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Multi-transmission-line-beam interactive system

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Abstract

We construct here a Lagrangian field formulation for a system consisting of an electron beam interacting with a slow-wave structure modeled by a possibly non-uniform multiple transmission line (MTL). In the case of a single line we recover the linear model of a traveling wave tube (TWT) due to J.R. Pierce. Since a properly chosen MTL can approximate a real waveguide structure with any desired accuracy, the proposed model can be used in particular for design optimization. Furthermore, the Lagrangian formulation provides for: (i) a clear identification of the mathematical source of amplification, (ii) exact expressions for the conserved energy and its flux distributions obtained from the Noether theorem. In the case of uniform MTLs we carry out an exhaustive analysis of eigenmodes and find sharp conditions on the parameters of the system to provide for amplifying regimes.

1 Introduction

We study here theoretical aspects of generation and amplification of microwave (millimeter waves) radiation by traveling wave tubes. Generally speaking, generation and amplification of electromagnetic radiation can be produced by enormous variety of devices of different designs depending on the frequency of the radiation and its power. For light such devices are lasers; we remind that the term laser is an acronym for Light Amplification by Stimulated Emission of Radiation. For microwaves, which are of our special interest here, there is a large class of amplifying devices including maser, a predecessor of the laser, magnetrons, klystrons, traveling wave tubes, crossed-field amplifiers and gyrotrons.

In the case of lasers, as suggested by its very name, the general principle of the amplification is based on the stimulated emission resulting from certain atomic transitions. "Lasers come in a great variety of forms, using many different laser materials, many different atomic systems, and many different kinds of pumping or excitation techniques. The beams of radiation that lasers emit or amplify have remarkable properties of directionality, spectral purity, and intensity.”, [Sieg, p. 2]. An important and defining property of laser radiation is its coherency, that is its monochromaticity. For amplification the coherency means that in a narrow frequency band the output signal, after being amplified, reproduces pretty accurately
the shape of the input signal but with a substantial increase in amplitude. Coherent amplification combined with a feedback allows to produce highly directional and highly monochromatic beams. Observe that atomic transitions of the laser medium constitute a fundamental basis of amplification, that is the amplification mechanism is fixed by the nature, so to speak. There is extended literature on the theory of lasers, see for instance [Fox, 4.1], [Loud], [Sieg]. Its basic phenomenological elements include: (i) Einstein’s treatment of the spontaneous and stimulated emission, [Fox, 4.1], and (ii) operation principle based on interaction between the laser (gain) medium and electromagnetic modes of a cavity containing this medium. More detailed and fundamental theory that can justify the laser phenomenology involves quantum optics (electronics), [Loud], [Fox].

In the case of microwaves the radiation is produced by microwave vacuum electronic devices, known formerly as microwave tubes. These devices use free electrons in a vacuum to convert energy from a DC power source to an RF (radio frequency) signal. In other words, as a result of interaction between the electron beam and properly designed structure the kinetic energy of the electrons is converted into electromagnetic energy stored in the field, [Gilm1], [Nus, 2.2], [SchaB, 4], [Tsim]. The key operational principle of any microwave device is a positive feedback interaction between coherent radiation by electrons radiating in phase on one hand and on the other hand electron bunching caused by radiation on the stream of electrons. The electron bunching associated with acceleration and deceleration of groups of electrons along the beam constitutes the physical mechanism of radiation generation and its amplification.

An important class of microwave devices uses as its operation principle the Cherenkov radiation generated by charged particles propagating in or near a medium supporting slow waves with phase velocity comparable with the particle velocity. Traveling wave tubes, the main subject of our studies here, belongs to this class.

Traveling wave tubes (TWT) are used widely in many areas including satellite communication and radar systems. Typical TWT consists of an elongated vacuum tube containing an electron beam which passes down the middle of an RF circuit (a slow-wave structure). The operation principle of a TWT is as follows. At one end of the TWT structure, the RF circuit is fed with a low-powered radio signal to be amplified. As the RF signal travels along the tube at near the same speed as the electron beam, the electromagnetic field acts upon the beam and causes electron bunching with consequent formation of the so-called space-charge wave. The electromagnetic field associated with the space-charge wave induces more current back into the RF circuit, thus enhancing the bunching, and so on. The EM field thus builds up and is amplified as it passes down the structure until a saturation regime is reached and a large RF signal is collected at the output. The role of the slow-wave structure is to slow down the electromagnetic wave to match up with the velocity of the electrons in the beam, usually a small fraction of the speed of light. Such a synchronism is required for effective in phase interaction between the structure and the beam with optimal extraction of the kinetic energy of the electrons. A typical slow-wave structure is the helix, which reduces the speed of propagation according to its pitch. Further details on the design and operation of TWT can be found in [Gilm1], [PierTWT], [Tsim], [Nus, 4].

An effective mathematical model for a TWT interacting with an electron beam was introduced by J. R. Pierce, [PierTWT], [Pier51]. This model is the simplest one that accounts for wave amplification along the structure, energy extraction from the electron beam and its conversion into microwave radiation in the TWT, see also [SchaB, 4], [Gilm1], [Gilm], [Tsim] and [Nus, 4]. In Section 3 we provide for precise description of the model as presented in [Pier51].
The mentioned presentation is a time domain model, in contrast to other presentations dealing with the frequency domain counterpart. Though simple, the Pierce model allows for adequate estimates of the gain and it was used effectively in designing working TWTs in the fifties. This model captures remarkably well significant features of wave amplification and the beam-wave energy transfer, and is still in use for basic design estimates.

The model presented by Pierce is one-dimensional and consists of (i) an ideal linear representation of the electron beam and (ii) a lossless transmission line (TL) representing the waveguide structure. The transmission line is assumed to be homogeneous, that is, with uniformly distributed capacitance and inductance. To overcome the Pierce theory limitations far more sophisticated nonlinear theories have been developed to model very involved physics of the electron beam and slow-wave structures, [SchaB], [Gilm], [Tsim]. Needless to say that those theories are far more complex and often require a massive computer work.

In this paper we advance the Pierce theory to a theory that, while keeping its simplicity and constructiveness, allows for more complex slow-wave structures. We start by developing a Lagrangian field framework for the original Pierce model. Such framework allows for extension of the model in two directions: a) we can replace the transmission line by a multi-transmission line (MTL) and b) we can dispense with the homogeneity assumption, thus considering general nonhomogeneous systems consisting of a multi-transmission line (MTL) coupled to an electron beam. We refer to such a system as a MTLB system. Extension to multiple transmission lines is motivated by the fact that general MTLs can approximate with desired accuracy real waveguided structures which can be homogeneous (uniform) as well as inhomogeneous (nonuniform), [Nit], [Paul], [SchwiE].

One of the advantages of the Lagrangian formulation is that conservation laws and explicit expressions for the conserved quantities and their fluxes can be obtained at once from the Noether theorem. We would like to point out that though conservation laws do follow from the Euler-Lagrange evolution equations there is no systematic way to extract them from those equations. In addition to that, since all the information about dynamics is encoded in the scalar Lagrange function we can trace the amplification mechanisms and the properties of the energy transfer from the electron beam to the microwave radiation to certain terms in the Lagrangian density.

For homogeneous MTLB systems, we study the amplification phenomenon by considering the exponentially growing eigenmodes and associated complex-valued wave numbers for the field equations, just as in the original Pierce theory, [PierTWT], [PierW], [Pier51]. We provide also a rigorous proof of the fact that, on the growing mode, the energy always flows in the expected direction, i.e. from the beam to the MTL. In this case, the eigenmodes analysis can be carried out analytically, providing for explicit expressions for their energy density and energy flux distributions as well as sufficient conditions for the existence of amplification regimes (growing modes). The analysis includes derivation of a special canonical form of the dispersion relation having a remarkable feature: one of its two terms depends only on the MTL, whereas another one depends only on the beam parameters. Such a special factorization and separation of variables simplifies the analysis significantly.

As to inhomogeneous MTLB systems, they are by far more involved compared to homogeneous ones. In particular, for periodic MTLB systems the dispersion relations are not polynomial and that requires to turn to the most general form of the Floquet theory, [YakSta1, II, III]. For this general case we provide the first step towards a systematic study, namely we transform the Euler-Lagrange field equations into the canonical Hamiltonian form using basics of the de Donder-Weyl theory, [Rund, 4.2]. This particular Hamiltonian form consists
of a system of equations which is of first order in the spatial variable, thus providing the basis for the most effective use of the Floquet theory, [YakSta1, II, III] in the study of periodic structures. Detailed development of the Floquet theory for periodic MTLB system requires to overcome a number of technical difficulties and it is left for future studies.

One of the features of the proposed here phenomenological approach is that it captures the electron bunching as a physical mechanism of amplification in some form. Consequently, our analysis is a valuable source of a solid information on the electron bunching.

The structure of the paper is as follows: In Section 2 we briefly summarize our main results. Section 3 is devoted to the description of Pierce’s model for beam-TL interaction as presented in [Pier51, I]. Section 4 deals with the Lagrangian approach to the model, including generalizations to both non-homogeneous and multiple transmission lines. In the following Section 5 we explore the amplification mechanism in the MTLB system as linked to instabilities in the dynamics of the beam. The appropriate mathematical setting, in particular the Hamiltonian structure of the model aimed at the study of eigenmodes in the periodic case is the subject of Section 6. In Section 7 we focus on the detailed study of growing modes for the homogeneous MTLB system. Section 8 deals with the questions of general energy conservation and energy transfer between the beam and the MTL on the growing mode. In Section 9 we make apparent how our general approach allows to easily recover some of the original Pierce’s results.

Finally, in Section 10 we collect some technically involved subjects which have been deferred there to avoid distracting the reader from the main flow of ideas.

2 Main results

One of the goals of this work is to identify the mathematical mechanism of amplification in MLTB systems. This goal has been accomplished by the construction of a Lagrangian field theory of MLTB systems that underlines their physical properties. Leaving detailed developments of this theory to the following sections we simply identify here the key term of the system Lagrangian responsible for amplification. This term quite expectedly is associated with the electron beam and is described by the following expression

$$\mathcal{L}_b = \frac{\xi}{2} \left( \partial_t q + u_0 \partial_z q \right)^2, \quad \xi = \frac{4\pi}{\omega_p^2 \sigma} > 0,$$  \hspace{1cm} (2.1)

where \( t \) and \( z \) are, respectively, time and longitudinal variable, \( q = q(t,z) \) is the charge ("smoothed-out jelly of charge", [Pier51, I]) flowing through the beam. \( \sigma \) and \( u_0 \) stand, respectively, for the cross section and the electron velocity and \( \omega_p \) is the plasma frequency. According to the general theory, we can identify the kinetic and potential energies of the beam by expanding the expression (2.1), that is

$$\mathcal{L}_b = \frac{\xi}{2} \left( \partial_t q \right)^2 + \xi u_0 \partial_t q \partial_z q + \frac{\xi}{2} u_0^2 \left( \partial_z q \right)^2,$$

where the potential energy of the beam \( -\frac{\xi}{2} (u_0 \partial_z q)^2 \) is a negative quantity. This is a marked feature distinguishing MLTB from common oscillatory systems, in which the potential energy is always positive. The negative sign of this potential energy term is ultimately responsible for system instability and consequent amplification.
Indeed, a typical oscillatory system has a positive potential energy manifested in forces that move the system toward its equilibrium state. The simplest examples are given by a linear mass-spring system or its electric analog – a simple electric LC oscillatory circuit. The corresponding Lagrangians are

\[ L_1(x, x') = \frac{1}{2} m x'^2 - \frac{1}{2} k x^2; \quad L_2(q, q') = \frac{1}{2} L q'^2 - \frac{1}{2} C q^2, \]

where \( m \) is the mass of the point, \( k \) is the elastic Hooke constant of the spring and \( q \) is the charge in the capacitor. Such forces result in a stable motion with oscillatory energy transfer between its kinetic and potential forms. A qualitatively different picture occurs when the potential energy is negative, as in \( L_b \). In this case resulting forces move the system away from the equilibrium at an exponentially growing rate. Such situation corresponds to having a negative Hooke constant \( k \) in \( L_1 \) or a negative capacitance in \( L_2 \) above. Interestingly, Pierce has observed an effective negative capacitance in his studies of a transmission line interacting with the electron beam, [Pier51].

Another marked feature of the term \( L_b \) in (2.1) is its degeneration as quadratic form manifested as a perfect square trinomial expression or, alternatively, as a precise gyrotrropic term. According to the general theory of unstable regimes, [YakSta1], this kind of degeneration is a necessary condition for instability arising under proper perturbations. From the point of view of the second order partial differential equation describing the beam dynamics this degeneracy is manifested as parabolicity compared to hyperbolicity occurring for common wave motion.

The power and efficiency of the Lagrangian approach is further demonstrated by an exhaustive analysis of amplification regimes for a general homogeneous MTLB system, including precise conditions under which amplification takes place. In particular, if \( 0 \leq v_1 \leq v_2 \leq \ldots \leq v_n \) denote the characteristic velocities of the MTL as an independent system, we show that there is always an amplifying regime if \( u_0 \leq v_1 \). If \( u_0 > v_1 \), we show that amplification occurs only for sufficiently small \( \xi \) in (2.1). We also provide a transparent form of the dispersion relation for a general homogeneous MTLB system, including possible degenerations, as well as an asymptotic analysis of the amplification factor as the beam parameter \( \xi \) defined in (2.1) becomes arbitrarily small or large. The limits \( \xi \to 0 \) and \( \xi \to \infty \) correspond to high, respectively small electron density of the beam. In [Pier51], Pierce deals with large values of \( \xi \), which allows him to simplify the dispersion relation to an exactly solvable third degree equation for the forward eigenmodes. We review Pierce’s result in the light of our approach.

Yet another benefit of our Lagrangian approach is an exhaustive analysis of the energetic issues, including the overall energy conservation and energy transfer between the MTL and the beam. This analysis yields explicit expressions for the power \( P_{B\to MTL} \) flowing from the beam to the MTL for an exponentially growing solution of the form

\[ Q(z, t) = \hat{Q} e^{-i(\omega t - k_0 z)}, \quad q(z, t) = \hat{q} e^{-i(\omega t - k_0 z)}, \quad \text{Im } k_0 < 0, \]

where \( Q \) is the coordinate describing the MTL and \( q \) is the one describing the beam. Namely, the following formula holds

\[ \left\langle P_{B\to MTL} \right\rangle (z) = -\left[ \omega \xi |k_0|^2 |\hat{q}|^2 (\text{Re } v_0 - u_0) \text{Im } v_0 \right] e^{-2(\text{Im } k_0) z}, \quad v_0 = \frac{\omega}{k_0}. \]  

We show that in the above formula the constant in front of the exponential is indeed \textit{positive}, meaning that the energy flows from the beam to the MTL. Formula (2.3) indicates also that
the power transferred to the MTL increases exponentially in the direction of the electron flow. The opposite is true of the evanescent wave when the power flows to the beam and decreases exponentially in the +z direction.

2.1 Negative potential energy and general gain media

Looking at the above analysis we can identify two main features of the Lagrangian providing for the amplification in the MTLB system. The first one is the fact that the beam potential energy is negative and unbounded from below. This feature of the electron beam Lagrangian clearly indicates that the model is an ideal one with the negative potential energy term representing effectively an inexhaustible source of energy. This energy can be converted into another form of energy such as energy of electromagnetic radiation. Such ideal model can be suitable for describing the amplification and gain up to the point of saturation. The saturation can conceivably be modeled phenomenologically by introducing an additional positive potential energy term into the beam Lagrangian represented by a higher order polynomial with a small coefficient. That would make the theory nonlinear, of course.

The second feature of the Lagrangian providing for amplification is a particular degeneracy of the expression (2.1) for the Lagrangian and its role in the system stability. More precisely, such term makes the system unstable under proper perturbations, as discussed in detail in Section 5.1. It is a well known fact from the Floquet theory of periodic Hamiltonian systems that such degeneration is indeed necessary in order to have unstable perturbations, [YakSta1].

The association of the amplification and gain with a negative potential energy term in a system Lagrangian can be a general way to model gain media. Interestingly, the phenomenon of negative energy waves in inhomogeneous plasmas is well known and understood at phenomenological level, see for instance, [Bellan 7.7], [Hase 1.3], [Meln 3.1]. The explanation provided in the cited references is essentially that in the approximate phenomenological model the wave-energy density corresponds to the change in the total system energy density in a more detailed theory. Such negative energy waves typically occur when the system is near equilibrium with a steady-state flow velocity and there exists a mode that reduces the average kinetic energy of the particles to a value below the initial equilibrium value. Importantly, concepts of negative energy waves and gain media are intimately related to the instability.

