Adaptive Control with Guaranteed Transient Behavior and Zero Steady-State Error for Systems with Time-Varying Parameters

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Abstract—It is nontrivial to achieve global zero-error regulation for uncertain nonlinear systems. The underlying problem becomes even more challenging if mismatched uncertainties and unknown time-varying control gain are involved, yet certain performance specifications are also pursued. In this work, we present an adaptive control method, which, without the persistent excitation (PE) condition, is able to ensure global zero-error regulation with guaranteed output performance for parametric strict-feedback systems involving fast time-varying parameters in the feedback path and input path. The development of our control scheme benefits from generalized $t$-dependent and $x$-dependent functions, a novel coordinate transformation and “congelation of variables” method. Both theoretical analysis and numerical simulation verify the effectiveness and benefits of the proposed method.

Index Terms—Guaranteed performance, uncertain nonlinear systems, adaptive control, global property

I. INTRODUCTION

We consider the following SISO nonlinear systems with fast time-varying parameters $\Theta$

$$\begin{cases}
\dot{x}_1 = \phi_1^T(x_1)\theta(t) + x_2 \\
\vdots \\
\dot{x}_i = \phi_i^T(x_i)\theta(t) + x_{i+1} \\
\vdots \\
\dot{x}_n = \phi_n^T(x_n)\theta(t) + b(t)u \\
y = x_1
\end{cases}$$

(1)

where $x_i = [x_1, \ldots, x_i]^T \in \mathbb{R}^i$ is the state vector, $u \in \mathbb{R}$ is the input, $y \in \mathbb{R}$ is the output. The regressors $\phi_i : \mathbb{R}^i \rightarrow \mathbb{R}^q$, $i = 1, \ldots, n$, are smooth mappings and satisfy $\phi_i(0) = 0$. $\theta(t) \in \mathbb{R}^n$ and $b(t) \in \mathbb{R}$ satisfy the following Assumptions 1-2.

Assumption 1 (Bounded parameters): The parameter $\theta(t)$ is piecewise continuous and $\theta(t) \in \Theta_0$, for all $t \geq 0$, where $\Theta_0$ is a compact set. The “radius” of $\Theta_0$ is assumed to be known, while $\Theta_0$ can be unknown.

Assumption 2 (Sign-definite parameter): The control gain $b(t)$ is bounded away from zero in the sense that there exists a constant $\ell_b$, such that $\text{sgn}(\ell_b) = \text{sgn}(b(t)) \neq 0$ and $0 < |\ell_b| \leq |b(t)|$, for all $t \geq 0$. The sign of $b(t)$ is known and does not change.

Stabilization of system (1) satisfying Assumptions 1-2 is originally investigated in [1]-[3], where it is shown that asymptotic stability can be achieved by the so-called "congelation of variables" method and both full state feedback and partial state feedback approaches are considered. By "congelation of variables" it means that the time-varying $\theta(t)$ can be substituted by constant $\ell_\theta$ ($\ell_\theta$ can be regarded as the average of $\theta(t)$) to avoid unnecessary time derivatives while not destroying the certainty equivalence principle [6]. It is noted that if the parameter $\theta(t)$ in (1) is unknown but constant, numerous adaptive control results have been reported in literature during the past decades, including the well-known adaptive backstepping control, robust and adaptive control, adaptive observers, immersion and invariance adaptive control, neural adaptive control, etc. (see [6]-[12] and the references therein).

However, real-world engineering systems with fast time-varying parameters are frequently encountered. For instance, the value of a circuit resistor might change with temperature, some morphing aerial vehicles are normally designed with varying structures and parameters in order to complete some specific tasks, where the parameters might change with time or system states swiftly [13], [14]. For this type of systems, traditional adaptive methods might not be able to ensure desired control performance in terms of transient behavior and convergence accuracy, or even unable to maintain system stability. Efforts have been made (see, for instance [15] and [16]) in developing adaptive control methods with the aid of the persistence of excitation (PE) to achieve exponential stability of linear time-varying systems. In [17], it is shown that the PE condition is not necessary to stabilize a linear time-varying system. The results in [18] and [19] implement the asymptotic/exponential tracking of robotic systems with/without time-varying parameters. In [20]-[22], along with observer-based adaptive control, projection algorithm is proposed to ensure the boundedness of slow time-varying parameter estimate. In the context of adaptive control for time-varying nonlinear systems, the work [23] explores a soft sign function based approach to deal with unknown time-varying parameters. Recently, an elegant method based on "congelation of variables" is proposed in [11]-[13] to asymptotically stabilize a class of nonlinear system with fast time-varying parameters, which is further extended to address multi-
agent systems in [4] and [5]. Thus far, meaningful results on adaptive control of systems with unknown and fast time-varying parameters are still limited, rendering the underlying problem interesting yet challenging.

In this note, we address the stabilization problem of fast time-varying system as described in [1] and our goal is to achieve zero-error full state regulation and at the same time maintaining global output performance, i.e., regulating each state to zero asymptotically and meanwhile confining the convergence process of the output within an prescribed boundary. Our development consists of three major steps: i) disassociating the recursive controller design from the initial condition of system [1] via two generalized functions and a novel coordinate transformation; ii) designing adaptive laws to estimate fast time-varying parameters involved in the constrained systems; and iii) separating the lumped nonlinear terms and exploiting additional nonlinear damping in each virtual control input to finally offset the undesired perturbations caused by unknown time-varying control gain. With this comprehensive treatment, output convergence transient behavior is well preset and asymptotic (zero-error) regulation is achieved in the presence of mismatched time-varying uncertainties.

Unlike most prescribed performance control methods that only achieve uniformly ultimately bounded (UUB) for nonlinear systems with unknown but constant parameters [30–41], the proposed method ensures zero-error stabilization and global output performance for systems with fast time-varying parameters and mismatched uncertainties.

II. PRELIMINARIES

A. Two Useful functions & Coordinate transformation

Before presenting the control algorithm, we introduce two useful functions and a novel coordinate transformation, which plays important roles in control design.

Definition 1: The generalized performance function \( \beta(t) \) satisfies the following properties:

- \( \beta(t) : [0, \infty) \rightarrow \mathbb{R}^+ \) is a \( n \)-times differentiable function;
- \( \beta(0) = 1 \) and \( \lim_{t \to +\infty} \beta(t) < 1 \);
- \( \beta(t) \in \mathcal{L}_\infty \) and \( \dot{\beta}(t) \in \mathcal{L}_\infty \), \( \forall t \in [0, +\infty) \).

