Cosine Series Representation

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This short paper is based on Chung et al. (2010), where the cosine series representation (CSR) is used in modeling the shape of white matter fiber tracts in diffusion tensor imaging (DTI) and Wang et al. (2018), where the method is used to denoise EEG. The proposed explicit analytic approach offers far superior flexibility in statistical modeling compared to the usual implicit Fourier transform methods such as the discrete cosine transforms often used in signal processing. The MATLAB codes and sample data can be obtained from http://brainimaging.waisman.wisc.edu/~chung/tracts.

1 Hilbert space theory in $[0,1]$

Consider the $i$-th functional time series data

$$\zeta_i(t) = \mu_i(t) + \epsilon_i(t), \quad (1)$$

where $t$ is time variable. We assume the functional data is scaled in such a way that they are defined in $[0,1]$. Formulating the Fourier analysis in a unit interval makes the numerical implementation more convenient. $\epsilon_i$ is a zero mean noise at each fixed $t$, i.e.,

$$\mathbb{E}\epsilon_i(t) = 0.$$

$\mu_i$ is an unknown smooth functional signal to be estimated. It is reasonable to assume that

$$\zeta_i, \mu_i \in \mathcal{L}^2[0,1],$$

the space of square integrable functions. Note any function $f \in \mathcal{L}^2[0,1]$ satisfies the condition

$$\int_0^1 f^2(t) \, dt < \infty.$$
This condition is needed to guarantee the convergence in the Fourier series expansion. Instead of estimating $\mu_i$ in $L^2[0, 1]$, we approximate $\mu_i$ as a linear combination of some known basis functions $\psi_0, \psi_1, \cdots, \psi_k$ by projecting data to the finite subspace spanned by $\psi_0, \psi_1, \cdots, \psi_k$.

Assume $\psi_l$ is orthonormal basis in $[0, 1]$ with respect to the inner product

$$\langle f, g \rangle = \int_0^1 f(t)g(t) \, dt.$$  

Thus, it is required to have

$$\int_0^1 \psi_l(t)\psi_m(t) = \delta_{lm},$$

the Kronecker delta with 1 if $l = m$ and 0 if $l \neq m$. With respect to the inner product, the norm $\| \cdot \|$ is then defined as

$$\| f \| = (\langle f, f \rangle)^{1/2} = \left[ \int_0^1 f^2(t) \, dt \right]^{1/2}.$$

We estimate $\mu_i$ in the subspace $\mathcal{H}_k$ spanned by up to the $k$ orthonormal basis functions:

$$\mathcal{H}_k = \left\{ \sum_{l=0}^k c_l \psi_l(t) : c_l \in \mathbb{R} \right\} \subset L^2[0, 1].$$

The least squares estimation (LSE) of $\mu_i$ in $\mathcal{H}_k$ is given by

$$\hat{\mu}_i = \arg \min_{f \in \mathcal{H}_k} \| f - \zeta_i(t) \|^2. \quad (2)$$

**Theorem 1** The minimization of (2) is given by

$$\hat{\mu}_i = \sum_{l=0}^k \langle \zeta_i, \psi_l \rangle \psi_l, \quad (3)$$

where the $l$-th degree Fourier coefficient $\langle \zeta_i, \psi_l \rangle$ is given by the inner product

$$\langle \zeta_i, \psi_l \rangle = \int_0^1 \zeta_i(t)\psi_l(t) \, dt. \quad (4)$$

**Proof.** We need to find function

$$f(t) = \sum_{l=0}^k c_l \psi_l(t)$$
that is the closest to \( \zeta_i \). The distance between \( f \) and \( \zeta_i \) is given by

\[
I(c_0, c_1, \cdots, c_k) = \int_0^1 \left| \sum_{l=0}^k c_l \psi_l(t) - \zeta_i(t) \right|^2 dt,
\]

which is a \( k+1 \) dimensional function in unknown parameter space \((c_0, c_1, \cdots, c_k) \in \mathbb{R}^{k+1}\). Since \( I \) is a quadratic function in \((c_0, c_1, \cdots, c_k)\), it has the global minimum at

\[
\frac{\partial I}{\partial c_0} = \frac{\partial I}{\partial c_1} = \cdots = \frac{\partial I}{\partial c_k} = 0.
\]

The algebraic derivation is straightforward and we have

\[
e_l = \langle \zeta_i, \psi_l \rangle
\]

for all \( l = 0, 1, \cdots, k \). \( \Box \)

The expansion (3) is called the Fourier series. As \( k \to \infty \), the expansion converges to \( \zeta_i \), i.e.,

\[
\zeta_i(t) = \sum_{l=0}^\infty \langle \zeta_i, \psi_l \rangle \psi_l.
\]

2 Cosine series representation

The cosine series representation (CSR) is first introduced in Chung et al. (2010). Unlike the discrete cosine transform used in signal processing, CSR is based on the least squares estimation technique.

