Network Massive MIMO for Cell-Boundary Users: From a Precoding Normalization Perspective

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Abstract—In this paper, we propose network massive multiple-input multiple-output (MIMO) systems, where three radio units (RUs) connected via one digital unit (DU) support multiple user equipments (UEs) at a cell-boundary through the same radio resource, i.e., the same frequency/time band. For precoding designs, zero-forcing (ZF) and matched filter (MF) with vector or matrix normalization are considered. We also derive the formulae of the lower and upper bounds of the achievable sum rate for each precoding. Based on our analytical results, we observe that vector normalization is better for ZF while matrix normalization is better for MF. Given antenna configurations, we also derive the optimal switching point as a function of the number of active users in a network. Numerical simulations confirm our analytical results.

Index Terms—Network massive MIMO, cloud BS, cell-boundary users, capacity bound, precoding, normalization.

I. INTRODUCTION

Multiple-input multiple-output (MIMO) wireless communication techniques have evolved from single-user to multiple-user systems [1]. To approach the capacity of the MIMO broadcast channel, the authors in [2], [3] proposed simple zero-forcing (ZF) based-linear algorithms, where the transmitter and the receivers are equipped with multiple antennas. The optimality of the linear algorithm was intensively investigated in [4] with an assumption of an infinite number of antennas at the receiver. The authors in [4] proved that a simple linear beamforming (coordinated beamforming in the paper) asymptotically approaches the sum capacity achieved by dirty paper coding (DPC).

Recently, massive MIMO (a.k.a. large-scale MIMO) has been proposed to further maximize network capacity and to conserve energy [5]–[7]. In [5], the authors showed that a single-cell system with an unlimited number of antennas at the transmitter is always advantageous. The authors in [6] investigated the energy and spectral efficiency for massive MIMO systems for a single-cell environment. It was shown, however, that the capacity gains obtained by multiuser MIMO processing degrade severely in multi-cell environments. To further maximize the network capacity, several network MIMO algorithms with multiple receive antennas have been proposed [8], [9]. These systems assume, however, that the network supports maximum of three users through a relatively small number of transmit antennas [1].

Massive MIMO systems in multi-cell environments were also studied in [7], [10]. As investigated in [7], [10], multi-cell massive MIMO has some critical issues, such as pilot contamination, that have to be resolved before it can be applied in practice. In [10], the authors showed theoretically and numerically the impact of pilot contamination, proposing a multi-cell minimum mean square error (MMSE) based precoding algorithm to reduce both intra- and inter-cell interference. In [10], matched filter (MF) precoding was used. Inter-user interference is eventually eliminated once the transmitter has large enough number of antennas. The assumption of an infinite number of antennas at the transmitter, however, is not, in practice, really feasible. This issue was studied in [11], [12]. The author in [12] concluded that the proposed architecture achieves the same spectral efficiency with ten times less antennas than previously proposed systems [7]. Note that the number of antennas is still quite large considering the current radio frequency (RF) techniques.

In this paper, we consider a network architecture (named cloud BS) to implement feasible massive MIMO systems. The network massive MIMO system consists of multiple radio units (RUs) connected with one another by optical fibers, and further connected to a centralized digital unit (DU), as illustrated in Figs. 1 and 2. Through the optical fibers, each RU can share the power, data messages, and channel state information; thus network massive MIMO with relatively small number of antennas can be treated as a single-cell massive MIMO with large-scale antennas. For algorithm designs, we consider ZF and MF precodings with two normalization techniques (vector/matrix normalizations). Most prior work on multiuser MIMO algorithms has paid little attention to this

1Note that more than three users can be supported if there is a common message, i.e., for a clustered broadcast channel.
2To avoid confusion, we use RU instead of base station hereafter.
problem statement in respect to precoding normalization methods and beamforming techniques for cell-boundary users. In Section III we analyze i) rate lower and upper bounds, ii) ergodic performance of ZF- and MF-precoding, and iii) network performance for cell-boundary users. Numerical results and discussions are presented in Section IV. Section V presents our conclusions.

II. SYSTEM MODEL AND PROBLEM STATEMENT

In this section, we introduce the basic notation used in this paper and the network massive MIMO system model[2].