Remark 1: There are many (in fact, infinite number of) functions that satisfy the aforementioned properties. For example,

\[
\beta(t) = \begin{cases} 
(1 - \beta_\infty) \left( \frac{T-t}{T} \right)^n + \beta_\infty, & 0 \leq t < T \\
\beta_\infty, & t \geq T;
\end{cases}
\]

where \( \beta_\infty = \lim_{t \to +\infty} \beta(t) \), \( T > 0 \) is a constant and \( n \) is the system order. Note that the performance function is not necessarily monotonically decreasing, which might be advantageous in various applications, e.g., when the system time-varying parameter changes strongly or the system is perturbed by some calibration so that a large error would enforce a large input action.

Definition 2: The generalized normalized function \( \psi(x) \) satisfies the following properties:

- \( \psi(x) : \mathbb{R} \to (-1, 1) \) is a monotonically increasing and \( n \)-times differentiable function;
- \( \lim_{x \to \pm\infty} \psi(x) = \pm 1 \) and \( \psi(0) = 0 \);
- \( \psi'(x) \) is bounded below by a positive constant over \([0, \infty)\), where \( \psi'(x) = \frac{d\psi}{dx} \).

Remark 2: We list two choices for \( \psi(x) \) as follows:

\[
\psi(x) = \frac{x}{\sqrt{x^2+1}}, \quad \psi(x) = \tanh(x), \quad (3)
\]

and for the above two choices, we have:

\[
\psi'(x) = \frac{1}{(x^2+1)^{3/2}}, \quad \psi'(x) = \text{sech}^2(x). \quad (4)
\]

Denoting the inverse function by \( \psi^{-1} \), it is seen that

\[
\psi'(x) > 0, \quad \psi_x = \frac{\psi_x}{x} > 0, \quad \psi^{-1}(\beta(0)) = \psi^{-1}(1) = +\infty. \quad (5)
\]

Making use of such \( \beta(t) \) and \( \psi(x) \), we construct the following coordinate transformation function to enable the properties on \( z \) and \( x \) as stated in Lemma 1

\[
z(\beta, \psi) = \frac{\beta(t)\psi(x)}{\beta^2(t) - \psi^2(x)}. \quad (6)
\]

Lemma 1: For any \( \beta(t) \) as defined in Section II and \( z \) as defined in (6), if \( \forall t \geq 0, z \in \mathcal{L}_\infty \), then it holds that \(-\psi^{-1}(\beta) < x < \psi^{-1}(\beta)\).

Proof: We first consider the moment when \( t = 0 \). According to \( \beta(0) = 1 \) and \( \psi(x) \in (-1, 1) \), we know that \( \beta(0) - \psi(|x(0)|) > 0 \), i.e., \( |x(0)| < \psi^{-1}(\beta(0)) \). Next, we continue the proof by contradiction. Note that \( z \in \mathcal{L}_\infty \) implies \( \beta(t) - \psi(x) \neq 0 \). Assume that \( \exists t \in (0, \infty) \) such that \( |x(t)| \geq \psi^{-1}(\beta(t)) \), i.e., \( \beta(t) - \psi(|x(t)|) \leq 0 \). As a result, by recalling that \( \beta(0) - \psi(x(0)) > 0 \), we have \( \exists t_1 \in (0, t] \) causes \( \psi(|x(t_1)|) = \beta(t_1) \), and therefore yields an unbounded \( z_1 \), which, however, contradicts the premise \( z \in \mathcal{L}_\infty \). This completes the proof.

This coordinate transformation introduced in (6) appears as a more straightforward approach compared to the tuning function modified transformation [34] and the multiple cascade transformation [28], by reason of its simple structure, smoothness and nonsingularity.

B. Control Objective

The control objective is to design an adaptive control law such that the closed-loop system is asymptotically stable, while the system output is always confined within a prescribed performance funnel \( F_\beta(t) \). Furthermore, the boundary of \( F_\beta(t) \) is \( \beta(t) \), which can be pre-defined at user’s will, irrespective of initial conditions.

Remark 3: If we choose a function \( \beta(t) \) with an exponential decay rate, e.g., \( \beta(t) = (1 - \beta_\infty)e^{-\epsilon t} + \beta_\infty \). By qualitative analysis, \( \psi^{-1}(\beta) \) is a function that increases monotonically as \( \beta \to \infty \), and \( \beta(t) \) is a function that decays exponentially as \( t \to \infty \), thus \( \psi^{-1}(\beta) \) is a function that decays exponentially as \( t \to \infty \) and \( \psi^{-1}(\beta(0)) \to \infty \). Therefore, \( |x| < \psi^{-1}(\beta) \) implies that there exist some positive constants \( l_1, l_2 \) and \( \epsilon \) such that \( |x(t)| < l_1 e^{-l_2 t} + \epsilon \) for any \( x(0) \), resulting in that the system output converges at least \( e^{-l_2 t} \) exponentially fast to

\[1\]Property 1 and Property 2 of \( \psi(x) \) ensure that \( \psi_x \) is positive and invertible for all \( x \in \mathbb{R} \).
the corresponding set. Similarly, if we choose $\beta(t)$ as defined in (2), one can find that the system output converges to a prescribed set at a prescribed time $T$, a favorable feature in practice.

III. Motivating Example

Consider the following first-order system

$$\dot{x} = b(t)u + \theta(t)x$$

(7)

where $x$ is the state, $u$ is the control input, $\theta(t) \in \mathbb{R}$ satisfies Assumption 1, and $b(t) \in \mathbb{R}$ satisfies Assumption 2.

By using the coordinate transformation (6), we can convert (7) into the following z-dynamics

$$\dot{z} = \Pi(x,t)\dot{x} + \Psi(x,t)$$

(8)

with

$$\Pi(x,t) = \frac{(\beta^2(t) - \psi^2(x))^2}{(\beta^2(t) - \psi^2(x))^2}$$

$$\Psi(x,t) = \frac{\beta(t)\psi(x)(\beta^2(t) - \psi^2(x)) - 2\beta^2(t)\beta(t)\psi(x)}{(\beta^2(t) - \psi^2(x))^2},$$

where $\Psi$ and $\Pi$ are known time-varying smooth functions and are bounded as long as $z$ is bounded. In addition, $\Pi > 0$ for $\forall z \in \mathcal{L}_\infty$. These facts ensure the controllability of (8).

