Wearable Slot Antenna for Biomedical Applications: Mutual Coupling and External Interference

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Abstract. A small slot antenna has desirable characteristics for radio communications and location of an internal transceiver in-vivo medical applications. The effect of coupling between two identical antennas on the human torso was measured between 2.1 GHz antennas on the skin surface. The effect of an external field was measured as a function of the angle in the horizontal plane to quantify noise isolation. The perimeter separation loss was approximately 0.25 dB/mm. The external radio source induces currents in the soft conducting tissue resulting in a sinc radiation pattern for the antenna/body combination with a front-to-back ratio of approximately 12 dB. As the UHF band is commonly used in many non-medical applications, there is concern that external radio sources can result in a reduced signal to noise ratio and perturbed field strength measurements on the skin.

Keywords
Wearable slot antenna, electromagnetic interference, noise, mutual coupling, surface fields

1. Introduction

A slot in a small conducting box has been suggested as a convenient wearable antenna for measuring the radio field on the skin surface from a subsurface transmitter [1]. This paper addresses two important characteristics of the antenna not yet reported, (a) the mutual coupling between two antennas on the skin surface, and (b) the effect of an in-band radio source on the field strength measurement and communications reliability.

The propagation of electromagnetic signals on and near the surface of anisotropic, lossy media is a challenging problem, particularly when there are multiple layers with different electromagnetic properties. A simple conducting wire at the air/material interface has a complex propagation coefficient related to the electromagnetic properties of the materials on both sides of the boundary. While a closed form solution exists for a plane interface with uniform properties [2], numerical modelling is required for more complex structures [3]. When the transmitter is embedded in the lossy media, then lateral wave propagation effects can be observed on the surface evidenced by significant interference nulls [2], [4]. This interference is caused by the difference in the path attenuation and the time of arrival for the direct ray through the material and a surface wave propagating across the surface. When one half space is layered, then trapped and ducted propagation paths can exist. This effect is commonly observed in clear air propagation caused by temperature and humidity variations in layers close to the earth’s surface [5].

The simplest approach for wearable antennas on the skin is to assume that the soft tissue is uniform [6], [7]. Fermat’s principle of least time on propagation paths is characterized by the phase constant $\beta$. The radio ray paths with the least travel time have the strongest signal. For lossy media (defined when the conductivity $\sigma$ is not zero), the complex propagation coefficient $\gamma$ is given by [8] as

$$\gamma = (\alpha + j\beta) = j\omega\mu\sigma - \varepsilon\omega\mu$$

where $\alpha$ is the attenuation coefficient, $\omega$ is the angular frequency of the radiation, and $\varepsilon$ and $\mu$ are the electric permittivity and magnetic permeability of the medium, respectively. The larger the ratio $R$,

$$R = \frac{\sigma}{\omega\varepsilon}$$

the higher the conductivity of the medium, the smaller is the skin depth ($\delta = 1/\alpha$) and the slower the phase velocity ($u$).

Human soft tissue is inhomogeneous and multi-layered. The electromagnetic properties are different for every layer and every individual. These properties are also frequency dependent. On the torso, the common tissue layers are skin, fat, muscle and the internal organs, and all have different electromagnetic properties [9]. Table 1 outlines some published data on these layers at UHF including the conductivity and permittivity values, their respective $R$ values, skin depth $\delta$ and phase velocity $u$. At UHF frequencies, these parameters lie in the intermediate range between a good conductor and a low loss dielectric so that lateral wave propagation is possible.

The significant difference (5 times the skin depth and 20% lower phase velocity) between the fat and the other
The electromagnetic properties of soft human tissue in-vivo at UHF (from [9]) with the skin depth and phase velocity.

| Tissue type | $\sigma$ (S/m) | $\varepsilon_r$ | $\delta$ (m) | $\mu$ $10^8$ m/s |
|-------------|---------------|----------------|-------------|-----------------|
| skin        | 1.46          | 38.0           | 0.33        | 0.022           | 0.486           |
| fat         | 0.10          | 5.28           | 0.17        | 0.118           | 1.305           |
| muscle      | 1.74          | 52.72          | 0.28        | 0.022           | 0.413           |
| organs      | 2.04          | 53.88          | 0.32        | 0.019           | 0.408           |

Tab. 1. Electromagnetic properties of soft human tissue in-vivo at UHF (from [9]) with the skin depth and phase velocity.

Tissues indicates that ducting of the radio signals in this layer is possible. For every individual, the layers vary in thickness and orientation making numerical modelling a significant challenge.

Wearable antennas have many communications applications in sport, entertainment and in the medical industry as body sensor networks. Since the antenna operates in close proximity or directly attached to the human body, compact design is important for wearer comfort and ease of use [10], [11]. Common requirements of a wearable antenna include a low profile, lightweight, mechanical flexibility, time-saving fabrication, and low cost [12].

