Manufacturing and Characterization of Three-Axis Magnetic Sensors Using the Standard 180 nm CMOS Technology

Chi-Han Wu 1, Po-Jen Shih 2, Yao-Chuan Tsai 3 and Ching-Liang Dai 1,*

1 Department of Mechanical Engineering, National Chung Hsing University, Taichung 402, Taiwan; abc7755998@smail.nchu.edu.tw
2 Department of Biomedical Engineering, National Taiwan University, Taipei 106, Taiwan; pjshih@ntu.edu.tw
3 Department of Bio-Industrial Mechatronics Engineering, National Chung Hsing University, Taichung 402, Taiwan; yctsaii@dragon.nchu.edu.tw
* Correspondence: cldai@dragon.nchu.edu.tw; Tel.: +886-4-2284-0433; Fax: +886-4-2287-7170

Abstract: A three-axis micro magnetic sensor (MS) is developed based on the standard 180 nm complementary metal oxide semiconductor (CMOS) technology. The MS designs two magnetic sensing elements (MSEs), which consists of an x/y-MSE and a z-MSE, to reduce cross-sensitivity. The x/y-MSE is constructed by an x-MSE and an y-MSE that are respectively employed to detect in the x- and y-direction magnetic field (MF). The z-MSE is used to sense in the z-direction MF. The x/y-MSE, which is constructed by two magnetotransistors, designs four additional collectors that are employed to increase the sensing current and to enhance the sensitivity of the MS. The Sentaurus TCAD software simulates the characteristic of the MS. The measured results reveal that the MS sensitivity is 534 mV/T in the x-direction MF, 525 mV/T in the y-direction MF and 119 mV/T in the z-axis MF.

Keywords: micro magnetic sensor; three-axis sensing; high sensitivity; CMOS; MEMS

1. Introduction

Magnetic sensors (MSs) have a wide range of uses and can be applied in control and monitoring devices, industrial products, material testing, manufacturing systems and biomedical engineering. For instance, Lu [1] used a microelectromechanical system (MEMS) MS to develop a location tracking system that was applied in real-time tracking of the vessel location and organ shape for surgical navigation. The location tracking system provided a magnetic field (MF) to a patient and employed the MS node to detect the organ location of the patient. Lee [2] utilized a 3 × 3 array of Hall MSs to construct an MF measurement system for MF mapping. The system could measure the MF distributions of a group of multi-magnets. Oh [3] adopted a small magnet, an MEMS MS and a breathing output component to compose a respiratory monitoring and training system that was used to train and monitor the breath of patients for radiotherapy. Lara-Castro [4] employed a micromachined MS to constitute a portable signal conditioning system for industrial applications to measure ferromagnetic material characteristic. The MS was composed of a silicon resonator, an aluminum loop and four piezoresistors. Krishnapriya [5] manufactured a micro MS with a planar micro coil for biomedical application to detect biomolecules and pathogens. An integrated microfluidic platform, proposed by Feng [6], was assembled by a micro MS and a micro-spiral planar coil. The microfluidic platform was used to detect and manipulate magnetic beads. Zhang [7] designed a magnetic scanning device with a digital micro MS. The magnetic scanning device, which was a non-destructive test, could detect the leakage of MF from steel corrosion in reinforced concrete. An unmanned aerial vehicle (UAV) navigation system, presented by Vetrella [8], was
built using an MEMS MS, a global positioning system receiver, an inertial sensor, a navigation algorithm and a vision component. The system was utilized to control and stabilize the UAV flight.

The advantages of micro MS are that is has a small volume, easy integration and a high performance. Recently, the MEMS technology was adopted to design and fabricate various micro MSs. Table 1 summarizes the sensing principle and type for various micro magnetic sensors. For example, an MS, as proposed by Niekiel [9], was fabricated using the MEMS technology. The MEMS MS, which was a magnetic-resonant type, contained a piezoelectric resonator that integrated permanent magnets. Chen [10] developed a MEM MS, which was a giant magneto-impedance type, using surface micromachining. The magnetic sensing material for the sensor was the copper-based amorphous wire. The sensor had a micro coil which surrounded the copper-based amorphous wire, and the output signal for the sensor was produced by the micro coil. The sensitivity of the MEMS MS was 130 mV/Oe. A MEMS MS that was a magnetic-resonant type was presented by Okada [11]. The sensor had a resonator and a micro bridge. The resonator was a silicon bridge with a PZT thin film, and the micro bridge was silicon with a FePd film. When an MF was applied to the micro bridge it produced a deflection to act on the resonator, so that the stiffness of the resonator changed, which lead to a variation in the resonance frequency of the resonator. Guo [12] made an MEMS MS that was a fluxgate type using the micromachining with chemical wet etching. The MEMS MS had a double-layer magnetic core that was deposited by the melt-spinning. Nejad [13] designed an MS. The relation between the mechanical displacement and magnetostatic force for the sensor was analyzed by the governing equation. The MS was made by micromachining, and the MS beams were a triple-layer of gold/Ni/gold. A micromachined MS, developed by Bahreyni [14], was fabricated by bulk micromachining. The sensor, which was a magnetic-resonant type, had an electrostatic resonator. When an MF was supplied to the MS, the resonant frequency of the electrostatic resonator produced a change. The MS did not exhibit hysteresis because it did not use any magnetic materials. Tseng [15] used the CMOS process to fabricate a one-axis micro MS, and the MS sensitivity was 354 mV/T. Sileo [16] employed the microfabrication to make a three-axis Hall MS. The Hall MS structure was constituted by the AlGaAs/InGaAs/GaAs multilayered material. The MS sensitivity was 0.03 V/T. A three-axis MS, presented by Zhao [17], was manufactured through the MEMS technology. The MS had four magnetic transistors and a Hall element. The x-axis, y-axis and z-axis sensitivity for the MS were 77.5 mV/T, 78.6 mV/T and 77.4 mV/T, respectively. Yeh [18] manufactured a three-axis MS, which was a magnetic-piezoelectric type, based on the MEMS technology. The MS structure had a silicon diaphragm and a magnetic nickel thick film and a piezoelectric lead zirconate titanate (PZT) thin film located on the silicon diaphragm. When the magnetic nickel film was excited by an alternating current MF, the PZT film and the silicon diaphragm generated a vibration and displacement. The PZT film produced an output voltage. The x-axis, y-axis and z-axis sensitivity for the MS were 0.156 mV/Oe, 0.156 mV/Oe, and 0.035 mV/Oe, respectively. These sensors [9–15] were one-axis MSs. The sensors [16–18] were three-axis MSs. This work develops a three-axis micro MS, in which its sensitivity exceeds that of the sensors [16,17].