A. System Model: Network Massive MIMO

Consider a cooperative network massive MIMO system as shown in Figs. 1 and 2. One DU controls three RUs and K users. Each RU is connected with one another by optical fibers. We assume that each RU has N_t transmit antennas and each user equipment (UE) is equipped with one receive antenna. We also assume that the channel is flat fading and the elements of a channel matrix are modeled as independent complex Gaussian random variables with zero mean and unit variance. The channel between cloud BS (one DU and three RUs) and the k-th user is denoted by an 1 × M row vector h_k^T (k = 1, 2, · · · , K), where M represents the number of cloud BS antennas 3N_f. A K × M channel matrix H between cloud BS and all UEs consists of channel vectors h_k^T. Let g_k denote the column vector of transmit precoding and s_k represent the transmit symbol for the k-th UE. Also, let n_k be the additive white Gaussian noise vector. Then, the received signal at the k-th UE is expressed by

\[ y_k = \sqrt{P}h_k^T g_k s_k + \sum_{\ell=1, \ell \neq k}^{K} \sqrt{P}h_\ell^T g_\ell s_\ell + n_k \]

where P denotes the total network transmit power across three RUs.

B. Problem Statement: Precoding Normalization Perspective

Eq. (1) contains the desired signal, interference, and noise terms. To eliminate the interference term, to maximize the signal-to-noise-ratio (SNR), we use the following precoding.

ZF : F = H^T (HH^T)^{-1} = [f_1, f_2, · · · , f_K] 
MF : F = H^* = [f_1, f_2, · · · , f_K]

where F is a precoding matrix consisting of each column vector f_k.

To satisfy the power constraint, we need to normalize the precoding matrix. In this paper, as mentioned earlier, two methods, i.e., vector/matrix normalizations, are considered. The normalized transmit beamforming vectors (columns of a precoding matrix) with vector/matrix normalizations are given as g_k = f_k/∥f_k∥ and g_k = f_k/∥F∥_F, respectively.

1) ZF/MF with vector normalization: The received signal at the k-th UE can be expressed as follows:

\[ y_k = \sqrt{P}h_k^T \frac{f_k}{\sqrt{K}∥f_k∥} s_k + \sum_{\ell=1, \ell \neq k}^{K} \sqrt{P}h_\ell^T \frac{f_\ell}{\sqrt{K}∥f_\ell∥} s_\ell + n_k. \]

(2)

2) ZF / MF with matrix normalization: Similarly, we can rewrite the received signal with matrix normalization as below:

\[ y_k = \sqrt{P}h_k^T \frac{f_k}{∥F∥_F} s_k + \sum_{\ell=1, \ell \neq k}^{K} \sqrt{P}h_\ell^T \frac{f_\ell}{∥F∥_F} s_\ell + n_k. \]

(3)
III. ASYMPTOTIC RATE BOUNDS: ZF AND MF CASES

In this section, we derive the capacity bounds and show which normalization method is suitable for ZF- and MF-type precoding. Based on our analytical results, we will also show which precoding technique is desired for cell-boundary users.

A. Capacity Bound

Using Jensen’s Inequality of convex and concave functions, we can get the capacity’s lower and upper bounds as follows:

\[
\log_2 \left( 1 + \frac{1}{E \left( \frac{I + N}{S} \right)} \right) \leq \log_2 \left( 1 + \frac{S}{I + N} \right) \leq \log_2 \left( 1 + E \left( \frac{S}{I + N} \right) \right) .
\]  

(4)

B. Ergodic performance of ZF precoding

1) Vector normalization-lower bound: From (2), we can derive the lower bound of vector normalization in the ZF case as follows:

\[
\mathbb{E}_{ZF} \left\{ \frac{1 + N}{S} \right\} = \mathbb{E} \left\{ \frac{P \sum_{\ell=1, \ell \neq k}^K h_k^T f_\ell \sqrt{K} \| f_\ell \|}{P h_k^T f_k \sqrt{K} \| f_k \|} \right\}^2 + 1 \right\}
\]

\[
= \mathbb{E} \left\{ \frac{1}{P} \left( \frac{K \| f_k \|^2}{P} \right) \right\} = \frac{1}{P} \mathbb{E} \left\{ \frac{K}{P} \right\} = \frac{1}{P} \frac{K}{M - K + 1}
\]

(5)

where (a) follows the property of ZF precoding. By using ZF, the interference term is perfectly eliminated and \( h_k^T f_k \) is equal to one in (a). Equality (b) results from the diversity order of ZF, as shown in [13].

2) Matrix normalization-upper bound: From (2), the upper bound of matrix normalization in the ZF case can be expressed as

\[
\mathbb{E}_{ZF} \left\{ \frac{S}{1 + N} \right\} = \mathbb{E} \left\{ \frac{P}{P \sum_{\ell=1, \ell \neq k}^K h_k^T f_\ell \sqrt{K} \| f_\ell \|} \right\}^2 + 1 \right\}
\]

\[
= \mathbb{E} \left\{ \frac{1}{P} \left( \frac{K \| f_k \|^2}{P} \right) \right\} \approx \frac{1}{P} \frac{K}{M - K}
\]

(6)

where (c) can be obtained by

\[
\mathbb{E} \{ \| F \|^2 \} = \mathbb{E} \left\{ \text{tr} \left( H^* (HH^*)^{-1} (HH^*)^{-1} H \right) \right\} = \mathbb{E} \left\{ \text{tr} \left( (HH^*)^{-1} (HH^*)^{-1} \right) \right\} = \frac{K}{M - K}
\]

using the property of Wishart matrix of random matrix theory [14].