It is instructive to compare and contrast the developed here approach for modeling the gain medium by a negative potential energy term with the conventional approach that represent the gain medium as a system with negative absorption. As an important and relevant example of later let us consider collisionless plasma in a weak external electric field $E = E_0 e^{-i(\omega t - kx)}$ described in [LiP, 3]. The interactions in such a plasma are non-local and consequently the plasma permittivity depends on the both on $\omega$ and $k$ (the so-called spatial dispersion), and it has non-zero imaginary part resulting in dissipation. The imaginary part of permittivity and the energy dissipation are given respectively by formulas

$$
\epsilon'' = -\frac{4\pi^2 e^2 m}{k^2} \left[ \frac{\partial f}{\partial p} \right]_{v=\omega/k}; \quad Q = \frac{\omega}{8\pi} \epsilon'' |E|^2 = -|E|^2 \frac{2\pi e^2 \omega}{2k^2} \left[ \frac{\partial f}{\partial p} \right]_{v=\omega/k},
$$

where $m, e$ are the electron mass (respectively charge) and $f$ is the momentum distribution function of the stationary plasma. If the plasma is isotropic (that is, the distribution function of momenta only depends on $|p|$ or, in the one-dimensional case, is an even function), it can be shown that $Q > 0$, ([LiP 3.30]), and consequently the plasma absorbs energy from the field, a phenomenon called Landau damping. However, in the presence of anisotropy, the sign
of $\left[ \frac{\partial f}{\partial p} \right]_{k=\omega/k}$ and hence that of $Q$ might be reversed yielding a net flow of energy from the electrons to the field and providing an example of gain medium. It is intuitively clear from (2.3) that the net energy flux depends on the relative number of electrons with the velocity larger/smaller than the phase velocity of the wave.

Main differences between our approach for modeling gain in the MTLB system and the conventional approach for modeling gain in the plasma example described above are as follows. The conventional approach is based fundamentally on the concept of open system and the gain medium is not modeled explicitly but rather by its effect on the system. In our approach the beam interacting with the electric field form a conservative system and the gain medium is modeled explicitly as the beam term in the system Lagrangian with a negative potential component. Yet another difference is that, in the MTLB system, the gain occurs for the space charge wave velocities larger or smaller than the wave phase velocity.

In fact, a causal dissipative system can always be extended uniquely to a properly constructed conservative system, [FigSch1], [FigShi1], [FigSch2]. It is an interesting question then whether one can carry out similar construction for the gain medium. Answering this question is not in the scope of this paper but we intend to look at this subject in our future work.

3 Pierce’s model

In [Pier51], J.R. Pierce presented a linear, one-dimensional model for the description of the interaction of an electron beam with a surrounding waveguide. The model is based on the following assumptions.

Assumption I. The modulation of both the electron velocity and the current on the beam (so called a.c. components) are small compared to the average or unperturbed velocity and current.

This assumption justifies the linearization of the equations around the unperturbed regime. Let the total velocity of the electrons be $u_0 + v$, where $u_0$ is the average velocity and $v$ is a small perturbation. Analogously, let $\rho_0 + \rho$ be the total electron density (per unit volume) where $\rho_0$ is the unperturbed density and $\rho$ is the perturbation. Let $\sigma$ be the cross section of the beam. Then, the total current flowing is $I_T = I_0 + I_b$, where $I_0 = \sigma \rho_0 u_0$ is the d.c. current and the perturbation is given by

$$I_b = \sigma (\rho u_0 + v \rho_0 + \rho v). \quad (3.1)$$

Linearization around the d.c. regime gets rid of the term $\rho v$, which is quadratic in the perturbations. Thus we take

$$I_b = \sigma (\rho u_0 + v \rho_0) \quad (3.2)$$

in what follows. The linearized conservation of charge equation reads

$$\frac{\partial \rho}{\partial t} + \frac{\partial i}{\partial z} = \frac{\partial \rho}{\partial t} + \frac{1}{\sigma} \frac{\partial I_b}{\partial z} = 0, \quad (3.3)$$

where $t$ represents time, $z$ is the longitudinal variable and $i$ is the current density, $i = I_b/\sigma$.

Assumption II. The beam is thought of as a continuous medium (electron jelly) with no internal stress and a unique volumetric force acting along it, namely the one resulting from the axial component of the electric field associated to the signal on the waveguide.
It is further assumed that the charge/mass ratio in the electron jelly is precisely $e/m$, $e = -|e|$ being the electron charge and $m$ being the electron mass. Therefore, if $E = E_z$ is the axial component of the field, the motion equation for the medium reads:

$$\frac{\partial v}{\partial t} + (u_0 + v)\frac{\partial v}{\partial z} = \frac{e}{m} E, \quad (3.4)$$

where, on the left-hand side, we have used the usual Eulerian expression for the acceleration in terms of the velocity field $v(z,t)$. Upon linearization, the term $v\frac{\partial v}{\partial z}$ is dropped, thus yielding

$$\frac{\partial v}{\partial t} + u_0 \frac{\partial v}{\partial z} = \frac{e}{m} E. \quad (3.5)$$

Notice that in Pierce’s original paper, [Pier51, I], the charge of the electron is denoted by $-e$, whereas here it is just $e$.

Actually, the full blown Pierce model, as presented in the book [PierTWT], also includes the effect of electron-electron repulsion in the beam (so called space charge effects); see also [Gilm], [Tsim]. Here we do not include such effect for the sake of simplicity, but we advance that this can be done and we plan to report on this issue in the future.

Taking the derivatives of (3.2) with respect to $t$ and $z$ we obtain the following expressions for $\partial v/\partial t$ and $\partial v/\partial z$:

$$\frac{\partial v}{\partial t} = \frac{1}{\sigma \rho_0} \frac{\partial I_b}{\partial t} - \frac{u_0}{\rho_0} \frac{\partial \rho}{\partial t}, \quad \frac{\partial v}{\partial z} = \frac{1}{\sigma \rho_0} \frac{\partial I_b}{\partial z} - \frac{u_0}{\rho_0} \frac{\partial \rho}{\partial z}. \quad (3.6)$$

We use (3.3) to express $\partial \rho/\partial t$ in terms of $\partial I_b/\partial z$ in the first of the above relations and differentiate the resulting relation with respect to $t$ thus yielding

$$\frac{\partial^2 v}{\partial t^2} = \frac{1}{\sigma \rho_0} \frac{\partial^2 I_b}{\partial t^2} + \frac{u_0}{\sigma \rho_0} \frac{\partial^2 I_b}{\partial z \partial t}. \quad (3.7)$$

Next, we differentiate the second relation in (3.6) with respect to $t$, expressing again $\partial \rho/\partial t$ in terms of $\partial I_b/\partial z$. We obtain

$$\frac{\partial^2 v}{\partial z \partial t} = \frac{1}{\sigma \rho_0} \frac{\partial^2 I_b}{\partial z \partial t} + u_0 \frac{\partial^2 I_b}{\partial z^2}. \quad (3.8)$$

On the other hand, differentiating (3.5) with respect to $t$ we get

$$\frac{\partial^2 v}{\partial t^2} + u_0 \frac{\partial^2 v}{\partial z^2} = \frac{e}{m} \frac{\partial E}{\partial t}. \quad (3.9)$$

Finally, we replace the second derivatives in (3.9) through their expressions in (3.7) and (3.8), yielding a second order equation for the beam current

$$\partial_t^2 I_b + 2u_0 \partial_t \partial_z I_b + u_0^2 \frac{\partial^2 I_b}{\partial z^2} = \sigma \frac{e}{m} \rho_0 \partial_t E. \quad (3.10)$$

(here and in what follows, we use $\partial_t^2$ for $\partial/\partial t^2$, $\partial_z^2$ for $\partial/\partial z^2$, etc. for the sake of brevity). Next, Pierce considers the reciprocal action of the electron beam on the transmission line (TL).
Assumption III. The action of the beam onto the waveguide amounts to a shunt current instantaneously induced on the line. This current is equal in absolute value and opposite to the current on the beam.

According to this assumption, the usual transmission line (telegraph) equations are modified so as to include an additional source term, \[ \text{[Pier51, I]} \],

\[
\partial_z I = -C \partial_t V - \partial_z I_b, \quad \partial_z V = -L \partial_t I. \tag{3.11}
\]

Here, as usual, \( I = I(t, z) \) and \( V = V(t, z) \) denote respectively the current through the inductive element and the voltage on the shunt capacitive element of the TL, \( C > 0 \) and \( L > 0 \) are respectively the shunt capacitance and inductance per unit of length. Note also that in the equations (3.11) \( \partial_z I \) and \( \partial_z V \) are respectively the current through the shunt capacitive element and the voltage drop on the inductive element of the TL per unit length. The addition of the source term \( -\partial_z I_b \) can be justified under the assumption of quasi-stationarity of the process: the charge wave on the beam "mirrors" onto the line. One of the lumped elements in the discretization of such excited TL is represented in Fig. 3.1. Induced current can be thought of as a distributed shunt current source.

The axial component of the electric field associated to the waveguide is related to the TL voltage:

\[
E(t, z) = -\partial_z V(t, z). \tag{3.12}
\]

Plugging the above expression into (3.10), we arrive at the equation

\[
\partial_t^2 I_b + 2u_0 \partial_t \partial_z I_b + u_0^2 \partial_z^2 I_b = -\sigma \frac{e}{m} \rho_0 \partial_t \partial_z V. \tag{3.13}
\]

Thus, according to \[ \text{[Pier51, I]} \], the equations (3.11) and (3.13) constitute a model of the interactive TL-beam (TLB) system.

Some comments are in order. In more recent literature, improved versions of the linear Pierce model have been considered, see e.g. \[ \text{[Nus 4]} \]. These versions account for finer features such as bunching saturation, or retain the nonlinearity present in the original versions of equations (3.1) and (3.4), etc. Although such enriched models are undoubtedly more realistic and numerical computations based on them might provide a better agreement with experiment, they hardly allow for analytical treatment. In particular, they do not possess
a Lagrangian structure. Pierce’s model, though simple, already captures the mechanism of amplification and, as mentioned in the Introduction, can be generalized to the case of MTLB systems, and allows for a thorough mathematical analysis in all cases. Taking into account the fact that real wave guides can be approximated, in principle, by an MTL with any degree of accuracy, [Nit], [Paul], [SchwiE], such generalization opens new perspectives in design optimization, which is the ultimate goal of our study.

4 Lagrangian formulation of Pierce’s model

In this section we construct a Lagrangian field theory underlying the Pierce model. The Lagrangian theory provides a deeper insight into mathematical mechanism of amplification and energy transfer from the electron beam to the radiation.

4.1 The Lagrangian

The linear system of equations (3.11)-(3.13) arises as Euler-Lagrange equations associated to certain quadratic Lagrangian. To see this, let us first introduce the charge variables $Q$ and $q$ related respectively to the currents $I$ and $I_b$ by

$$I = \partial_t Q, \quad I_b = \partial_t q. \quad (4.1)$$

Thus the variables $Q, q$ represent the amount of charge traversing the cross-section of the line (respectively the beam) at the point $z$ within the time interval $(t_0, t)$, where $t_0$ is some fixed reference time. Then the TLB system (3.11) and (3.13) takes the form

$$\partial_z Q = -CV - \partial_z q, \quad \partial_z V = -L\partial^2_t Q, \quad (4.2)$$

$$(\partial_t + u_0 \partial_z)^2 q = -\frac{\sigma \omega_p^2}{4\pi} \partial_z V, \quad (4.3)$$

where $\omega_p$ is the plasma frequency defined (in Gaussian units) by

$$\omega_p^2 = \frac{4\pi e \rho_0}{m}. \quad (4.4)$$

[DayNP 2.2].

Since it is not any harder to deal with inhomogeneous (in particular, periodic) TLs, we suppose from now on that $C$ and $L$ can be position dependent, that is

$$C = C(z), \quad L = L(z). \quad (4.5)$$

Notice that the first equation in (4.2) readily implies the following representation for $V$

$$V = -C^{-1} \partial_z (Q + q). \quad (4.6)$$

Inserting the above expression for $V$ into the second equation in (4.2) and into the equation (4.3) yield the following TLB evolution equations for the charges:

$$L\partial^2_t Q - \partial_z \left[ C^{-1} \partial_z \right] (Q + q) = 0, \quad (4.7)$$
The Lagrange equations take the form
\[ L = \frac{1}{2} \left( \partial_t q + u_0 \partial_z q \right)^2 - \frac{1}{2} C^{-1} \left( \partial_z q + \partial_z B, C^{-1} [\partial_z Q + \partial_z qB] \right) + \frac{\xi}{2} \left( \partial_t q + u_0 \partial_z q \right)^2. \] (4.9)

Indeed, for a general Lagrangian density \( L = L(t, z, Q, \partial_z Q, \partial_t q, \partial_z q) \), the Euler-Lagrange equations take the form
\[ \partial_t \frac{\partial L}{\partial(\partial_t q)} + \partial_z \frac{\partial L}{\partial(\partial_z Q)} - \frac{\partial L}{\partial q} = 0, \quad \partial_t \frac{\partial L}{\partial(\partial_t q)} + \partial_z \frac{\partial L}{\partial(\partial_z q)} - \frac{\partial L}{\partial q} = 0. \] (4.10)

A straightforward computation confirms that the application of equation (4.10) to the Lagrangian defined by (4.9) indeed yields the TLB evolution equations (4.7) and (4.8). Pierce’s original equations are obtained as a particular case, when \( C, L \) are constant along the line.

As to the units of \( L \), they are energy/length, as expected for a Lagrangian density. In this respect, we remind that we are using Gaussian units and charge \( q^2 = \text{force} \times \text{length}^2 \), in agreement with the Gaussian version of Coulomb’s law, \( F = q_1 q_2 / r^2 \).

Let us make a final observation: It is assumed that the current induced by the beam onto the TL is due to the fact that the charge on the beam perfectly “mirrors” onto the waveguide. This assumption can be justified as an approximation in the “quasistatic” regime, in the spirit of Ramo’s Theorem, \( \text{[Ra], [Tsim]} \). According to some authors, e.g. R. Kompfner, \( \text{[Kom]} \) or J.H. Booske, \( \text{[Nus, 4]} \), in dealing with real devices a coefficient \( \kappa \in (0, 1) \) must be included in front of \( \partial_z I_b \) in (3.11) (accordingly in (4.2)) to account for the real induced current, the case \( \kappa = 1 \) being regarded as ideal. The Lagrangian approach can easily handle the general case. However, in order to keep the exposition as simple as possible, we only consider the ideal situation.

### 4.2 Generalization to multiple transmission lines

It is known that fairly general wave guides can be well approximated by multiple transmission lines, MTL, \( \text{[Paul]} \). The corresponding generalization of Pierce’s model is straightforward thanks to our Lagrangian formulation. Indeed, suppose that we have \( n + 1 \) conductors, one of them being grounded, say the \( (n + 1) \)-th. We denote by \( V(z, t) = \{V_i(z, t)\}_{i=1 \ldots n} \) the \( n \)-dimensional vector-column of voltages on the first \( n \) conductors with respect to the ground and by \( I(z, t) = \{I_i(z, t)\}_{i=1 \ldots n} \) the vector-column of currents flowing on them and set
\[ Q(z, t) = \{Q_i(z, t)\}_{i=1 \ldots n}, \quad Q_i(z, t) = \int_0^t I_i(z, s) ds. \]

Let \( L = L(z) \), \( C = C(z) \) be the \( n \times n \) matrices of self- and mutual inductance and capacity. As it is well known, they are positive symmetric (Hermitian). A natural generalization of (4.9) is provided by
\[ L = \frac{1}{2} \left\{ (\partial_t Q, L \partial_t Q) - (\partial_z Q + \partial_z qB, C^{-1} [\partial_z Q + \partial_z qB]) \right\} + \frac{\xi}{2} \left( \partial_t q + u_0 \partial_z q \right)^2, \] (4.11)
where \(( \, , \, )\) stands for the scalar product in \(\mathbb{R}^n\) and \(B\) is the \(n\)-dimensional vector-column with all components being the unity, that is
\[
B = (1, 1, \ldots, 1)^T.
\]

The corresponding Euler-Lagrange second order system is
\[
L \partial_t^2 Q - \partial_z \left[ C^{-1}(\partial_z Q + \partial_z q B) \right] = 0; \tag{4.13}
\]
\[
\xi \left[ \partial_t^2 q + 2u_0 \partial_t \partial_z q + u_0^2 \partial_z^2 q \right] - (B^T, \partial_z \left[ C^{-1}(\partial_z Q + \partial_z q B) \right]) = 0.
\]

The generalized telegraph equations, equivalent to the first equation above, adopt the form
\[
\partial_z I = -C \partial_t V - \partial_z I_b B; \quad \partial_z V = -L \partial_t I. \tag{4.14}
\]

Our choice of the vector \(B\) assumes, besides perfect induction, a symmetry in the interaction between the beam and the different lines. A more realistic approach might include coefficients \(\kappa_i \in (0, 1)\) in vector \(B\) to account for non-symmetric interaction. As we already mentioned in Subsection 4, such effects can be easily handled by our approach.