Table 1. Summary of sensing principle and type for various micro magnetic sensors.

| Authors     | Sensing Principle | Type  |
|-------------|-------------------|-------|
| Niekiel [9] | magnetic-resonance | 1-axis |
| Chen [10]   | giant magneto-impedance | 1-axis |
| Okada [11]  | magnetic-resonance | 1-axis |
| Guo [12]    | fluxgate          | 1-axis |
| Nejad [13]  | magnetic-resonance | 1-axis |
| Bahreyni [14] | magnetic-resonance | 1-axis |
| Tseng [15]  | magnetic-transistor | 1-axis |
The CMOS technology is adopted to make micro elements [19,20], micro actuators [21–24] and micro sensors [25–30]. The benefits of MSs that are made using the CMOS technology have low noise, high performance and easy mass-production. We employ this technology to design and manufacture a three-axis micro MS. To enhance the sensitivity and decrease the cross-sensitivity, the MS is built by two magnetic sensing elements (MSEs) that are an x/y–MSE and a z–MSE. The x/y–MSE senses in the x− and y−direction MF, and the z–MSE detects in the z−direction MF.

2. Design of Magnetic Sensor

The three-axis micro MS includes an x/y–MSE and an z–MSE. The x/y–MSE is combined by an x–MSE and a y–MSE. The x–MSE and y–MSE are used to detect the MF in the x−direction and the y−direction, respectively. The z–MSE is employed to measure in the z−direction MF.

Figure 1a displays the x/y–MSE structure, where E1a, E2a, E3a, and E4a are emitters; B1a, B2a, B3a and B4a are bases; CA1, CA2, CA3 and CA4 are additional collectors; C is a collector; STI is shallow trench isolation. The STI is utilized to restrict the current moving direction and to decrease the current leakage. The x−MSE that is a magnetic transistor is constructed by the emitters E1a and E3a, the additional collectors CA1 and CA3, the bases B1a and B3a, and the collector C. The y−MSE is a symmetric structure with the x−MSE, and the y−MSE is composed of the emitters E2a and E4a, the additional collectors CA2 and CA4, the bases B2a and B4a, and the collector C.

The y−MSE cross-sectional view is illustrated in Figure 1b. When the bias applies to the collector C, additional collectors (CA2 and CA4) and bases (B2a and B4a) and carriers move from the emitters (E2a and E4a) to the collector C, bases (B2a and B4a) and additional collectors (CA2 and CA4). There is an MF in the y−direction. The current and MF generate a Lorentz force that acts on carriers, resulting in carriers (on the right in Figure 1b) are bended downward. Most carriers migrate to the collector C, such that the current of the additional collector CA4 decreases. Opposite, carriers (on the left in Figure 1b) are bended upward by the Lorentz force. Most carriers migrate to the additional collector CA2, leading to an increase in the current of the additional collector CA2. Therefore, when the y−direction MF applies to the x/y−MSE, the x/y−MSE generates a voltage difference between the additional collectors CA2 and CA4. The output voltage (Vo) of the MS is obtained by the voltage difference of the additional collectors CA2 and CA4 when applying a y−direction MF to the x/y−MSE.

As shown in Figure 1b, the element structure of the x−MSE is similar to that of the y−MSE. The structure of the x−MSE, which consists of the emitters (E1a and E3a), the bases (B1a and B3a), the additional collectors (CA1 and CA3) and the collector C, is along the y−axis direction. When the bias applies to the collector C, additional collectors (CA1 and CA3) and bases (B1a and B3a) and carriers move from the emitters (E1a and E3a) to the collector C, bases (B1a and B3a) and additional collectors (CA1 and CA3). Suppose that there is an MF in the x−direction, then the current and the MF generate a Lorentz force that acts on carriers, such that carriers, which move from the emitter E1a to the base B1a and collector C, are bended downward. Most carriers migrate to the collector C, leading to a decrease in the current of the additional collector CA1. Opposite, carriers that move from the emitter E3a to the collector C and base B3a are deflected upward by the Lorentz force. Most carriers migrate to the additional collector CA3, such that the current of the additional collector CA3 increases. Therefore, when the x−direction, MF applies to the x/y−MSE, the x/y−MSE produces a voltage difference between the additional collectors CA1 and CA3. The MS Vo is obtained by the voltage difference of the additional collectors CA1 and CA3 when applying the x−direction MF to the x/y−MSE.
Figure 1. (a) x/y magnetic sensing element (MSE) structure; (b) Cross-sectional view for x/y–MSE along line AA.

Figure 2 displays the z–MSE structure of the MS. The z–MSE is composed of one emitter E, four bases and eight collectors. The STI oxide is used to restrict the current moving direction and reduces the current leakage. When the bias applies to the bases and the collectors, carriers migrate from the emitter to the bases and collectors. There is an MF in the z–direction. The current and the MF produces a Lorentz force that acts on carriers, such that carriers are bended toward the collectors (C_{1b}, C_{3b}, C_{5b} and C_{7b}), resulting in the currents of the collectors (C_{1b}, C_{3b}, C_{5b} and C_{7b}) being higher than that of the collectors (C_{2b}, C_{4b}, C_{6b} and C_{8b}). A voltage difference is generated between the collectors. The Vo of the MS is obtained by the voltage differences in series when applying a z–direction MF to the z–MSE.
Figure 2. The z–MSE structure, where E is emitter; B1b, B2b, B3b and B4b are base; C1b, C2b, C3b, C4b, C5b, C6b, C7b and C8b are collectors.