There are infinitely many possible orthonormal basis in interval \([0, 1]\). Here we explain a spectral approach for obtaining orthonormal basis in \([0,1]\). Consider the space of square integrable functions in \([0, 1]\) denoted by \( L^2[0, 1] \). Let us solve the eigenfunction

\[
\Delta \psi + \lambda \psi = 0
\]

in \( L^2[0, 1] \) with 1D Laplacian \( \Delta = \frac{d^2}{dt^2} \). We are basically solving the 2nd order differential equation

\[
\frac{d^2 \psi(t)}{dt^2} + \lambda \psi(t) = 0.
\]

We can show that eigenfunctions \( \psi_0, \psi_1, \psi_2, \cdots \) form an orthonormal basis in \( L^2[0, 1] \). Note that if \( \psi_l \) is an eigenfunction, any multiple of \( \psi_l \) is also an eigenfunction. Thus, it is expected the eigenfunctions are properly normalized such that

\[
\int_0^1 \psi_l^2(t) \ dt = 1.
\]
The eigenfunctions satisfying (5) is then given by the usual Fourier sine and cosine basis

\[
\psi_0(t) = 1, \psi_l = \sqrt{2} \sin(l\pi t), \sqrt{2} \cos(l\pi t)
\]  

with the corresponding eigenvalues \( \lambda_l = l^2 \pi^2 \). We can check (7) are solutions by differentiating the eigenfunctions twice.

There are two eigenfunctions corresponding to the same eigenvalue. The multiplicity of eigenfunctions only happen if there is a symmetry in the domain of the Laplace eigenvalue problem (5). The constant \( \sqrt{2} \) is introduced to make the eigenfunctions orthonormal in \([0, 1]\).

Using both sine and cosine basis is not computationally efficient. Instead of solving (5) in the domain \([0, 1]\), if we add an additional constraint to eigenfunctions, we can get rid of the sine basis. Consider solving the problem in the larger unbounded domain \( \mathbb{R} \) with the periodic constraint

\[
\psi(t + 2) = \psi(t).
\]  

Suppose \( \psi \) is the solution in \([0, 1]\). The period 2 constraint forces \( \psi \) to be only valid in the intervals \( \cdots, [-2, -1], [0, 1], [2, 3], \cdots \). There are gaps in \( \cdots, (-1, 0), (1, 2), (3, 4), \cdots \), where there is no solution. We can fill the gap by padding with some arbitrary function. However, if we pad the gaps with any function, it may result in the Gibbs phenomenon (ringing artifacts) at the boundary of the intervals \( \cdots, 2, 1, 0, 1, 2, \cdots \) (Chung et al., 2007). To avoid the Gibbs phenomenon, we force the function to be continuous at the boundary by putting the additional constraint of evenness, i.e.,

\[
\psi(t) = \psi(-t).
\]  

If \( \psi(t) \) is the eigenfunction well defined in \( \cdots, [-2, -1], (-1, 0) [0, 1], (1, 2), [2, 3], \cdots \) we must have

\[ \cdots, \psi(t - 2), \psi(-t), \psi(t), \psi(-t + 2), \psi(t + 2), \cdots \]

The only eigenfunctions satisfying the two constraints (8) and (9) are the cosine basis

\[
\psi_0(t) = 1, \psi_l(t) = \sqrt{2} \cos(l\pi t)
\]  

with the corresponding eigenvalues \( \lambda_l = l^2 \pi^2 \) for integers \( l > 0 \). Then using the cosine basis only, any \( f \in L^2[0, 1] \) can be represented as

\[
f(t) = \sum_{l=0}^{k} c_l \psi_l(t) + \epsilon(t),
\]
Figure 1: A. fMRI time series at two different voxels at time point $t_1$. B. fMRI time series at voxel 1 (red) and 2 (black) shown for all 1200 time points. C. Normalized and scaled time series at voxel 1 and its cosine series representation with degree $k = 59$. D. Normalized and scaled time series at voxel 2 and its cosine series representation with degree $k = 59$. Such high-frequency denoising is often necessary for functional signals like fMRI and EEG.
where $c_l$ is the Fourier coefficients and $\epsilon$ is the residual error for using only $k$-th degree expansion.

