microstrip line, whose dimensions can be tuned after the microfluidic channel substrate is printed and characterized mechanically and electrically.

The channels have then been sequentially filled with DI water, to observe their loaded response. The water-loaded response can then be used as the calibration of the sensor, prior to characterizing different fluid samples [16]. Fig. 11 shows the simulated and measured $S_{21}$ response for DI water loading in different channels. From the sequential channel-loading, it can be seen that injecting water in a certain channel introduces a minimal influence on $f_r$ of the neighbouring channels. Along with the three fundamental resonant frequencies, around 1.45, 2.5 and 3.4 GHz, the higher-order resonance of the longest stub is observed around 3.9 GHz in Fig. 11, but only when channel 1 is filled with water. This indicates that all three stubs are indeed within the bandwidth limits, previously discussed in Section II-B.

IV. PERMITTIVITY MEASUREMENTS AND SENSOR EVALUATION

A. Permittivity Extraction and Accuracy Evaluation

The next step in evaluating the sensor is to characterize its accuracy in measuring the real permittivity of an MUT. The formula linking $\Delta f_r$ to $\Delta \varepsilon_r$ has been extracted using empirical fitting of the CST-simulated response of the sensor.
As the simulations and measurements were found to be in good agreement, as observed in Fig. 10 and 11, the measured response is expected to be in agreement with the full-wave simulation. A $\Re\{\varepsilon_r\}$ sweep from 1 to 90 was simulated. Fig. 12 and 13 show the $\varepsilon_r$ and $\Delta f_r$ relation for channels 1 and 3, respectively.

For channel 1, the relation observed in Fig. 12(a) has been fitted empirically using Matlab Curve-fitting toolbox. For channel 1, $\varepsilon_r$ is given by

$$\varepsilon_1 = n \times \frac{m_1 \Delta f_1^2 + m_2 \Delta f_1^{1.675} + m_3 \Delta f_1 + m_4}{m_5 \Delta f_1^3 + m_6 \Delta f_1 + m_7}$$

where $m_1$-$m_7$ were obtained using curve fitting as: $m_1=16.35$; $m_2=-17.55$; $m_3=1.764$; $m_4=0.2324$; $m_5=6.048$; $m_6=-2.075$; $m_7=0.1778$, resulting in a high $R^2 = 0.995$.

In Fig. 12(a), it can be observed that the fitted values (dashed line) agree closely with the CST-simulated resonances (discrete points). $n$ in (5) is a tuning variable to be used after filling the channels with DI water, to calibrate the sensor’s response after filling all the neighbouring channels with the MUT. In Fig. 12(b), it can be observed that the fitted relation exhibits a very high accuracy in evaluating $\varepsilon_r$ from $f_r$.

The same approach was repeated for the other channels, for channels 2 and 3, the relation is given by

$$\varepsilon_3 = p \times \frac{q_1 \Delta f_3^2 + q_2 \Delta f_3^{1.675} + q_3 \Delta f_3 + q_4}{q_5 \Delta f_3^3 + q_6 \Delta f_3 + q_7},$$

where $q_1$-$q_7$ were evaluated using the same curve-fitting approach based on the simulated values in Fig. 13(a).

As observed in Fig. 13(a) and (b), the very high curve-fitting $R^2=0.9992$ results in a near-zero error in evaluating $\Re\{\varepsilon_r\}$ from $\Delta f_r$.

Prior to filling the channels with the MUT, all channels were filled with DI water for calibration, and to evaluate the tuning parameters $n$ and $p$. Binary mixtures of IPA and water, at 100%, 75%, 50%, 25% and 0% IPA-to-water volume ratios were placed in each of the channels and $f_r$ was measured using the VNA. Fig. 14 shows the variation in $f_r$ for across the three channels, showing a very similar frequency-shift across all channels in response to the the variation in $\varepsilon_r$ of the different IPA/water ratios.