The performance of the x/y–MSE was simulated using the Sentaurus TCAD software. The x/y–MSE model (Figure 1a) was set, and the Delaunay triangulation method was used to mesh the x/y–MSE model. The Poisson electron hole approach was employed to evaluate the electrical and MF coupling effect for the x/y–MSE. The Bank–Rose approach was employed to compute the distribution of carrier density for the x/y–MSE. Figure 3 presents the simulated Vo for the MS in the x–direction MF, where VB is the bias of the bases; VC is the bias of the collector and VCA is the bias of the additional collectors. In the evaluation, the collector C and additional collectors (CA1, CA2, CA3 and CA4), respectively, connected with a resistance of 1 kΩ. The bases bias was 2.5 V, and the collector bias was 5 V. The bias of the additional collectors was with different voltages of 0.5, 1, 1.5 and 2 V. The x/y–MSE was applied by the x–direction MF, and the x/y–MSE Vo was the voltage difference of the additional collector AC1/AC3. The evaluated results depicted that the MS Vo changed from −116 mV at −200 mT to 116 mV at 200 mT when VB = 2.5 V, VC = 5 V and VCA=2 V. The slope of the curve (at VB = 2.5 V, VC = 5 V and VCA=2 V) was 580 mV/T, so the evaluated sensitivity of the x/y–MSE was 580 mV/T at VB = 2.5 V, VC = 5 V and VCA = 2 V in the x–direction MF.
Figure 3. Simulated output voltage (Vo) in the x−direction MF.

The x/y−MSE was applied by the y−direction MF, and the Vo of the x/y−MSE was the voltage difference of the additional collector AC2/AC4. Figure 4 shows the x/y−MSE Vo in the y−direction MF, where Vb is the bias of the bases; Vc is the bias of the collector and Vca is bias of the additional collectors. The results presented that the MS Vo increased from −116 mV at −200 mT to 116 mV at 200 mT when Vb = 2.5 V, Vc = 5 V and Vca = 2 V. The slope of the curve (at Vb = 2.5 V, Vc = 5 V and Vca =2 V) was 580 mV/T, so the evaluated sensitivity for the x/y−MSE was 580 mV/T at Vb = 2.5 V, Vc = 5 V and Vca = 2 V in the y−direction MF. The evaluated results of the x/y−MSE in the y−direction MF was the same with that in the x−direction MF because the x/y−MSE structure was a symmetric.

Figure 4. Simulated Vo in the y−direction MF.
The performance of the z-MSE was simulated using the Sentaurus TCAD. The z-MSE model was established according to the structure in Figure 2, and the z-MSE Vo was evaluated using the same simulation approach. In the simulation, the bases and collectors connected with a resistance of 1 kΩ, respectively. The bias of the bases was 1.5 V, the bias of the collectors was 5 V. The z-MSE was supplied by the z-direction MF, and the z-MSE Vo was computed by the Sentaurus TCAD. Figure 5 shows the z-MSE Vo in the z-direction MF, where \( V_b \) is bias of the bases and \( V_c \) is bias of the collector. The evaluated results depicted that the z-MSE Vo changed from \(-26\) mV at \(-200\) mT to \(26\) mV at \(200\) mT when \( V_b = 1.5 \) V and \( V_c = 5 \) V. The slope of the curve at \( V_b = 2.5 \) V and \( V_c = 5 \) V was \(130\) mV/T, so the evaluated sensitivity of the z-MSE was \(130\) mV/T at \( V_b = 1.5 \) V and \( V_c = 5 \) V.

**Figure 5.** Simulated Vo in the z-direction MF.

In order to characterize the MS cross-sensitivity, the cross-sensitivity of x/y-MSE and z-MSE were simulated using the Sentaurus TCAD with the same simulation approach. The bias of bases for the x/y-MSE was \(2.5\) V, and the bias of additional collectors for the x/y-MSE was \(2\) V. The bias of collector for the x/y-MSE was \(5\) V. At the same time, the bias of bases for the z-MSE was \(1.5\) V, and the bias of collectors for the z-MSE was \(5\) V. First, the Vo of the x/y-MSE and z-MSE was computed when applying an x-direction MF to the MS. Figure 6 shows the evaluated output for the MS, where \( V_{out}(x,x) \) is the x-axis Vo for x/y-MSE in the x-direction MF; \( V_{out}(y,x) \) is the y-axis Vo for x/y-MSE Vo in the x-direction MF; \( V_{out}(z,x) \) is the z-MSE Vo in the x-direction MF. The results presented that the \( V_{out}(y,x) \) and \( V_{out}(z,x) \) values approximated to zero and the \( V_{out}(x,x) \) had a high response, so the MS cross-sensitivity in the x-direction MF was very small. Then, the Vo of the x/y-MSE and z-MSE was calculated when applying a y-direction MF to the MS. Figure 7 shows the evaluated output for the MS, where \( V_{out}(x,y) \) is the x-axis Vo for x/y-MSE in the y-direction MF; \( V_{out}(y,y) \) is the y-axis Vo for x/y-MSE in the y-direction MF; \( V_{out}(z,y) \) is the z-MSE Vo in the y-direction MF. The results revealed that the \( V_{out}(x,y) \) and \( V_{out}(z,y) \) values approximated zero, and the \( V_{out}(y,y) \) had a high response. The MS cross-sensitivity in the y-direction MF was very small. Finally, the Vo of the x/y-MSE and z-MSE was simulated when applying a z-direction MF to the MS. Figure 8 shows the evaluated output for the MS, where \( V_{out}(x,z) \) is the x-axis Vo for x/y-MSE in the z-direction MF; \( V_{out}(y,z) \) is the y-axis Vo for x/y-MSE in the z-direction MF; \( V_{out}(y,z) \) is the z-MSE Vo in the y-direction MF.
Vo in the z-direction MF. The results depicted that the $V_{out}(y,y)$ had a high response, and the $V_{out}(x,y)$ and $V_{out}(z,y)$ values approximated zero. The MS cross-sensitivity in the z-direction MF was very small.

