Adjoint variable method for transient nonlinear electroquasistatic problems

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Abstract
Many optimization problems in electrical engineering consider a large number of design parameters. A sensitivity analysis identifies the design parameters with the strongest influence on the problem of interest. This paper introduces the adjoint variable method as an efficient approach to study sensitivities of nonlinear electroquasistatic problems in time domain. In contrast to the more common direct sensitivity method, the adjoint variable method has a computational cost nearly independent of the number of parameters. The method is applied to study the sensitivity of the field grading material parameters on the performance of a 320 kV cable joint specimen, which is modeled as a finite element nonlinear transient electroquasistatic problem. Special attention is paid to the treatment of quantities of interest, which are evaluated at specific points in time or space. It is shown that the method is a valuable tool to study this strongly nonlinear and highly transient technical example.

Keywords Adjoint variable method · Nonlinear electroquasistatic problem · Sensitivity analysis · Time domain

1 Introduction
When developing electrical equipment, engineers optimize initial design proposals by carefully identifying a number of design parameters related to, e.g., material properties and geometric dimensions. In doing so, they rely on rules of thumb, know-how and previous experience, existing standards and, increasingly, simulation and optimization tools. Numerical optimization is used to simultaneously improve—possibly conflicting—quantities of interest (QoIs), robustness and costs. While stochastic optimization plays a major role, derivative-based deterministic optimization algorithms are becoming, again, increasingly interesting [12]. Their advantages over stochastic methods are a faster convergence, i.e. less expensive optimization runs, and efficient coupling with mesh refinement and reduced order models. However, in case of derivative-based approaches, the problem of efficient gradient computation arises. The most common methods for gradient computation, e.g., finite differences and the direct sensitivity method (DSM), are not well suited for applications with many design parameters because their computational costs scale with the number of parameters [16,18]. The adjoint variable method (AVM), on the other hand, has computational costs that are almost independent of the number of parameters [3,16]. So far, the AVM has been applied mainly in the analysis of electric networks [8,18] and, since the 2000s, more and more often in the context of electromagnetic simulation [1,9,15]. Zhang et. al. recently introduced the AVM for linear electroquasistatic (EQS) problems in frequency domain [23]. However, many problems in the field of high-voltage (HV) engineering require an investigation in time domain since they are exposed to transient overvoltages and contain strongly nonlinear materials [11,21]. In this work, the AVM is formulated and solved numerically for the nonlinear transient EQS problem. Additionally, a method for sensitivity calculation of (QoIs) evaluated at a given point in time is presented, since the AVM naturally only considers time-integrated (QoIs). The AVM is validated using an ana-
lytical example. Subsequently, a nonlinear resistively graded 320 kV high-voltage direct current (HVDC) cable joint under impulse operation serves as a prominent technical example. It is shown that the AVM is capable of computing the sensitivities of this highly transient nonlinear problem with reasonable computational effort. This is an important step toward gradient-based optimization of electric devices in HV engineering.

2 Electroquasistatic problem

The EQS problem in time domain reads

\[\begin{align*}
- \nabla (\sigma \nabla \phi) - \nabla \left( \partial_t (\varepsilon \nabla \phi) \right) &= 0 \\
t &\in [0, T], \; r \in \Omega ; \\
\phi &= \phi_{\text{fixed}} \\
t &\in [0, T], \; r \in \Gamma_e ; \\
- (\sigma \nabla \phi + \partial_t (\varepsilon \nabla \phi)) \cdot n &= 0 \\
t &\in [0, T], \; r \in \Gamma_m ; \\
\phi &= \phi_0 \\
t &= 0, \; r \in \Omega ,
\end{align*}\] (1a-d)

where \(\phi\) is the electric potential and \(\sigma\) and \(\varepsilon\) represent the electric conductivity and permittivity, respectively. The time variable is denoted by \(t\) and the position vector by \(r\). \(\Omega\) is the computational domain and \(T\) is the terminal simulation time. \(\phi_{\text{fixed}}\) are the fixed voltages at the electrodes, \(\Gamma_e \neq \emptyset\), and \(n\) is the unit vector at the magnetic boundaries, \(\Gamma_m = \partial \Omega \setminus \Gamma_e\). The initial condition is denoted by \(\phi_0\). In case of a field-dependent conductivity or permittivity, i.e., \(\sigma = \sigma (E (r, t), r)\) and \(\varepsilon = \varepsilon (E (r, t), r)\), (1) becomes nonlinear.