Using a simple polynomial curve fitting, it can be observed that the measured results are highly linear with $R^2 > 0.999$. The measured $\Delta f_r$ for DI water-loading was used to obtain the tuning parameters $n$, $p$, and $r$, for channels 1, 3, and 2, respectively. The fitted equations obtained from the 3D simulations were then used to evaluate the permittivity of the liquids based on the measured results, and the tuning parameters given by $n=1.11$ , $p=1.55$ , and $r=0.542$, for channels 1, 3, and 2, respectively. Fig. 15 shows the RF-measured $\Re\{\varepsilon_r\}$ of the binary mixtures, as evaluated by proposed the microwave microfluidic sensor.

In order to assess the accuracy of the measured values, the coaxial-probe-measured values for binary water/IPA mixtures from [14] were used. The Debye model was then used to calculate $\varepsilon_r$ for the frequencies of interest [36]. Fig. 15 shows the error percentage in the measured $\varepsilon_r$ compared to the coaxial-probe measurements in [14], fitted using the Debye
Apart from the 100%-IPA case, it can be observed that the sensor achieves at least 90% accuracy, which increases as the MUT approaches the permittivity of water. This is attributed to the utilized DI water-based calibration, which improves the sensor’s accuracy for $\epsilon_r > 20$. The maximum error of the proposed sensor is line with recently reported 3D-printed sensors [14], [22], which can only characterize a single MUT, compared in more detail in Section IV-C.

B. Microfluidic Temperature Sensing

As the permittivity of fluids is highly dependent on temperature, the proposed microfluidic sensor could be utilized to observe temperature variations through $\Delta f_r$. In this case, the MUT, e.g., water, will act as the temperature sensing element. Therefore, the temperature sensitivity could be improved by using an MUT with a high temperature $\epsilon_r$-dependence. As the previous experiments were carried out at room temperature, the 3D printed sensor was placed on a hot-plate as shown in Fig. 16(a). Due to the heat dissipation around the sensor, the temperature of the DI water sample inside the channel will be different from that of the hot plate’s surface. Therefore, differential temperature measurements were take, where $\Delta f_r$ is characterized relative to the hot-plate’s $\Delta T$.

The temperature has been varied from 60 to 85°C. The temperature variation $\Delta T$ has been increased in 5°C increments. As observed across the three channels, in Fig. 16(b)-(d), all stubs exhibited a linear relation between temperature and $f_r$, showing the suitability of the sensor. Across all three channels, the sensor exhibits a varying sensitivity ranging from $\Delta f_r = 2.4$ to 5 MHz/°C. Compared to a recent microwave microfluidic temperature sensor, realized for evaluating the temperature of a single MUT [37], the proposed sensor, with a water-filled channel, achieves over 10-fold improvement in the sensitivity to temperature, from approximately 0.25 MHz/°C in [37], to 2.4–5 MHz/°C. This highlights the benefits of utilizing a parallel-plate-based sensor with a high sensitivity, where the MUT’s $\epsilon_r$-temperature-dependence could be used in a high-sensitivity readout circuit based on the proposed sensor structure.

The temperature was not raised beyond 90°C due to the rated temperature of the cables (85°C) and the glass-transition temperature of the SLA substrate. Nevertheless, in a real-world temperature sensing application, commercially-available high-temperature 3D-printed resins such as Roboze EXTEN could enable 3D-printed substrates to retain their properties as high as 247°C. Moreover, using a read-out circuitry integrated on the same substrate, the need for external cabling and connectors is eliminated further improving the reliability and accuracy of the sensor, while reducing the test cost and time.

C. Discussion and Comparison with Recent Sensors

Observing the sensor’s accuracy in characterizing the relative permittivity of binary fluid mixtures across all channels, the proposed design can be applied to a range of applications where a single-connection/single-instrument setup could be used to characterize multiple fluids. Therefore, while single-connection microfluidic sensor calibration methods were proposed [33], there has been no complementary sensor design to-date, with the ability to characterize more than two materials with a single connection and read-out circuit.