3) Performance comparison of ZF: To find which normalization technique is better in ZF, we compare the lower rate bound of vector normalization (\( \mathcal{R}_{ZF_{vec}}^L \)) with the upper rate bound of matrix normalization (\( \mathcal{R}_{ZF_{mat}}^U \)) in ZF; thus the gap is given by

\[
\mathcal{R}_{ZF_{vec}}^L - \mathcal{R}_{ZF_{mat}}^U = \log_2 \left( 1 + \frac{P(M - K + 1)}{K} \right) - \log_2 \left( 1 + \frac{P(M - K)}{K} \right)
\]

(7)

\[
= \log_2 \left( \frac{K + P(M - K) + P}{K + P(M - K)} \right) = \log_2 \left( 1 + \frac{P}{K + P(M - K)} \right) \geq 0.
\]

From (7), we could conclude that, in the ZF case, vector normalization is always better than matrix normalization.

C. Ergodic performance of MF precoding

1) Matrix normalization-lower bound: From (3), the lower rate bound of matrix normalization is given as follows:

\[
\mathbb{E}_{MF} \left\{ \frac{I + N}{S} \right\} = \mathbb{E} \left\{ \frac{P \sum_{\ell=1, \ell \neq k}^K h_k^T f_\ell \sqrt{K} \| f_\ell \|}{P h_k^T f_k \sqrt{K} \| f_k \|} \right\}^2 + 1 \right\}
\]

\[
= \mathbb{E} \left\{ \frac{P(K - 1)}{P(K - 1) + 1} \right\} = \frac{P(K - 1) + K}{P(M + 1)}
\]

(8)

Note that MF precoding asymptotically eliminates the interference term in (1) when the cloud BS has reasonably large scale transmit antennas. Applying the properties of random vectors and the law of large numbers to (d), \( h_k^T f_k \) is equal to one in (a). Equality (b) results from the diversity order of MF, as shown in [13].

2) Vector normalization-upper bound: In a similar way, we can get the upper rate bound of vector normalization as follows:

\[
\mathbb{E}_{MF} \left\{ \frac{S}{I + N} \right\} = \mathbb{E} \left\{ \frac{P h_k^T f_k \sqrt{K} \| f_k \|}{P \sum_{\ell=1, \ell \neq k}^K h_k^T f_\ell \sqrt{K} \| f_\ell \|} \right\}^2 + 1 \right\}
\]

\[
= \mathbb{E} \left\{ \frac{P h_k^T f_k \sqrt{K} \| f_k \|}{P h_k^T f_k \sqrt{K} \| f_k \|} \right\} = \frac{P h_k^T f_k \sqrt{K} \| f_k \|}{P h_k^T f_k \sqrt{K} \| f_k \|}
\]

(9)

\[
= \frac{P h_k^T f_k \sqrt{K} \| f_k \|}{P h_k^T f_k \sqrt{K} \| f_k \|} = \frac{P(K - 1) + K}{P(M + 1)}.
\]

3) Performance comparison of MF: We also compare the rate bounds, and the gap is given by

\[
\mathcal{R}_{MF_{vec}}^L - \mathcal{R}_{MF_{mat}}^U = \log_2 \left( 1 + \frac{P(M + 1)}{P(K - 1) + K} \right) - \log_2 \left( 1 + \frac{P(M + 1)}{P(K - 1) + K} \right)
\]

(11)

\[
= 0.
\]

where \( \mathcal{R}_{MF_{vec}}^L \) and \( \mathcal{R}_{MF_{mat}}^U \) denote the lower rate bound of vector normalization and the upper rate bound of vector normalization, respectively. From (11), we confirm that, for MF precoding, matrix normalization is always better than vector normalization.
Fig. 3. The difference of the gradient between ZF and MF at $K_{cross}$ when $P$ is very small (almost zero) and $M$ is much larger than $P$. The difference is always positive (> 0).

D. Low SNR Analysis

In Section III, we investigated which normalization is appropriate for ZF and MF precoding techniques. In this section, we will analyze which precoding technique is better for cell-boundary users, i.e., low signal-to-interference-plus-noise (SINR) users.