Observe that if we remove the beam from the system by setting \(q = 0\), our model is in full agreement with well established models for the interaction of several lines, derived from Maxwell’s equations under reasonable assumptions. See, for example, \([\text{Nit}, 2]\), \([\text{Paul}, 1.4.1]\) for models of interacting TLs.

To summarize: from now on, by MTLB system we mean the field Lagrangian system governed by the Lagrangian \(L\) in (4.11) and the corresponding Euler-Lagrange field equations (4.13).

5 The beam as a source of amplification. The role of instability

Evidently, the beam is the sole source of energy in the MTLB system and the ultimate responsible for the presence of exponentially growing modes. In this section we identify and analyze the mathematical mechanism underlying amplification.

To trace the amplification to the beam we view the Lagrangian (4.11) as a perturbation of the Lagrangian \(L_b\) for the isolated beam defined by
\[
L_b = \frac{1}{2} (\partial_t q + u_0 \partial_z q)^2 = \frac{1}{2} \left[ (\partial_t q)^2 + 2u_0 \partial_t \partial_z q + u_0^2 (\partial_z q)^2 \right]. \tag{5.1}
\]

We introduce the equivalent Lagrangian \(\tilde{L} = \frac{1}{\xi} L\), where \(L\) is as in (4.11), i.e.
\[
\tilde{L} = L_b + \varepsilon L' = \frac{1}{2} (\partial_t q + u_0 \partial_z q)^2 + \frac{\varepsilon}{2} \left\{ (\partial_t Q, L \partial_t Q) - (\partial_z Q + \partial_z q B, C^{-1}[\partial_z Q + \partial_z q B]) \right\}, \tag{5.2}
\]
and \(\varepsilon = 1/\xi\). Small values of \(\xi\) defined by (4.8) and consequently large values \(\varepsilon\) correspond to strong coupling and regimes where the beam effectively feeds its energy into transmission lines in the form of EM field. The EM field energy gain originates in the beam as an infinite reservoir of the potential energy \(-\frac{1}{2}(u_0 \partial_z q)^2\). Importantly, the potential energy is negative unlike in oscillatory systems. For small coupling as we will show no energy transfer might occur from the beam to the EM field. This perturbation analysis suggests to consider first the beam as an isolated system.
5.1 Charge wave dynamics

In this subsection, we investigate beam charge dynamics as an isolated system, described by (5.1). We already mentioned the role of the term $u_0^2 (\partial_z q)^2$ as a source of energy. This term is responsible for the system instability manifesting itself by exponentially growing solutions of the associated E-L equations. The gyrotropic term $u_0 \partial_t q \partial_z q$ in the Lagrangian provides for stabilizing effect. As we will see, for the Lagrangian (5.1) the balance between instability and stability is struck exactly in the margin. Namely, a small perturbation of this Lagrangian can make the system either stable or unstable.

The beam Lagrangian $L_b$ is quadratic in $(\partial_t q, \partial_z q)$, see Section 10.2, and has the following structure

$$L_b = \frac{1}{2} \alpha (\partial_t q)^2 + \theta \partial_t q \partial_z q - \frac{1}{2} \eta (\partial_z q)^2 = (\partial_t q, \partial_z q)^T M (\partial_t q, \partial_z q),$$

(5.3)

where

$$M = \begin{bmatrix} \alpha & \theta \\ \theta & -\eta \end{bmatrix} = \begin{bmatrix} 1 & u_0 \\ u_0 & u_0^2 \end{bmatrix}, \quad \text{(thus } \alpha = 1, \theta = u_0, \eta = -u_0^2).$$

(5.4)

The corresponding Euler-Lagrange equation (10.16) is

$$(\partial_t + u_0 \partial_z)^2 q = 0.$$  

(5.5)

Applying the general formulas (8.2)-(8.3) for the energy $H$ and its flux $S$ we obtain

$$H_b [q] = \frac{1}{2} (\partial_t q)^2 - \frac{u_0^2}{2} (\partial_z q)^2,$$

(5.6)

$$S_b [q] = \partial_t q (u_0 \partial_t q + u_0^2 \partial_z q) = u_0 \partial_t q (\partial_t q + u_0 \partial_z q) = u_0 (\partial_t q)^2 + u_0^2 \partial_t q \partial_z q.$$  

(5.7)

Since $L_b$ does not depend on time explicitly, conservation of energy takes place, (10.56):

$$\frac{\partial H_b}{\partial t} + \frac{\partial S_b}{\partial z} = 0.$$  

(5.8)

5.1.1 Eigenmodes and stability issues.

Since the beam parameters are constant in space we can make use of the dispersion relation to study the eigenmodes. Thus, if we try solutions of the form $q(z, t) = e^{-i(\omega t - k z)}$ in (5.5), we get

$$\omega^2 - 2u_0 \omega k + u_0^2 k^2 = (\omega - u_0 k)^2 = 0,$$

(5.9)

hence $k_\omega = \omega / u_0$ is a double real root. The corresponding eigenmodes are $q_1(z, t) = e^{i(k_\omega z - \omega t)}$ and $q_2(z, t) = z e^{i(k_\omega z - \omega t)}$ or their real valued counterparts

$$v_1(z, t) = \cos (k_\omega z - \omega t), \quad v_2(z, t) = z \cos (k_\omega z - \omega t).$$

The associated energy flux is, according to (5.7),

$$S_b [v_1] = 0, \quad S_b [v_2] = -u_0^2 z \omega \sin (k_\omega z - \omega t) \cos (k_\omega z - \omega t).$$  

(5.10)

To make useful inference related to conservation laws it is common to use the following time-averaging operation. Namely, for a (locally integrable) function $f$ defined on $[0, \infty)$ we introduce

$$\langle f \rangle = \lim_{T \to \infty} \frac{1}{T} \int_0^T f(t) \, dt.$$  

(5.11)
This time-averaging operation has the following properties. If $f$ is a smooth and bounded function on $[0, \infty)$, then
\[
\left\langle \frac{df}{dt} \right\rangle = \lim_{T \to \infty} \frac{1}{T} \int_0^T \frac{df}{dt} \, dt = \lim_{T \to \infty} \frac{1}{T} \left[ f(T) - f(0) \right] = 0. \tag{5.12}
\]
Differentiation with respect to parameters commutes with the time-averaging operation. Namely, if $f$ also depends (smoothly) on some parameter $z$ the following identity holds
\[
\langle \partial_z f \rangle = \partial_z \langle f \rangle. \tag{5.13}
\]
Taking time average on both sides of the conservation law (5.8) and using the above properties of averaging, we conclude that
\[
\langle S_b[v_2] \rangle(z) = \text{const.}
\]
On the other hand, it follows from (5.10) that $\langle S_b[v_2] \rangle(0) = 0$. Hence $\langle S_b[v_2] \rangle(z) = 0$.

From the stability point of view, this situation is a very degenerate one. To illustrate this point, let us introduce a special form perturbation in the beam dispersion relation (5.9):
\[
\omega^2 - 2\alpha u_0 \omega k + u_0^2 k^2 = 0 \quad \text{or} \quad (\omega - \alpha u_0 k)^2 = (\alpha^2 - 1) u_0^2 k^2,
\]
where $\alpha$ is a real number. Our situation corresponds to $\alpha = 1$. Let us consider the behavior close to $\alpha = 1$. The quadratic equation above has the following solutions
\[
k_\omega(\alpha) = \frac{\omega}{u_0} (\alpha \pm \sqrt{\alpha^2 - 1}).
\]
Notice that if $\alpha^2 < 1$ the above solutions become complex conjugate, whereas if $\alpha^2 > 1$ they are real distinct. $\alpha = 1$ corresponds to a double real solution, already showing the degeneracy.

An important subject of our interest is the analysis of MTL structures in which the parameters vary periodically in $z$. The Floquet theory and, in particular, the Floquet multipliers are the mathematical objects that deal with such situations par excellence. As explained above, we may consider the coupled system as a perturbation of the beam. Consequently, it is instructive to take a look at the isolated beam in the light of Floquet theory with arbitrary period (eventually dictated by the period of the structure).

The Floquet multipliers with period unity are $\rho_\omega(\alpha) = e^{i k_\omega(\alpha)}$. It is clear that in the case $\alpha^2 < 1$ they are symmetrically located with respect to the unit circle in the complex plane. The solution corresponding to the multiplier outside the circle is a growing wave, whereas the one corresponding to the multiplier inside the circle is an evanescent one. In the opposite case $\alpha^2 > 1$, both roots are located on the unit circle, and the corresponding modes are purely oscillatory. We refer to these two qualitatively different perturbations as respectively unstable and stable. Aiming at amplification by coupling the beam to a MTL, the unstable situation is the one to be favoured.

The special perturbation of the beam equation considered above was for illustration purposes to see the degenerate stability properties of the system under perturbation of its parameters. For the MTLB system, however, it is the MTL the one that plays the role of perturbation. In Section 7, we prove that the desired instability and resulting amplification for spatially homogeneous MTL is achieved by sufficiently strong coupling (small values of $\xi$). An extension of this result to the case of periodic MTLs is left for a forthcoming publication.
Additional insight into the mathematical mechanism of amplification associated with instability can be gained by looking at the nature of the partial differential equations involved. Indeed, the equation

$$\left(\frac{\partial^2}{\partial t^2} + 2\alpha u_0 \frac{\partial}{\partial t} + u_0^2 \frac{\partial^2}{\partial z^2}\right) q = 0$$ \quad (5.14)

is hyperbolic if $\alpha^2 > 1$. In this case, there are two propagation velocities $v^\pm(\alpha)$ of the same sign, namely

$$v^\pm(\alpha) = \frac{\omega}{k^\pm(\alpha)} = -u_0(\alpha \mp \sqrt{\alpha^2 - 1}),$$

and the general solution has the form

$$q(z, t) = q_1(z - v^+ t) + q_2(z - v^- t).$$

Therefore, any solution which is bounded in time (as it is the case for harmonic in time solutions) is automatically bounded in space. In other words, no harmonic in time regime can be exponentially growing in space.

In the critical case, $\alpha = 1$, the equation is of the parabolic type. Changing variables $(z, t) \rightarrow (\xi, \eta)$ with $\xi = x - u_0 t$, $\eta = ax + bt$ with $b + au_0 \neq 0$, it can be easily checked that the general solution in this case is

$$q(z, t) = zF(z - u_0 t) + G(z - u_0 t) = \tilde{t}F(z - u_0 t) + \tilde{G}(z - u_0 t),$$

where $F, G, \tilde{F}, \tilde{G}$ are arbitrary functions. In particular any travelling wave with velocity $u_0$ is a solution. Again here we see that bounded in time dependence can be accompanied by at most linear growth in space.

If $\alpha^2 < 1$ we are dealing with the elliptic case where there is no propagation. This is the only case allowing for exponential amplification. Indeed, a linear change of variables $(z, t) \rightarrow (\xi = az + bt, \eta = cz + dt)$ transforms the equation into the Laplace equation

$$u_{\xi\xi} + u_{\eta\eta} = 0,$$

which admits real solutions of the form $u(\xi, \eta) = e^{k\xi} \cos(\omega\eta), u(\xi, \eta) = e^{k\xi} \sin(\omega\eta)$, etc.

## 6 Hamiltonian structure of the MTLB system

In order to study the MTLB system, in particular the associated modes, their stability and the amplification phenomenon, we make use of the Hamiltonian structure associated to the Lagrangian (5.2). More precisely, we use a version of Hamiltonian formalism that treats the space and time variables on the same footing, known as de Donder-Weyl formalism. For reader’s convenience, we have gathered the basic information about this topic in Section 10.1. As usual, the passage from Lagrangian to Hamiltonian point of view allows to cast the second-order Euler-Lagrange system of equations (4.13) in the form of a first order system, either with respect to $t$ or with respect to $z$.

To comply with notations of Section 10.1 from now on we put $q_1 = Q$, $q_2 = q$, $q = (q_1, q_2)^T$. The Lagrangian $\tilde{L}$ in (5.2) is quadratic in its variables $(\partial_t Q, \partial_t q, \partial_z Q, \partial_z q)$, that is in $(\partial_t q, \partial_z q)^T$ in the new notation. Indeed,

$$\tilde{L} = \frac{1}{2} \partial_t q^T \alpha \partial_t q + \partial_t q^T \theta \partial_z q - \frac{1}{2} \partial_z q^T \eta \partial_z q,$$ \quad (6.1)
where
\[ \alpha = \begin{bmatrix} \epsilon L & 0 \\ 0 & 1 \end{bmatrix}, \quad \theta = \begin{bmatrix} 0 & 0 \\ 0 & u_0 \end{bmatrix}, \quad \eta = \begin{bmatrix} \epsilon C^{-1} & \epsilon C^{-1}B \\ \epsilon B^T C^{-1} & \epsilon B^T C^{-1}B - u_0^2 \end{bmatrix}, \]
\( \text{(6.2)} \)

or, using a block matrix,
\[ \tilde{\mathcal{L}} = \frac{1}{2} u^T M_L u, \quad \text{with} \quad M_L = \begin{bmatrix} \alpha & \theta \\ \theta & -\eta \end{bmatrix}, \quad u = \begin{bmatrix} \partial_t q \\ \partial_z q \end{bmatrix}. \]
\( \text{(6.3)} \)

Let us introduce the vector of canonical momenta \( p = (p_z, p_\epsilon)^T \), related to the vector \( u \) above by means of
\[ p = M_L u, \]
\( \text{(6.4)} \)
where \( M_L \) is as in \( \text{(6.3)} \). In the following result, we express the dynamics of our system in terms of the variables \( p_z \) and \( \partial_t q \).

**Theorem 6.1** The second order Euler-Lagrange system \( \text{(4.13)} \) is equivalent to the \( 2n \)-first order system
\[ \tilde{J} \partial_z V = i \partial_t \tilde{M} V, \quad V = \begin{bmatrix} p_z \\ \partial_t q \end{bmatrix}, \]
\( \text{(6.5)} \)

where
\[ \tilde{J} = \begin{bmatrix} 0 & i1 \\ i1 & 0 \end{bmatrix}, \quad \tilde{M} = \tilde{M} (z) = \begin{bmatrix} -\eta (z)^{-1} \theta & \eta (z)^{-1} \theta \\ \theta \eta (z)^{-1} \theta & -\alpha (z) - \theta \eta (z)^{-1} \theta \end{bmatrix}. \]
\( \text{(6.6)} \)

**Proof.** The derivation of de Donder-Weyl version of Hamilton equations in the variable \( z \) for general quadratic Lagrangians is described in Section \( \text{(10.2)} \). In particular, for \( \tilde{\mathcal{L}} \) defined by \( \text{(6.1)} \) the Hamiltonian \( H_{\text{DW}} (p) \) does not depend explicitly on \( q \) and
\[ H_{\text{DW}} (p) = \tilde{\mathcal{L}} (u), \]
where \( p \) is linked to \( u \) as in \( \text{(6.4)} \). Equivalently,
\[ H_{\text{DW}} (p) = \frac{1}{2} p^T M_L^{-1} p. \]
\( \text{(6.7)} \)

According to \( \text{(10.38)} \), \( \text{(10.39)} \), in the variables \( (p_z, \partial_t q)^T \), the corresponding first-order system is precisely \( \text{(6.5)} \), with \( \tilde{J} \) and \( \tilde{M} \) as in \( \text{(6.6)} \).

We recall that \( \tilde{J} \) and \( \tilde{M} \) are, respectively, antihermitian and hermitian, i.e.,
\[ \tilde{J}^* = -\tilde{J}, \quad \tilde{M}^* = \tilde{M}. \]

Consider now a time harmonic solution of the form \( q(z, t) = \hat{q} (z) e^{-i \omega t} \). In this case,
\[ V(z, t) = \begin{bmatrix} p_z \\ \partial_t q \end{bmatrix} = \hat{V} (z) e^{-i \omega t}, \quad \hat{V} (z) = \begin{bmatrix} \hat{p}_z (z) \\ -i \omega \hat{q} (z) \end{bmatrix} \]
\( \text{(6.8)} \)
and the Hamiltonian equation \( \text{(6.5)} \) is reduced to
\[ \tilde{J} \partial_z \hat{V} = \omega \tilde{M} \hat{V}. \]
\( \text{(6.9)} \)

Notice that the equation \( \text{(6.9)} \) for \( \hat{V} \) is Hamiltonian according to the definition in Section \( \text{(10.3)} \) and the conservation law \( \text{(10.52)} \) applies, yielding
\[ \hat{V}^* \tilde{J} \hat{V} = i \left[ \hat{p}_z \ ( -i \omega \hat{q} (z)) + (-i \omega \hat{q} (z))^* \hat{p}_z \right] = 2i \Re \{ ( -i \omega \hat{q} (z))^* \hat{p}_z \} \]
\[ = -2i \omega \Im \{ ( \hat{q} (z))^* \theta ( -i \omega \hat{q} (z) - \eta (z) \partial_z \hat{q} (z)) \} = \text{constant}. \]
\( \text{(6.10)} \)

Later on, in Section \( \text{(8.1)} \) we will see how the above conservation law relates to energy flux constancy.
7 Amplification for the homogeneous case

This subsection is devoted to the analysis of the amplification regime associated with a single exponentially growing mode in the case of an homogeneous MTLB system, that is with parameters not varying with $z$. For real $\omega$ we seek solutions of (4.13) in the form

$$Q(z, t) = \hat{Q} e^{-i(\omega t - kz)}, \quad q(z, t) = \hat{q} e^{-i(\omega t - kz)},$$

where $\hat{q}$ and $k$ are complex constants and $\hat{Q}$ is a complex vector. We show that, under certain conditions, there is a solution with genuinely complex, that is, non-real wave number $k$.