The standard two-dimensional (2D) axisymmetric finite element (FE) problem of (1) is formulated by discretizing \(\phi (r, t) \approx \sum_j u_j N_j\), where \(N_j (r)\) are linear nodal FE shape functions. The degrees of freedom are \(u_j (t)\), which are assembled in the vector \(u\). The semi-discrete version of (1) according to the Ritz–Galerkin procedure reads

\[K_{\sigma} u + \partial_t (K_{\varepsilon} u) = 0,\] (2)

with

\[\begin{align*}
[K_{\sigma}]_{ij} &= \int_{\Omega} \sigma \nabla (N_j) \cdot \nabla (N_i) \, d\Omega \quad i, j = 1, ..., N_N , \\
[K_{\varepsilon}]_{ij} &= \int_{\Omega} \varepsilon \nabla (N_j) \cdot \nabla (N_i) \, d\Omega \quad i, j = 1, ..., N_N ,
\end{align*}\] (3)

where \(N_N\) denotes the number of nodes. For the time discretization, the implicit Euler time stepping scheme is used. The Newton method is applied in every time step to handle the material nonlinearities.

3 Adjoint method for nonlinear EQS problems

Numerical optimization studies the effects of multiple design parameters, \(p = \{p_1, ..., p_j, ..., p_{N_p}\}\), on the (QoIs), \(G_k (\phi, p)\), \(k = 1, ..., N_{\text{QoI}}\). In each FE simulation, one parameter combination \(p_0\) is adopted, and the (QoIs) are analyzed by post-processing the electric scalar potential. Common design parameters are, in particular, material parameters and the dimensions of the geometry. Taking the cable joint of Sect. 4.2 as an example, possible (QoIs) are, e.g., the maximum tangential field stress at material interfaces, or the electric losses during impulse operation [11,22].

The AVM is a method for gradient or sensitivity calculation, which is particularly efficient when the number of parameters, \(N_p\), is significantly larger than the number of QoIs, \(N_{\text{QoI}}[3,16]\). Sensitivities describe how and how strong a given QoI \(G_k\) is affected by a design parameter \(p_j\), i.e.

\[
\frac{dG_k}{dp_j} (p_0) = \frac{\partial G_k}{\partial \phi} (p_0) + \frac{\partial G_k}{\partial \phi} \frac{d\phi}{dp_j} (p_0), \tag{5}
\]

where \(p_0\) is the active parameter configuration. In case of nonlinear media, the sensitivity of the electric potential with respect to the parameter, \(\frac{d\phi}{dp_j}\), is typically unknown. The idea of the AVM is to avoid the computation of \(\frac{d\phi}{dp_j}\) by a clever modification of the (QoIs) [3,16]: The (QoIs) are expressed in terms of a functional \(G_k\), which is integrated over the temporal and spatial computational domain, \([0, T] \times \Omega\). Additionally, the nonlinear EQS problem (1) is embedded, multiplied by a test function \(w_k (r, t)\), i.e.

\[
G_k (\phi, p) = \int_{0}^{T} \int_{\Omega} g_k (\phi, r, t, p) \, d\Omega \, dt \\
\quad - \int_{0}^{T} \int_{\Omega} w_k (r, t) \cdot (- \nabla (\sigma \nabla \phi) - \nabla (\partial_t (\varepsilon \nabla \phi))) \, d\Omega \, dt .
\] (6)