Motivated by (11), we design $u = \hat{\rho}u$, with $\hat{\rho}$ being an “estimate” of $\rho$, and $\bar{u}$ being the compensating signal to be specified later, then (8) can be written as

$$\dot{z} = \Pi(\hat{u} + \bar{u} + \theta(t) - \bar{\ell}_b)x + (\ell(t) - \ell_b)\bar{u}$$

(9)

where $\hat{\ell}$ is an “estimate” of $\ell$, $\hat{\Pi}/\Pi \in \mathcal{L}_\infty$ for $\forall z \in \mathcal{L}_\infty$. Note that $\ell_0$ and $\ell_b$ are unknown constants, which can be regarded as the “average” of $\ell(t)$ and $b(t)$, respectively. Consider the Lyapunov function candidate

$$V = \frac{1}{2}z^2 + \frac{1}{2}\ell_0(\ell_b - \hat{\ell})^2 + \frac{\ell_b}{2\gamma_0} \left( \ell_b - \hat{\ell} \right)^2$$

10.

Then, the derivative of (10) along the trajectory of (7) becomes

$$\dot{V} = \Pi(z\hat{u} + \theta(t) - \bar{\ell}_b)x + (\ell(t) - \ell_b)\bar{u}$$

(11)

where $\Delta_\theta = \theta(t) - \ell_0$ and $\Delta_b = b(t) - \ell_b$. The last two lines of (11) will be canceled by the following adaptive laws:

$$\dot{\theta}(x, \beta) = \gamma_\theta z\Pi x$$

(12)

$$\dot{\hat{\rho}} = -\gamma_\rho\text{sgn}(\ell_b)z\Pi\bar{u}$$

(13)

Remark 4: Note that $z(x, \beta)$ as defined in (6) is a smooth function and $x = 0 \iff z = 0$, thus we can directly express $x$ as $x = W(x, \beta)z$ by using Hadamard’s Lemma (see [11, 43], [26]), where $W(x, \beta)$ is a bounded smooth mapping for every bounded $z$. As a matter of fact, here $W = \frac{z}{\sqrt{\beta^2 - \gamma^2}} = \frac{\rho^2 - \gamma^2}{\beta^2 - \gamma^2} \in \mathbb{R}^+.$

According to Remark 1 and formula (5), the perturbation terms in the first line of (11) can be rewritten as

$$\dot{\theta}(x, \beta) = \left( \theta + \frac{\Psi(x)}{\Pi} \right) W(x, \beta)z^2$$

(14)

$$\Delta_\theta \dot{x} = \Delta_\theta W(x, \beta)z^2.$$

By applying Young’s inequality, then

$$\left( \theta + \frac{\Psi(x)}{\Pi} \right) W(x, \beta)z^2 \leq \frac{1}{2} \left( \theta + \frac{\Psi(x)}{\Pi} \right)^2 W^2z^2 + \frac{1}{2}z^2,$$

(15)

$$\Delta_\theta W(x, \beta)z^2 \leq \frac{1}{2} \delta_{\Delta_\theta} W^2z^2 + \frac{\delta_{\Delta_\theta}}{2}z^2,$$

(16)

where $\delta_{\Delta_\theta} \geq |\Delta_\theta|$ is the “radius” of the compact set of $\theta(t)$.

Now consider $\bar{u}$ with a nonpositive nonlinear gain as

$$\bar{u} = -\left( k_1 + \frac{1}{2}(\delta_{\Delta_\theta} + 1) + \frac{W^2}{2} \left( \theta + \frac{\Psi(x)}{\Pi} \right)^2 \right) z$$

(17)

where $k > 0$.

We are now in the position to state the following theorem.

Theorem 1: System (4) with the control law (17) and the parameter update laws (12) and (13) is globally asymptotically stable. Furthermore, the state $x(t)$ is always confined within the prescribed performance funnel $F_{\beta} := \{(t, x) \in \mathbb{R}_{\geq 0} \times \mathbb{R} | |x(t)|/\psi^{-1}(\beta(t)) < 1\}$, and ultimately converges to zero. Furthermore, $\lim_{t \to \infty} \theta$ and $\lim_{t \to \infty} \beta$ exist (although not necessarily equal to $\ell_0$ and $1/\ell_b$, respectively). In addition, the control input and update laws remain uniformly bounded over $(0, \infty)$.

Proof: Substituting (17) into (11), yields

$$\dot{V} \leq -kz^2 + \Pi(z\Delta_\theta x - \frac{1}{2}\delta_{\Delta_\theta} W^2z^2 - \frac{\delta_{\Delta_\theta}}{2}z^2)$$

$$+ \Pi(z\hat{\theta}x - \frac{\beta_0}{\beta}z - \frac{W^2}{2} \left( \theta - \frac{\beta_0}{\beta} \right)^2 z^2 - \frac{1}{2}z^2)$$

(18)

Then, substituting (17) into (13) yields

$$\dot{\hat{\theta}}(t) = \gamma_\theta \Pi_{x}(t)\gamma^2z^2,$$

where $\Pi > 0$ and $k(\gamma, x, \beta, \hat{\theta}) > 0$. When $b(t) > 0$, according to Assumption 2, we can obtain $0 < b(t) < b(t)$ and thus $\text{sgn}(\ell_b(t)) > 0$ and $\Delta_\theta > 0$, implying that $\hat{\rho}(t) \geq 0$. It follows from $\hat{\rho}(t) > 0$ that $\hat{\rho}(t) > 0$, and therefore $\Delta_\theta \hat{\rho}u = -\gamma_{\Delta_\theta} \Delta_\theta \hat{\rho}u < 0$. Similarly, when $b(t) < 0$, according to Assumption 2, we can obtain $\Delta_\theta < 0$, $\text{sgn}(\ell_b(t)) < 0$, $\dot{\theta}(t) \leq 0$ and therefore $\Delta_\theta \hat{\rho}u = -\gamma_{\Delta_\theta} \Delta_\theta \hat{\rho} < 0$ by selecting the initial condition $\rho(0) < 0$. Recalling (12) and (13), and noting the fact $-\Pi \Delta_\theta \hat{\rho}u \leq 0$, it can be concluded that for any bounded initial $z(0)$, $V(t) \leq \dot{V}(0)$, which yields $z(t), \hat{\theta}, \hat{\rho}$, and $W(x, \beta)$ are bounded.
The boundedness of $\Pi$, $1/\Pi$, $\Psi$ and $\kappa$ is guaranteed by the boundedness of $z$ and $\beta(t)$, it follows from (9) and (13) that $\dot{z} \in L_\infty$ and $z \in L_2$. Therefore, invoking Barbalat’s lemma one can conclude that $\lim_{t \to \infty} z(t) = 0$, which further indicates that $\lim_{t \to \infty} x(t) = 0$, hence the closed-loop system (7) is asymptotically stable. Furthermore, by using Lemma 1, we have $x(t) \in F_\beta = \{(t,x) \in \mathbb{R}_+ \times \mathbb{R} | x(t) \in \mathbb{R}^1 \}$. 