Table I compares the fabricated prototype to recently reported microwave microfluidic sensors, either utilizing additive manufacturing or capable of sensing multiple fluids. While dual-mode resonators were proposed in [38] and [39], the proposed sensor is the only scalable multi-fluid microwave sensor proposed in literature. In addition to being the only reported sensor with >2 microfluidic channels and resonant elements, the proposed sensor’s accuracy and sensitivity is comparable to state-of-the-art sensors which utilize additive manufacturing. Moreover, the proposed sensor maintains the lowest complexity in its fabrication process, which can also scale in volume, similar to inkjet printed microfluidics [35]. To explain, while several sensors utilized 3D-printed channels
Their integration was either done manually using copper tape [14], unsuitable for scalable manufacturing [16], or with complex high-frequency PCB on an expensive substrate [22]. On the other hand, the direct-written printed microstrip line could be easily adjusted post-fabrication to account for tolerances in the microfluidic channel. Moreover, the simple open-ended stub design, combined with direct-write printing, enable the operation frequency of the sensor to be adapted in a simple and scalable manner, without requiring a substrate re-design as in [14]. Moreover, while a four-band sensor was proposed based on multiple complimentary SRRs [40], the sensor was only used for characterizing a single MUT.

V. Conclusion

In this paper, a scalable microwave microfluidic sensor design was proposed for characterizing the real relative permittivity of multiple fluid samples. The simple design utilizing the parallel-plate capacitance of microstrip stub enables simple analytical and numerical analysis, and is enabled by hybrid low-cost additive manufacturing. The proposed sensor has been experimentally characterized based on binary water/IPA mixtures showing at least 90% accuracy in determining the real-permittivity across all channels. The accuracy and sensitivity of the proposed sensor are in line with state-of-the-art sensors while significantly improving the throughput and testing speed, owing to the utilization of three different channels. The proposed sensor was also demonstrated as a microfluidic temperature sensor exhibiting a high linearity and better sensitivity compared to similar microfluidic temperature sensors. The proposed design is expected to enable future low-cost fast-turnaround microfluidic sensors owing to its highly simplified microstrip structure, and due to utilizing low-cost hybrid 3D and dispenser printing.

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Janet Shi received a B.Sc. degree from the Nanjing Institute of Technology, China in 2008, an M.Sc. degree in Microelectromechanical systems in 2012, and a Ph.D. in 2017 at the University of Southampton. He was appointed as a Research Fellow in the School of Electronics and Computer Science (ECS) in 2017. His research interest covers a broad range of research areas, including antennas and microwave systems, RF front-end design, and antennas for 5G and beyond.

Mahmoud Wagih (GS’18, M’21) received his B.Eng. (Hons.) from the University of Southampton in September 2018, and his Ph.D. in April 2021. In 2017 he worked as a Research Assistant at the University of Southampton. In 2018, he was a Hardware Engineering Intern at Arm, and, in 2020, a Research Intern at Arm, Cambridge, U.K. He is currently a Research Fellow at the University of Southampton, U.K., leading a Flexible Innovation Fund project on Green RF Sensing. His interests broadly cover antennas and microwave systems in energy harvesting, sensing, and wearable applications. He has over 40 refereed journal and conference publications, and has delivered several invited webinars on these topics.

Dr. Wagih is a member of the Institute of Engineering and Technology (MIET). He was the recipient of the Best Undergraduate Project Prize, School Winner Doctoral Research Award, Best in Faculty Doctoral Research Award, and the Dean’s Award for Early Career Researchers, in 2018-2021, at the University of Southampton. He was selected for the IEEE International Microwave Symposium Project Connect in 2019. He received the Best Student Paper Award at the IEEE Wireless Power Transfer Conference, 2019, the Best Oral Paper at PowerMEMS, 2019, and was a Best Student Paper Finalist at IEEE WPTC, 2021. He won the MTT-S Best 3MT Presentation Prize (second place) at the IEEE Microwave Week, 2020, and was a session co-chair at the European Conference on Antennas and Propagation, 2021. He acts as a reviewer for nine TRANSACTIONS and journals. He is an affiliate member of the IEEE MTT-S, Wireless Sensors and IoT Technical Committee (TC-26).