For any \(\phi\) solving (1), the additional term is zero and the test function can be chosen freely. The goal of the AVM is to choose the test function in such a way, that the sensitivity of the extended QoI no longer contains the unknown term \(\frac{\phi}{p_j}\). After a lengthy derivation, it can be shown that the unknown term is eliminated if the test function is chosen as the so-
called adjoint variable, i.e. the solution of the adjoint problem [3,16]. The adjoint problem for EQS problems with nonlinear materials reads

\[- \text{div}(\sigma \text{ grad}(w_k)) + \text{div}(\varepsilon_\ell \partial_t \text{ grad}(w_k)) = \frac{dg_k}{d\phi},\]

\(t \in [0, T], \ r \in \Omega;\)

\(w_k = 0, \ t \in [0, T], \ r \in \Gamma_c;\)

\(- (\sigma_\ell \text{ grad}(w_k) - \varepsilon_\ell \partial_t \text{ grad}(w_k)) \cdot n = 0,\]

\(t \in [0, T], \ r \in \Gamma_m;\)

\(w_k = 0, \ t = T, \ r \in \Omega,\]

(7a)

(7b)

(7c)

(7d)

where all quantities are evaluated at the active parameter configuration \(p_0\). Note the plus sign in front of the term with the time derivative in (7a) instead of the minus sign in (1a) and the terminal condition (7d) instead of the initial condition (1d), which indicate that the adjoint problem needs to be integrated backwards in time or the time reversing variable transformation \(\hat{t} = T - t\) must be applied [2].

The adjoint problem is a linear partial differential equation (PDE) that naturally includes the tensorial material linearizations \(\sigma_\ell = \frac{d\sigma}{dp}\) and \(\varepsilon_\ell = \frac{d\varepsilon}{dp}\) [7]. Through that, it implicitly depends on the solution of the EQS problem, i.e., \(\sigma_\ell(E)\) and \(\varepsilon_\ell(E)\). Therefore, in order to solve the adjoint problem in backward mode, the EQS problem must first be solved conventionally, i.e., in forward mode, and its solution stored for all time steps. In case of FE simulations, this can lead to a significant memory overhead [2,6]. For strategies on how to reduce the memory requirement, see for example [2,6].

Once the solution of the electric potential and all adjoint variables, \(w_k\), are available, all sensitivities can be computed directly by

\[\frac{dG_k}{dp_j}(p_0) = \int_0^T \int_\Omega \frac{d\theta}{dp_j} + \text{grad} (w_k),\]

\[- \int_\Omega \text{ grad} (w_k) \cdot \frac{dD}{dp_j} \text{ d}\Omega \bigg|_{t=0},\]

(8)

where the derivative \(\frac{dD}{dp_j}(t = 0)\) is obtained by differentiating the initial condition (1d). Again, all quantities are evaluated for the active parameter configuration \(p_0\).

### 3.1 Finite element discretization

The derivative of the electric scalar potential to the parameter \(p_j\) and the adjoint variable are discretized using linear FE nodal shape functions, i.e.

\[\frac{d\phi}{dp_j}(r, t) \approx \sum_{r=1}^{N_{\text{node}}} u'_r(t) N_r(r) \quad w(r, t) \approx \sum_{r=1}^{N_{\text{node}}} u_r(t) N_r(r),\]

and the time axis is discretized using \(N_t\) samples, i.e., \(t \in \{t_1 = 0, ..., t_{N_t} = T\}\). The semi-discrete version of the adjoint problem (7) then reads

\[K_{\sigma_\ell} w - K_{\varepsilon_\ell} \frac{dw}{dt} = q,\]