To show the asymptotic constancy of $\theta$ and $\hat{\rho}$, recalling (12), (13), (18) and the fact that $z \in L_2$, we have $\theta \in L_1$ and $\hat{\rho} \in L_1$. Then, by using the argument similar to Theorem 3.1 in [27], it is concluded that $\theta$ and $\hat{\rho}$ have a limit as $t \to \infty$. Furthermore, it is seen from (12), (14) and (17) that the update laws $\theta \in L_\infty$, $\hat{\rho} \in L_\infty$, and the control input $u = \hat{\rho} \bar{u} \in L_\infty$. This completes the proof.

IV. DESIGN FOR HIGH-ORDER TIME-VARYING SYSTEMS

Motivated by the design process for the first-order system, we now explore its applicability to more general high-order system as described in [1]. For such strict-feedback system, we use classical backstepping method [25], with additional special treatment in each step, as detailed in what follows:

Step 1: Let $\alpha_1 = x_2 - z_2$ and according to (6), we can convert $\dot{x}_1 = \phi_1^T \theta(t) + x_2$ into the following $z_1$-dynamics

$$\dot{z}_1 = \Pi \left( \alpha_1 + z_2 + \hat{\phi}_1^T \theta(t) + \Psi \right) \Pi^T,$$

$$\dot{z}_1 = \Pi \left( \alpha_1 + z_2 + \hat{\phi}_1^T \theta(t) + \Psi \right) \Pi^T + \hat{\phi}_1^T (\ell \theta - \hat{\theta}),$$

where $\Pi(x_1,t)$ and $\Psi(x_1,t)$ are given below equation (8), and $\ell \theta \in \mathbb{R}^q$ can be expressed in unknown constant vector. By Hadamard’s Lemma, one can express the regressor $\phi_1$ as $\phi_1(x_1) = \Phi_1(x_1) |x_1|$, where $\Phi_1(x_1) \in \mathbb{R}^q$ is smooth mapping. The third line of (19) will be treated by the following tuning function

$$\tau_1(x_1, \beta) = \Gamma z_1 \Phi_1 \Pi x_1$$

(20)

where $\Gamma = \Gamma^T \in \mathbb{R}^{q \times q}$ is the positive adaptation gain. Consider the Lyapunov function candidate

$$V_1 = \frac{1}{2} z_1^2 + \frac{1}{2} (\ell \theta - \hat{\theta})^T \Gamma^{-1} (\ell \theta - \hat{\theta}),$$

then, by recalling Remark 4

$$\dot{V}_1 = \Gamma z_1 \Phi_1 \Pi x_1$$

(22)

where $\Psi_{x_1} = \Psi/x_1$ is positive and invertible for all $x_1 \in \mathbb{L}_\infty$. Invoking Young’s inequality, yields

$$z_1 \phi_1^T \Delta \theta = z_1 \Phi_1^T \Delta \theta x_1 \leq \frac{\delta_\Delta}{2} \Phi_1^T \Phi_1 W_1^2 z_1^2 + \frac{\delta_\Delta}{2} z_1^2$$

(23)

where $W_1$ is shown in Remark 4. The virtual control law $\alpha_1$ is designed as

$$\alpha_1(x_1, \beta^{(1)}, \hat{\theta}) = - \frac{1}{\Pi} (k_1 + \zeta_1) z_1 - \frac{\Psi_{x_1}}{\Pi} x_1 - \hat{\phi}_1^T \theta$$

(24)

where $\beta^{(1)} = \beta \beta^T$, $k_1 > 0$, and

$$\zeta_1 = \frac{1}{2} \left( \frac{\alpha_1}{\epsilon_\psi} + \delta_\Delta \Pi \Phi_1^T \Pi \Phi_1 W_1^2 + \Pi \delta_\Delta + (n-1) \Delta_\Delta \right)$$

is the nonlinear damping gain with $\epsilon_\psi > 0$ and $\delta_\Delta$ being the “radius” of the compact set of $\theta(t)$. Then $\Pi(x_1,t) \in \Pi_+ \Phi \in \mathbb{R}^q$ and $W_1 \in \mathbb{R}$ are computable functions. The resulting $V_1$ is

$$\dot{V}_1 \leq -k_1 z_1^2 + \Pi z_2 z_2 - \frac{(n-1)}{2} \Delta \Delta_2 z_2^2 - \frac{1}{2 \epsilon_\psi} z_2^2$$

(25)

$$+ (\epsilon_\psi - \delta_\Delta) \Gamma^{-1} (\zeta_1 - \hat{\theta}).$$

The second term $\Pi z_2 z_2$ in the right hand side of (25) can be canceled at the next step.

Step 2: Recall $\dot{x}_2 = x_3 + \phi_2^T (\phi_2 \phi_2 \theta)$ and let $\alpha_2 = x_3 - z_3$, we rewrite $\dot{x}_2 = \dot{x}_2 - \alpha_2$ as

$$\dot{z}_2 = \alpha_2 + z_3 - \frac{\partial_\alpha}{\partial x_1} x_2 - \frac{\partial_\alpha}{\partial \theta} \theta - \frac{\partial_\alpha}{\partial \beta} \beta - \frac{\partial_\alpha}{\partial \beta} \beta$$

(26)

$$+ \phi_2^T \theta(t) - \frac{\partial_\alpha}{\partial x_1} \phi_1^T \theta(t).$$

Define $w_2(x_2, \hat{\theta}, \beta^{(1)}) = \phi_2 - \frac{\partial_\alpha}{\partial x_1} \phi_1$, then the second line of (26) can be rewritten as

$$\dot{w}_2^T \theta(t) = \dot{w}_2^T \hat{\theta} + w_2^T (\theta(t) - \ell \theta) + w_2^T (\ell \theta - \hat{\theta}).$$

(27)

Denote $\theta(t) - \ell \theta$, and according to Assumption 1, there exist a known constant $\delta_\Delta$ such that $\Delta \Delta \geq |\Delta \theta|$. Also note that $z_1$ and $\alpha_1 (x_1, \beta^{(1)}, \hat{\theta})$ are smooth and $\alpha_1 (0, \beta^{(1)}, \hat{\theta}) = 0$, and the $\theta$- and $\beta^{(1)}$-dependent change of coordinates between $\dot{z}_2$ and $\dot{x}_2$ are smooth, invertible, and $z_3 = 0 \Leftrightarrow \dot{z}_2 = 0$. Using Hadamard’s Lemma, one can directly express $w_2$ as $w_2 = W_2^T (x_2, \beta^{(1)}, \theta) z_2$, where $W_2(x_2, \beta^{(1)}, \theta) \in \mathbb{R}^{2 \times q}$ is a smooth mapping. Therefore, one can calculate that

$$z_2 w_2^T \theta(t) - \ell \theta \right) = z_2 \Delta_\Delta \Delta_2 W_2^T \hat{\theta}$$

$$\leq \frac{1}{2} \delta_\Delta \Delta_2 W_2^2 z_2^2 + \frac{\delta_\Delta}{2} z_2^2$$