Let us recall that the eigenvelocities of the MTL are the roots of the equation

$$|C^{-1} - v^2 L| = 0,$$

[Paul, Nit]. Since both $L$ and $C$ are positive definite, the symmetric $n \times n$ matrix $L^{-1/2}C^{-1}L^{-1/2}$ has positive eigenvalues $0 < \lambda_1 \leq \lambda_2 \leq \ldots \leq \lambda_n$, where multiple eigenvalues are repeated according to their multiplicity. Then the MTL has characteristic velocities are precisely

$$\pm v_i, \text{ where } v_i^2 = \lambda_i.$$

(7.2)

**Theorem 7.1** Let $u_0, \xi > 0$. If either

(i) $0 < u_0 \leq v_1$ or

(ii) $v_1 < u_0$ and $\xi > 0$ is sufficiently small,

then for each real $\omega$ there are exactly two genuinely complex conjugate values $k_0$ and $k_0^*$ such that (7.1) is a non-trivial solution of equations (4.13).

Hence, assuming $\text{Im} k_0 < 0$ we have the associated solution

$$Q(z, t) = A(z)e^{-i\omega t}e^{-(\text{Im} k_0)z}, \quad q(z, t) = B(z)e^{-i\omega t}e^{-(\text{Im} k_0)z}, \quad A(z), B(z) \neq 0,$$

(7.3)

that grows exponentially in the $+z$ direction, whereas the solution associated with $k_0^*$ decays exponentially.

We sketch the proof, deferring the mathematical details to section 10.5. Substituting the expressions (7.1) into the system (4.13), we obtain the following linear algebraic system of $n + 1$ equations for $\hat{Q}, \hat{q}$:

$$\begin{bmatrix}
-\nu^2 L + C^{-1} \\
D^T \\
d - \xi(v - u_0)^2
\end{bmatrix}
\begin{bmatrix}
\hat{Q} \\
\hat{q}
\end{bmatrix}
= \begin{bmatrix}
0 \\
0
\end{bmatrix}, \quad \text{where } \nu = \frac{\omega}{k}
$$

(7.4)

and

$$D = (D_i), \quad D_i = \sum_j (C^{-1})_{ij}, \quad d = \sum_i D_i.
$$

(7.5)

For the sake of brevity, we denote

$$A(v) = -v^2 L + C^{-1}, \quad \tilde{A}(v) = \begin{bmatrix}
A(v) \\
D^T \\
d - \xi(v - u_0)^2
\end{bmatrix}.
$$

(7.6)
The system (7.4) has nontrivial solutions if and only if $|A(v)| = 0$. The corresponding polynomial equation of degree $2n + 2$ is the dispersion relation of our system written in terms of the velocity $v$. In Section [10.5] we prove in full detail that the equation $|A(v)| = 0$ has exactly one pair of complex conjugate solutions if either (i) or (ii) holds. Here we outline the main ideas of the proof.

First of all, if $|A(v)| \neq 0$, the following canonical factorization takes place

$$|\tilde{A}(v)| = |A(v)| \left[ d - \xi (v - u_0)^2 - D^T (A(v))^{-1} D \right].$$

(7.7)

The values of $v$ such that $|A(v)| = 0$ are precisely the eigenvelocities $\pm v_i$ of the waveguide. Therefore, the roots of $|\tilde{A}(v)| = 0$ different from $\pm v_i$, $i = 1, 2, \ldots, n$ are the roots of the equation

$$- \xi (v - u_0)^2 = R(v), \text{ where } R(v) = D^T (A(v))^{-1} D - d,$$

(7.8)

in which the two components of the system enter separately. The rational function $R(v)$ in (7.8) contains the relevant information about the MTL whereas the left hand side depends only on the beam parameters. In what follows we refer to function $R(v)$ as MTL characteristic function. It can be explicitly written in terms of the characteristic velocities:

$$R(v) = \sum_{i=1}^{n} \frac{D_i^2}{v_i^2 - v^2} - d,$$

(7.9)

where $\tilde{D}_i$ are constants related to $D_i$, see Section [10.5]. The graph of the MTL characteristic function $R$ is symmetric with respect to the vertical axis and is made up of branches, a central one with the minimum at $(0,0)$, a number of increasing branches for $v > 0$ and decreasing for $v < 0$. One can readily see that $\lim_{v \to -\infty} R(v) = -d$. In addition to that, the graph of $R$ has vertical asymptotes at $v = \pm v_i$ if at least one of the associated $\tilde{D}_j$ does not vanish. The number of the asymptotes varies between $2$ and $2n$. The left-hand side in (7.8) is a parabola with vertex at $(u_0,0)$.

Figure [7.1] shows the graph of $R$ and that of the parabola $y = -\xi (v - u_0)^2$ with the following inductance and capacity matrices

$$L = \begin{bmatrix} 4 & 1 & 1/2 \\ 1 & 5 & 2 \\ 1/2 & 2 & 2 \end{bmatrix}; \quad C = \begin{bmatrix} 2 & 1 & 2 \\ 1 & 4 & 0 \\ 2 & 0 & 1 \end{bmatrix}.$$

The approximate values of the characteristic velocities are: $v_1 = 0.18357$ and $v_2 = 0.42383$. In Figure [7.1] (a), $u_0 = 0.18$ and $\xi = 2$; in Figure [7.1] (b), $u_0 = 0.8$ and $\xi = 18$.

It is important to observe that the parabola always intersects all the branches of $R$ except for the central one. For small $\xi$ each branch is intersected only once, and consequently the number of real roots of the equation (7.8) is exactly the number of asymptotes, as in Figure [7.1] (a) above. For large $\xi$ however the number of real roots can exceed the number of asymptotes as in Figure [7.1] (b), where a large value of $\xi$ produces three points of intersection with the far right branch of the graph of $R$. Moreover, if $u_0 \leq v_1$ (geometrically, the vertex of the parabola lies between the vertical axis and the first asymptote), then clearly the number of real roots equals the number of asymptotes irrespective of the value of $\xi > 0$. These facts can be proved rigorously based on monotonicity properties, but their geometric interpretation
Figure 7.1: (a) $u_0 < v_1$: the parabola $y = -\xi(v - u_0)^2$ (dashed line) intersects each branch of $y = R(v)$ (where $R$ is as in (7.9)) just once. Four real roots. (b) $u_0 > v_1$: for large $\xi$, the parabola intersects one of the branches of $y = R(v)$ three times. Six real roots.

is so transparent that a quick look at Figure 7.1 is quite convincing. In the generic case, the roots of the dispersion relation $|\tilde{A}(v)| = 0$ are exactly those of equation (7.8), but in general some of the $v_i$ can also be roots. Whenever some $v_i$ is a real root (maybe multiple) of $|\tilde{A}(v)| = 0$, the number of asymptotes in the graph of $R$ is reduced by the corresponding amount. The same is true of the number of real roots of (7.8) under either (i) or (ii). This fact follows from factorization (7.7). The main point is that in all cases the total number of real roots of $|\tilde{A}(v)| = 0$ is $2n$ if either condition (i) or (ii) above holds. We thus conclude that under (i) or (ii) there is necessarily a unique pair of complex conjugate roots. The detailed proof of these facts is provided in Section 10.5.

It is not difficult to estimate how small $\xi$ should be in condition (ii) above. In the case $n = 1$ and $u_0 > v_1$ there is a precise criterion on $\xi$, namely amplification takes place if

$$\xi < \xi_0 := \frac{L\gamma^2}{1 - \gamma^{2/3}}, \quad \gamma = \frac{v_1}{u_0} = \frac{1}{u_0\sqrt{LC}}. \quad \text{(7.10)}$$

A simple sufficient condition can be also given for $n > 1$. For example, one can just impose that the left branch of the parabola at $v = 0$ be flatter than the flattest point of the graph of $R$ on $(v_1, u_0)$. This leads to

$$\xi < \tilde{\xi}_0 := \frac{\min_{v \in (v_1, u_0)} R'(v)}{2u_0}. \quad \text{(7.11)}$$

Observe that both $\xi_0$ and $\tilde{\xi}_0$ vanish as $u_0 \to \infty$, as expected. The value of $\tilde{\xi}_0$ is not sharp but we did not make an effort to find one.

Thus, under the above assumptions the system exhibits spatially exponentially growing, as well as exponentially decaying time harmonic regimes. Using the terminology of dynamical systems, if we restrict to time harmonic evolutions $z \to X$, where $X = \{e^{-i\omega t} \tilde{Q}, \quad \tilde{Q} \in \mathbb{C}^{n+1}\}$, there is a subspace of data inducing an individual exponential dichotomy for $X$, both for forward and backward in $z$ evolutions, [Ha, XIII]. The subspace is
determined by the solutions \( \tilde{Q} = (\tilde{Q}, \tilde{q}) \) of the system (7.4) with \( v \) being the corresponding complex solution of \( |A(v)| = 0 \).

7.1 Asymptotic behavior of the amplification factor as \( \xi \to 0 \) and as \( \xi \to \infty \).

Let \( k_0 \) denote the complex root with \( \text{Im} k_0 < 0 \) whose existence we proved in the previous section under appropriate conditions. It is interesting to study the asymptotics of the "amplification factor" – \( -\text{Im} k_0 \) as the beam parameter \( \xi \to 0 \), as well as its behavior when \( \xi \to \infty \).

A careful analysis shows (see Section 10.5) that, if we denote by \( v_0 = \omega/k_0 \) the corresponding velocity with \( \text{Im} v_0 > 0 \), then

\[
\text{Im} v_0 = \sqrt{K' \xi + o(\xi)} = \sqrt{K'} \sqrt{\xi} + o(\sqrt{\xi}) \quad \text{as} \quad \xi \to 0,
\]

where \( K' \) depends only on \( L, C, u_0 \). As a consequence,

\[
-\text{Im} k_0 = \frac{\text{Im} v_0}{|v_0|^2} \sim \frac{K''}{\sqrt{\xi}} \quad \text{as} \quad \xi \to 0; \quad K'' > 0.
\] (7.12)

The conclusion is that, in this model, the amplification factor can be indefinitely improved by reducing \( \xi \). According to (4.8), this amounts to increasing \( \sigma \rho_0 \), the linear electron density of the beam.

On the other hand, the limit \( \xi \to \infty \) makes sense only if \( 0 < u_0 \leq v_1 \). In the case of one line and \( u_0 = v_1 \), it can be proved that

\[
-\text{Im} k_0 = \frac{\text{Im} v_0}{|v_0|^2} \sim \frac{K''}{\sqrt{\xi}} \quad \text{as} \quad \xi \to \infty; \quad K'' > 0,
\] (7.13)

see Section 10.5.

The regime considered by Pierce corresponds to the latter situation, in which there are two real solutions (for \( v \)) close to \( \pm u_0 \), and two complex conjugate with real part close to \( u_0 \); see Section 9. The situation is similar for \( u_0 < v_1 \), but in this case \( -\text{Im} k_0 \) has a finite positive limit as \( \xi \to \infty \).

8 Energy conservation and transfer

The conservation laws for our system can be obtained via Noether theorem, [GelFom, 38.2-3], [Gold, 13.7].

Theorem 8.1 Conservation of energy for the system (4.13) holds in the form

\[
\partial_t H + \partial_z S = 0,
\] (8.1)

where the total energy \( H \) and the total energy flux \( S \) are given by

\[
H = \frac{1}{2} \partial_t q^T \alpha \partial_t q + \frac{1}{2} \partial_z q^T \eta \partial_z q;
\] (8.2)

\[
S = \partial_t q^T \theta \partial_t q - \partial_t q^T \eta \partial_z q = \partial_t q^T (\theta \partial_t q - \eta \partial_z q) = \partial_t q^T p_z.
\] (8.3)
Proof. The Lagrangian density $\mathcal{L}$ does not depend explicitly on $t$ (this is a consequence of the closedness of the system), therefore by the fields version of Noether theorem, \cite[38.2-3]{GelFom}, \cite[13.7]{Gold}, conservation of energy (8.1) holds, with energy and the energy flux densities given by

$$H = \sum_j \frac{\partial \mathcal{L}}{\partial (\partial_t q_j)} \partial_t q_j - \mathcal{L}, \quad S = \sum_j \frac{\partial \mathcal{L}}{\partial (\partial_z q_j)} \partial_t q_j.$$  

(8.4)

A straightforward computation yields the expressions of $H$ and $S$ given in (8.2), (8.3). In (8.3), $p_z$ is the canonical momentum defined in (10.17), Section 10.2.

Consider now a real time harmonic eigenmode

$$q(t, z) = \text{Re} \left\{ \hat{q}(z) e^{-i\omega t} \right\},$$  

(8.5)

which solves the Euler-Lagrange equation (10.16). Notice that $\langle q \rangle(z) = 0$, where $\langle \cdot \rangle$ is the time average operation defined in (5.11). However, if

$$a(t) = \text{Re} \left\{ \hat{a} e^{-i\omega t} \right\},$$  

(8.6)

and $b(t)$ is defined by a similar formula then we have

$$\langle ab \rangle = \frac{1}{2} \text{Re} \left\{ \hat{a}^* \hat{b} \right\}.  

(8.7)

Applying the averaging operation $\langle \cdot \rangle$ to the conservation law (8.1) for a time harmonic eigenmode $q$ as in (8.5) and using (5.12) and (5.13), we obtain

$$\partial_z \langle S \rangle(z) = 0 \quad \text{implying} \quad \langle S \rangle(z) = \text{constant}.$$  

(8.8)

On the other hand, $S$ defined by (8.3) can be written as the product of two real time harmonic functions:

$$S(t, z) = \text{Re} \left\{ \hat{A}(z) e^{-i\omega t} \right\} \text{Re} \left\{ \hat{B}(z) e^{-i\omega t} \right\},$$  

(8.9)

where

$$\hat{A}(z) = -i\omega \hat{q}(z); \quad \hat{B}(z) = -i\omega \theta \hat{q}(z) - \eta \partial_z \hat{q}(z).$$  

(8.10)

Using (8.7) we obtain the energy flux conservation law in the form

$$\langle S \rangle(z) = \frac{1}{2} \text{Re} \left\{ \langle (-i\omega \hat{q})^* \rangle \langle -i\omega \theta \hat{q} - \eta \partial_z \hat{q} \rangle \right\} = \frac{1}{2} \text{Re} \left\{ \langle (-i\omega \hat{q})^* \hat{p}_z \rangle \right\} = \text{constant}.$$  

(8.11)

Constancy of $\langle S \rangle(z)$ is related to the constancy of the symplectic square of the solution of the Hamiltonian system satisfied by

$$\hat{V}(z) = \begin{bmatrix} \hat{p}_z \\ -i\omega \hat{q} \end{bmatrix},$$  

(8.12)

see formula (6.10). Indeed,

$$V^* \bar{J} V = 2i \text{Re} \left\{ \langle (-i\omega \hat{q}(z))^* \rangle \langle -i\omega \theta \hat{q}(z) - \eta \partial_z \hat{q}(z) \rangle \right\} = \text{constant} = 4i \langle S \rangle(z).$$  

(8.13)
8.1 Energy exchange between subsystems

This section deals with the balance of energy between the two subsystems making up our system: the beam and the MTL. As already pointed out by Pierce in \[\text{Pier51, p. 635}\] an amplification regime assumes that the energy extracted from the beam is stored in the EM field. In other words, the net flux of energy must have a definite sign. Pierce tacitly considers this condition as an additional one to be imposed on top of other conditions ensuring the existence of an exponentially growing solution. We show below that in fact this condition is automatically satisfied for exponentially growing solutions.

When computing the energy flux between the beam and the MTL we take advantage of our Lagrangian setting. This setting allows for a systematic derivation of expressions for energies and fluxes satisfying \textit{a priori} the fundamental conservation laws. We proceed using the results from Section \[\text{10.4}\] for a more general coupled system.