**Figure 6.** Simulation of output for the MS in the x-direction MF.

**Figure 7.** Simulation of output for the MS in the y-direction MF.
3. Manufacturing of Magnetic Sensor

The three-axis MS contained an x/y–MSE and a z–MSE. The x/y–MSE (Figure 1) was composed of four emitters, four bases, four additional collectors and one collector, where the emitters, collector, additional collectors were n-type silicon with doping phosphorus and the bases were p-type silicon with doping boron. The deep n-well layer was used to restrict the current movement range and to decrease current leakage.

As shown in Figure 2, the z–MSE was composed of four bases, eight collectors and one emitter, where the bases were p-type silicon with doping boron and the collectors and emitters were n-type silicon with doping phosphorus. According to the x/y–MSE structure (Figure 1) and the z–MSE structure (Figure 2), the x/y–MSE and z–MSE layouts were devised. In accordance with the x/y–MSE and z–MSE layout, the three-axis MS was fabricated based on the standard CMOS process from Taiwan Semiconductor Manufacturing Company (TSMC) [31]. Figure 9 shows a picture of the three-axis MS after the CMOS process. The three-axis MS chip (Figure 9) included an z–MSE and an x/y–MSE. Figure 10 demonstrates a picture of the three-axis MS chip during the wire-bonding. The three-axis MS chip was bonded on a printed circuit board by a wire-bonder for measuring its characteristics.
4. Results

The three-axis MS was measured utilizing a digital multimeter, two power supplies, a Gauss meter and an MF generator. Figure 11 demonstrates the setup for measuring the three-axis MS characteristic. The MS chip was placed in the MF generator. The power supply inputted power to the MF generator. The Gauss meter calibrated the MF magnitude that was produced by the MF generator. The MF generator supplied various MFs for the MS measurement. The power supply provided the bias for the MS. The digital multimeter measured the Vo of the three-axis MS.
The x/y–MSE characteristic was tested in the x–direction MF. The MS chip (Figure 11) was placed in the MF generator. An MF in the x–direction that was provided by the MF generator was applied to the x/y–MSE. A bias of 2.5 V was supplied to the bases of x/y–MSE. The additional collectors were without bias. The collector and each additional collector, respectively, was connected with a resistance of 1 kΩ. The different voltages including 1, 2, 3, 4 and 5 V were supplied to the collector of x/y–MSE. The x/y–MSE Vo that was the voltage difference between the addition collectors CA1 and CA3 was recorded by the digital multimeter. The tested results for the x/y–MSE without the additional collectors bias in the x–direction MF are shown in Figure 12, where Vb is the bias of the bases and Vc is the bias of the collector. When Vb = 2.5 V and Vc = 1 V, the x/y–MSE Vo varied from −9.1 mV at −200 mT to 9.2 mV at 200 mT. When Vb = 2.5 V and Vc = 3 V, the x/y–MSE Vo increased from −50.5 mV at −200 mT to 50.6 mV at 200 mT. When Vb = 2.5 V and Vc = 5 V, the x/y–MSE Vo changed from −57.2 mV at −200 mT to 57.1 mV at 200 mT. The curve slope was 286 mV/T at Vb = 2.5 V and Vc = 5 V, so the x/y–MSE sensitivity was 286 mV/T at Vb = 2.5 V and Vc = 5 V in the x–direction MF.
The x/y-MSE had the addition collectors that were used to increase the moving current and enhance the x/y-MSE sensitivity. To characterize the function of additional collectors, the x/y-MSE was measured under different biases of additional collectors. A bias of 2.5 V was provided to the bases of the x/y-MSE, and a bias of 5 V was applied to the collector of the x/y-MSE. The difference voltages that included 0.5, 1, 1.5 and 2 V were applied to the additional collectors of the x/y-MSE. The Vo of the x/y-MSE was recorded by a digital multimeter. Figure 13 displays the measured Vo for the x/y-MSE with the additional collectors bias in the x-direction MF, where $V_B$ is bias of the bases; $V_C$ is bias of the collector and $V_{CA}$ is bias of the additional collectors. When $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 0.5$ V, the x/y-MSE Vo varied from −60.3 mV at −200 mT to 60.2 mV at 200 mT. When $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 1$ V, the x/y-MSE Vo increased from −84.3 mV at −200 mT to 84.4 mV at 200 mT. When $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 2$ V, the x/y-MSE Vo changed from −112 mV at −200 mT to 112 mV at 200 mT. The linear regression method was used to fit the curve at $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 2$ V. The results showed that the regression line had a slope of 534 mV/T and a coefficient of determination $R^2 = 0.9984$, so the sensitivity of the x/y-MSE was 534 mV/T at $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 2$ V in the x-direction MF. The output linearity for the x-MSE was 99%. In a comparison of results in Figures 12 and 13, the Vo of the x/y-MSE with the additional collectors bias exceeds that of the x/y-MSE without the additional collectors bias. In the x-direction MF, the sensitivity of the x/y-MSE (534 mV/T at $V_B = 2.5$ V, $V_C = 5$ V and $V_{CA} = 2$ V) with the addition of collectors bias is higher than that of the x/y-MSE (286 mV/T at $V_B = 2.5$ V and $V_C = 5$ V) without the additional collectors bias.