(28)

$$\Delta \Delta \Delta_2 = \left( W_2^2 + 1 \right) z_2^2 + \frac{\Delta \Delta \Delta_2}{2} z_2^2$$

where $\frac{\Delta \Delta \Delta_2}{2} z_2$ is used and $|W_2|_F = \sqrt{\sum_{i=1}^{n} \sum_{j=1}^{n} (W_2 i j)^2}$ denotes the Frobenius norm. Choosing the Lyapunov function candidate $V_2 = V_1 + \frac{1}{2} z_2^2$, its derivative along the trajectories of (11) is

$$\dot{V}_1 \leq -k_1 z_1^2 + \Pi z_2 z_2 - \frac{(n-1)}{2} \Delta \Delta_2 z_2^2 - \frac{1}{2 \epsilon_\psi} z_2^2$$

$$+ z_2 \alpha_2 - z_2 \left( - \frac{\partial_\alpha}{\partial x_1} x_2 - \frac{\partial_\alpha}{\partial \theta} \theta - \frac{\partial_\alpha}{\partial \beta} \beta - \frac{\partial_\alpha}{\partial \beta} \beta \right)$$

(29)

$$+ \frac{\delta_\Delta}{2} \left( W_2^2 + 1 \right) z_2^2 + \frac{\Delta \Delta \Delta_2}{2} z_2^2 + z_2 w_2^T \theta + z_2 z_3$$

$$+ (\epsilon_\psi - \delta_\Delta) \Gamma^{-1} (\zeta_1 - \hat{\theta}).$$

According to (29), we design the tuning function as

$$\tau_2 (x_2, \beta^{(1)}, \hat{\theta}) = \tau_1 + \Pi w_2 z_2$$

(30)
In addition, the virtual control law \( \alpha_2 \) is constructed as
\[
\alpha_2(x, \hat{\beta}^{(2)}, \hat{\theta}) = -\Pi z_1 - (k_2 + \zeta_2)z_2 - w_1^T \hat{\theta}
+ \frac{\partial \alpha_1}{\partial x_1} x_2 + \frac{\partial \alpha_1}{\partial \beta} \hat{\beta} + \frac{\partial \alpha_1}{\partial \theta} \tau_2,
\]
where \( \hat{\beta}^{(2)} = [\beta, \hat{\beta}, \hat{\theta}]^T, k_2 > 0 \), and \( \zeta_2(x, \hat{\beta}^{(1)}, \hat{\theta}) \) is the nonlinear damping gain, as follows
\[
\zeta_2 = \frac{1}{2} \left( \delta_{\Delta_\theta}[W_2]^2 + (n - 1)\delta_{\Delta_\theta} + \frac{1}{\epsilon_{\psi}} \right).
\]
After some simplifications and using (30) and (31), we express (29) as
\[
\hat{V}_2 \leq -k_1 z_1^2 - k_2 z_2^2 - \frac{1}{2} \left( (n - 2)\delta_{\Delta_\theta} + \frac{1}{\epsilon_{\psi}} \right) z_2^T z_2
+ z_2 z_3 + \left( z_2 \frac{\partial \alpha_1}{\partial \theta} + (\ell_\theta - \hat{\theta})^T \Gamma^{-1} \right) (\tau_2 - \hat{\theta}).
\]
where \( z_2 z_3 \) can be canceled at the next step.

**Step 3:** Introducing \( \alpha_3 = x_4 - z_4 \) and according to \( z_4 = x_3 - \alpha_2 \), we can transform \( \dot{x}_3 = x_4 + \phi_3(x_3)\theta(t) \) to the following \( z_3 \)-dynamics
\[
\dot{z}_3 = \alpha_3 + z_4 - \frac{\partial \alpha_2}{\partial x_1} x_2 - \frac{\partial \alpha_2}{\partial x_2} x_3 - \frac{\partial \alpha_2}{\partial \beta} \hat{\beta} - \frac{\partial \alpha_2}{\partial \theta} \hat{\theta}
+ \frac{\partial \alpha_2}{\partial \theta} \hat{\theta} + w_3^T (\ell_\theta - \hat{\theta})
+ w_3^T (\theta(t) - \ell_\theta).
\]
Now we choose the Lyapunov function candidate \( V_3 = V_2 + \frac{1}{2} z_3^2 \), then
\[
\dot{V}_3 \leq -k_3 z_1^2 - k_2 z_2^2 - \frac{1}{2} \left( (n - 2)\delta_{\Delta_\theta} + \frac{1}{\epsilon_{\psi}} \right) z_2^T z_2
+ z_3 z_4 + \left( z_2 \frac{\partial \alpha_1}{\partial \theta} + (\ell_\theta - \hat{\theta})^T \Gamma^{-1} \right) (\tau_2 - \hat{\theta})
- z_3 \left( \sum_{j=1}^{2} \frac{\partial \alpha_2}{\partial x_j} x_{j+1} + \sum_{j=1}^{2} \frac{\partial \alpha_2}{\partial \theta} \beta(j+1) + \frac{\partial \alpha_2}{\partial \theta} \hat{\theta} \right)
+ z_3 \alpha_3 + z_2 z_3 + z_3 w_3^T \hat{\theta} + z_3 w_3^T (\ell_\theta - \hat{\theta})
+ z_3 w_3^T (\theta(t) - \ell_\theta).
\]
where \( w_3(x, \hat{\beta}^{(2)}, \hat{\theta}) = \phi_3 - \frac{\partial \alpha_2}{\partial x_1} \phi_1 - \frac{\partial \alpha_2}{\partial x_2} \phi_2 \in \mathbb{R}^q \) is the new regressor vector, and it can be verified that \( w_3(0, \hat{\beta}^{(2)}, \hat{\theta}) = 0 \).