First, we should split the Lagrangian into two parts \(\mathcal{L} = \mathcal{L}_1 + \mathcal{L}_2\) corresponding to the MTL and the beam. Namely,

\[
\mathcal{L}_1(Q_t, Q_z) = \frac{1}{2} \left( \partial_t Q, L \partial_t Q \right)^2 - \frac{1}{2} \left( \partial_z Q, C^{-1} \partial_z Q \right)^2; \quad (8.14)
\]

\[
\mathcal{L}_2(q_t, q_z) = \xi \left( \frac{1}{2} \left( \partial_t q + u_0 \partial_z q \right)^2, \text{ where } \partial_z Q = \partial_z Q + B \partial_z q. \right.
\]

The above Lagrangian has the structure of \[\text{(10.53)}, \text{ with } B = (1, 1...1)^T. \] Our first result concerning energy flows is contained in the following

\textbf{Theorem 8.2} The instantaneous power by unit length supplied by the beam to the MTL is given by

\[
P_{B\rightarrow MTL} = \partial_t \left[ \frac{1}{2} \left( CV, V \right) + \frac{1}{2} \left( LI, I \right) \right] + \partial_z (I, V) \quad (8.15)
\]

where, as usual, \((, )\) stands for the scalar product.

\textbf{Proof.} According to \[\text{(10.62)}, \text{ the power } P_{B\rightarrow MTL} \text{ flowing from the beam to the (unit length of) the MTL is given by}

\[
P_{B\rightarrow MTL} = - \partial_t \partial_t \partial_z Q^2 = \partial_z Q^T C^{-1} B \partial_z q = \partial_z I_0 \sum_i D_i \partial_z Q_i, \quad (8.16)
\]

where \(D_i = \sum_j (C^{-1})_{ij}. \)

Using \[\text{(4.14)}, \text{ we recast the expression for } P_{B\rightarrow MTL} \text{ in terms of currents and voltages. Indeed, the voltage } V \text{ is given by}

\[
V = -C^{-1} (\partial_z Q + \partial_z q). \quad (8.17)
\]

Then we notice that

\[
\sum_i D_i \partial_z Q_i = \sum_i \sum_j (C^{-1})_{ij} (\partial_z Q_i + \partial_z q B_i) = -\sum_j V_j
\]
and hence, according to (4.14),

$$P_{B\rightarrow MTL} = -\sum_j \partial_z I_b V_j = -(\partial_z I_b, V) = (C\partial_t V, V) + (\partial_z I, V) = (8.18)$$

$$= \partial_t \left[ \frac{1}{2} (CV, V) \right] + \partial_z (I, V) - (I, \partial_z V) = \partial_t \left[ \frac{1}{2} (CV, V) \right] + (L\partial_t I, I) + \partial_z (I, V) =$$

$$= \partial_t \left[ \frac{1}{2} (CV, V) + \frac{1}{2} (LI, I) \right] + \partial_z (I, V).$$

The first two terms in (8.15) correspond to $\partial_t H$ where

$$H = \frac{1}{2} (CV, V) + \frac{1}{2} (LI, I)$$

is the density of the total energy stored in the shunt capacitors and the inductances per unit length. The last term in $P_{B\rightarrow MTL}$ represents the divergence of the energy flux, $S = (I, V)$. In the particular case of one line, we recover the usual expressions for the corresponding quantities:

$$P_{B\rightarrow MTL} = \partial_t \left[ \frac{1}{2} CV^2 \right] + \partial_t \left[ \frac{1}{2} LI^2 \right] + \partial_z (IV).$$

Our next result deals with the direction on the (time averaged) power flow.

**Theorem 8.3** Let $k_0, v_0$ denote the complex values of the wave number and the velocity for the unique exponentially growing solution according to Theorem 7.1. Then, the following formula holds for the time average of the power:

$$\langle P_{B\rightarrow MTL} \rangle (z) = -\left[ \omega |k_0|^2 |\hat{q}|^2 (\text{Re} v_0 - u_0) \text{Im} v_0 \right] e^{-2(\text{Im} k_0)z}. \quad (8.21)$$

Moreover, $\langle P_{B\rightarrow MTL} \rangle (z) > 0$ for all $z$. Thus, the power on the growing solution flows from the beam to the MTL.

**Proof.** First, observe that for real time harmonic solutions $Q$ and $q$ of the form

$$Q = \text{Re} \left( \hat{Q} e^{i(kz-\omega t)} \right), \quad q = \text{Re} \left( \hat{q} e^{i(kz-\omega t)} \right), \quad (8.22)$$

where $\hat{Q}, \hat{q}$ are complex constants, the expression for $P_{B\rightarrow MTL}$ can be written in the form

$$P_{B\rightarrow MTL} = \partial_z Q^T C^{-1} B \partial_z q = \text{Re}(\hat{a}(z) e^{-i\omega t}) \text{Re} (\hat{b}(z) e^{-i\omega t}), \quad (8.23)$$

where

$$\hat{a}(z) = ike^{ikz} (\hat{Q} + B\hat{q})^T C^{-1}; \quad \hat{b}(z) = \omega k\hat{q} e^{ikz} B. \quad (8.24)$$

Applying formula (8.7) for time average, we get

$$\langle P_{B\rightarrow MTL} \rangle (z) = \frac{\omega}{2} e^{-2(\text{Im} k_0)z} \text{Im} \left\{ |k|^2 \left( \hat{Q} + B\hat{q} \right)^T C^{-1} B\hat{q} \right\}. \quad (8.25)$$

Suppose now that $k_0$ is the complex root providing amplification, that is, in the notation of Subsection 7.1 $k_0 = \omega/v_0$ with $\text{Im} k_0 < 0$. Then, $v_0$ is a root of the system (7.4) and therefore, in the notation of Section 7 and returning to the variable $k$,

$$k_0^2 (\hat{Q}^T D + \hat{d}) = \xi (\omega - k_0 u_0)^2 \hat{q}. \quad (8.26)$$
Taking complex conjugate in the above equation and observing that
\( C^{-1}B = D \) and \( B^T C^{-1}B = d \), we can rewrite (8.25) in the form

\[
\langle P_{B \rightarrow \text{MTL}} \rangle (z) = \frac{\omega \xi}{2} e^{-2(\text{Im} k_0)z} \text{Im} \left\{ \frac{|k_0|^2}{k_0^*} (\omega - u_0 k_0^*)^2 |q|^2 \right\}.
\]

(8.26)

In terms of velocities, we have

\[
\text{Im} \left( \frac{k_b - k_0^*}{k_0^*} \right)^2 = \text{Im} \left( \frac{v_0^*}{u_0} - 1 \right)^2 = -\frac{2}{u_0^2} (\text{Re} v_0 - u_0) \text{Im} v_0.
\]

(8.27)

Since we are assuming \( \text{Im} v_0 > 0 \), we see from formula (8.26) that \( \langle P_{B \rightarrow \text{MTL}} \rangle (z) \geq 0 \) for all \( z \) exactly if \( \text{Re} v_0 \leq u_0 \). But this is always the case, as it follows from (10.71) and (10.73). Formula (8.21) follows at once from (8.26) and (8.27).

Observe also that, since \( \text{Im} k_0 < 0 \), formula (8.21) implies that \( \langle P_{B \rightarrow \text{MTL}} \rangle \) increases in the \(+z\) direction. For the evanescent wave, corresponding to the value \( k_0^* \), we have exactly the opposite situation: the energy flows from the MTL to the beam and the power flux decreases in the \(+z\) direction.

### 9 The Pierce model revisited

Let us examine Pierce’s original results in the light of our general theory. They correspond to \( n = 1 \), hence \( d = D = C^{-1} \) and the dispersion relation \( |\tilde{A}(v)| = 0 \) becomes

\[
(-v^2 L + C^{-1}) \left[ C^{-1} - \xi (v - u_0)^2 \right] - C^{-2} = 0,
\]

(9.1)

which, in terms of \( k = \omega/v \), reads

\[
-L \omega^2 k^2 + \xi (\omega - k u_0)^2 (L C \omega^2 - k^2) = 0.
\]

(9.2)

After elementary algebraic transformations the above equation turns into

\[
u_0^2 k^4 - 2 u_0 \omega k^3 + \left[ 1 + \frac{L}{\xi} - L C u_0^2 \right] \omega^2 k^2 + 2 L C u_0 \omega^3 k - L C \omega^4 = 0,
\]

(9.3)

which is precisely the fourth order equation in [Pier51, (1.16)].

The TL has only two characteristic velocities, namely \( \pm v_1 = \pm 1/\sqrt{L C} \) which are not solutions of (9.1). The graph of the characteristic function \( \tilde{R} \) has only two vertical asymptotes at \( v = \pm v_1 \). The special regime considered in [Pier51] corresponds to taking large \( \xi \), and \( u_0 = v_1 \). As we know, in this case amplification occurs for any \( \xi > 0 \). For small values of the parameter

\[
k_p = \frac{\omega_p}{u_0} = \frac{1}{u_0} \sqrt{\frac{4\pi}{\sigma \xi}},
\]

(9.4)

Pierce asserts that \( k \simeq k_b = \omega/u_0 \) for the forward unattenuated wave. In terms of velocities this means that for large values of \( \xi \) the positive real solution \( v_1^+ \) is very close to \( u_0 \). The
Figure 9.1: Pierce’s dispersion relation for \( u_0 = v_1 \): For large \( \xi \), the parabola \( y = -\xi(v-u_0)^2 \) is very narrow and intersects the graph of \( y = R(v) \) close to the asymptotes: \( v_1^+, v_1^- \approx u_0 \).

The graph in Figure 9.1 refers to this situation and it clearly shows that indeed \( v_1^+ \approx u_0 \) and \( -v_1^- \approx -u_0 \) for large \( \xi \) (the parabola becomes very narrow and the right and left branches of the graph of \( R \) are intersected close to the asymptotes).

Consequently, the identity

\[
2 \text{Re} v_0 + v_1^+ + v_1^- = 2u_0
\]

implies that \( \text{Re} v_0, \text{Re} v_0^* \approx u_0 \). Therefore, three solutions have real part close to \( u_0 \) and the remaining real solution is close to \(-u_0\). The latter corresponds to the backward wave. In terms of the wavenumber, three solutions have real part close to \( k_b \). By looking for solutions (in \( k \)) to the fourth order equation (9.3) in the form

\[
k = k_b + i\delta,
\]

with small (compared to \( k_b \)) complex \( \delta \), Pierce gets rid of the backward wave. The dispersion relation (9.2) in terms of \( \delta \) reads

\[
(i\delta)^3 \left( 2 + i\delta k_b^{-1} \right) = -L\xi^{-1}k_b^2 \left( 1 + i\delta k_b^{-1} \right)^2.
\]

Neglecting \( i\delta/k_b \), we arrive at Pierce’s third degree equation for \( \delta \):

\[
\delta^3 = -\frac{Lk_b^2\xi^{-1}}{2}i,
\]

which has three complex roots,

\[
\delta_1 = ci, \quad \delta_2 = c \left( -\sqrt{3} - i \right)/2, \quad \delta_3 = c \left( \sqrt{3} - i \right)/2, \text{ where } c = \sqrt[3]{Lk_b^2\xi^{-1}/2},
\]

corresponding respectively to the unattenuated wave faster than the natural phase velocity of the circuit \( (v_1^+ > v_1 = u_0) \), the increasing and the decreasing waves.

It is clear from the analysis in Section 10.5 that in case of several identical, non-interacting TLs, only two asymptotes are present in the graph of \( R \). This fact suggests that we can replace such a system by a single effective line, with modified parameters, interacting with the beam.
Indeed, let \( C = \hat{C} \text{Id}_n \), \( L = \hat{L} \text{Id}_n \) with \( n \geq 2 \). Then, there are exactly two characteristic velocities \( \pm v_1 \) where \( v_1 = 1/\sqrt{\hat{L}\hat{C}} \). According to Section 10.5 the latter are necessarily characteristic velocities of the entire system, of multiplicity \( n - 1 \) each. Using the notation from that section, we have
\[
D = \hat{C}^{-1}(1,1,...1)^T, \quad d = n\hat{C}^{-1}, \quad \tilde{D} = \hat{L}^{-1/2}\hat{C}^{-1}(1,1,...1)^T.
\]

The MTL characteristic function \( R(v) \) has the explicit expression
\[
R(v) = \frac{n\hat{L}^{-1}\hat{C}^{-2}}{v_1^2 - v^2} - n\hat{C}^{-1}.
\] (9.9)

If we choose \( \tilde{C} = \hat{C}/n \) and \( \tilde{L} = n\hat{L} \), the above function coincides with the characteristic function for one line with parameters \( \tilde{C} \) and \( \tilde{L} \),
\[
R(v) = \frac{\tilde{L}^{-1}\tilde{C}^{-2}}{v_1^2 - v^2} - \tilde{C}^{-1}.
\]

Since amplification depends only on the complex root of the dispersion relation, which is a root of the canonical dispersion relation, amplification factors also coincide.

Actually, a more general assertion holds:

**Theorem 9.1** Let \( C \) and \( L \) be the capacity, respectively inductance matrices of an \( n \)-lines MTL. If
\[
LC = v_1^{-2} \text{Id},
\] (9.10)
then the canonical dispersion relation of the system consisting of the MTL and a given beam coincides with the canonical dispersion relation of the system consisting of a single transmission line with parameters \( \tilde{L}, \tilde{C} \) defined by
\[
\tilde{C}^{-1} = \sum_{i,j=1}^{n} (C^{-1})_{i,j}, \quad \tilde{L} = v_1^{-2}\tilde{C}^{-1}
\]
and the same beam. Consequently, the amplification factors of both systems coincide.

It should be noted that the multiple line system and the reduced (one line) system above are not equivalent in all respects. Actually, the multiline system admits oscillatory modes with eigenvelocity \( \pm v_1 \), whereas the equivalent one-line system does not. However, the exponentially growing and evanescent modes coincide, as well as the two purely oscillatory modes with eigenvelocities different from \( \pm v_1 \). The proof of the above theorem is a straightforward generalization of the case of identical lines and we omit it.

A different reduction can be achieved by suitably modifying the beam. Suppose we have \( n \) identical, uncoupled lines as before. Dividing the dispersion relation (9.9) by \( n \), we conclude that the interaction of the system with a beam with parameters \( (\xi, u_0) \) is equivalent to the interaction of one line with parameters \( \tilde{L}, \tilde{C} \) with a beam with parameters \( (\xi/n, u_0) \). The asymptotic formula (7.12) then implies that the amplification factor grows like \( \sqrt{n} \) as \( n \to \infty \).
10 Mathematical subjects

10.1 de Donder-Weyl version of the Hamiltonian formalism

In this section we introduce basic settings of the de Donder-Weyl (DW) version of the Hamilton equations which treats the time and space variable in equal manner just as the Lagrangian approach which constitutes its basis. The DW theory is a generalization of the standard Hamiltonian formalism and the Hamilton-Jacobi theory, [Rund 4.2] that has the advantage of requiring a finite-dimensional phase space. We do not use any significant results of the DW theory but rather take advantage of its set up that allows to treat the time \( t \) and the space variable \( z \) on equal footing. We remind that the standard Hamilton-Jacobi theory gives preferential treatment to time \( t \).

Let us consider a system \( q = \{q_j(t, z), \ j = 1, \ldots, n\} \) of real valued fields depending on time \( t \) and one-dimensional space variable \( z \). Suppose it has a Lagrangian density of the form

\[
L = L(t, z, q, q_t, q_z),
\]

where \( q_t = \partial_t q \), \( q_z = \partial_z q \).

The corresponding Euler-Lagrange equations are, [Gelfond 4.16]

\[
\frac{\partial L}{\partial q} - \partial_t \frac{\partial L}{\partial q_t} - \partial_z \frac{\partial L}{\partial q_z} = 0.
\]

Evidently, (10.2) is a system of second order partial differential equations for \( q \) as a function of \( t, z \). It can be recast as a first order partial differential system with respect to time \( t \) or with respect to the space variable \( z \) using a generalization of the standard Hamiltonian formalism known as de Donder-Weyl (DW) theory. Thus, following the DW theory we introduce two canonical momenta densities \( p_t \) and \( p_z \) and the DW Hamiltonian density \( H \) by the formulas

\[
p_t = \frac{\partial L}{\partial q_t}(t, z, q, q_t, q_z),
\]

\[
p_z = \frac{\partial L}{\partial q_z}(t, z, q, q_t, q_z),
\]

\[
H_{DW} = H_{DW}(t, z, q, p_t, p_z) = p_t^T q_t + p_z^T q_z - L(t, z, q, q_t, q_z),
\]

where \( q_t \) and \( q_z \) are supposed to be found from respective equations (10.3)-(10.4) and to be substituted in the right-hand side for the second equation in (10.5). Then the corresponding DW version of the Hamilton equations are

\[
\partial_t q = \frac{\partial H_{DW}}{\partial p_t}(t, z, q, p_t, p_z),
\]

\[
\partial_z q = \frac{\partial H_{DW}}{\partial p_z}(t, z, q, p_t, p_z),
\]

\[
\partial_t p_t + \partial_z p_z = - \frac{\partial H_{DW}}{\partial q}(t, z, q, p_t, p_z),
\]

and this system of \( 3n \) first order equations is equivalent to the Euler-Lagrange system (10.2).