The x/y-MSE characteristic was tested in the y-direction MF. As shown in Figure 11, the MS chip was placed in the MF generator. An MF in the y-direction that was produced by the MF generator was supplied to the x/y-MSE. The collector and each additional collector connected with a resistance of 1 kΩ, respectively. A bias of 2.5 V was applied to the
The additional collectors were without bias. The difference in voltages that had 1, 2, 3, 4 and 5 V were provided to the collector of the x/y–MSE. The Vo of the x/y–MSE that was the voltage difference between the addition collectors CA2 and CA4 and were recorded by the digital multimeter. The tested results for the x/y–MSE without the additional collectors bias in the y–direction MF is shown in Figure 14, where VB is the bias of the bases and VC is the bias of the collector. When VB = 2.5 V and VC = 1 V, the x/y–MSE Vo varied from −8.9 mV at −200 mT to 9 mV at 200 mT. When VB = 2.5 V and VC = 3 V, the x/y–MSE Vo increased from −50.9 mV at −200 mT to 50.8 mV at 200 mT. When VB = 2.5 V and VC = 5 V, the x/y–MSE Vo changed from −56.9 mV at −200 mT to 56.8 mV at 200 mT. The slope of curve was 283 mV/T at VB = 2.5 V and VC = 5 V, so the sensitivity of the x/y–MSE was 283 mV/T in the y–direction MF.

To understand the function of the additional collectors, the x/y–MSE was tested under the different biases of additional collectors. A bias of 2.5 V was applied to the bases of the x/y–MSE, and a bias of 5 V was supplied to the collector of the x/y–MSE. The different voltages that had 0.5, 1, 1.5 and 2 V were provided to the additional collectors of the x/y–MSE. The Vo of the x/y–MSE was measured by the digital multimeter. Figure 15 displays the measured Vo for the x/y–MSE with the additional collectors bias in the y–direction MF, where VB is the bias of the bases; VC is the bias of the collector and VCA is the bias of the additional collectors. When VB = 2.5 V, VC = 5 V and VCA = 0.5 V, the x/y–MSE Vo varied from −59.5 mV at −200 mT to 59.6 mV at 200 mT. When VB = 2.5 V, VC = 5 V and VCA = 1 V, the x/y–MSE Vo increased from −83.6 mV at −200 mT to 83.5 mV at 200 mT. When VB = 2.5 V, VC = 5 V and VCA = 2 V, the x/y–MSE Vo changed from −111.5 mV at −200 mT to 111.5 mV at 200 mT. The linear regression method was employed to fit the curve at VB = 2.5 V, VC = 5 V and VCA = 2 V. The results depicted that the regression line had a slope of 525 mV/T and a coefficient of determination R² = 0.9982, so the sensitivity of the x/y–MSE was 525 mV/T at VB = 2.5 V, VC = 5 V and VCA = 2 V in the y–direction MF. The output linearity for the y–MSE was 99%. A comparison of results in Figures 14 and 15, the Vo of the x/y–MSE with the additional collectors bias exceeds that of the x/y–MSE without the additional collectors bias. In the y–direction MF, the sensitivity of the x/y–MSE (525 mV/T
at $V_B = 2.5\ V$, $V_C = 5\ V$ and $V_{CA} = 2\ V$) with the addition collector bias exceeds that of the x/y–MSE (283 mV/T at $V_B = 2.5\ V$ and $V_C = 5\ V$) without the additional collectors bias.

The z–MSE characteristic was tested in the z–direction MF. The MF generator produced a z–direction MF that applied to the z–MSE. The power supply provided a bias of 1.2 V to the bases of the z–MSE. The difference voltages that had 1, 2, 3, 4 and 5 V were applied to the collectors of the z–MSE. The bases and collectors were connected with a resistance of 1 kΩ, respectively. The Vo of the z–MSE was detected by the digital multimeter. Figure 16 displays the measured Vo for the z–MSE at $V_B = 1.2\ V$ in the z–direction MF, where $V_B$ is bias of the bases and $V_C$ is bias of the collectors. When $V_B = 1.2\ V$ and $V_C = 1\ V$, the z–MSE Vo increased from $-6.2\ mV$ at $-200\ mT$ to $6.1\ mV$ at $200\ mT$. When $V_B = 1.2\ V$ and $V_C = 3\ V$, the z–MSE Vo changed from $-12.4\ mV$ at $-200\ mT$ to $12.4\ mV$ at $200\ mT$. When $V_B = 1.2\ V$ and $V_C = 5\ V$, the z–MSE Vo varied from $-15.2\ mV$ at $-200\ mT$ to $15.1\ mV$ at $200\ mT$. The slope of curve was $75\ mV/T$ at $V_B = 1.2\ V$ and $V_C = 5\ V$, so the sensitivity of the x/y–MSE was $75\ mV/mT$ at $V_B = 1.2\ V$ and $V_C = 5\ V$ in the z–direction MF. To characterize the influence of the bases voltage for the z–MSE, the bias of the bases was increased to 1.5 V. The difference voltages that included 1, 2, 3, 4 and 5 V were supplied to the collectors of the z–MSE. Figure 17 shows the measured Vo for the z–MSE at $V_B = 1.5\ V$ in the z–direction MF, where $V_B$ is the bias of the bases and $V_C$ is the bias of the collectors. When $V_B = 1.5\ V$ and $V_C = 1\ V$, the z–MSE Vo changed from $-7.5\ mV$ at $-200\ mT$ to $7.6\ mV$ at $200\ mT$. When $V_B = 1.5\ V$ and $V_C = 3\ V$, the z–MSE Vo varied from $-17.3\ mV$ at $-200\ mT$ to $17.3\ mV$ at $200\ mT$. When $V_B = 1.5\ V$ and $V_C = 5\ V$, the z–MSE Vo increased from $-24.2\ mV$ at $-200\ mT$ to $24.1\ mV$ at $200\ mT$. The linear regression method was used to fit the curve at $V_B = 1.5\ V$ and $V_C = 5\ V$. The results showed that the regression line had a slope of $119\ mV/T$ and a coefficient of determination $R^2 = 0.9995$, so the sensitivity of the z–MSE was $119\ mV/T$ at $V_B = 1.5\ V$ and $V_C = 5\ V$ in the z–direction MF. The output linearity for the z–MSE was 99%. A comparison of results in Figures 16 and 17, the Vo of the z–MSE at $V_B = 1.5\ V$ exceeds that of the z–MSE at $V_B = 1.2\ V$. In the z–direction MF, and the sensitivity of the z–MSE increases from $75\ mV/T$ at $V_B = 1.2\ V$ and $V_C = 5\ V$ to $119\ mV/T$ at $V_B = 1.5\ V$ and $V_C = 5\ V$. 