Both $R_{ZFvec}^L$ and $R_{MFmat}^L$ are concave functions. Also, unlike $R_{ZFvec}^L$, $R_{MFmat}^L$ is a monotonic increasing function; thus, two cross points exist: one is when the number of users $K$ is one, the other is when the number of users $K$ has the following value with a large $M$ approximation:

$$K_{cross} = \frac{P(M + 1)}{1 + P}$$  (12)

where, $K_{cross}$ denotes the crossing point as functions of the number of cloud BS antennas ($M$) and the total transmit power ($P$). In (12), intuitively, $K_{cross}$ should be greater than zero and integer. Therefore, at the low SNR regime, (12) goes to zero, thus the limit becomes one with a $K_{cross}$ constraint. This means that as SNR decreases the cross point shifts to the left.

At $K_{cross}$, we check the difference of the gradient between ZF and MF. If the gradient of ZF is larger than that of MF, the rate of ZF with vector normalization is larger than that of MF when $K \geq K_{cross}$. In the other case, the rate of MF with matrix normalization is larger than that of MF when $K \geq K_{cross}$. The difference of the gradient between ZF and MF is expressed as (10) on the bottom, where $G_{MFmat}^L$ denotes the gradient of the $R_{MFmat}^L$ curve at $K_{cross}$. Similarly, $G_{ZFvec}^L$ is the gradient of the $R_{ZFvec}^L$ curve at $K_{cross}$. In general, cell-boundary users have relatively low SINR and, as we assumed, the cloud BS has large-scale antennas, meaning $M$ is much larger than $P$. Therefore, if $K_{cross}$ exists, (10) is always positive. We also confirm this through numerical comparisons as shown in Fig.3. From this observation, we realize that MF precoding is suitable for cell-boundary users if the number of active users is larger than $K_{cross}$.

IV. DISCUSSION

For numerical comparisons, we assume that each RU has eight transmit antennas; thus the cloud BS has a total of 24 antennas. Note that any number of antennas can be used and this constraint is not really related to our system. This assumption is based on a 3GPP LTE-advanced’s parameter; Release 10 supports eight Node B antennas [15]. Instead of increasing the number of antennas at each transmitter, we propose using the more feasible cloud concept. Indeed, having more than eight transmit antennas would be difficult due to pilot overhead and other system constraints. We are not arguing here that we have to have only eight antennas, rather we show the gain of massive MIMO that can be achieved through a simple cooperation with relatively small number of antennas at each RU.

In Fig. 4, we compare the achievable sum rates of ZF precoding with MF precoding when the total transmit SNR is 0 dB. As was shown in Section III, ZF with vector normalization is better. In contrast, MF with matrix normalization is better at getting an improved sum rate performance. Fig. 5

![Achievable Rate Comparison (M=24 / Total SNR=0 dB)](image)

Fig. 4. Achievable rate vs. the number of cell-boundary users, where $M = 24$, $K = [1, 24]$, and total SNR = 0 dB.

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We derive this by using two rate bounds equations and omit the derivations here.

$$G_{MFmat}^L - G_{ZFvec}^L = 4(M + 1)P - M^2P^2 + (MP + P - 1)\sqrt{MP(MP - 4) - 4P} \over (M + 1)(2MP + P + 1)$$

$$= -((MP - 2)^2 - 4(P + 1)) + (MP + P - 1)\sqrt{(MP - 2)^2 - 4(P + 1)} \over (M + 1)(2MP + P + 1)$$  (10)
Fig. 5. Achievable rate vs. the number of cell-boundary users, where \( M = 24 \), \( K = [1, 24] \), and total SNR = -5 dB.

illustrates the achievable sum rates of ZF- and MF-precoding with -5 dB transmit SNR. The result is similar to that of Fig. \[4\] As we mentioned in Section III-D, in the low SNR regime (meaning that users are located at cell-boundary), using MF precoding is generally better when the number of active users is larger than \( K_{\text{cross}} \). Also, we realize through Figs. \[4\] and \[5\] that as SNR decreases \( K_{\text{cross}} \) shifts to the left. The derived \( K_{\text{cross}} \) expression is also verified through numerical results. We summarize our conclusions in Tables \[I\] and \[II\].

V. CONCLUSION

In this paper, we proposed network massive MIMO systems supporting multiple cell-boundary UEs. For precoding designs, we first derived the achievable rate bounds of zero-forcing (ZF) and matched filter (MF) with vector or matrix normalization. Through analytical and numerical results, we confirmed that vector normalization is better for ZF while matrix normalization is better for MF. We also investigated the optimal mode switching point as functions of the number of active users in a network and the total transmit power. In future work, we will consider limited cooperation among RUs and cooperation delay.

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