Distribution of voltage fluctuations in a current-biased conductor

M. Kindermann,1 Yu. V. Nazarov,2 and C. W. J. Beenakker1

1Instituut-Lorentz, Universiteit Leiden, P.O. Box 9506, 2300 RA Leiden, The Netherlands
2Department of Nanoscience, Delft University of Technology, Lorentzweg 1, 2628 CJ Delft, The Netherlands

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We calculate the fluctuating voltage \( V(t) \) over a conductor driven out of equilibrium by a current source. This is the dual of the shot noise problem of current fluctuations \( I(t) \) in a voltage-biased circuit. In the single-channel case the distribution of the accumulated phase \( \Phi = (e/\hbar) \int Vdt \) is the Pascal (or binomial waiting-time) distribution — distinct from the binomial distribution of transferred charge \( Q = \int I dt \). The weak-coupling limit of a Poissonian \( P(\Phi) \) is reached in the limit of a ballistic conductor, while in the tunneling limit \( P(\Phi) \) has the chi-square form.

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The current–voltage or charge–phase duality plays a central role in the theory of single-electron tunneling through tunnel junctions of small capacitance [9]. At the two extremes one has a voltage-biased junction (in which the voltage is kept fixed by a source with zero internal resistance, while the current fluctuates) and a current-biased junction (fixed current from a source with infinite internal resistance, fluctuating voltage). In the current-biased case the Coulomb blockade induces periodic oscillations in the voltage [2], while in the voltage-biased case the Coulomb blockade is inoperative.

Quantum mechanically, the duality appears because current \( I \) and voltage \( V \) are noncommuting operators [3]. This is conveniently expressed by the canonical commutator \([\Phi, Q] = ie\) of the transferred charge \( Q = \int_0^t I(t)dt \) and accumulated phase \( \Phi = (e/\hbar) \int_0^t V(t)dt \) (in a given detection time \( \tau \)). Moments of charge and phase determine the measured correlators of current and voltage, respectively [4].

While all moments of \( Q \) in a voltage-biased conductor are known [4], the dual problem (moments of \( \Phi \) under current bias) has only been studied for the first two moments [4,6,7]. In the absence of Coulomb-blockade effects, the first two moments in the dual problems are simply related by rescaling \( I(t) \to V(t) \times G \) (with \( G \) the conductance). One might surmise that this linear rescaling carries over to higher moments, so that the dual problems are trivially related in the absence of the Coulomb blockade. However, the rescaling (as derived for example in Ref. [8]) follows from a Langevin approach that is suspect for moments higher than the second — so that one might expect a more complex duality relation.

The resolution of this issue is particularly urgent in view of recent proposals to measure the third cumulant of shot noise in a mesoscopic conductor [5,9,10,11]. Does it matter if the circuit is voltage biased or current biased, or can one relate one circuit to the other by a linear rescaling? That is the question addressed in this paper.

We will demonstrate that, quite generally, the rescaling breaks down beyond the second moment. We calculate all moments of the phase (hence all correlators of the voltage) for the simplest case of a single-channel conductor (transmission probability \( \Gamma \)) in the (zero-temperature) shot-noise limit. In this case the charge \( Q \equiv qe \) for voltage bias \( V_0 \equiv h\phi_0/e\tau \) is known to have the binomial distribution

\[
P_{\phi_0}(q) = \binom{\phi_0}{q} \Gamma^q (1-\Gamma)^{\phi_0-q},
\]

(1)

We find that the dual distribution of phase \( \Phi \equiv 2\pi\phi \) for current bias \( I_0 \equiv eq_0/\tau \) is the Pascal distribution [3]

\[
P_{q_0}(\phi) = \binom{\phi}{q_0} \Gamma^{q_0} (1-\Gamma)^{\phi-q_0}.
\]

(2)

(Both \( q \) and \( \phi \) are integers for integer \( \phi_0 \) and \( q_0 \).)