Using the analysis similar to that used in (27) - (28), one can express \( w_3 \) as \( w_3 = W_3^T (\bar{\alpha}_3, \hat{\beta}^{(2)}, \hat{\theta}) \bar{w}_3 \), where \( W_3 \in \mathbb{R}^{3\times q} \) is a smooth mapping. Therefore, we obtain an upper bound of the last line of (35), as follows
\[
z_3 w_3^T (\theta(t) - \ell_\theta) \leq \frac{\delta_{\Delta_\theta}}{2} (|W_3|^2 + 1) z_2^2 + \frac{\delta_{\Delta_\theta}}{2} z_2^2 z_2.
\]
Then, we design the following tuning function and virtual control law, respectively
\[
\tau_3(x, \hat{\beta}^{(2)}, \hat{\theta}) = \tau_2 + \Gamma w_3 z_3
\]
\[
\alpha_3(x, \hat{\beta}^{(3)}, \hat{\theta}) = -z_2 - (k_3 + \zeta_3)z_3 - w_3^T \hat{\theta} + \frac{\partial \alpha_2}{\partial \theta} z_3
+ \sum_{j=1}^{2} \frac{\partial \alpha_2}{\partial x_j} x_{j+1} + \sum_{j=0}^{2} \frac{\partial \alpha_2}{\partial \theta} \beta(j+1) + \frac{\partial \alpha_2}{\partial \theta} \Gamma z_2 w_3,
\]
where \( \hat{\beta}^{(3)} = [\beta, \hat{\beta}, \hat{\theta}]^T, k_3 > 0 \), and
\[
\zeta_3 = \frac{1}{2} \left( \delta_{\Delta_\theta}[W_3]^2 + (n - 2)\delta_{\Delta_\theta} + \frac{1}{\epsilon_{\psi}} \right).
\]
Now, in virtue of (37) and (38), we can rewrite \( V_3 \) as
\[
\dot{V}_3 \leq -\sum_{j=1}^{3} k_j z_j^2 + z_3 z_4 - \frac{1}{2} \left( (n - 3)\delta_{\Delta_\theta} + \frac{1}{\epsilon_{\psi}} \right) z_2^T z_2
+ \left( z_2 \frac{\partial \alpha_1}{\partial \theta} + z_3 \frac{\partial \alpha_2}{\partial \theta} + (\ell_\theta - \hat{\theta})^T \Gamma^{-1} \right) \tau_3 - \hat{\theta},
\]
where \( z_3 z_4 \) can be canceled at the next step.

**Step n:** This step is different from the previous steps. On one hand, the actual control law and update law of \( \hat{\theta} \) should be designed in this step. On the other hand, we need to extend the congelation of variables for time-varying parameters in the feedback path to the scenario that time-varying parameters in the input path.

To proceed, we rewrite \( \dot{x}_n = \phi_n^T \theta(t) + b(t) \) as
\[
\dot{z}_n = w_n^T \theta(t) + b(t) + \frac{\partial \alpha_n}{\partial \theta} \frac{\partial \alpha_n}{\partial \theta} - \sum_{j=1}^{n-1} \frac{\partial \alpha_n}{\partial \theta} x_{j+1}
- \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \theta} \beta(j+1)
\]
where \( w_n = \phi_n - \frac{\partial \alpha_n}{\partial x_1} \phi_1 \). The main different will start from the following design. For the next developments we need
the following intermediate result by means of $u = \hat{\rho} \dot{u}$

$$z_n \dot{z}_n = z_n \dot{w}_n^\top \theta + z_n \dot{w}_n^\top (\theta(t) - \ell_b) + z_n \dot{w}_n^\top (\ell_{\theta} - \hat{\theta}) + z_n \ddot{u} + z_n (b(t) - \ell_b) \ddot{u} + z_n \ell_{\theta} \left( \frac{1}{\ell_b} - \hat{\rho} \right) \ddot{u}$$

$$- z_n \frac{\partial \alpha_n}{\partial \theta} \ddot{\theta} - z_n \sum_{j=1}^{n-1} \frac{\partial \alpha_n}{\partial \theta} x_{j+1}^t - z_n \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \theta} x_{j+1} - z_n \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \beta} \ddot{\beta}(j+1).$$

where $\ell_b$ is an unknown constant which can be regarded as the average of $b(t)$, $\hat{\rho}$ is an “estimate” of $1/\ell_b$ and denote $b(t) - \ell_b$ by $\Delta_b$. Note that we need $\delta_{\Delta_b}$ to construct the nonlinear damping gain to cancel the effect of unknown $\theta(t)$, as our previous steps do. However, the same method cannot be used directly for dealing with $b(t)$ since the perturbation term $z_n \Delta_b \ddot{\theta} \ddot{u}$ is coupled with the control input. Here we apply a special way to cope with the unknown time-varying quantities, i.e., designing $\ddot{u}$ skillfully to ensure the perturbation term $z_n (b(t) - \ell_b) \ddot{u}$ in the second of (43) is always negative.

Consider the Lyapunov function candidate

$$V_n = V_{n-1} + \frac{|b_n|}{2 \gamma \rho} \left( \frac{1}{\ell_b} - \hat{\rho} \right)^2$$

then,

$$V_n = V_{n-1} - \frac{|b_n|}{\gamma \rho} \left( \frac{1}{\ell_b} - \hat{\rho} \right)^2$$

$$\leq \frac{-n}{2} \sum_{j=1}^{n-1} k_j \dot{z}_j^2 - \frac{1}{2} \left( \delta_{\Delta_b} + \frac{1}{\epsilon_{\rho}} \right) \dot{z}_{n-1}^2 - \frac{1}{2} \left( \delta_{\Delta_b} + \frac{1}{\epsilon_{\rho}} \right) \ddot{\theta}^2$$

$$+ \sum_{j=1}^{n-1} \frac{\partial \alpha_n}{\partial \theta} z_{j+1} + \frac{\partial \alpha_n}{\partial \theta} \dot{\theta} \Gamma_n^t \left( \Gamma_n \ddot{\theta} \right)$$

$$+ z_n - z_n \Delta_b \ddot{u} + z_n \dot{w}_n^\top \theta + z_n \dot{w}_n^\top (\ell_{\theta} - \hat{\theta}) + z_n \ddot{u} + z_n \Delta_b \ddot{\theta} \ddot{u} - z_n \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \theta} x_{j+1}^t - z_n \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \beta} \ddot{\beta}(j+1)$$

$$- \sum_{j=1}^{n-1} \frac{\partial \alpha_n}{\partial \theta} x_{j+1} - \sum_{j=0}^{n-1} \frac{\partial \alpha_n}{\partial \theta} \dot{\theta}$$

$$+ z_n \ell_{\theta} \left( \frac{1}{\ell_b} - \hat{\rho} \right) \ddot{u} - \frac{|b_n|}{\gamma \rho} \left( \frac{1}{\ell_b} - \hat{\rho} \right) \ddot{u}$$

$$= -\kappa \rho \left( \frac{1}{\ell_b} - \hat{\rho} \right)^2$$

where $w_n = \phi_n - \sum_{j=1}^{n-1} \frac{\partial \alpha_n}{\partial \beta} \ddot{\beta}_j$. Now, to cancel the third and last lines of (43), we design the update laws for the parameters $\dot{\theta}$ and $\dot{\rho}$, as follows