The position of the antenna on the body depends on the application. For example:

- The communication channel between the body and an off-body transceiver requires minimal contact and electromagnetic isolation from the skin. This type of remote human monitoring is of interest in sport, military and medical applications [13].
- On-body (OB) communications antennas are directly attached to the human body and operate in close proximity with other body worn or implanted antennas [14], [15].
- In-body (IB) antennas are those imbedded or implanted inside the body. They communicate through the human tissue and are used in medical applications [16].

One example of an on-body antenna is a linearly polarized slot antenna where the slot is pressed against the skin for communications with an internal transmitter [1]. One might assume that this antenna, when tuned for soft tissue contact, would be immune from external sources and other antennas mounted on the body. In this paper we report the performance of this type of antenna on the body. These effects have not been explored in detail previously.

An object in the near field of an antenna becomes part of the radiating structure. If the material is lossy, then the antenna efficiency is reduced. If the object is large, then the radiation pattern has an increased aperture size and so a larger number of side-lobes. Both effects are not well characterized as the human torso is multilayered with skin, fat, muscle and organs.

Given the common use of the 2.45 GHz UHF band for medical applications, as well as for Wi-Fi and Bluetooth communications, the influence of external sources on the total field measured was investigated numerically and experimentally. To avoid stray noise sources, the investigations were conducted at 2.1 GHz using a slightly modified slot antenna tuned for this frequency. This frequency is outside of the internationally defined UHF ISM frequency band.

This paper addresses two important questions:

1. Can the mutual coupling between two antennas on the torso result in interference and inaccurate measurements of the signal level? and
2. What is the effect of external sources on the signal received by a surface mounted, inwardly directed, slot antenna?

Two sets of measurements were made in-vivo on a male participant standing vertically.

2. Methods and Procedures

2.1 Antenna Design

The inward-looking slot antenna shown in Fig. 1 [1], was reconfigured to resonate at 2.1 GHz ($S_{11} = -24$ dB) when pressed against the soft tissue of the human torso in-vivo. The antenna box was machined aluminum with a resonant slot internally coupled with a monopole probe made from brass sheet. The antenna input impedance was approximately $50 \Omega$ and all measurements were made using a vector network analyzer.

The path loss from an internal radio transmitter can be determined from the line-of-sight path, but with surface propagation also a possibility. If the material is relatively low loss, an interference pattern can be observed on the surface as lateral wave effect [3], [4].

![Fig. 1. The slot antenna consisted of a machined aluminum box with a brass, monopole feed. The antenna was optimized in software to resonate at 2.1 GHz when placed on the human torso: (a-b) box and slot dimensions; (c) monopole dimensions; (d) antenna prototype with a 50Ω SMA connector.](image-url)
The effectiveness of the slot antenna shown in Fig. 1 has previously been reported in on/off-body applications. Some examples include the relationship between the radiofrequency absorption and antenna efficiency at 2.45 GHz, for different locations of the antenna on the human body [17], the power absorbed by various body tissues, and the total power absorbed in adult male and female body models [18]. The electric field distribution on the skin surface was measured on multi-layered porcine tissue [19], where two identical slot antennas tuned at 2.45 GHz were pressed against the tissue. The S\textsubscript{21} parameter was reported at both epidermal and subdermal locations of the porcine tissue.

### 2.2 Measurements and Modelling

Two antennas were placed directly on human skin (Fig. 2). The measurement was made on the torso of a male participant (age 69 years old, weight 91 kg, height 1.83 m and BMI 27.2 kg/m\textsuperscript{2}). The measurements were conducted in accordance with Griffith University’s ethics research committee approval (GU ref no:2018/601). Table 2 lists the marked locations on the skin with a description and thickness values.

Ultrasound imaging using a GE LOGIQe apparatus, was used to determine the soft tissue thickness values of skin and fat of the trunk at all 7 locations listed in Tab. 2. This provided precise morphological information on the anatomical structure of the participant.

All positions were located under the ribcage and above the iliac crest to avoid scattering from bone. The reflection coefficient S\textsubscript{11} was less than –5 dB at the resonant frequency of 2.1 GHz for all locations. In Fig. 2, the operator hand holding the antenna in position did not influence the S\textsubscript{11} measurements nor provided discomfort to the participant.

#### 2.2.1 Mutual Coupling

The mutual coupling measurements were recorded at distances between two identical slot antennas. One antenna was located at location 1 for the first measurement, and then at location 2. The second antenna 2 was placed systematically at the other locations. The two antennas were always parallel and in direct contact with the skin (Fig. 2) and oriented for vertical polarization.

The mutual coupling between the slot antennas was modelled numerically using the electromagnetic FDTD software suite CST Microwave Studio\textsuperscript{®}. The multilayer human tissue was modelled as a planar surface using the tabulated electromagnetic properties of human tissue at 2.45 GHz [9]. The body layers included: front surface skin, fat, muscle, soft tissue, fat, back surface skin, and were modelled with the actual thickness obtained from the ultrasound measurements (Fig. 3). The muscle and the inner soft tissue were modelled with a thickness of 100 mm, which was demonstrated to be effectively infinite for S\textsubscript{11} measurements (skin depth approximately 23 mm).
2.2.2 External Interference

To imitate the effect of an external noise source, a standard gain horn antenna (vertical polarization, center frequency 2.5 GHz, aperture size 230 mm × 150 mm) was directed towards a slot antenna positioned centrally on the front of the torso (Fig. 4). The antenna was attached at position 2 (Tab. 2) using a Velcro™ band (this was shown to have no effect on the measurements results).