In the more general case we have found that the distributions of charge and phase are related in a remarkably simple fashion for \( q, \phi \to \infty \):

\[
\ln P_q(\phi) = \ln P_q(\phi) + O(1).
\]

(3)

(The remainder \( O(1) \) equals \( \ln(q/\phi) \) in the shot-noise limit.) This manifestation of charge-phase duality, valid with logarithmic accuracy, holds for any number of channels and any model of the conductor.

We have obtained these results in the course of an extensive study of the quantum statistics of fluctuating currents and voltages in electrical circuits containing mesoscopic conductors. Our method is the non-equilibrium Keldysh action technique. Within this technique, the electronic degrees of freedom are traced out and the quantum dynamics is represented by path integrals over fluctuating phases on the Keldysh contour [12]. Before presenting the derivation we give an intuitive physical interpretation.

The binomial distribution [5] for voltage bias has the interpretation [12] that electrons hit the barrier with frequency \( eV_0/\hbar \) and are transmitted independently with probability \( \Gamma \). For current bias the transmission rate is fixed at \( I_0/e \). Deviations due to the probabilistic nature of the transmission process are compensated for by an adjustment of the voltage drop over the barrier. If the transmission rate is too low, the voltage \( V(t) \) rises so that electrons hit the barrier with higher frequency. The
number of transmission attempts (“trials”) in a time $\tau$ is given by $(e/h) \int_0^\tau V(t) dt \equiv \phi$. The statistics of the accumulated phase $\phi$ is therefore given by the statistics of the number of trials needed for $I_0\tau/e$ successful transmission events. This stochastic process has the Pascal distribution $\mathcal{P}$.  

Starting point of our derivation is a generalization to time-dependent bias voltage $V(t) = (\hbar/e)\Phi(t)$ of an expression in the literature $[\text{5, 14}]$ for the generating functional $Z[\Phi(t), \chi(t)]$ of current fluctuations:

$$Z[\Phi, \chi] = \left\langle \sum \prod \exp \left\{ \frac{i}{e} \int dt \left[ \Phi(t) + \frac{\chi(t)}{\hbar} \right] I(t) \right\} \right\rangle.$$  

(4)

(The notation $\sum \prod$ denotes time-ordering of the exponentials in ascending (descending) order.) Functional derivatives of the Keldysh action $\ln Z$ with respect to $\chi(t)/e$ produce irreducible correlators of the current operator $I(t)$ to any order desired. This expression provides an example of the “tracing out” of degrees of freedom: The average is over electronic degrees of freedom, the electrical current being the only operator coupled to the phase.

To make the transition from voltage to current bias we introduce a second conductor $B$ in series with the mesoscopic conductor $A$ (see Fig. 1). The generating functional $Z_{A+B}$ of current fluctuations in the circuit is a (path integral) convolution of $Z_A$ and $Z_B$,

$$Z_{A+B}[\Phi, \chi] = \int D\Phi_1 D\chi_1 Z_A[\Phi_1, \chi_1] Z_B[\Phi - \Phi_1, \chi - \chi_1].$$  

(5)

One can understand this expression as the average over fluctuating phases $\Phi_1, \chi_1$ at the node of the circuit shared by both conductors.

In general the functional dependence of $Z_A, Z_B$ is rather complicated and non-local in time, but we have found an interesting and tractable low-frequency regime: The non-locality may be disregarded for sufficiently slow realizations of the fluctuating phases. In this regime the functional $Z$ can be expressed in terms of a function $S$,

$$\ln Z[\Phi(t), \chi(t)] = \int dt S(\Phi(t), \chi(t)).$$  

(6)

The path integral (6) can be taken in saddle-point approximation, with the result

$$S_{A+B}(\Phi, \chi) = S_A(\Phi_s, \chi_s) + S_B(\Phi - \Phi_s, \chi - \chi_s).$$  

(7)

Here $\Phi_s$ and $\chi_s$ stand for the (generally complex) values of $\Phi_1$ and $\chi_1$ at the saddle point (where the derivatives with respect to these phases vanish).