(9)

with

\[[K_{\sigma_\ell}]_{rs} = \int_\Omega \text{ grad} (N_r) \cdot \sigma_\ell \text{ grad} (N_s) \text{ d}\Omega \quad r, s = 1, ..., N_N;\]

\[[K_{\varepsilon_\ell}]_{rs} = \int_\Omega \text{ grad} (N_r) \cdot \varepsilon_\ell \text{ grad} (N_s) \text{ d}\Omega \quad r, s = 1, ..., N_N;\]

\[[q]_r = \int_\Omega \frac{\partial g}{\partial u_r} \text{ d}\Omega \quad r = 1, ..., N_N.\]

Finally, the semi-discrete version for the sensitivity calculation reads

\[\frac{dG_k}{dp_j}(p_0) = \int_0^T \int_\Omega \frac{d\theta}{dp_j} - u^T K_{\sigma_\ell} w + u^T K_{\varepsilon_\ell} \frac{dw}{dt} \bigg|_{t=0} + (u')^T K_{\varepsilon_\ell} w \bigg|_{t=0},\]

(13)

with

\[[K_{\sigma_\ell}]_{rs} = \int_\Omega \frac{\partial \sigma}{\partial p_j} \text{ grad} (N_r) \cdot \text{ grad} (N_s) \text{ d}\Omega \quad r, s = 1, ..., N_N;\]

\[[K_{\varepsilon_\ell}]_{rs} = \int_\Omega \frac{\partial \varepsilon}{\partial p_j} \text{ grad} (N_r) \cdot \text{ grad} (N_s) \text{ d}\Omega \quad r, s = 1, ..., N_N;\]

\[[q]_r = \int_\Omega \frac{\partial g}{\partial u_r} \text{ d}\Omega \quad r = 1, ..., N_N.\]

In the scope of this work, the time integral of (13) is computed using trapezoidal integration and the time derivative in (9) is approximated using the implicit Euler method.

### 3.2 Treatment of pointwise QoIs

As can be seen from (6), the AVM is naturally suited for integrated (QoIs). Often, however, it is desired to analyze (QoIs) that are evaluated at certain points in space or time. The evaluation at a certain position or time can be expressed by Dirac delta functions inside the functional \(g_k\). To illustrate
the effects this has on the AVM, the electric potential evaluated at a specified position \( r_{\text{ref}} \) and time \( t_{\text{ref}} \) is considered as an example, i.e.

\[
G_k = \int_0^T \int_\Omega g_k \, d\Omega \, dt = \int_0^T \int_\Omega \delta(r - r_{\text{ref}}) \delta(t - t_{\text{ref}}) \phi \, d\Omega \, dt ,
\]

where \( \delta \) is the Dirac delta function. The right-hand side of the adjoint problem is then given by

\[
\frac{dg}{d\phi} = \delta(r - r_0) \delta(t - t_{\text{QoI}}) ,
\]

and after discretization in space one finds

\[
q(t) = [0 \ldots 0 \, 1 \ldots 0 \ldots 0]^T \delta(t - t_{\text{QoI}}) .
\]

In (19), the spatial integration during the derivation of the FE formulation has converted \( \delta(r - r_0) \) into a unit excitation at the corresponding node. The temporal Dirac function \( \delta(t - t_{\text{QoI}}) \) on the other hand must be approximated during numeric integration. In the context of this work, this is done by hat functions with an area of one, i.e.,

\[
q(t_n) = [0 \ldots 0 \, 1 \ldots 0 \ldots 0]^T \frac{1}{\Delta_{\text{imp}}} \delta_{\text{ref}}^n ,
\]

where \( \Delta_{\text{imp}} \) denotes the time step size right before and after \( t_{\text{ref}} \). The approximation of the Dirac impulse with the help of other functions, e.g., a normal distribution, led to similar results. However, the approximation by a hat function is easy to implement and has the clear advantage of a compact support.

### 4 Results

In this section, the AVM (7) for transient EQS problems is validated. In the first step, the layered resistor of Fig. 1a is considered, and the FE adjoint and analytic sensitivities are compared. In a second step, the method is applied to a nonlinear 320 kV cable joint specimen and the results are validated using results obtained by the DSM as a reference.