![Figure 15. Measured Vo for the x/y–MSE with bias of the additional collectors in the y–direction MF.](image-url)
Figure 16. Measured Vo for the z−MSE at Vb = 1.2 V in the z−direction MF.

Figure 17. Measured Vo for the z−MSE at Vb = 1.5 V in the z−direction MF.

An excellent three-axis MS must have a low cross-sensitivity. The cross-sensitivity of the MS was investigated. First, an x−direction MF was applied to the MS. A bias of 2.5 V was supplied to the bases of x/y−MSE. A bias of 5 V was provided to the collector of x/y−MSE, and a bias of 2 V was applied to the additional collectors of x/y−MSE. At the same time, a bias of 1.5 was supplied to the bases of z−MSE, and a bias of 5 V was provided to the collectors of z−MSE. The digital multimeter measured the Vo of the x/y−MSE and z−MSE. Figure 18 displays three-axis Vo for the MS in the x−direction magnetic, where Vout(x,x) is the x−axis Vo for the x/y−MSE in the x−direction MF; Vout(y,x) is the y−axis Vo
for the x/y−MSE in the x−direction MF and $V_{out}(z,x)$ is the z−MSE Vo in the x−direction MF. The $V_{out}(y,x)$ and $V_{out}(z,x)$ values in Figure 18 are very small. The slope of the curve $V_{out}(y,x)$ was 25.4 mV/T, and the slope of the curve $V_{out}(z,x)$ was 11.2 mV/T. In the x−direction MF, the MS had a cross-sensitivity of 25.4 mV/T (y−axis output) and a cross-sensitivity of 11.2 mV/T (z−axis output). The MS sensitivity in the x−direction MF was 534 mV/T, so the MS cross-sensitivity in x−direction MF was less than 4.8%.

![Figure 18](image_url)

**Figure 18.** Measured Vo for the MS in the x−direction MF.

A y−direction MF was provided to the MS. A bias of 2.5 V was provided to the bases of x/y−MSE. A bias of 5 V was supplied to the collectors of x/y−MSE, and a bias of 2 V was applied to the additional collectors of x/y−MSE. At the same time, a bias of 1.5 was applied to the bases of z−MSE, and a bias of 5 V was supplied to the collectors of z−MSE. The digital multimeter recorded the Vo of the x/y−MSE and z−MSE. Figure 19 displays three-axis Vo for the MS in the y−direction magnetic, where $V_{out}(x,y)$ is the x−axis Vo for the x/y−MSE in the y−direction MF; $V_{out}(y,y)$ is the y−axis Vo for the x/y−MSE in the y−direction MF and $V_{out}(z,y)$ is the z−MSE Vo in the y−direction MF. The $V_{out}(x,y)$ and $V_{out}(z,y)$ values in Figure 19 are very low. The slope of the curve $V_{out}(x,y)$ was 24.5 mV/T, and the slope of the curve $V_{out}(z,y)$ was 12 mV/T. In the y−direction MF, the MS had a cross-sensitivity of 24.5 mV/T (x−axis output) and a cross-sensitivity of 12 mV/T (z−axis output). The MS sensitivity in the y−direction MF was 525 mV/T, so the MS cross-sensitivity in y−direction MF was less than 4.7%.
A z-direction MF was supplied to the MS. A bias of 2.5 V was applied to the bases of x/y-MSE. A bias of 5 V was provided to the collector of x/y-MSE, and a bias of 2 V was supplied to the additional collectors of x/y-MSE. At the same time, a bias of 1.5 V was supplied to the bases of z-MSE, and a bias of 5 V was applied to the collectors of z-MSE. The digital multimeter detected the Vo of the x/y-MSE and z-MSE. Figure 20 displays three-axis Vo for the MS in the z-direction magnetic, where $V_{out}(x,z)$ is the x-axis Vo for the x/y-MSE in the z-direction MF; $V_{out}(y,z)$ is the y-axis Vo for the x/y-MSE in the z-direction MF and $V_{out}(z,z)$ is the z-MSE Vo in the z-direction MF. The $V_{out}(x,z)$ and $V_{out}(y,z)$ values in Figure 20 are very small. The slope of the curve $V_{out}(x,z)$ was 3.4 mV/T, and the slope of the curve $V_{out}(y,z)$ was 3.2 mV/T. In the z-direction MF, the MS had a cross-sensitivity of 3.4 mV/T (x-axis output) and a cross-sensitivity of 3.2 mV/T (y-axis output). The MS sensitivity in the z-direction MF was 119 mV/T, so the MS cross-sensitivity in z-direction MF was less than 2.9%.
Table 2 summarizes the performances of the magnetic sensor. The area of the x/y–MSE is $80 \times 80 \ \mu m^2$, and the area of z–MSE is $120 \times 120 \ \mu m^2$. The measurement range of magnetic field for the x–MSE, y–MSE and z–MSE is $\pm 200 \ mT$. The sensitivity of the x–MSE is $534 \ mV/T$, and the sensitivity of the y–MSE is $525 \ mV/T$. The sensitivity for the z–MSE is $119 \ mV/T$. The cross-sensitivity of the x–MSE is less than 4.8%. The cross-sensitivity for the y–MSE is less than 4.7%, and the cross-sensitivity for the z–MSE is less than 2.9%. The output linearity for the x–MSE (coefficient of determination $R^2 = 0.9984$), y–MSE ($R^2 = 0.9982$) and z–MSE ($R^2 = 0.9995$) is 99%. The power consumption of the x/y–MSE is 6 mW, and the power consumption of the z–MSE is 4 mW.