Eqs. (5) and (6) are quite general and now we apply them to the specific circuit of Fig. 1. We assume that the mesoscopic conductor $A$ (conductance $G$) is in series with a macroscopic conductor $B$ with frequency dependent impedance $Z(\omega)$. Any electromagnetic environment with a Gaussian Keldysh action can be represented by an external impedance.) The circuit is driven by a voltage source with voltage $V_0$. To avoid Coulomb blockade effects $[\text{1}]$, we assume that $|Z(\omega)| \ll \hbar/e^2$ at frequencies $\hbar\omega \gtrsim \max(eV, kT)$ (with $T$ the temperature). The zero-frequency impedance $Z(0)\equiv Z_0 \equiv \pm \hbar/e^2$ can have any value.

Both the voltage drop $V$ at the mesoscopic conductor and the current $I$ through the conductor fluctuate in time for finite $Z_0$, with averages $\overline{\tilde{I}} = V_0 G(1 + Z_0 G)^{-1}$, $\overline{\tilde{V}} = V_0 (1 + Z_0 G)^{-1}$. Voltage bias corresponds to $Z_0 G \ll 1$ and current bias to $Z_0 G \gg 1$, with $I_0 = V_0/Z_0$ the imputed current. There are three characteristic time scales: $h/(\max(eV, kT)), \ e/I$, and the $RC$-time of the circuit. (The inverse of the $RC$-time is the frequency below which $Z \approx Z_0$.) The low-frequency regime on which we concentrate is reached for current and voltage fluctuations that are slow on any of these time scales.

We assume that the temperature is sufficiently low ($kT \ll e\overline{\tilde{V}}$) to neglect thermal noise relative to shot noise. (Generalization to finite temperature is straightforward, but will not be done here for simplicity.) The low-temperature, low-frequency Keldysh action of the external impedance is simply $S_B(\phi, \chi) = i\phi \tilde{\chi} / 2\pi z_0$, while the action $S_A$ of the mesoscopic conductor is given by

$$S_A(\Phi, \chi) = \frac{\Phi}{2\pi} S(i\chi), \quad S(\xi) = \sum_{n=1}^N \ln[1 + (e^{\xi} - 1) T_n].$$  

(8)

The $T_n$’s are the transmission eigenvalues, with $\sum_n T_n = Gh/e^2 \equiv g$ the dimensionless conductance. (The formulas are written without including spin or other degeneracies $[\text{2}]$.)

We seek the cumulant generating function of charge

$$\mathcal{F} = \ln \left( \sum_{q=0}^{\infty} e^{q\xi} P(q) \right) = \sum_{p=1}^{\infty} \frac{\langle q^p \rangle}{p!} \xi^p.$$  

(9)
where $\langle q^p \rangle$ is the $p$-th cumulant of the charge transferred during the time interval $\tau$. It is related to the Keldysh action (10) by

$$\mathcal{F}(\xi) = \tau S_{A+B}(eV_0/h, -i\xi).$$

We also require the cumulant generating function of phase, $\mathcal{G}(\xi)$. Since $V = V_0 - Z_0 I$ (in the absence of thermal noise from the external impedance), it is related to $\mathcal{F}(\xi)$ by a change of variables (from $q$ to $\phi = \phi_0 - qz_0$).

The relation is

$$\mathcal{G}(\xi) = \sum_{p=1}^{\infty} \frac{\langle q^p \rangle \xi^p}{p!} = \phi_0 \xi + \mathcal{F}(-z_0 \xi).$$

In the limit $Z_0 \to 0$ of voltage bias the saddle point of the Keldysh action is at $\Phi_s = \Phi$, $\chi_s = \chi$, and from Eqs. (9), (10), and (11) one recovers the results of Ref. 3: The cumulant generating function $\mathcal{F}_0(\xi) = \tau S_A(eV_0/h, -i\xi) = \phi_0 \mathcal{S}(\xi)$ and the corresponding probability distribution

$$P_{\phi_0}(q) = \lim_{x\to-q} \frac{d^q}{dq^q} \prod_{n=1}^{N} \left[1 + (x - 1)T_n\right]^{\phi_0}.$$ (12)

The parameter $\phi_0 = eV_0\tau/h$ is the number of attempted transmissions per channel, assumed to be an integer $\gg 1$. The first few cumulants are $\langle q \rangle_0 = \phi_0 g_0$, $\langle q^2 \rangle_0 = \phi_0 \sum_n T_n(1-T_n)$, $\langle q^3 \rangle_0 = \phi_0 \sum_n T_n^3(1-T_n)(1-2T_n)$. In the single-channel case ($N = 1$, $T_1 \equiv 1$) the distribution (13) has the binomial form (14).