$$\dot{\theta} = \gamma_n \tau_{n-1} + \Gamma w_n z_n$$

$$\dot{\rho} = -\gamma_n \rho \dot{\theta}$$

and $\alpha_i = \dot{w}_i = \tau_i = \Omega = 0$ if $z_i = 0$. Note also that the coordinate transformation

$$z_1 = \frac{\beta(t)\phi(x_1)}{\beta^2(t) - \psi^2(x_1)}$$

and $z_i = x_i - \alpha_{i-1}$ $(i > 1)$ is also smooth, invertible and $x_i = 0 \Leftrightarrow z_i = 0$. According to Hadamard’s Lemma, $w_n(x, \beta(n), \theta)$ and $\Omega$ can be expressed as $w_n = \frac{\partial \theta}{\partial \theta} w_n \Delta_b \ddot{u}$ and $\Omega = \Delta_b \ddot{u}$, respectively, with $z = \{z_1, \cdots, z_n\}^t$, $W_n \in \mathbb{R}^{n \times q}$ and $\Omega \in \mathbb{R}^n$ being smooth mappings.

Applying Young’s inequality with $\epsilon \rho > 0$, yields

$$z_n \Delta_b \ddot{u} \leq \frac{\delta_{\Delta_b}}{2} \left( |W_n|^2 + 1 \right) \dot{z}_n^2 + \frac{\delta_{\Delta_b}}{2} \dot{z}_{n-1}^2$$

Finally, we choose the actual control law $u = \hat{\rho} \dot{u}$ such that the time-varying perturbed term $z_n \Delta_b \ddot{u}$ is nonpositive

$$\left\{ \begin{array}{l}
\dot{\theta} = -\kappa (x, \beta, \cdots, \beta(n), \dot{\theta}) z_n \\
\kappa = k_n + \frac{1}{2} \left( \delta_{\Delta_b} |W_n|^2 + \delta_{\Delta_b} + \frac{1}{\epsilon_{\rho}} + \epsilon_{\rho} |\Omega|^2 \right)
\end{array} \right.$$ (50)

where $k_n > 0$. Inserting (47)-(50) into (46), yields

$$\dot{V}_n \leq -\sum_{j=1}^{n} k_j z_j^2 - \kappa \Delta_b \dot{\rho} z_n^2.$$ (51)

V. Stability Analysis

Firstly, it can be shown that $\dot{\rho}(t)$ in the right hand side of (51) is a monotonic increasing (or decreasing) function by calculating equation (43) as $\dot{\rho} = \gamma_n \rho \dot{\theta}$ if $\kappa < 0$. In addition, one can select $\dot{\rho}(0) > 0$ when $0 < \ell_b \leq b(t)$ (in this case, $\Delta_b > 0$) to make sure that $\dot{\rho}(t) > 0$, thereby obtaining $-\kappa \Delta_b \dot{\rho} z_n^2 < 0$. Similarly, one can select $\dot{\rho}(0) < 0$ when $b(t) \leq \ell_b < 0$ (in this case, $\Delta_b < 0$) to make sure that $\dot{\rho}(t) < 0$, thereby obtaining $-\kappa \Delta_b \dot{\rho} z_n^2 < 0$ again. Therefore, formula (51) can be simplified as $\dot{V}_n \leq -\sum_{j=1}^{n} k_j z_j^2$ which guarantees that $z, \dot{\theta}$ and $\dot{\rho}$ are bounded for all $t \geq 0$.

Next, in view of Remark 4 and the boundedness of $z$, it follows that $W_1, 1/II$ and $\Pi$ are bounded, and therefore $\tau_1$ and $\alpha_1$ are bounded, which further proves the boundedness of $x_2$ along with the coordinate transformation $x_2 = x_2 + \tau_1$ and the boundedness of $w_2$ due to (41). Hence $W_2, \tau_2$ and $\alpha_2$ are also bounded. Following this line of argument, the boundedness of state $x_1$, virtual control $\alpha_i$ $(i = 3, \cdots, n - 1)$, and the actual control input $u$ are ensured. In addition, it is seen from (47) and (48) that $\dot{\theta} \in \mathbb{L}_\infty$ and $\dot{\rho} \in \mathbb{L}_\infty$. To show the asymptotic constancy of $\dot{\theta}$ and $\dot{\rho}$, it follows from $\dot{V}_n \leq -\sum_{j=1}^{n} k_j z_j^2$ that $z_j \in \mathbb{L}_\infty$, then from (47) and (48) we get $\dot{\theta} \in \mathbb{L}_1$ and $\dot{\rho} \in \mathbb{L}_1$ by using the argument similar to Theorem 3.1 in [27]. It is concluded that $\dot{\theta}$ and $\dot{\rho}$ have a limit as $t \to \infty$.

Finally, it follows from (19), (26), (34), (43) and (51) that $\dot{z} \in \mathbb{L}_\infty$ and $z \in \mathbb{L}_2 \cap \mathbb{L}_\infty$ then, using Barbalat’s Lemma yields $\lim_{t \to \infty} \dot{x}(t) = 0$, which further indicates that $\lim_{t \to \infty} x(t) = 0$. Therefore, the closed-loop system is asymptotically stable. By virtue of Lemma 1, we get the output $x_1(t)$ is always constrained within the prescribed performance funnel $F_{\beta} := \{(t, x_1) \in [0, \infty) \times \mathbb{R} | |x_1(t)| / \psi^{-1}(\beta(t)) < 1\}$. 
The above facts prove the following result:

**Theorem 2:** Suppose that the design procedure is applied to the nonlinear system (1) with time-varying parameters. Then, the closed-loop system is asymptotically stable and the system output $x_1(t)$ is always confined within the prescribed performance funnel $F_\beta := \{(t, x_1) \in \mathbb{R}_{\geq 0} \times \mathbb{R} | |x_1(t)|/\psi^{-1}(\beta(t)) < 1\}$ and ultimately decays to zero. Furthermore, $\lim_{t \to \infty} \hat{\theta}$ and $\lim_{t \to \infty} \hat{\rho}$ exist but they are not necessarily equal to $\ell_\theta$ and $1/\ell_\rho$. In addition, the control input and update laws remain uniformly bounded over $[0, \infty)$.