Table 2. Summary of the performances for the magnetic sensor.

| Characteristic                  | x/y–MSE | y–MSE | z–MSE |
|--------------------------------|---------|-------|-------|
| $V_B$ voltage                  | 2.5 V   | 2.5 V | 1.5 V |
| $V_C$ voltage                  | 5 V     | 5 V   | 5 V   |
| $V_{CA}$ voltage               | 2 V     | 2 V   |      |
| Area                           | $80 \times 80 \ \mu m^2$ combination with x–MSE | $120 \times 120 \ \mu m^2$ |
| Measurement range              | $\pm 200 \ mT$ | $\pm 200 \ mT$ | $\pm 200 \ mT$ |
| Sensitivity                    | $534 \ mV/T$ | $525 \ mV/T$ | $119 \ mV/T$ |
| Cross-sensitivity              | $<4.8\%$ | $<4.7\%$ | $<2.9\%$ |
| Output linearity               | 99%     | 99%   | 99%   |
| Power consumption              | 6 mW    | 6 mW  | 4 mW  |

These micro magnetic sensors, proposed by Niekiel [9], Okada [11], Nejad [13], Bahreyni [14], were magnetic-resonant types that had a high sensitivity. The magnetic-resonant magnetic sensors required suspension structures and a high-actuated voltage to produce actuation and sensing, so the sensors had the disadvantages of a complicated fabrication, high-actuated voltage, high power consumption, and easy interference by environmental vibration. In this work, the micro magnetic sensor was without a suspension structure and was fabricated using the commercial CMOS process, so the sensor had the...
advantages of a low power consumption, easy fabrication and easy mass-production. Tseng [15] developed a one-axis micro-magnetic sensor using the CMOS process, and the sensor was a magnetic transistor type. The sensitivity of the magnetic sensor was 354 mV/T. Sileo [16] fabricated a Hall magnetic sensor, and its sensitivity was 0.03 V/T. Zhao [17] proposed a three-axis micro magnetic sensor manufactured using the MEMS technology. The magnetic sensor was composed of four magnetic transistors and a Hall element. The x-axis, y-axis and z-axis sensitivity for the magnetic sensor were 77.5 mV/T, 78.6 mV/T and 77.4 mV/T, respectively. In this work, the magnetic sensor was a magnetic transistor type, and the sensitivity of the sensor exceeded that of Tseng [15], Sileo [16] and Zhao [17].

5. Conclusions

The design and manufacturing of a three-axis MS base on the standard 180 nm CMOS technology were implemented. The MS was composed of an x/y–MSE and a z–MSE, where the z–MSE detected the MF in the z–direction and the x/y–MSE sensed the MF in the x– and y–direction. The x/y–MSE, which consisted of two magnetotransistors, designed four additional collectors that enhanced the mobility of carriers in the p-substrate and increased the MS sensitivity. The Sentaurus TCAD simulated the MS characteristic. The simulated results revealed that the MS sensitivity was 580 mV/T in the x–direction MF, 580 mV/T in the y–direction MF and 135 mV/T in the z–direction MF. The MS was an easy fabrication because it was without any post-CMOS process. The measured results depicted that the MS sensitivity was 534 mV/T in the x–direction MF, 525 mV/T in the y–direction MF and 119 mV/T in the z–direction MF. The measured results of the MS was in agreement with the measured results of the MS. Experiments showed that the MS cross-sensitivity in the x–direction was less than 4.8%, and its cross-sensitivity in the y–direction was less than 4.7%. The MS cross-sensitivity in the z–direction MF was less than 2.9%. Thereby, the MS had a low cross-sensitivity and an excellent sensitivity.

Author Contributions: Conceptualization, P.-J.S. and C.-L.D.; Data curation, P.-J.S., Y.-C.T. and C.-L.D.; Formal analysis, C.-H.W.; Funding acquisition, P.-J.S. and C.-L.D.; Investigation, C.-H.W.; Methodology, C.-H.W. and C.-L.D.; Project administration, C.-L.D.; Resources, P.-J.S. and Y.-C.T.; Software, C.-H.W.; Validation, Y.-C.T. and C.-L.D.; Writing—original draft, C.-H.W.; Writing—review and editing, C.-L.D. All authors have read and agreed to the published version of the manuscript.

Funding: This research was funded by the Ministry of Science and Technology of the Republic of China, grant number MOST 110-2221-E-005-047.

Institutional Review Board Statement: Not applicable.

Informed Consent Statement: Not applicable.

Data Availability Statement: Data sharing not applicable.

Acknowledgments: The authors would like to thank National Center for High-performance Computing (NCHC) for simulation; Taiwan Semiconductor Research Institute (TSRI) for fabrication and the Ministry of Science and Technology (MOST) of the Republic of China for financially supporting this research under Contract Nos. MOST 110-2221-E-005-047.

Conflicts of Interest: The authors declare no conflict of interest.

References

1. Lu, J.; Zhang, L.; Maeda, R. Real-time tracking of organ-shape and vessel-locations for surgical navigation using MEMS tri-axis magnetic sensors. Med. Eng. Phys. 2021, 93, 42–48.
2. Lee, C.Y.; Lin, Y.Y.; Kuo, C.K.; Fu, L.M. Design and application of MEMS-based Hall sensor array for magnetic field mapping. Micromachines 2021, 12, 299.
3. Oh, Y.; Jung, Y.J.; Choi, S.H.; Kim, D.W. Design and evaluation of a MEMS magnetic field sensor-based respiratory monitoring and training system for radiotherapy. Sensors 2018, 18, 2742.
4. Lara-Castro, M.; Herrera-May, A.L.; Juarez-Aguirre, R.; Lopez-Huerta, F.; Ceron-Alvarez, C.A.; Cortes-Mestizo, I.E.; Morales-Gonzalez, E.A.; Vazquez-Leal, H.; Dominguez-Nicolás, S.M. Portable signal conditioning system of a MEMS magnetic field sensor for industrial applications. *Microsystem Technol.* 2017, 23, 215–223.

5. Krishnapriya, S.; Komaragiri, R.; Suja, K.J. Fabrication, characterization, and modelling of a novel via-less single metal level magnetic microcoil sensor for biosensing applications. *Sens. Actuators A* 2019, 290, 190–197.

6. Feng, Z.; Zhi, S.T.; Guo, L.; Wei, M.C.; Zhou, Y.; Lei, C. A novel integrated microfluidic platform based on micro-magnetic sensor for magnetic bead manipulation and detection. *Microfluid. Nanoﬂuid.* 2018, 22, 86.