After these preparations we are now ready to generalize all of this to finite $Z_0$, and in particular to derive the dual distribution of phase (15) under current bias. The key equation that allows us to do that follows directly from Eqs. (13) and (14):

$$\mathcal{F}(\xi) = \frac{\phi_0}{z_0} [\xi - \sigma(\xi)], \quad \sigma + z_0 \mathcal{S}(\xi) = \xi.$$ (13)

The implicit function $\sigma(\xi)$ (which determines the saddle point of the Keldysh action) provides the cumulant generating function of charge $\mathcal{F}$ for arbitrary series resistance $z_0 = (e^2/h)Z_0$. One readily checks that $\mathcal{F}(\xi) = \phi_0 \mathcal{S}(\xi)$ in the limit $z_0 \to 0$, as it should.

By expanding Eq. (13) in powers of $\xi$ we obtain a relation between the cumulants $\langle q^p \rangle$ of charge at $Z_0 \neq 0$ and the cumulants $\langle q^p \rangle_0$ at $Z_0 = 0$. For example, to linear order we find $\langle q \rangle = (1 + z_0 g)^{-1}\langle q \rangle_0$, which amounts to the obvious statement that the mean current $I$ is rescaled by a factor $1 + Z_0 G$ as a result of the series resistance. The Langevin approach discussed in the introduction predicts that the same rescaling applies to the fluctuations. Indeed, to second order we find $\langle q^2 \rangle = (1 + z_0 g)^{-3}\langle q^2 \rangle_0$, in agreement with Ref. 1.

However, if we go to higher cumulants we find that other terms appear, which cannot be incorporated by any rescaling. For example, Eq. (13) gives for the third cumulant

$$\langle q^3 \rangle = \frac{\langle q^3 \rangle_0}{(1 + z_0 g)^4} - \frac{3z_0 g}{(1 + z_0 g)^6} \left(\frac{\langle q^2 \rangle_0}{q_0}\right)^2.$$ (14)

The first term on the right-hand side has the expected scaling form, but the second term does not. This is generic for $p \geq 3$: $\langle q^p \rangle = (1 + z_0 g)^{-p-1}\langle q^p \rangle_0$ plus a non-linear (rational) function of lower cumulants [15].

All terms are of the same order of magnitude in $z_0 g$, so one can not neglect the non-linear terms.

Turning now to the limit $z_0 g \to \infty$ of current bias, we see from Eq. (14) that $\mathcal{F} \to \mathcal{F}_\infty$ with

$$\mathcal{F}_\infty(\xi) = q_0 \xi - q_0 \mathcal{S}_\infty(\xi/z_0)$$

defined in terms of the functional inverse $\mathcal{S}_\infty$ of $\mathcal{S}$. The parameter $q_0 = \phi_0 / z_0 = I_0 \tau / e$ (assumed to be an integer $\gg 1$) is the number of charges transferred by the imposed current $I_0$ in the detection time $\tau$. Transforming from charge to phase variables by means of Eq. (14), we find that $\mathcal{G} \to \mathcal{G}_\infty$ with

$$\mathcal{G}_\infty(\xi) = \phi_0 \xi - \phi_0 \mathcal{S}_\infty(\xi).$$ (16)

In the single-channel case Eq. (14) reduces to $\mathcal{G}_\infty(\xi) = -q_0 \ln[1 + (1 - e^{-\xi} - 1)]$, corresponding to the Pascal distribution (2). The first three cumulants are $\langle \phi \rangle = q_0 / \Gamma$, $\langle \phi^2 \rangle = (q_0 / \Gamma^2)(1 - \Gamma)$, $\langle \phi^3 \rangle = (q_0 / \Gamma^3)(1 - \Gamma)(2 - \Gamma)$.