One can solve the system (10.3)-(10.4) for \( q_t \) and \( q_z \) in terms of the momenta, obtaining representations

\[
q_t = G_t(t, z, q, p_t, p_z), \quad q_z = G_z(t, z, q, p_t, p_z),
\]

(10.9)
for some functions $G_t$ and $G_z$. Solving for $p_t$ in the first and for $p_z$ in the second, we get

$$p_t = K_t(t, z, q, q_t, p_z)$$  
$$p_z = K_z(t, z, q, p_t, q_z),$$  

(10.10)

for some functions $K_t$ and $K_z$.

To obtain the first order partial differential equations with respect to $t$ we consider the pair $p_t, q$ and using equations (10.6) and (10.8) we get

$$\frac{\partial}{\partial t} q = \frac{\partial H_{DW}}{\partial p_t} (t, z, q, p_t, p_z) = F_q(t, z, q, q_t, p_t),$$  

(10.11)

$$\frac{\partial}{\partial t} p_t = -\frac{\partial}{\partial z} p_z - \frac{\partial H_{DW}}{\partial q} (t, z, q, p_t, p_z) = F_p(t, z, q, q_z, q_t, p_t, p_t, p_z),$$  

where the expressions $F_q$ and $F_p$ are obtained by replacing $p_z$ in (10.11) by its representation (10.10). Observe that the system of partial differential equations (10.11) for $p_t$ and $q$ is of the first order with respect to time $t$.

To obtain the first order partial differential equations with respect to $z$ we consider the pair $p_z, q$ and proceed just as in the previous case with using the equations (10.7) and (10.8) to get

$$\frac{\partial}{\partial z} q = \frac{\partial H_{DW}}{\partial p_z} (t, z, q, p_z) = \tilde{F}_q(t, z, q, q_t, p_z),$$  

(10.12)

$$\frac{\partial}{\partial z} p_z = -\frac{\partial}{\partial t} p_t - \frac{\partial H_{DW}}{\partial q} (t, z, q, p_t, p_z) = \tilde{F}_p(t, z, q, q_t, q_z, q_t, p_z, p_z),$$  

where the expressions $\tilde{F}_q$ and $\tilde{F}_p$ are determined by replacing $p_t$ in the relevant expressions in (10.12) by its representation (10.10). Observe that the system of partial differential equations (10.12) for $p_z$ and $q$ is of the first order with respect to the space variable $z$.

Summing up, we have proved the following

**Theorem 10.1** The second order Euler-Lagrange system (10.2) is equivalent to either the first order system (10.11) for $q$ and $p_t$ or the first order system (10.12) for $q$ and $p_z$.

### 10.2 Quadratic Lagrangian densities

In this section we present some results concerning a special family of Lagrangians, namely those quadratic in the derivatives (and independent both of coordinates and the fields). This kind of Lagrangians often appear in practice, in particular in the TL-beam interaction system. Thus, let us consider a quadratic Lagrangian density of the form

$$L(q_t, q_z) = \frac{1}{2} \partial_t q^T \alpha \partial_t q + \partial_t q^T \theta \partial_z q - \frac{1}{2} \partial_z q^T \eta \partial_z q,$$  

(10.13)

where $q = \{q_j(t, z), j = 1, \ldots n\}$ are real valued fields depending on time $t$ and one-dimensional space variable $z$, $q_t = \partial_t q$, $q_z = \partial_z q$ and $\alpha(t, z)$, $\eta(t, z)$, $\theta(t, z)$ are symmetric $n \times n$ matrices with real entries, that is

$$\alpha^T = \alpha, \quad \eta^T = \eta, \quad \theta^T = \theta.$$  

(10.14)
The Lagrangian density (10.13) can be recast into the following form, involving a block matrix:

\[ \mathcal{L} = \frac{1}{2} u^T M_L u; \quad M_L = \begin{bmatrix} \alpha & \theta \\ \theta & -\eta \end{bmatrix}, \quad u = \begin{bmatrix} \partial_t q \\ \frac{}{} \partial_z q \end{bmatrix}. \quad (10.15) \]

The Euler-Lagrange equation (10.2) for this Lagrangian is

\[ \left[ \partial_t \alpha \partial_t + \partial_t \theta \partial_z + \partial_z \theta \partial_t - \partial_z \eta \partial_z \right] q = 0. \quad (10.16) \]

Now we would like to use the DW Hamiltonian approach from the previous section to recast the second order differential \( n \times n \) system (10.16) into first order ones with respect to \( t \) and with respect to \( z \) as well. With that in mind we introduce the canonical momenta as in (10.3)-(10.4)

\[ p_t = \frac{\partial \mathcal{L}}{\partial \dot{q}_t} = \alpha \partial_t q + \theta \partial_z q, \quad p_z = \frac{\partial \mathcal{L}}{\partial \dot{q}_z} = \theta \partial_t q - \eta \partial_z q, \quad (10.17) \]

which can be recast as

\[ p = \begin{bmatrix} p_t \\ p_z \end{bmatrix} = \begin{bmatrix} \alpha & \theta \\ \theta & -\eta \end{bmatrix} \begin{bmatrix} \partial_t q \\ \partial_z q \end{bmatrix} = M_L u, \quad (10.18) \]

or

\[ \begin{bmatrix} \partial_t q \\ \partial_z q \end{bmatrix} = u = M_L^{-1} p = M_L^{-1} \begin{bmatrix} p_t \\ p_z \end{bmatrix}. \quad (10.19) \]

Notice that the difference in signs in expressions for momenta \( p_t \) and \( p_z \) in (10.17) is due to difference in signs for matrices \( \alpha \) and \( \eta \) as they enter the expressions for the kinetic and potential energies in the Lagrangian density defined by (10.13).

Solving equations (10.18) for \( \partial_t q \) and \( \partial_z q \) we obtain

\[ \partial_t q = \alpha^{-1} (p_t - \theta \partial_z q), \quad \partial_z q = \eta^{-1} (\theta \partial_t q - p_z). \quad (10.20) \]

Using (10.13) and (10.17) we get the following identity

\[ p_t^T \partial_t q + p_z^T \partial_z q = \partial_t q^T p_t + \partial_z q^T p_z = \partial_t q^T (\alpha \partial_t q + \theta \partial_z q) + \partial_z q^T (\theta \partial_t q - \eta \partial_z q) = 2 \mathcal{L}. \quad (10.21) \]

Then in view of (10.21) the general DW Hamiltonian \( \mathcal{H}_{DW} \) defined by (10.5) takes here the form

\[ \mathcal{H}_{DW} = p_t^T \partial_t q + p_z^T \partial_z q - \mathcal{L} = \mathcal{L} = \frac{1}{2} \partial_t q^T \alpha \partial_t q + \partial_t q^T \theta \partial_z q - \frac{1}{2} \partial_z q^T \eta \partial_z q. \quad (10.22) \]

Another way to obtain a representation for the DW Hamiltonian is to use (10.19) yielding

\[ \mathcal{H}_{DW} = p^T u - \frac{1}{2} u^T M_L u = u^T M_L u - \frac{1}{2} u^T M_L u = \frac{1}{2} u^T M_L u = \frac{1}{2} p^T M_L^{-1} p. \quad (10.23) \]

Observe that the DW Hamiltonian \( \mathcal{H} \) equals the Lagrangian \( \mathcal{L} \) at the corresponding point, that is

\[ \mathcal{H}_{DW} = \mathcal{H}_{DW}(p) = \mathcal{L}(u) = \mathcal{L}, \text{ where } p = M_L u \quad (10.24) \]
(actually, this is a general property of the Legendre transform of homogeneous quadratic polynomials). The equation (10.8) takes here the form

$$\partial_t p_t + \partial_z p_z = 0. \quad (10.25)$$

To obtain the first order equations with respect to $t$ we pick the pair $p_t, q$. We use equations (10.25) and (10.20) for respectively $\partial_t p_t$ and $\partial_t q$. We eliminate $p_z$ in (10.25) by using its representation (10.17) getting the system

$$\begin{align*}
\partial_t p_t &= -\partial_z p_z = -\partial_z \theta \partial_t q + \partial_z \eta \partial_z q = -\partial_z \theta \alpha^{-1} (p_t - \theta \partial_t q) + \partial_z \eta \partial_z q, \quad (10.26) \\
\partial_t q &= \alpha^{-1} (p_t - \theta \partial_t q). \quad (10.27)
\end{align*}$$

Observe that we used equation (10.27) to get the right-hand side of equation (10.26). The above system can be written in matrix form

$$\begin{align*}
\partial_t \begin{bmatrix} p_t \\ q \end{bmatrix} &= \begin{bmatrix} -\partial_z \theta \alpha^{-1} & \partial_z \eta \partial_z + \partial_z \theta \alpha^{-1} \theta \partial_z \\ \alpha^{-1} & -\alpha^{-1} \theta \partial_z \end{bmatrix} \begin{bmatrix} p_t \\ q \end{bmatrix}. \quad (10.28)
\end{align*}$$

One can recast the above system into a canonical Hamiltonian form by using the following symplectic matrix

$$J = \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix}, \quad J^2 = -1, \quad J = -J^T. \quad (10.29)$$

Namely,

$$\begin{align*}
\partial_t V &= JM_{Ht} V, \quad V = \begin{bmatrix} p_t \\ q \end{bmatrix}, \quad (10.30)
\end{align*}$$

where

$$M_{Ht} = \begin{bmatrix} \alpha^{-1} & -\alpha^{-1} \theta \partial_z \\ \partial_z \theta \alpha^{-1} & -\alpha^{-1} \theta \partial_z \theta \alpha^{-1} \theta \partial_z - \partial_z \eta \partial_z \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \partial_z \theta & 1 \end{bmatrix} \begin{bmatrix} \alpha^{-1} & 0 \\ 0 & -\partial_z \eta \partial_z \end{bmatrix} \begin{bmatrix} 1 & \theta \partial_z \\ 0 & 1 \end{bmatrix}. \quad (10.31)$$

To obtain the first order equations with respect to $z$ we pick the pair $p_z, q$. We use equations (10.25) and (10.20) for respectively $\partial_z p_z$ and $\partial_z q$. We eliminate $p_t$ in (10.25) by using its representation (10.17) getting the system

$$\begin{align*}
\partial_z p_z &= -\partial_t p_t = -\partial_t (\alpha \partial_t q + \theta \partial_t q) = -\partial_t \alpha \partial_t q - \partial_t \eta \theta^{-1} (\theta \alpha \partial_t q - p_z), \quad (10.32) \\
\partial_z q &= \eta^{-1} (\theta \partial_t q - p_z). \quad (10.33)
\end{align*}$$

Observe that we used equation (10.33) to get the right-hand side of equation (10.32). The above system can be written as

$$\begin{align*}
\partial_z \begin{bmatrix} p_z \\ q \end{bmatrix} &= \begin{bmatrix} \partial_t \theta \eta^{-1} & -\partial_t \alpha \partial_t - \partial_t \theta \eta^{-1} \theta \partial_t \\ -\eta^{-1} & \eta^{-1} \theta \partial_t \end{bmatrix} \begin{bmatrix} p_z \\ q \end{bmatrix}. \quad (10.34)
\end{align*}$$

The system (10.34) can be transformed into the following canonical Hamiltonian form

$$\begin{align*}
\partial_z V &= JM_{Hz} V, \quad V = \begin{bmatrix} p_z \\ q \end{bmatrix}, \quad (10.35)
\end{align*}$$
where

\[ M_{Hz} = \begin{bmatrix} -\eta^{-1} & \eta^{-1} \theta \partial_t \\ -\partial_t \eta^{-1} & \partial_t \alpha \partial_t + \partial_t \theta \eta^{-1} \theta \partial_t \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \partial_t \theta & 1 \end{bmatrix} \begin{bmatrix} -\eta^{-1} & 0 \\ 0 & \theta \eta^{-1} \partial_t \theta \end{bmatrix} \begin{bmatrix} 1 & -\theta \partial_t \\ 0 & \partial_t \end{bmatrix}. \] (10.36)

Comparing expressions (10.31) and (10.36) we observe a noticeable difference in signs that is explained by the difference in signs in the expressions for the kinetic and potential energies in the Lagrangian density defined by (10.13).

We can transform the system (10.35)-(10.36) further yet into another form intimately related to the energy conservation law. For that we begin with the identity

\[ M_{Hz} = \begin{bmatrix} -\eta^{-1} & \eta^{-1} \theta \partial_t \\ -\partial_t \eta^{-1} & \partial_t \alpha \partial_t + \partial_t \theta \eta^{-1} \theta \partial_t \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \partial_t \theta & 1 \end{bmatrix} \begin{bmatrix} -\eta^{-1} & 0 \\ 0 & \theta \eta^{-1} \partial_t \theta \end{bmatrix} \begin{bmatrix} 1 & -\theta \partial_t \\ 0 & \partial_t \end{bmatrix}. \] (10.37)

Based on (10.37), the system (10.35)-(10.36) can be recast into the following "Hamiltonian" form

\[ \tilde{J} \partial_t V = i \partial_t \tilde{M} V, \quad V = \begin{bmatrix} p_z \\ \partial_t q \end{bmatrix}, \] (10.38)

where

\[ \tilde{J} = \begin{bmatrix} 0 & i \mathbf{1} \\ i \mathbf{1} & 0 \end{bmatrix}, \quad \tilde{M} = \begin{bmatrix} -\eta^{-1} & \eta^{-1} \theta \\ \theta \eta^{-1} & -\alpha - \theta \eta^{-1} \theta \end{bmatrix}. \] (10.39)

When deriving the Hamiltonian equation (10.38)-(10.39) we used the following identity relating \( \tilde{J} \) and \( J \) defined in (10.29)

\[ \begin{bmatrix} 1 & 0 \\ 0 & \partial_t \end{bmatrix} J \begin{bmatrix} 1 & 0 \\ 0 & -\partial_t \end{bmatrix} = -i \partial_t \begin{bmatrix} 0 & i \mathbf{1} \\ \mathbf{1} & 0 \end{bmatrix} = -i \partial_t \tilde{J} \] (10.40)

Note that the matrices \( \tilde{J} \) and \( \tilde{M} \) are respectively antihermitian and hermitian, that is

\[ \tilde{J}^* = -\tilde{J}, \quad \tilde{M}^* = \tilde{M}. \] (10.41)

Notice also that the definitions of \( V \) and \( \tilde{J} \) in (10.38)-(10.39) imply the identity

\[ V^* \tilde{J} V = i [p_z^* \partial_t q + (\partial_t q)^* p_z] = 2i \text{Re} \{ (\partial_t q)^* p_z \}, \] (10.42)

which via the theory of Hamiltonian equations can be associated with the energy conservation law as we show below.

### 10.3 Canonical and Hamilton equations

In this section we provide a concise review of canonical and Hamilton equations following [YakSta1, II.3.1-4]. By **canonical** we call an equation of the form

\[ \tilde{J} \frac{dz}{dt} = \tilde{H} (t) z, \] (10.43)
where $\tilde{H}(t)$ is a $2n \times 2n$ symmetric matrix valued function with real entries and $\tilde{J}$ is a constant $2n \times 2n$ nondegenerate skew-symmetric matrix with real entries, that is

$$\tilde{H}^T(t) = \tilde{H}(t), \quad \tilde{J}^T = -\tilde{J}, \quad |\tilde{J}| \neq 0. \quad (10.44)$$

The matrix $\tilde{H}(t)$ in (10.43) is a called "Hamiltonian" of the equation. A standard form of $2n \times 2n$ nondegenerate skew-symmetric matrix $J$ is

$$J_{2n} = \begin{bmatrix} 0 & -1_n \\ 1_n & 0 \end{bmatrix}. \quad (10.45)$$

The canonical equation (10.43) can be always reduced to the special form

$$J_{2n} \frac{dx}{dt} = H(t)x, \quad (10.46)$$

by means of a linear change of variables, i.e.

$$x = Sz, \quad \tilde{J} = S^T J_{2n} S, \quad \tilde{H}(t) = S^T H(t) S \quad (10.47)$$

for some real nondegenerate $2n \times 2n$ matrix $S$.