If we put the constraint of oddness, i.e., $\psi(t) = -\psi(-t)$, we have sine basis
\[
\psi_l(t) = \sqrt{2} \sin(l\pi t).
\]

3 Least squares estimation

In practice, functional time series are observed at discrete time points $t_1, t_2, \cdots, t_n$:
\[
\zeta_i(t_j) = \mu_i(t_j) + \epsilon_i(t_j), \quad j = 1, \cdots, n.
\] (10)

The underlying mean functions $\mu_i(t)$ are estimated as
\[
\hat{\mu}_i(t) = \sum_{l=0}^{k} c_{li} \psi_l(t),
\]
where the Fourier coefficients $(c_{0i}, c_{1i}, \cdots, c_{ki})$ for the $i$-th time series is estimated using LSE as follows. At each $t_j$, we have
\[
\begin{pmatrix}
\zeta_i(t_1) \\
\zeta_i(t_2) \\
\vdots \\
\zeta_i(t_n)
\end{pmatrix}
= Y_i
= \begin{pmatrix}
\psi_0(t_1) & \psi_1(t_1) & \cdots & \psi_k(t_1) \\
\psi_0(t_2) & \psi_1(t_2) & \cdots & \psi_k(t_2) \\
\vdots & \vdots & \ddots & \vdots \\
\psi_0(t_n) & \psi_1(t_n) & \cdots & \psi_k(t_n)
\end{pmatrix}
\begin{pmatrix}
c_{i0} \\
c_{i1} \\
\vdots \\
c_{ik}
\end{pmatrix},
\]
where the design matrix of basis functions $\Psi$ is given by
\[
\Psi_{n \times (k+1)} = 
\begin{pmatrix}
\psi_0(t_1) & \psi_1(t_1) & \cdots & \psi_k(t_1) \\
\psi_0(t_2) & \psi_1(t_2) & \cdots & \psi_k(t_2) \\
\vdots & \vdots & \ddots & \vdots \\
\psi_0(t_n) & \psi_1(t_n) & \cdots & \psi_k(t_n)
\end{pmatrix}.
\]

The unknown coefficient vector $C_i$ is estimated by multiplying $\Psi^T$ on the both sides:
\[
\Psi^T Y_i = \Psi^T \Psi C_i.
\]

If $n \gg k$, $\Psi^T \Psi$ is full rank and invertible. Thus, we have
\[
C_i = (\Psi^T \Psi)^{-1} \Psi^T Y_i.
\] (11)
Although we will not show, it is further possible to discretize and reshape the basis \( \psi_l \) in such a way that

\[ \Psi^\top \Psi = I \]

the identity matrix. The detail is given in (Chung, 2012). This will make the numerical implementation of the Fourier series expansion computationally much more efficient for large \( k \). The proposed least squares estimation technique avoids using the often used implicit Fourier transform (FT) (Batchelor et al., 2006; Bulow, 2004; Gu et al., 2004).

If we have \( p \) number of time series, it requires solving equation (11) \( p \) number of times. For really large LSE problems, this is computationally very inefficient. A more efficient way is to estimate the all the coefficients using a single LSE and invert \( \Psi^\top \Psi \) only once by solving

\[ Y_{n \times p} = \Psi_{n \times k} C_{k \times p}, \]

where

\[ Y_{n \times p} = \begin{bmatrix} Y_1, Y_2, \ldots, Y_p \end{bmatrix}, \]
\[ C_{k \times p} = \begin{bmatrix} C_1, C_2, \ldots, C_p \end{bmatrix}. \]

Subsequently, all the coefficients are simultaneously estimated in the least squares fashion as

\[ \hat{C} = (\Psi^\top \Psi)^{-1} \Psi^\top Y. \]

4 Additional notes

It is also possible to have a slightly different but equivalent model that is easier to use in statistical inference. Assuming Gaussianity of data, \( \epsilon_i(t) \) in (1) is a Gaussian stochastic process, which is simply a collection of random variables. Then \( \epsilon_i(t) \) can be expanded using the given basis \( \psi_l \) as follows.

\[ \epsilon_i(t) = \sum_{l=0}^{k} Z_l \psi_l(t) + e_i(t), \]

where \( Z_l \sim N(0, \tau_l^2) \) are possibly correlated Gaussian random variables and \( e_i \) is the residual error that can be neglected in practice if enough number of basis are used. This is the consequence of the Karhunen-Loeve expansion (Adler, 1990; Dougherty, 1999; Kwapien and Woyczynski, 1992; Yaglom, 1987).
Note. Karhunen-Loeve expansion states that $\epsilon_i(t)$ can be decomposed as

$$
\epsilon_i(t) = \sum_{l=0}^{m} Z_l \phi_l(t)
$$

for uncorrelated Gaussian random variables $Z_l$ and some orthonormal basis $\phi_l(t)$. The algebraic determination of $Z_l$ and $\phi_l(t)$ are not hard. Since $\phi_l$ and $\psi_l$ are different basis, $\phi_l$ can be represented as a linear combination of $\psi_l$. Rewriting $\phi_l$ in terms of $\psi_l$ will make the Gaussian random variables $Z_l$ correlated.

At the end, we can write model (1) as

$$
\zeta_i(t) = \sum_{l=0}^{k} X_l \psi_l(t) + e_i(t),
$$

where $X_l$ are correlated Gaussian random variables.

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