7. Zhang, H.; Liao, L.; Zhao, R.Q.; Zhou, J.T.; Yang, M.; Xia, R.C. The non-destructive test of steel corrosion in reinforced concrete bridges using a micro-magnetsensor. *Sensors* 2016, 16, 1439.

8. Vetrella, A.R.; Fasano, G.; Accardo, D.; Moccia, A. Differential GNSS and vision-based tracking to improve navigation performance in cooperative multi-UAV systems. *Sensors* 2016, 16, 2164.

9. Niekiel, F.; Su, J.X.; Bodduluri, M.T.; Liseck, T.; Blohm, L.; Pieper, I.; Wagner, B.; Lofink, F. Highly sensitive MEMS magnetic field sensors with integrated powder-based permanent magnets. *Sens. Actuators A* 2019, 297, 111560.

10. Chen, J.W.; Li, J.H.; Li, Y.Y.; Chen, Y.L.; Xu, L.X. Design and fabrication of a miniaturized GMI magnetic sensor based on amorphous wire by MEMS technology. *Sensors* 2018, 18, 732.

11. Okada, N.; Sasabuchi, T.; Koike, K.; Mineta, T. MEMS magnetic sensor with bridge-type resonator and magnetostriective thin film. *Electron. Commun. Jpn.* 2018, 101, 90–95.

12. Guo, L.; Wang, C.; Zhi, S.T.; Feng, Z.; Lei, C.; Zhou, Y. Wide linearity range and highly sensitive MEMS-based micro-fluxgate sensor with double-layer magnetic core made of Fe-Co-B amorphous alloy. *Micromachines* 2018, 8, 352.

13. Nejad, S.N.; Mansour, R. Development of MEMS reed magnetic sensors. *IEEE Trans. Magn.* 2016, 52, 4000407.

14. Bahreyni, B.H.; Shafai, C. A resonant micromachined magnetic field sensor. *IEEE Sens. J.* 2007, 7, 1326–1334.

15. Tseng, J.Z.; Wu, C.C.; Dai, C.L. Modeling and manufacturing of a micromachined magnetic sensor using the CMOS process without any post-process. *Sensors* 2014, 14, 6722–6733.

16. Sileo, L.; Todaro, M.T.; De Vittorio, M.; Passaseo, A. Fully integrated three-axis Hall magnetic sensor based on micromachined structures. *Microelectron. Eng.* 2010, 87, 1217–1219.

17. Zhao, X.F.; Bai, Y.J.; Deng, Q.; Ai, C.P.; Yang, X.H.; Wen, D.Z. Research of the monolithic integrated 3-D magnetic field sensor based on CMOS technology. *Sensors J.* 2017, 17, 5849–5856.

18. Yeh, P.C.; Duan, H.; Chung, T.K. A novel three-axial magnetic-piezoelectric MEMS AC magnetic field sensor. *Micromachines* 2019, 10, 710.

19. Yang, M.Z.; Wu, C.C.; Dai, C.L.; Tsai, W.J. Energy harvesting thermoelectric generators manufactured using the complementary metal oxide semiconductor process. *Sensors* 2013, 13, 2359–2367.

20. Dai, C.L.; Chiou, J.H.; Lu, M.S.C. A maskless post-CMOS bulk micromaching process and its application. *J. Micromech. Microeng.* 2005, 15, 2366–2371.

21. Cheng, Y.C.; Dai, C.L.; Lee, C.Y.; Chen, P.H.; Chang, P.Z. A circular micromirror array fabricated by a maskless post-CMOS process. *Microsystem Technol.* 2005, 11, 444–451.

22. Dai, C.L.; Chen, H.L.; Chang, P.Z. Fabrication of a micromachined optical modulator using the CMOS process. *J. Micromech. Microeng.* 2001, 11, 612–613.

23. Muñoz-Gamarra, J.L.; Uranga, A.; Barniol, N. CMOS-NEMS copper switches monolithically integrated using a 65 nm CMOS technology. *Micromachines* 2016, 7, 30.

24. Dai, C.L.; Peng, H.J.; Liu, M.C.; Wu, C.C.; Hsu, H.M.; Yang, L.J. A micromachined microwave switch fabricated by the complementary metal oxide semiconductor post-process of etching silicon dioxide. *Ipn. J. Appl. Phys.* 2005, 44, 6804–6809.

25. Huang, H.; Wang, D.; Xu, Y. A monolithic CMOS magnetic Hall sensor with high sensitivity and linearity characteristics. *Sensors* 2015, 15, 27359–27373.

26. Hsieh, C.H.; Dai, C.L.; Yang, M.Z. Fabrication and characterization of CMOS-MEMS magnetic microsensors. *Sensors* 2013, 13, 14728–14739.

27. Dai, C.L.; Kao, P.H.; Tai, Y.W.; Wu, C.C. Micro FET pressure sensor manufactured using CMOS-MEMS technique. *Microelectron. J.* 2008, 39, 744–749.

28. Yang, M.Z.; Dai, C.L. Ethanol microsensors with a readout circuit manufactured using the CMOS-MEMS technique. *Sensors* 2015, 15, 1623–1634.

29. Shen, W.C.; Shih, P.J.; Tsai, Y.C.; Hsu, C.C.; Dai, C.L. Low-concentration ammonia gas sensors manufactured using the CMOS-MEMS technique. *Micromachines* 2020, 11, 92.

30. Dai, C.L.; Chen, Y.C.; Wu, C.C.; Kuo, C.F. Cobalt oxide nanosheet and CNT micro carbon monoxide sensor integrated with readout circuit on chip. *Sensors* 2010, 10, 1753–1764.

31. Chen, W.R.; Tsai, Y.C.; Shih, P.J.; Hsu, C.C.; Dai, C.L. Magnetic micro sensors with two magnetic field effect transistors fabricated using the commercial complementary metal oxide semiconductor process. *Sensors* 2020, 20, 4731.