For the general multi-channel case a simple expression for $P_{\phi_0}(\phi)$ can be obtained in the ballistic limit (all $T_n$’s close to 1) and in the tunneling limit (all $T_n$’s close to 0). In the ballistic limit one has $\mathcal{G}_\infty(\xi) = q_0 \xi / N + q_0 (N - g)(\xi / N - 1)$, corresponding to a Poisson distribution in the discrete variable $N = q_0 - q = 0, 1, 2, \ldots$. In the tunneling limit $\mathcal{G}_\infty(\xi) = -q_0 \ln(1 - \xi / g)$, corresponding to a chi-square distribution $P_{\phi_0}(\phi) \propto \phi^{q_0 - 1} e^{-\phi \xi}$ in the continuous variable $\phi > 0$. In contrast, the charge distribution $P_{\phi_0}(q)$ is Poissonian both in the tunneling limit (in the variable $q$) and in the ballistic limit (in the variable $N = q_0 - q$).

For large $q_0$ and $\phi$, when the discreeteness of these variables can be ignored, we may calculate $P_{\phi_0}(\phi)$ from $\mathcal{G}_0(\xi)$ in saddle-point approximation. If we also calculate $P_{\phi_0}(q)$ from $\mathcal{F}_0(\xi)$ in the same approximation (valid for large $\phi_0$ and $q$), we find that the two distributions have a remarkably similar form:

$$P_{\phi_0}(q) = N_{\phi_0}(q) \exp[\pi \Sigma(2\pi \phi_0 / \tau, q / \tau)],$$ (17)

$$P_{\phi_0}(\phi) = N_{\phi_0}(\phi) \exp[\pi \Sigma(2\pi \phi / \tau, q_0 / \tau)].$$ (18)

The same exponential function

$$\Sigma(x, y) = S_A(x, -i\xi_s) - y \xi_s$$ (19)

appears in both distributions (with $\xi_s$ the location of the saddle point). The pre-exponential functions $N_{\phi_0}$ and $N_{\phi_0}$ are different, determined by the Gaussian integration around the saddle point. Since these two functions vary only algebraically, rather than exponentially, we conclude
that Eq. (3) holds with the remainder $\mathcal{O}(1) = \ln(q/\phi)$ obtained by evaluating $\ln[2\pi(\partial^2\Sigma/\partial x^2)^{1/2}(\partial^2\Sigma/\partial y^2)^{-1/2}]$ at $x = 2\pi\phi/\tau$, $y = q/\tau$.

The distributions of charge and phase are compared graphically in Fig. 2 in the tunneling limit $\Gamma \ll 1$. We use the rescaled variable $x = q/(q_0)$ for the charge and $x = \phi/(\phi_0)$ for the phase, and take the same mean number $N = q_0 = \phi_0\Gamma$ of transferred charges in both cases. We plot the asymptotic large-$N$ form of the distributions,

$$
P_{\text{charge}}(x) = \frac{N^2}{2\pi}{x^{-1/2}}e^{N[x-1-x\ln x]},$$

$$
P_{\text{phase}}(x) = \frac{N}{2\pi}{x^{-1}}e^{N(1-x+x\ln x)},$$

(corresponding to the Poisson and chi-square distribution, respectively. Since the first two moments are the same, the difference appears in the non-Gaussian tails. The difference should be readily visible as a factor of two in a measurement of the third cumulant: $\langle x^3 \rangle = N^{-2}$ for the charge and $\langle x^3 \rangle = 2N^{-2}$ for the phase.

In summary, we have demonstrated theoretically that electrical noise becomes intrinsically different when the conductor is current biased rather than voltage biased. While the second cumulants can be related by a rescaling with the conductance, the third and higher cumulants can not. Experiments are typically carried out in an intermediate regime, and for that purpose we have derived explicit expressions [such as Eq. (14)] for the crossover from voltage to current bias. The non-linear term that breaks the rescaling is of the same order of magnitude as the linear term, so it can not be neglected. From a fundamental point of view, the limit of full current bias is of particular interest. The counterpart of the celebrated binomial distribution of transferred charge [3] turns out to be the Pascal distribution of phase increments.

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\end{thebibliography}