**Remark 6:** The proposed controller primarily consists of three units: robust unit, $\theta(t)$-adaptive unit and $b(t)$-adaptive unit. Note that the $\theta(t)$-adaptive unit is completely equivalent to the design of update laws in classical adaptive control since we use the unknown constant $\ell_\theta$ to replace $\theta(t)$. The time-varying perturbation term $\Delta_\theta(t)$ caused by $\theta(t) - \ell_\theta$ is allocated to the robust unit for processing. This is an easy-to-understand and easy-to-implement solution, in other words, the proposed controller is simple in structure and user-friendly in design. In addition, $b(t)$-adaptive unit is deliberately designed for unknown and time-varying control gain, whose main purpose is to ensure the perturbation term $z_\Delta \beta \hat{u} \leq 0$, thereby avoiding the adaptive parameter drift caused by unknown gains.

**Remark 7:** The control scheme involves the selection of $\{k_i \}_{i=1}^n > 0$, $\bar{\theta}(0) \geq 0$, $\hat{\rho}(0) > 0$, $\delta_{\Delta_\theta} > 0$, $\epsilon_\psi > 0$, and $\Gamma > 0$, which theoretically can be chosen quite arbitrarily by users. Certain compromise between convergence rate and control effort needs to be made when making the selection for those parameters for a given system. For example, the parameters $k_i$ and $\delta_{\Delta_\theta}$ are proportional to convergence rate and control effort in this paper, and thus reducing the input effort will cause the convergence rate to slow down. However, it is worth noting that the prescribed constraint rule will not be violated no matter how the parameters are selected.

**Remark 8:** Compared with the previous work [29] on adaptive exponential regulation for systems with time-invariant parameters, the proposed method provides a simpler solution, and without loss of final control accuracy, completely eliminates the necessity for the control gain to grow with time ceaselessly.

**Remark 9:** Different from traditional guaranteed performance control (see, for instance, [32-34]), that can only achieve bounded regulation and the size of the regulation residual set is reversely proportional to the control gain, such that higher final control precision is essentially at the price of large control gain, the proposed control method is able to steer each system state to zero asymptotically without the need for prohibitively large controller gain. Furthermore, no matter how small the control gain is, $-\psi^{-1}(\beta) < x_1(t) < \psi^{-1}(\beta)$ always holds.

**Remark 10:** Our control scheme benefits from Chen & Astolfi’s Method [11]–[13] in dealing with unknown time-varying parameters, furthermore, by introducing the performance function and employing a novel coordinate transformation, our control scheme is able to explicitly address global transient behavior of system output, together with its steady-state performance.

**VI. SIMULATION**

To verify the effectiveness of the proposed control method, we consider the following system:

$$\begin{align*}
\dot{x}_1 &= \theta(t)x_1 + x_2; \\
\dot{x}_2 &= b(t)u; \\
y(t) &= x_1
\end{align*}$$

with fast time-varying parameters:

$$b(t) = 2 + 0.1 \cos(x_1) + \text{sign}(x_1x_2)$$

$$\theta(t) = 2 + 0.8 \sin(t) + \sin(x_1x_2) + 0.2 \sin(x_1t) + \text{sign}(\sin(t)).$$

It is not difficult to verify that Assumptions 1-2 are satisfied. The control objective is to make the state $x_1$ moves back to zero at a prescribed rate no slower than exponential and ultimately converges to zero. Now we consider three controllers: Controller 1 is the adaptive controller proposed by Chen & Astolfi in [11]; Controller 2 is the semi-global adaptive prescribed performance controller which can be obtained by combining Controller 1 and the controller proposed in [32]; Controller 3 is the global controller proposed in Theorem 2. In fact, Controller 2 can be viewed as a special case of Controller 3. For fair comparison, we set $[x_1(0); x_2(0)] = [1; -1]$, $k_1 = k_2 = \gamma_\rho = 0.1$, $\delta_{\Delta_\theta} = 1$, $\Gamma = 0.1I$, $\theta(0) = 0$ and $\hat{\rho}(0) = 0.25$ for all controllers. In addition, we select $\beta(t) = 4e^{-0.4t} + 0.1$ and $z_1 = \text{atan}(\pi x_1/(2\beta))$ for Controller 2, and select $\beta_1(t) = 0.9e^{-0.4t} + 0.1$ for Controller 3.

The responses of the state signals are shown in Figs 1-2, and the responses of control input signals are shown in Fig 3. The evolution of adaptive parameters $\theta$ and $\hat{\rho}$ are shown in Fig. 4 and 5, respectively. In addition, we also illustrate the time-varying parameters $\theta(t)$ and $b(t)$ in Fig 6, which shows that the state-dependent parameters are fast time-varying and nondifferentiable. From these simulation results, we know that the proposed controllers outperforms the adaptive controller in [11], since the transient behavior of the system can be confined to a prescribed performance boundary. In particular, compared Controller 1 with Controllers 2-3, one can find a counterintuitive phenomenon, that is, based on the previous parameter selection, faster system response can be achieved without an increase in control effort. In short, all results show that the proposed methods are powerful enough to stabilize the nonlinear system with fast time-varying parameters.

Note that when $\theta(t)$ is an unknown constant and $b(t) = 1$, this model is a simplified version of the one studied by [29], where the exponential regulation is proposed for a class of strict-feedback systems with known control gain and unknown constants $\theta$.

Here $b(t)$ and $\theta(t)$ are fast time-varying parameters and they are only piecewise continuous yet $b(t)$ may undergo sudden changes. Therefore, some classical adaptive schemes [37, 38] are not available because those methods require the parameters be slow time-varying (i.e., there exists a parameter $\epsilon$ such that $|\theta(t)| < \epsilon$ and $|b(t)| < \epsilon$).
VII. CONCLUSION

This work presents an adaptive control strategy with guaranteed performance for strict-feedback nonlinear systems involving fast time-varying parameters. It is shown that with this strategy, not only each system state is regulated to zero asymptotically, but also the system output is strictly confined within an exponentially decaying boundary, making system output well behaved during transient period and steady-state phase. We start with a simple scalar system with time-varying parameters in the feedback path and input path to illustrate our core idea in addressing time-varying parameters and output performance constraint simultaneously. By using classical Backstepping technology and nonlinear damping, we then extend our method to higher-order system and remove the need for overparametrization. Furthermore, the diversity of performance function selection and the diversity of normalized function selection together with the independence on initial conditions imply the universal of our controller, and simulation comparisons confirm the effectiveness and benefits of these methods.

Prior to the work, the prevailing wisdom in adaptive control in the context of exponential stability for time-varying systems is that certain persistent excitation conditions (sufficiently rich signals) must be present. Here in this work we develop a method that achieves exponential convergence, pointwise in time, without the need for PE conditions. An interesting future research topic is to study the exponential stabilization of
nonlinear systems with unknown time-varying parameters and control coefficients.

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