We call an equation Hamiltonian if it is of the form (10.43) and (i) $\tilde{H}(t)$ is a Hermitian matrix with complex valued entries; (ii) $\tilde{J}$ is a constant nondegenerate antihermitian matrix, that is

$$\tilde{H}^*(t) = \tilde{H}(t), \quad \tilde{J}^* = -\tilde{J}, \quad |\tilde{J}| \neq 0. \quad (10.48)$$

Canonical equations are of course Hamiltonian. A Hamiltonian equation (10.43) can be always reduced by a transformation $x = Sz$ with a nondegenerate $S$ to the following special form

$$-iG_0 \frac{dx}{dt} = H_0(t)x, \quad (10.49)$$

where $H_0(t)$ is a Hermitian matrix and

$$G_0 = \begin{bmatrix} 1_p & 0 \\ 0 & -1_q \end{bmatrix}, \text{ where } p + q = 2n. \quad (10.50)$$

Any matrix solution $Z(t)$ to the Hamiltonian equation (10.43) satisfies the identity, [YakSta1, II.3.4]

$$Z(t)^* \tilde{J} Z(t) = \tilde{J}, \quad (10.51)$$

and for any two vector solutions $z_1(t)$ and $z_2(t)$ there holds

$$\left( z_1(t), \tilde{J}z_2(t) \right) = [z_1(t)]^* \tilde{J} z_2(t) = \text{constant}. \quad (10.52)$$

(so called Poincaré invariant).
10.4 Energy exchange between subsystems

In this section we derive a general formula for the energy flux between two systems constituting a closed conservative system described by the Lagrangian $\mathcal{L} = \mathcal{L}(q_t, q_z)$. With the MTLB Lagrangian in mind let us put $q = (Q, q)$ and assume that $\mathcal{L}$ can be split as

$$\mathcal{L} = \mathcal{L}_1 (\partial_t Q, \partial_z Q) + \mathcal{L}_2 (\partial_t q, \partial_z q), \quad (10.53)$$

where

$$\partial_z Q = \partial_z Q + B \partial_z q$$

and $B$ is a fixed matrix. The Lagrangian $\mathcal{L}$ of the general form $(10.53)$ describes two coupled interacting systems. The special form of coupling via the modified derivative $\partial_z Q$ in $(10.53)$ resembles the minimal coupling in the charge gauge theory. The variable $q$ plays the role of the gauge field potential and $B$ plays the role of coupling constant.

The corresponding Euler-Lagrange equations are $(10.54)$, $(10.55)$

$$\partial_t \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} + \partial_z \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} = 0, \quad (10.54)$$

$$\partial_t \frac{\partial \mathcal{L}_2}{\partial \partial_t q} + \partial_z \left[ \frac{\partial \mathcal{L}_2}{\partial \partial_z q} + \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} B \right] = 0, \quad (10.55)$$

where the derivative $\frac{\partial \mathcal{L}}{\partial Q}$ of the scalar function $\mathcal{L}$ with respect to a column-vector $Q$ is understood as a row-vector of the same dimension.

Recall now that the energy conservation law for the entire system has the form, [GelFom, 38.2-3], [Gold, 13.7]

$$\partial_t H + \partial_z S = 0, \quad (10.56)$$

where $H$ and $S$ are the energy and energy flux densities defined by

$$H = H_1 + H_2, \quad S = S_1 + S_2, \quad (10.57)$$

with the following expressions for the individual energies and energy fluxes

$$H_1 = \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} \partial_t Q - \mathcal{L}_1 (\partial_t Q, \partial_z Q), \quad S_1 = \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} \partial_t Q, \quad (10.58)$$

$$H_2 = \frac{\partial \mathcal{L}_2}{\partial \partial_t q} \partial_t q - \mathcal{L}_2 (\partial_t q, \partial_z q), \quad S_2 = \left[ \frac{\partial \mathcal{L}_2}{\partial \partial_z q} + \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} B \right] \partial_t q. \quad (10.59)$$

The above expressions imply the following identities for the first system

$$\partial_t H_1 = \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} \partial_t^2 Q + \partial_t \left( \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} \right) \partial_t Q - \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} \partial_t^2 Q - \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} \left( \partial_t^2 Q + B \partial_t^2 q \right) = \partial_t \left( \frac{\partial \mathcal{L}_1}{\partial \partial_t Q} \right) \partial_t Q - \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} \left( \partial_t^2 Q + B \partial_t^2 q \right), \quad (10.60)$$

$$\partial_z S_1 = \partial_z \left( \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} \right) \partial_t Q + \frac{\partial \mathcal{L}_1}{\partial \partial_z Q} \partial_t^2 Q. \quad (10.61)$$
The equations (10.60), (10.61), combined with the Euler-Lagrange equations (10.54), yield the following energy conservation law for the first system

\[
\partial_t H_1 + \partial_z S_1 = -\frac{\partial L_1}{\partial \partial_z Q} B \partial^2_{tz} q, \tag{10.62}
\]

where the right-hand side of (10.62) can be interpreted as the power flow density from the second system into the first one.

Carrying out similar computations for the second system we obtain

\[
\partial_t H_2 = \frac{\partial L_2}{\partial \partial_t q} \partial^2_{tz} q + \partial_t \left( \frac{\partial L_2}{\partial \partial_t q} \partial_t q - \frac{\partial L_2}{\partial \partial_z q} \partial^2_{tz} q \right) = \partial_t \left( \frac{\partial L_2}{\partial \partial_t q} \partial_t q - \frac{\partial L_2}{\partial \partial_z q} \partial^2_{tz} q, \tag{10.63}
\right.
\]

\[
\partial_z S_2 = \partial_z \left[ \frac{\partial L_2}{\partial \partial_z q} + \frac{\partial L_1}{\partial \partial_z Q} B \right] \partial_t q + \left[ \frac{\partial L_2}{\partial \partial_z q} + \frac{\partial L_1}{\partial \partial_z Q} B \right] \partial^2_{tz} q. \tag{10.64}
\]

Combining equations (10.63) and (10.64) with the Euler-Lagrange equations (10.55) for the second system we obtain the following conservation law

\[
\partial_t H_2 + \partial_z S_2 = \frac{\partial L_1}{\partial \partial_z Q} B \partial^2_{tz} q, \tag{10.65}
\]

where the right-hand side of (10.65) can be interpreted as the power density flow transferred from the first system into the second one.

Notice that relations (10.62) and (10.65) have right-hand sides of the same magnitude and opposite signs. This can be viewed as a manifestation of the conservation of energy for the entire system. Indeed we recover (10.56) by adding (10.62) and (10.65).

10.5 Amplification for homogeneous MTLB systems: proofs.

This section contains rigorous formulations and proofs of the assertions made in Section 7.

**Theorem 10.2** Let the hypotheses in Theorem 7.1 hold. Then, there is a unique pair of complex conjugate solutions \( v_0, v_0^* \) of the equation \( |\tilde{\Lambda}(v)| = 0 \), where \( \tilde{\Lambda}(v) \) is defined in (7.3), (7.6).

**Proof.** By our assumption, the equation \( |A(v)| = |v^2L + C^{-1}| = 0 \) has exactly \( 2n \) real roots, \( \pm v_1, \pm v_2, ... \pm v_n \), with \( v_i > 0 \) \( (\lambda_i = v_i^2) \). We assume in what follows that they are ordered: \( 0 < v_1 \leq v_2 \leq ... \leq v_n \) and each root is repeated a number of times equal to its multiplicity. If \( |A(v)| \neq 0 \), the following decomposition holds

\[
|\tilde{\Lambda}(v)| = |A(v)| \left[ d - \xi (v - u_0)^2 - D^T (A(v))^{-1} D \right]. \tag{10.66}
\]

This follows from the following more general fact: if \( M \) is a square block matrix of the form

\[
M = \begin{bmatrix}
A_1 & A_2 \\
A_3 & A_4
\end{bmatrix},
\]

34
where $A_1, A_4$ are square matrices with $|A_1| \neq 0$, then

$$|M| = |A_1| \left| A_4 - A_3 A_1^{-1} A_2 \right|,$$

see e.g. [Bern, Lemma 2.8.6, page 108]. Observe that in our case $A_2 = D$ is a column matrix and $A_3 = D^T$ is a row matrix. Then, if $|A(v)| \neq 0$, $v$ is a root of $|A(v)| = 0$ if and only if it is a root of the equation

$$-\xi(v - u_0)^2 = D^T (A(v))^{-1} D - d =: R(v).$$

The function $R(v)$ above turns out to have very nice properties. A well known fact from linear algebra concerning simultaneous diagonalization of two quadratic forms, one of which is positive, assures that there exists a non-degenerate matrix $P$ such that

$$P^T A(v) P = \text{diag} \left( v \right) := \begin{bmatrix} v_1^2 - v^2 & 0 & \cdots & 0 \\ 0 & v_2^2 - v^2 & \cdots & 0 \\ \vdots & \vdots & \ddots & 0 \\ 0 & 0 & 0 & v_n^2 - v^2 \end{bmatrix}.$$ 

Consequently,

$$D^T (A(v))^{-1} D = \tilde{D}^T \begin{bmatrix} 1 & 0 & \cdots & 0 \\ \frac{1}{v_1^2 - v^2} & 0 & \cdots & 0 \\ 0 & \frac{1}{v_2^2 - v^2} & \cdots & 0 \\ \vdots & \vdots & \ddots & 0 \\ 0 & 0 & 0 & \frac{1}{v_n^2 - v^2} \end{bmatrix} \tilde{D} = \sum_{i=1}^{n} \frac{\tilde{D}_i^2}{v_i^2 - v^2},$$

where $\tilde{D} = P^T D$. Therefore,

$$R(v) = \sum_{i=1}^{n} \frac{\tilde{D}_i^2}{v_i^2 - v^2} - d$$

is a rational function defined on the set $\{v : |A(v)| \neq 0\}$. It is immediately seen that $R$ is an even function, exhibiting vertical asymptotes at $v = \pm v_i$ if at least one of the $\tilde{D}_k$ associated to $v_i$ does not vanish ($v_i$ may be a multiple root). For $v > 0$, each branch between two consecutive asymptotes is increasing and they are decreasing for $v < 0$. Moreover, $\lim_{v \to \infty} R(v) = -d$. Also,

$$R(0) + d = D^T (A(0))^{-1} D = D^T C D = \sum_{i=1}^{n} \sum_{j=1}^{n} C_{ij} D_i D_j = \sum_{i=1}^{n} \sum_{j=1}^{n} C_{ij} \left[ \sum_{k=1}^{n} (C^{-1})_{ik} \right] \left[ \sum_{r=1}^{n} (C^{-1})_{jr} \right] = \sum_{k=1}^{n} \sum_{r=1}^{n} \left[ \sum_{i=1}^{n} (C^{-1})_{ik} \sum_{j=1}^{n} C_{ij} (C^{-1})_{jr} \right] = \sum_{k=1}^{n} \sum_{r=1}^{n} \left[ \sum_{i=1}^{n} (C^{-1})_{ik} \delta_{ir} \right] = \sum_{k=1}^{n} \sum_{r=1}^{n} (C^{-1})_{rk} = \sum_{k=1}^{n} D_k = d;$$

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hence $R(0) = 0$. Since $C^{-1}$ is non-degenerate, $D \neq 0$. Moreover, since the matrix $P$ is non-degenerate, we have $\tilde{D} \neq 0$. Therefore, the graph has at least two vertical asymptotes and always exhibits a central symmetric branch with the minimum at the point $(0,0)$. The number of real roots of the equation

$$-\xi(v - u_0)^2 = R(v)$$

is the number of intersection points of the parabola $y = f(v) := -\xi(v - u_0)^2$ and the graph of $R$. For $\xi$ small, it is exactly the number of monotonic branches (all branches, except for the central one), which coincides with the number of asymptotes. This number is always between 2 and $2n$, depending on the number of vanishing $\tilde{D}_i$ and on the possible multiple roots; a precise description is given below. Moreover, it is easily seen that whenever $u_0 \in (0, v_1]$, the number of intersection points is equal to the number of asymptotes irrespective of the value of $\xi > 0$; see Figure 7.1 (a), whereas $\xi$ small is needed otherwise; indeed, in Figure 7.1 (b) a large value of $\xi$ produces three points of intersection with the far right branch of the graph of $R$, making the total number of intersection points exceed by two the number of asymptotes. If either (i) or (ii) holds, the intersections are transversal, hence the roots are simple. The previous assertions follow easily and rigorously from the monotonicity properties of $R$ and $f$ but their clear geometric meaning makes a lengthy proof unnecessary.

So far, we have considered the real roots of the equation $|\tilde{A}(v)| = 0$ in the set $\{v : \det A(v) \neq 0\}$. Next, we consider the possible roots of the equation in the complementary set $\{\pm v_1, \pm v_2, \ldots, \pm v_n\}$. Multiplying the matrix $\tilde{A}(v)$ by $\tilde{P}^T$ from the left and by $\tilde{P}$ from the right, where

$$\tilde{P} = \begin{bmatrix} P & 0 \\ 0 & 1 \end{bmatrix},$$

there follows that the equation $|\tilde{A}(v)| = 0$ is equivalent to the equation

$$\Delta(v) := \begin{vmatrix} v_1^2 - v^2 & 0 & \cdots & 0 \\ 0 & v_2^2 - v^2 & \cdots & : \\ \vdots & \vdots & \ddots & \vdots \\ 0 & \cdots & 0 & v_n^2 - v^2 \end{vmatrix} \tilde{D} = 0,$$

$$\tilde{D}^T \begin{vmatrix} d - \xi(v - u_0)^2 \\ v_1^2 - v^2 \\ \vdots \\ v_n^2 - v^2 \end{vmatrix} = 0,$$

where, as before, $\tilde{D} = P^T D$. Let us analyze under what condition $\pm v_1$ are roots of the equation $\Delta(v) = 0$. Expanding the determinant with respect to the last column, and then the $n$-th order minor corresponding to $\tilde{D}_i$ with respect to its $i$-th column, we get the expression

$$\Delta(v) = -\tilde{D}_1^2 \begin{vmatrix} v_2^2 - v^2 & 0 & \cdots & 0 \\ 0 & v_3^2 - v^2 & \cdots & : \\ \vdots & \vdots & \ddots & \vdots \\ 0 & \cdots & 0 & v_n^2 - v^2 \end{vmatrix} \begin{vmatrix} v_1^2 - v^2 & 0 & \cdots & 0 \\ 0 & v_3^2 - v^2 & \cdots & : \\ \vdots & \vdots & \ddots & \vdots \\ 0 & \cdots & 0 & v_n^2 - v^2 \end{vmatrix} = -\tilde{D}_2^2 \begin{vmatrix} d - \xi(v - u_0)^2 \\ v_1^2 - v^2 \\ \vdots \\ v_n^2 - v^2 \end{vmatrix} = -\tilde{D}_n^2 \begin{vmatrix} v_1^2 - v^2 & 0 & \cdots & 0 \\ 0 & v_2^2 - v^2 & \cdots & : \\ \vdots & \vdots & \ddots & \vdots \\ 0 & \cdots & 0 & v_{n-1}^2 - v^2 \end{vmatrix} + \begin{vmatrix} d - \xi(v - u_0)^2 \\ v_1^2 - v^2 \\ \vdots \\ v_n^2 - v^2 \end{vmatrix}$$

$$(10.67)$$
that is,

\[ \Delta(v) = \prod_{i=1}^{n} (v_i^2 - v^2) \left[ d - \xi(v - u_0)^2 \right] - \sum_{i=1}^{n} \tilde{D}_i^2 \prod_{j \neq i} (v_j^2 - v^2). \]  

(10.68)

We note in passing that the factorization (10.66) is easily obtained from the above expression by extracting the factor

\[ |A(v)| = \prod_{i=1}^{n} (v_i^2 - v^2) \]

under the assumption \( v \neq v_i \).

Assume first that \( \pm v_i \) are simple roots of \( |A(v)| = 0 \), that is, that the binomial \( v_i^2 - v^2 \) appears only once in the matrix diag\( (v) \). Then, there follows that \( \Delta(v_i) = 0 \) if and only if \( \tilde{D}_i = 0 \). Whenever this condition holds, the partial fraction \( \tilde{D}_i^2/(v_i^2 - v^2) \) in the expression of \( R \) disappears and the number of asymptotes is reduced by two. The number of real roots is thus increased by two \((\pm v_i)\) and reduced by two, leaving the total number of roots unaffected.