Two measurement sets were recorded:

(a) The transmission coefficient (S21) was recorded at 13 different separation distances (0.35 m–3.41 m).

(b) With the participant standing on a vertical axis, on a 360° rotator, the S21 (horn to slot) was recorded at each angle θ from 0° to 360° in 10° steps (see Fig. 5(a)). The distance between the wearable antennas to the rotation axis was 100 cm (Fig. 5(b)).

3. Results and Discussion

The mutual coupling simulations and experimental results are shown in Fig. 6. A slight offset in S21 is evident at different circumferential distances between the measured and modelled data. This can be linked to the complex human anatomy and the variations in the dielectric properties of the human tissue not captured in the simulation. This difference is most evident when one antenna is located at position 3. This is thought to be due to the complex tissue structure at this location. The coupling is postulated to be due to partly trapped radiation by the skin layers and so, to some extent, follows the curvature of the torso. The S21 values recorded varied approximately linearly with distance with an attenuation of −0.25 dB/mm (r² = 0.79). The attenuation coefficient is likely to vary between individuals because of anatomical differences (which is not the focus of this paper).

The effect of the external radiation source as a function of distance is shown in Fig. 7. The noise level was approximately −75 dB at 2.1 GHz. The received signal was recorded as a function of the human rotation angle θ about the vertical axis, where θ = 0° is the front of the human torso (see Fig. 8). While this measurement was not in the far-field zone of the horn, the signal strength has maximum power at the angle 0° i.e., when antenna on the body is facing the reverse direction. The combination of the human body and the slot antenna results in this principal radiation direction. The radiation pattern F(θ) of the slot antenna with the human body was fitted to the sinc aperture function [8] common in antenna radiation pattern analysis, but modified to cover the full 360° as

\[ F(\theta) = \sin^2 \left( 0.9 \pi \sin \left( \frac{\theta}{2} \right) \right). \]  

While the fit is approximate, the maximum radiation sensitivity lies in the angle range ±90° and the half power beam-width is approximately 110°. There is also some evidence of a possible lateral wave interference pattern, as minor sidelobes and nulls are present in the measured data. This lateral wave effect was previously demonstrated in field attenuation at 2.4 GHz [20]. The nulls in the theoretical curve cannot match the experimental measurements as the antenna aperture is not infinitely thin.
Cardiac, respiratory, muscular and digestive systems can be monitored using external sources such as mobile devices. These external sources, combined with the human normal living environment and likely use a mobile telephone, might be used for applications involving intra-torso cavity communications for other biomedical applications. The presence of the fat layer with a lower conductivity can impact significantly on the UHF signal quality used for radio pill location.

As the passage through the gut can take up to 72 hours, an ingested transmitter in radio-endoscope measurements are important for determining the location of movements associated with normal biological functions. The major source of external interference occurs when the antenna is closest to the noise source although the impact significantly on the UHF signal quality used for radio pill location.

In the current radio environment, the use of UHF signals in the ISM band (Industrial, Scientific and Medical) coincides with both the Wi-Fi and Bluetooth frequencies of radio pill transmitters in common medical use. It is highly likely that a patient will move past Wi-Fi terminals in the normal living environment. As the passage through the gut can take up to 72 hours, a wearable sensor array is required so that some degree of normal living is possible.

The major source of external interference occurs when the antenna is closest to the noise source although the slot antenna is inwardly directed. While the front to back ratio of the slot antenna is > 40 dB in air, when the antenna is pressed against the skin, induced currents result in significant field detection from external and adjacent sources. The presence of the fat layer with a lower conductivity was thought to provide a radio duct inside the body resulting in lateral wave propagation and interference in the mutual coupling measurements. The importance of this effect is evident on the human subject used in these measurements and is likely to be particularly important for morbidly obese individuals where the fat layers are much thicker. While the signal levels recorded in these measurements are relatively low, it is clear that the slot antenna is not completely isolated from the environment. It is recommended that during the measurement of signal strength for radio pill location, the nearby UHF sources of the radiation are removed. RSSI measurements based on the coded signal strength can minimize this effect.

Further study into the ducting effect through fat layers might be useful for applications involving intra-torso cavity communications for other biomedical applications.

4. Conclusion

An inward-looking, resonant, slot antenna was designed for impedance matching to the human torso at 2.1 GHz. Two important characteristics of this antenna were assessed experimentally on a human subject. These were the coupling between two identical slot antennas placed on the soft-tissue region of the torso, and the susceptibility of the antenna to an external noise source. Both measurements are important for determining the location of an ingested transmitter in radio-endoscope measurements.

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