#### 4.1 Analytical example

The first example is the layered resistor depicted in Fig. 1a. The upper electrode is excited with a sinusoidal voltage, i.e., \( U(t) = 1 \, \text{V} \cdot \sin(\omega t) \) with \( \omega = 2\pi \cdot 50 \, \text{Hz} \), and the bottom electrode is grounded. For \( t = 0 \), the potential is assumed to be zero everywhere. The conductivities, \( \sigma_1 = 10 \, \text{A/Vm} \) and \( \sigma_2 = 20 \, \text{A/Vm} \), and the permittivities, \( \varepsilon_1 = 40 \, \text{As/Vm} \) and \( \varepsilon_2 = 60 \, \text{As/Vm} \), of the two materials are constant. The EQS AVM is validated for an integrated QoI as well as for non-integrated QoI. More specifically, the (QoIs) are the electrical energy converted in the time span \([0 \ldots \frac{2\pi}{\omega} \ldots 0] \) s,

\[
W_{\text{el}} = \int_{0s}^{\frac{2\pi}{\omega}s} \int_\Omega \sigma(\Delta) \phi^2 \, d\Omega \, dt ,
\]

and the potential in the middle of the upper material, i.e., \( r_{\text{QoI}} = (0, \frac{d}{2}) \), evaluated at \( t_{\text{QoI}} = \frac{\pi}{\omega} \) s.

\[
\phi_{\text{ref}} = \int_{0s}^{\frac{2\pi}{\omega}s} \int_\Omega \delta(r - r_{\text{QoI}}) \delta(t - t_{\text{QoI}}) \phi \, d\Omega \, dt .
\]

First, the sensitivity \( \frac{d\phi_{\text{ref}}}{dt_1} \) of the reference potential with respect to the permittivity of the upper material is computed. The time axis is discretized using the step size \( \Delta_{\text{main}} \). Directly before and after \( t = t_{\text{QoI}} \), the step size is reduced to \( \Delta_{\text{imp}} = 10^{-8} \Delta_{\text{main}} \) in order to approximate the Dirac impulse. Figure 1b shows that the AVM is able to reproduce the analytic results of \( \frac{d\phi_{\text{ref}}}{dt_1} \) for a wide range of permittivity values. In Fig. 1c a first-order convergence of the relative error with respect to the number of time steps can be observed, which matches the order of the implicit Euler method.

Next, the sensitivity \( \frac{dW_{\text{el}}}{dt_1} \) of the electric energy with respect to the conductivity of the upper material is computed for conductivities ranging from 1 A/Vm to 1000 A/Vm (see Fig. 2a). As shown in Fig. 2b the results again converge linearly with respect to the number of time steps. The AVM
Transient standard 1.2/50 lightning impulse is given by \[14\] that is superimposed on the HVDC excitation of 320 kV. The overvoltage clipping of metal-oxide surge arresters [20,21] has, thus, been successfully validated for a transient EQS problem.

### 4.2 320 kV HVDC cable joint specimen

The development of HVDC cable systems is one of the greatest challenges of our time for the HV engineering community [5,10,17]. Cable joints are known to be the most vulnerable part of HVDC systems, as they must safely handle field strengths in the range of several kV/mm [4,5,13,19]. The electric field stress can be reduced by adding a layer of so-called FGM, which features a strongly nonlinear electric conductivity and balances the electric field, similar to the overvoltage clipping of metal-oxide surge arresters [20,21].