Let us next consider the case of a multiple root. Assume that \( v_i = v_{i+1} = \ldots = v_{i+k-1} \), hence the binomial \( v_i^2 - v^2 \) appears \( k \) times in diag\( (v) \), \( k > 1 \). Then \( \pm v_i \) are necessarily roots of \( \Delta(v) = 0 \), as it can be readily seen from (10.68). As for their multiplicity, there are two cases:

a) multiplicity = \( k \), if all of \( \tilde{D}_i, \tilde{D}_{i+1}, \ldots \tilde{D}_{i+k-1} \) are zero, since in this case all non-zero terms in (10.68) contain \( k \) times the factor \( v_i^2 - v^2 \);

b) multiplicity = \( k - 1 \), if not all of \( \tilde{D}_i, \tilde{D}_{i+1}, \ldots \tilde{D}_{i+k-1} \) are zero, since in this case the terms in (10.68) corresponding to the non-zero \( \tilde{D}_r \) with \( r \in \{i, i+1, \ldots i+k-1\} \) contain the factor \( v_i^2 - v^2 \) only \( k - 1 \) times, while the rest contain it \( k \) times.

In case (a), all the fractions with denominator \( v_i^2 - v^2 \) are missing in the rational function \( R \), with consequent reduction of the number of asymptotes (with respect to the total possible number \( 2n \)) by \( 2k \), which is precisely the number of additional roots, counting their multiplicity. Thus the total number of real roots is unaffected.

In case (b), there is one fraction with denominator \( v_i^2 - v^2 \). Thus the total number of asymptotes is reduced by \( 2k - 2 \), which is the number of additional roots, counting their multiplicity.

Summing up, the total number of real roots of \( \Delta(v) = 0 \) (counting their multiplicity) is exactly \( 2n \) under our assumptions. Since the total number of roots of \( \Delta(v) = 0 \) is \( 2n + 2 \), there is necessarily one and only one pair of complex conjugate roots, thus proving the assertion.

The following theorem deals with the behavior of amplification as \( \xi \to 0 \) and as \( \xi \to \infty \).

**Theorem 10.3** Let \( v_0, \overline{v_0} \) with \( \text{Im} v_0 > 0 \) denote the unique pair of complex conjugate roots of the equation \( \overline{A}(v) = 0 \) under the assumptions of Theorem 7.1. Let \( k_0 = \omega/v_0 \). Then,

\[ - \text{Im} k_0 \sim \frac{C}{\sqrt{\xi}} \quad \text{as} \ \xi \to 0, \ C > 0. \]  

(10.69)

Under the additional assumption \( u_0 = v_1 \) we also have

\[ - \text{Im} k_0 \sim \frac{C'}{\sqrt{\xi}} \quad \text{as} \ \xi \to \infty, \ C' > 0. \]  

(10.70)
The coefficients $a_{2n+2}, a_{2n+1}$ and $a_0$ can be easily computed in terms of the parameters. Indeed, $a_0 = |\tilde{A}(0)|$, which can be computed by adding the first $n$ rows of $\tilde{A}(0)$ and subtracting the result from the last. Recalling that $D_i = \sum_j (C^{-1})_{ij}$ and that $d = \sum_i D_i$, we obtain

$$a_0 = |\tilde{A}(0)| = \begin{vmatrix} C^{-1} & D \\ 0 & -\xi u_0^2 \end{vmatrix} = -\xi u_0^2 |C^{-1}|.$$ 

The only addends in $|\tilde{A}(v)|$ yielding powers $v^{2n+2}$ or $v^{2n+1}$ are those coming from the product $|-v^2 L + C^{-1}| [d - \xi (v - u_0)^2]$. Clearly, the relevant terms are

$$(-1)^n |L| v^{2n} [d - \xi (v - u_0)^2] + \ldots = (-1)^{n+1} \xi |L| v^{2n+2} + 2(-1)^n \xi u_0 |L| v^{2n+1} + \ldots$$

where the dots stand for lower order in $v$ terms. Consequently,

$$a_{2n+2} = (-1)^{n+1} \xi |L| ; \quad a_{2n+1} = 2(-1)^n \xi u_0 |L|.$$ 

Vieta’s formulas then imply

$$2 \text{Re} \, v_0 + \sum_{i=1}^n v_i^+ - \sum_{i=1}^n v_i^- = -\frac{a_{2n+1}}{a_{2n+2}} = 2u_0$$

(10.71)

$$(-1)^n |v_0|^2 \prod_{i=1}^n v_i^+ v_i^- = \frac{a_0}{a_{2n+2}} = (-1)^n \frac{u_0^2 |C^{-1}|}{|L|} = (-1)^n \frac{u_0^2}{|LC|}.$$ 

(10.72)

Next, we study the behavior as $\xi \to 0$ of both the sum and the product of the real roots. In the asymptotic formulas below, $K_1, K_2, K_1', K_2'$ etc. denote positive constants depending on $L, C, u_0$ but not on $\xi$. 

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Let \( n > 1 \) and suppose that the graph of \( R \) has more than two asymptotes. First of all, we note that, as \( \xi \to 0 \), the parabola becomes flat and the roots \( v_i^+ - v_i^- \) with \( i = 1, 2, \ldots, n - 1 \) become symmetric due to the symmetry of the graph of \( R \). More precisely, if we denote by \( \hat{v}_k^+, \hat{v}_k^- \) with \( k \in \{1, 2, \ldots, n - 1\} \) the abscissas of the points on the \( k \)-th right (respectively, \( k \)-th left) branch of the graph of \( R \) for which \( R(\hat{v}_k^+) = R(\hat{v}_k^-) = 0 \), then clearly \( v_k^+(\xi) \to \hat{v}_k^+ \) and \( v_k^-(\xi) \to \hat{v}_k^- \). Moreover, since the branches of \( R \) are strictly increasing for \( v > 0 \) and strictly decreasing for \( v < 0 \), \( v_k^+(\xi) - \hat{v}_k^+ \sim -A_k \xi, -v_k^-(\xi) + \hat{v}_k^- \sim B_k \xi \) as \( \xi \to 0 \), with \( A_k, B_k > 0 \). We also note the following fact, which is used in the proof of Section 8.1 and is a simple consequence of the lack of symmetry of the parabola \( y = -\xi(v - u_0)^2 \) with respect to the vertical axis: if \( v_k^+ - v_k^- \) is a pair of real roots not belonging to the set \( \{\pm v_1, \pm v_2, \ldots, \pm v_n\} \) (and there is at least one such pair, see the proof of Theorem 7.1), then

\[
v_k^+(\xi) - v_k^-(\xi) > 0. \tag{10.73}\]

Thus in particular \( B_k > A_k \) in the above asymptotic relations. This inequality can be easily seen on the graph and given a simple analytical proof.

The roots belonging to the set \( \{\pm v_1, \pm v_2, \ldots, \pm v_n\} \) are symmetric and do not contribute to their sum. Therefore,

\[
\sum_{i=1}^{n-1} v_i^+(\xi) - \sum_{i=1}^{n-1} v_i^-(\xi) = K_1 \xi + o(\xi) \text{ as } \xi \to 0. \tag{10.74}
\]

As for the product of roots, we have

\[
\prod_{i=1}^{n-1} v_i^+(\xi)v_i^-(\xi) = (-1)^n K_2 + K_3 \xi + o(\xi) \text{ as } \xi \to 0. \tag{10.75}
\]

If there are only two asymptotes, then \( v_1 = v_2 = \ldots = v_{n-1} \) and the left-hand side in (10.74) is zero. Also, the left-hand side in (10.75) is the constant \((-1)^n K_2\). Thus this case can be formally included in (10.74) and (10.75) by allowing \( K_1 \) and \( K_3 \) to vanish.

Let us now consider the extreme roots. As we noted, \( v_n^+, v_n^- \to +\infty \). More precisely, since we have

\[
-\xi(v_n^+ - u_0)^2 = R(v_n^+) \to -d \text{ as } \xi \to 0, \tag{10.76}
\]

then necessarily \( \lim_{\xi \to 0} \xi(v_n^+ - u_0)^2 = d > 0 \) and thus

\[
v_n^+(\xi) = \sqrt{\frac{d}{\xi}} + u_0 + E(\xi), \text{ where } E(\xi) = o\left(\frac{\sqrt{1}}{\xi}\right) \text{ as } \xi \to 0. \tag{10.77}
\]

We need further refinement in the asymptotics of \( E(\xi) \) as \( \xi \to 0 \). To this end, we recall that

\[
R(v) + d \sim -\frac{A}{v^2} \text{ as } v \to \infty, \ \ A > 0. \tag{10.78}
\]

Replacing \( v_n^+(\xi) \) in (10.76) by the expression (10.77) and using (10.78), we arrive at

\[
2E(\xi)\sqrt{d/\xi} + E(\xi)^2 \to A \text{ as } \xi \to 0,
\]
which implies
\[ E(\xi) = K_3\sqrt{\xi} + o(\sqrt{\xi}) \quad \text{as} \quad \xi \to 0. \]

Summing up, we have the following asymptotic representation for \( v_n^+ \):
\[ v_n^+(\xi) = \sqrt{\frac{d}{\xi}} + u_0 + K_3\sqrt{\xi} + o(\sqrt{\xi}) \quad \text{as} \quad \xi \to 0. \]  
(10.79)

An analogous representation takes place for \( v_n^- \):
\[ -v_n^-(\xi) = -\sqrt{\frac{d}{\xi}} + u_0 - K_3\sqrt{\xi} + o(\sqrt{\xi}) \quad \text{as} \quad \xi \to 0. \]  
(10.80)

Plugging (10.79), (10.80), (10.74) and (10.75) into (10.71) and (10.72) yields
\[ \Re v_0 = o(\sqrt{\xi}) \quad ; \quad |v_0|^2 = K_4\xi + o(\sqrt{\xi}) \quad \text{as} \quad \xi \to 0. \]  
(10.81)

As a consequence,
\[ \Im v_0 = \sqrt{K_4\xi + o(\xi)} = \sqrt{K_4\sqrt{\xi} + o(\sqrt{\xi})} \quad \text{as} \quad \xi \to 0 \]
and, finally,
\[ -\Im k_0 = \frac{\Im v_0}{|v_0|^2} \sim \frac{K_5}{\sqrt{\xi}} \quad \text{as} \quad \xi \to 0, \]
thus proving (10.69) for \( n > 1 \). If \( n = 1 \), (10.79) and (10.80) hold and plugging into (10.71) and (10.72) yields the same result.

We turn now to the proof of (10.70), restricting ourselves to the case of just one line; the case of several lines can be handled in a similar fashion. First of all, it is clear that \( v_1^+ \downarrow u_0 \) and \( -v_1^- \uparrow -u_0 \) as \( \xi \to \infty \). This can be rigorously proved in a way, similar to the above proof of the fact that \( v_n^+ \uparrow \infty, v_n^- \downarrow -\infty \) as \( \xi \to 0 \). Put
\[ v_1^+(\xi) = u_0 + G(\xi); \quad G(\xi) > 0, \quad G(\xi) \to 0 \quad \text{as} \quad \xi \to \infty. \]  
(10.82)

Near \( u_0 = v_1 \) we have
\[ R(v) \sim \frac{A}{u_0 - v} \quad \text{as} \quad v \to v_0^+ \quad \text{with} \quad A > 0. \]  
(10.83)

After use of (10.82) and (10.83), the equation
\[ -\xi(v_1^+ - u_0)^2 = R(v_1^+) \]
yields the following asymptotic relation:
\[ \xi G(\xi)^3 \sim A \quad \text{as} \quad \xi \to \infty, \]
that is,
\[ G(\xi) \sim \frac{K_1'}{\sqrt{\xi}} \quad \text{as} \quad \xi \to \infty. \]

An analogous formula takes place for the negative root:
\[ v_1^-(\xi) = u_0 + H(\xi) \quad \text{with} \quad H(\xi) \sim \frac{K_2'}{\sqrt{\xi}} \quad \text{as} \quad \xi \to \infty. \]
Therefore,
\[ v_1^+(\xi) - v_1^- (\xi) \sim \frac{K'_3}{\sqrt[3]{\xi}} \text{ as } \xi \to \infty. \]

Applying again (10.71) and (10.72) we obtain
\[
\text{Re } v_0 = u_0 + \frac{K'_4}{\sqrt[3]{\xi}} + o\left( \frac{1}{\sqrt[3]{\xi}} \right), \quad |v_0| \to \frac{1}{\sqrt{LC}} + \frac{K'_5}{\sqrt[3]{\xi}} + o\left( \frac{1}{\sqrt[3]{\xi}} \right) \text{ as } \xi \to \infty. \quad (10.84)
\]

Recall that \( v_1 = 1/\sqrt{LC} = u_0 \). The last two relations imply \( \text{Im } v_0 \sim K'_6/\sqrt[3]{\xi} \). Finally,
\[
- \text{Im } k_0 = \frac{\text{Im } v_0}{|v_0|^2} \sim \frac{K'_7}{\sqrt[3]{\xi}} \text{ as } \xi \to \infty.
\]

as was to be proved. ■

One can also verify that if \( u_0 < 1/\sqrt{LC} \) then both \( \text{Im } v_0 \) and \( \text{Im } k_0 \) have a finite, nonzero limit as \( \xi \to \infty \).

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**References**

[Bar] A. Barut, *Electrodynamics of and Classical Theory of Fields and Particles*, Dover, 1980.

[Bellan] Bellan P., *Fundamentals of Plasma Physics*, Cambridge Univ. Press, 2006.

[Bellm] Bellman R., *Introduction to Matrix Analysis*, 2nd ed., SIAM, 1997.

[Bern] Bernstein, D.S., *Matrix Mathematics: Theory, Facts, and Formulas*, Princeton University Press, 2009.

[DavNP] Davidson R., Physics of Nonneutral Plasmas, World Scientific, 2001.

[FigWeLC] Figotin A. and Welters A., *Lagrangian Treatment of Systems Composed of High-Loss and Lossless Components* (in preparation) 2013.

[FigSch1] A. Figotin and J. Schenker, *Spectral Theory of Time Dispersive and Dissipative Systems*, J. Stat. Phys., 118, Nos. 1/2, 199-263, (2005).

[FigShi1] Figotin A. and Shipman S., *Open Systems Viewed Through Their Conservative Extensions*, J. Stat. Phys., 125, No. 2, 363-413, (2006).

[FigSch2] A. Figotin and J. Schenker, *Hamiltonian Structure for Dispersive and Dissipative Dynamical Systems*, J. Stat. Phys., 128, No. 4, 969-1056, (2007).

[Fox] Fox M., *Quantum Optics - An Introduction*, Oxford Univ. Press, 2006.

[GelFom] Gelfand I. and Fomin S., *Calculus of Variations*, Prentice-Hall, 1963.

[Gilm] Gilmour A. S., *Principles of Traveling Wave Tubes*, Artech House, 1994.
Gilmour A. S., *Principles of Klystrons, Traveling Wave Tubes, Magnetrons, Cross-Field Amplifiers, and Gyrotrons*, Artech House, 2011

Goldstein H. et al., *Classical Mechanics*, 3d ed., Addison Wesley, 2000.

Hartman, P., *Ordinary Differential Equations*, SIAM Classics in Applied Mathematics, 2002.

Hasegawa A., *Plasma Instabilities and Nonlinear Effects*, Springer, 1975.

Kompfner, R., *Travelling-wave Tubes*, Reports on Progress in Physics 15, 275-327 (1952).

E.M. Lifshitz, L. P. Pitaevskii, *Physical Kinetics: Volume 10 (Course of Theoretical Physics)*, Butterworth-Heinemann, 1981.

Loudon R., *The Quantum Theory of Light*, 3rd ed., Oxford Univ. Press, 2000.

Melrose D., Instabilities in Space and Laboratory Plasmas, Cambridge Univ. Press, 1989.

Rund H., *The Hamilton-Jacobi theory in the calculus of variations*, D. Van Nostrand Co., 1966.

Nusinovich G. et al., Modern Microwave and Millimeter-Wave Power Electronics, Wiley, 2005.

Nitsch J. et al., *Radiating nonuniform transmission-line systems and the partial element equivalent circuit method*, Wiley, 2009.

Paul C., *Analysis of Multiconductor Transmission Lines*, 2nd ed., Wiley, 2008.

Pierce J., *Waves in Electron Streams and Circuits*, Bell Sys. Tech. Jour., 30, 626-651, (1951).

Pierce J., *Theory of the beam-type traveling-wave tube*, Proc. of the IRE 35, 111-123, (1947).

Pierce J., *Traveling-Wave Tubes*, D. van Nostrand, 1950.

Pierce J., *Almost All About Waves*, MIT, 1981.

Ramo S., *Currents induced by electron motion*, Proc. of the IRE 27, 584-585 (1939)

Schachter L., *Beam-Wave Interaction in Periodic and Quasi-Periodic Structures*, 2nd ed., Springer, 2011.

Schwinger J. et al., *Classical Electrodynamics*, Perseus Books, 1998.

Siegman A., *Lasers*, University Science Books, 1986.

Tsimring S., *Electron Beams and Microwave Vacuum Electronics*, Wiley, 2007.

Yakubovich V. and Starzhinskij V., *Linear Differential Equations with Periodic Coefficients*, Wiley, 1975.