The AVM is applied to a 320 kV HVDC cable joint specimen, which is adopted from [11] and shown in Fig. 3. The joint connects two copper conductors (domain 1) with an aluminum connector (domain 2). These domains are covered by a layer of conductive silicone rubber (SiR) (domain 3). The cable insulation consists of cross-linked polyethylene (XLPE) (domain 4) and the joint insulation of an insulating SiR (domain 5). Both insulation layers are separated by a nonlinear resistive FGM (domain 6, highlighted in green). The outer conductive SiR sheaths of the cable (domain 7) and the joint (domain 8) are on ground potential. Inside the FGM, elevated field stresses occur at the triple points, i.e., the contact points of FGM, insulating material and conductive SiR (indicated by red circles). The joint is subjected to an impulse overvoltage with an amplitude of \(U = 100\) kV that is superimposed on the HVDC excitation of 320 kV. The transient standard 1.2/50 lightning impulse is given by [14]

\[
U_{\text{imp}}(t) = \bar{U} \cdot \frac{\tau_2}{\tau_2 - \tau_1} \left( \exp \left( -\frac{t}{\tau_2} \right) - \exp \left( -\frac{t}{\tau_1} \right) \right),
\]

with \(\tau_1 = 1.2\ \mu s\) and \(\tau_2 = 50\ \mu s\). The excitation is applied to the conductive SiR covering the conductor and the conductor clamp, which is modeled as a perfect electric conductor.

The field dependence of the conductivity of the FGM is described by the analytic function

\[
\sigma(E) = a_1 \frac{1 + a_2 (E - a_3) a_2^{-1}}{1 + a_4 (E - a_3) a_4^{-1}},
\]

with the parameters \(a_1 = 10^{-10} \text{ A/Vm}, a_2 = 0.7 \cdot 10^6 \text{ V/m}, a_3 = 2.4 \cdot 10^6 \text{ V/m} \) and \(a_4 = 1864\). The simulation is performed with a mesh consisting of 51681 nodes and 106113 elements.

The results of the EQS AVM are validated against results of the DSM for two exemplary sensitivities. Again, both a time-integrated QoI and a QoI evaluated at a specific point in time are considered, i.e., the electric losses, \(W_{\text{el}}\), during the time span \([0 \mu s, \tau_{\text{rise}}]\) and the critical electric field stress, \(E_c\), in the proximity of the triple point next to the conductor clamp during peak excitation. The derivatives of the (QoIs) are computed with respect to the switching field strength, \(a_2\), which determines when the conductivity changes from the base conductivity \(a_1 = 10^{-10} \text{ S/m} \) into the strongly nonlinear region of (24) (see Fig. 4a).

Figure 4b shows the relative error of the derivatives for different numbers of time steps. A first-order convergence due to the Euler time-stepping scheme is observed for both quantities of interest(QoIs). More importantly, the computational cost is reasonable for both the time-integrated QoI and the QoI evaluated at a given point in time. With only 200 time steps, the relative errors are below one percent, requir-
ing a simulation time in the range of tens of minutes. The AVM was, thus, successfully validated for a strongly nonlinear and highly transient example. It was shown that with the method presented in Sect. 3.2 the AVM is no longer restricted to integrated quantities of interest (QoIs). Moreover, it was demonstrated that even for this very challenging example the computation time lies within reasonable limits, which is an important first step toward gradient-based optimization.

5 Conclusion

The adjoint variable method is a method for calculating gradients of selected quantities of interest with respect to a set of design parameters. It has computational costs nearly independent of the number of design parameters and is, thus, very efficient for problems where the number of parameters is larger than the number of quantities of interest. In this work, the adjoint variable method is adopted for transient electro-quasistatic problems with nonlinear material characteristics. The adjoint partial differential equation is presented and formulated as a two-dimensional axisymmetric finite element problem. It is shown, how to consider quantities of interest evaluated at specific points in space or time. After validating the method against an analytic example, the method is applied to a 320 kV high-voltage direct current cable joint featuring a layer of nonlinear field grading material which is exposed to an impulse overvoltage. The results of the adjoint variable method are validated using the direct sensitivity method, and it is shown that the computational costs of the adjoint variable method are, even for this strongly nonlinear technical example, within reasonable limits. This is an important step toward gradient-based optimization of high-voltage equipment.

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Declaration

Conflict of interest The authors declare no conflict of interest.

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